Operational issues with BPMs and BPIs

Known issues

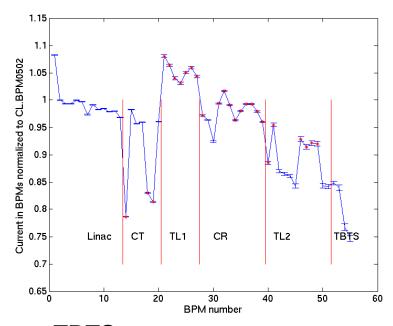
- Inconsistency between BPMs and BPIs
- Response of BPIs is non-linear along the pulse

Note – BPIs in delay loop have different processing electronics

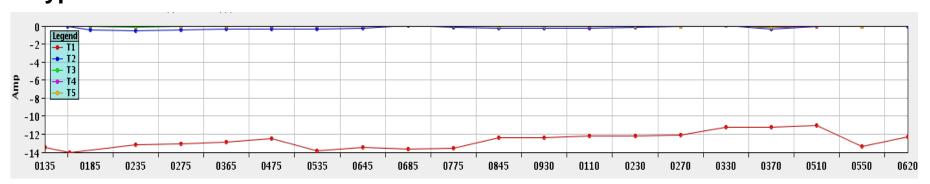
- Diagnostic data taken in December 2011 for delay loop BPIs only
- Will focus on these monitors
- However, conclusions may well be applicable to all BPIs to be confirmed

Example – transmission along machine

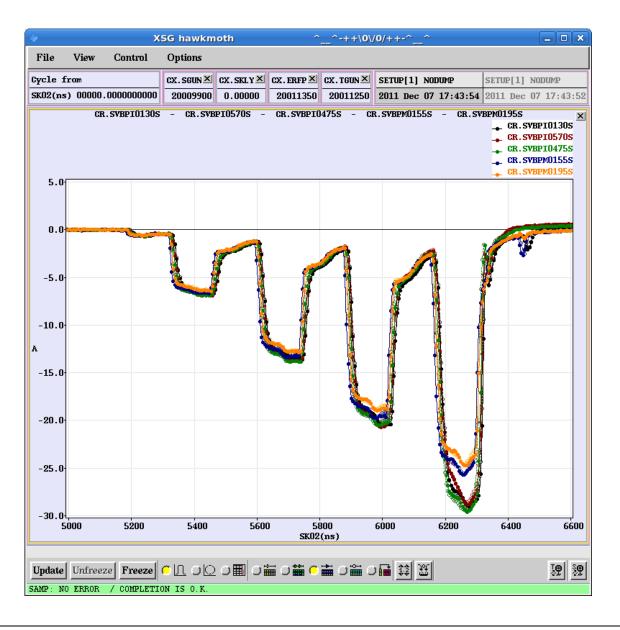
- Tobias 2011, analysis of current losses
 - Red asterisks mark BPIs, other monitors are BPMs

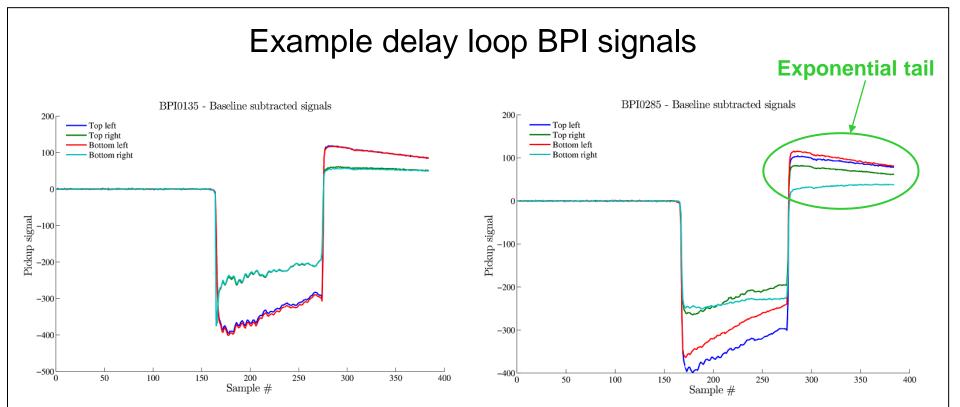


Typical transmission to TBTS



Example - non-linear BPIs vs. BPMs in combiner ring





Observation of delay loop signals December 2011

- Log raw electrode signals (decouple beam offset, current)
- Droop on individual strip signals seen to be approximately exponential
- Strips have different time constants
- Time constants are dependent on the signal level

Offline BPI signal processing

Signal processing is as follows:

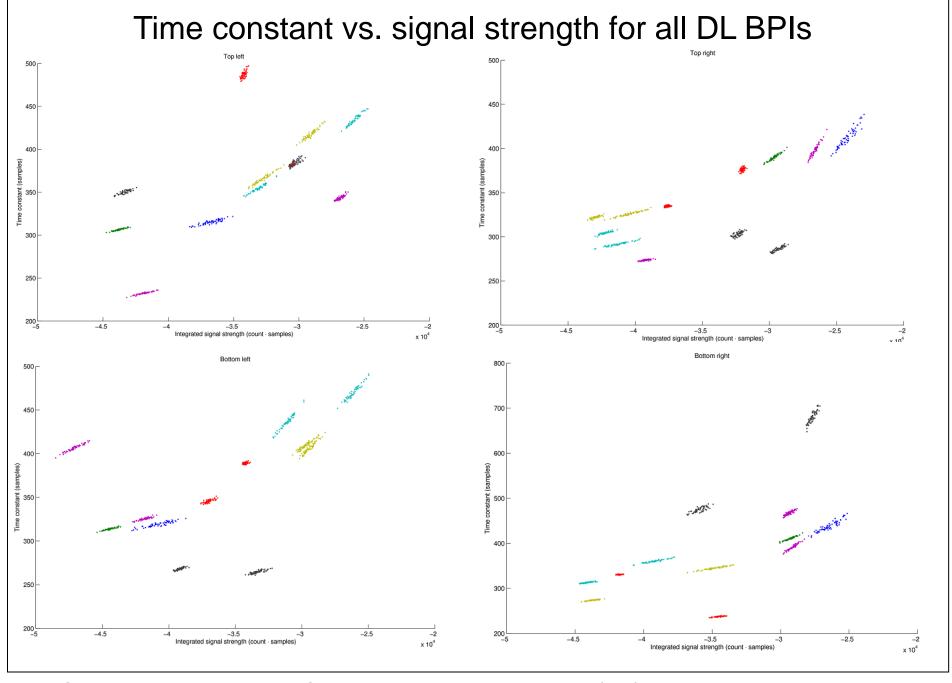
$$\Sigma = V_{TL} + V_{TR} + V_{BL} + V_{BR}$$

$$x = \frac{(V_{TL} + V_{BL}) - (V_{TR} + V_{BR})}{\Sigma}$$

$$y = \frac{(V_{TL} + V_{TR}) - (V_{BL} + V_{BR})}{\Sigma}$$

where the signals have had their pedestal subtracted

- Summation over signals with differing time constants
 - The droop on the current/position measurements is not a pure exponential
 - Any correction should be done on an individual electrode basis
- Used least square fitting to measure decay time constants
 - However, poor fits to signals with larger time constants due to limited range of exponential tail (see next plot)
 - No obvious parameterisation



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Consistency between BPM and BPI current measurement

- Not simply a calibration issue
 - The delay loop BPIs give decaying sum signals
 - Similar non-linearity observed in BPIs elsewhere
 - Typically, signals are averaged over a window
 - Absolute recorded current dependent on window position and size
- Ideally, would like to correct for non-linearity before calibration

Droop correction by IIR filter

Can correct an exponential droop with an IIR filter:

$$V_n' = V_n + \lambda \sum_{i=0}^{n-1} V_i$$

where V_n is the n^{th} sample and $\lambda = 1/\tau$ is the decay constant

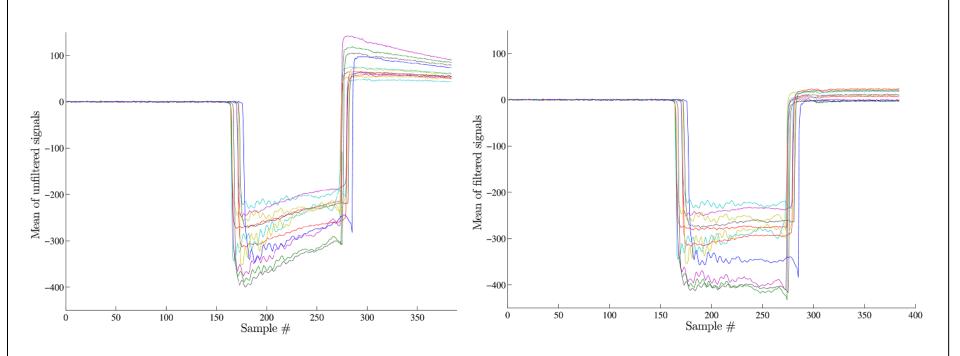
- Pragmatic approach to filtering
 - Least square fit to the signal exponential tail
 - Fit for each electrode on a pulse-to-pulse basis
 - Fast fitting by linearisation and direct calculation of time constant
 - For N samples of the exponential tail at times t_n , the least sq. estimate of τ is:

$$\tau = \frac{\operatorname{var}\left[t_n\right]}{\operatorname{cov}\left[\ln V(t_n), \ t_n\right]}$$

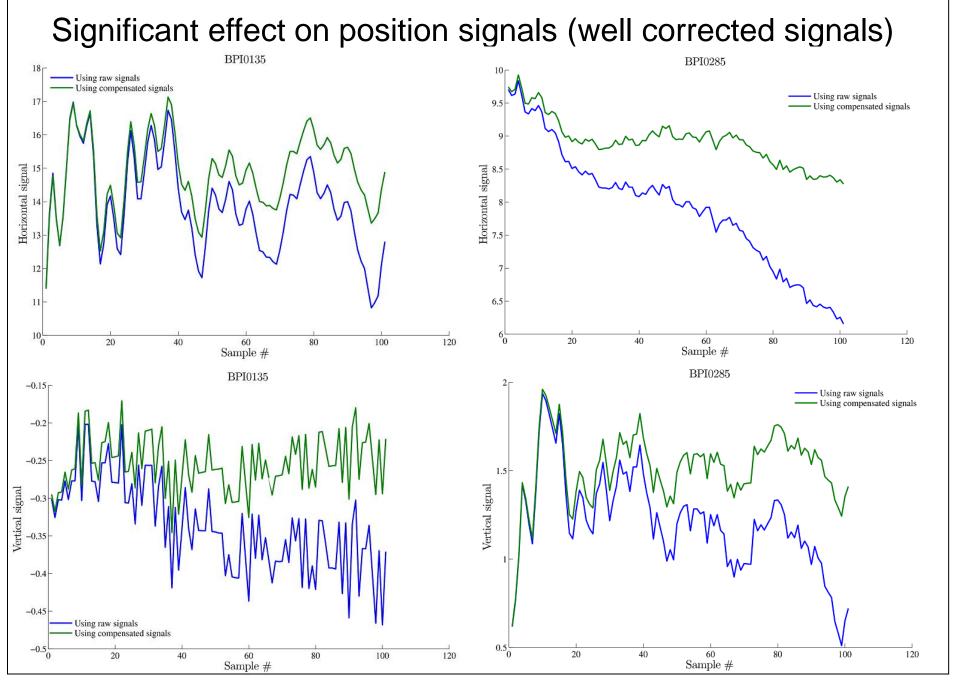
- Algorithm is fast
 - Would expect approximately double total signal processing latency

Offline processing using fast algorithm

Top-left electrode signals for all DL BPIs, averaged over ~50 pulses:



- Smaller time constants are well corrected, larger time constant signals aren't
 - Slowly decaying signals' tails do not go to zero
 - Again, due to insufficient length of exponential tail

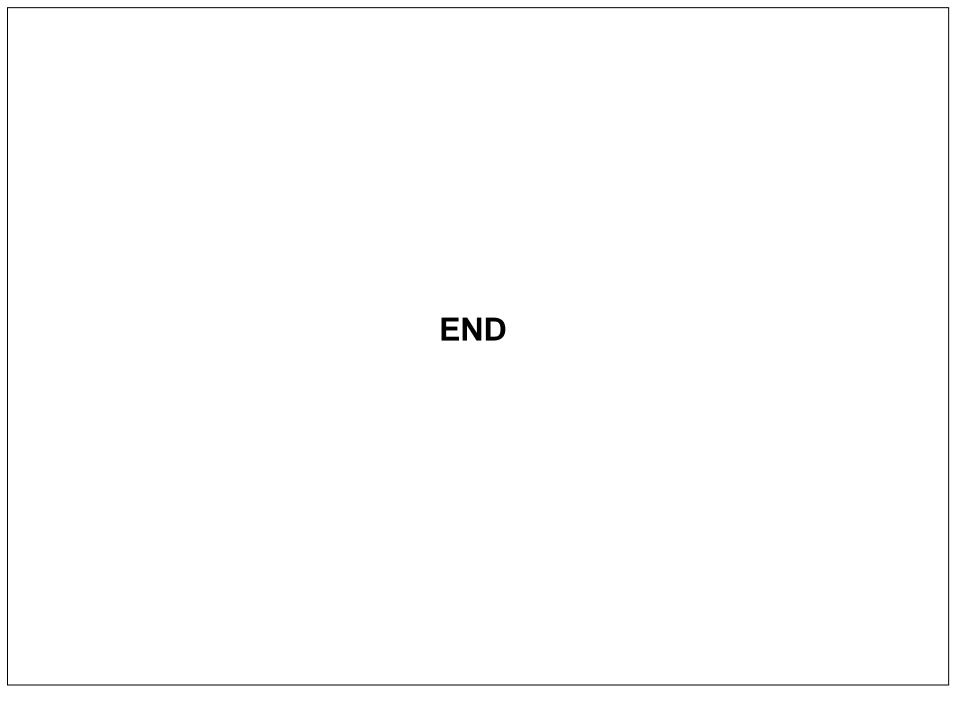


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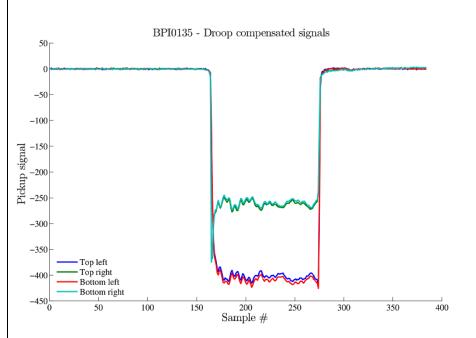
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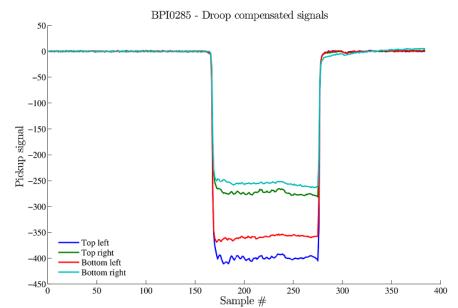
Conclusions and further work

- BPI non-linearity has significant impact on measured current and position
 - Desirable to correct this. IIR filter could potentially do so
- Parameterisation of each electrode's response non-trivial
 - Overcome by real-time fitting to exponential tail of signal
 - Currently, DL ADC gates do not sample enough tail for good correction
 - Shifting gate time to sample less baseline, more tail should work for DL
 - Extend gate length?
- Must consider robustness if implementing in, for example, BPI driver
 - Must work for different pulse lengths (pedestal subtraction, location of tail)
- Check applicability to TL1/CR/TL2 BPIs
 - Different processing electronics and higher beam current
 - Diagnostic data required to say more about these
- Any changes to BPI processing should be followed by calibration



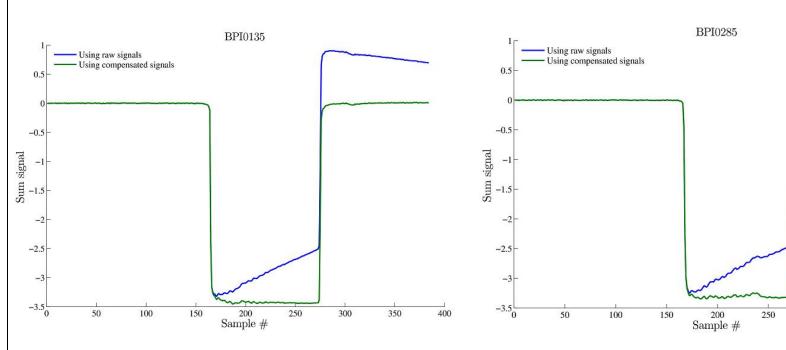
Example filtered signals





Effect on sum signals

Note: calibrated empirically with respect to treated signals



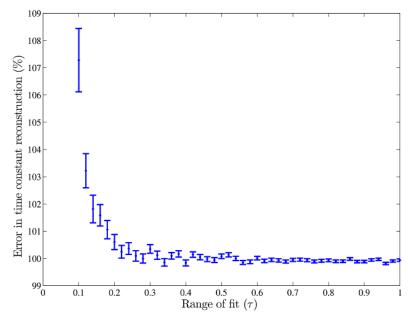
300

350

400

Monte Carlo exponential fits

- Fast fit to simulated, noisy exponential
 - RMS noise 2% of initial signal level
 - Vary range of simulated data used in fit from 10 100% of the time constant
 - Plot the fitted time constant as percentage of real value (with std. error)



- Bad fit when droop is comparable to noise or any systematic 'overshoot'
- When fit range is small compared to time constant, fit tends to over-estimate

Possible solutions

- Range of tail is 10 30% of the various observed time constants
 - Clearly too small a range for reliable time constant fitting
 - Contributing to (or explaining) the apparent nonlinearity in time constant vs. signal level

1. Shift the gate timing to observe more tail

- Currently ~150 samples of baseline before signal which could be used
- Gives tail range of 40 100% of time constant, and likely accurate fitting
- Would this ruin pedestal subtraction? How does the current algorithm work?

2. Extend the gate

- Could we do the same as we did for the CR and observe more tail?
- 3. Parameterise...

Possibility of parameterisation (for DL)

- Much of the apparent nonlinearity (time constant vs. signal level) may be an artefact due to poor fits
- Observing longer tail may allow decent parameterisation
- Need to parameterise time constant vs. signal level by calibration
- In absence of variable high current supply, must be beam based:
 - Short pulse would allow more tail to be observed
 - Generate two or three 1.5 GHz beams with different capture efficiencies
 - Inject satellites and inject mains into DL
 - Should give a good range of signal levels