

# Effect of cable on pulse shape

- **Already studied (in 1998?) for present electronics**
- **The formulas in the time domain (modification to step pulse) are well known ( for example in 1964 <http://lss.fnal.gov/archive/other/lbl-cc-2-1b.pdf> )**
- **A step pulse becomes as function of time  $[1. - \text{erf}( 0.6745*\text{sqrt}(t_0/2t)]$  therefore the cable effect depends on a single parameter  $t_0$**
- **An estimate for the parameter  $t_0$  is  $t_0=4.56 \times 10^{-3} * A^2 * l^2 / f$  where  $A$  is the cable attenuation in db/m,  $l$  is the cable length in m at a frequency  $f$  in MHz and  $t_0$  in ns. This formula is not perfect since  $A^2/f$  is not constant for the cable we use KX3B.**
- **In 1998(?)  $t_0$  obtained by fitting a step pulse + 12.8 m of our cable measured with a scope  $\Rightarrow t_0=1.024\text{ns}$**

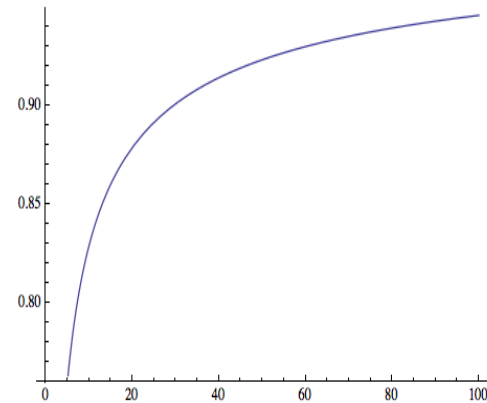
# Some Mathematica plots(I) (fortran to old to get plots!)

effect on step function

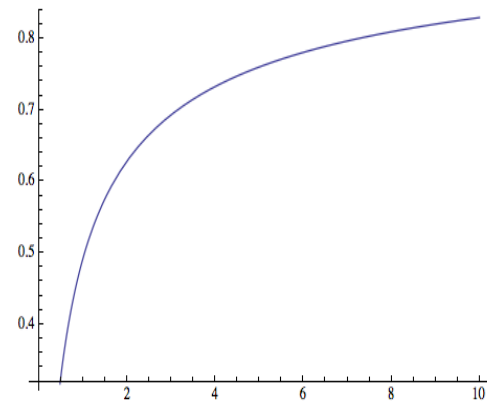
```
Cable[x_] = 1 - Erf[0.6745 * (Sqrt[(0.5 + 1.024 / x)])]
```

$$1 - \text{Erf}\left[0.482633 \sqrt{\frac{1}{x}}\right]$$

```
Plot[Cable[x], {x, 0., 100.}]
```



```
Plot[Cable[x], {x, 0., 10.}]
```



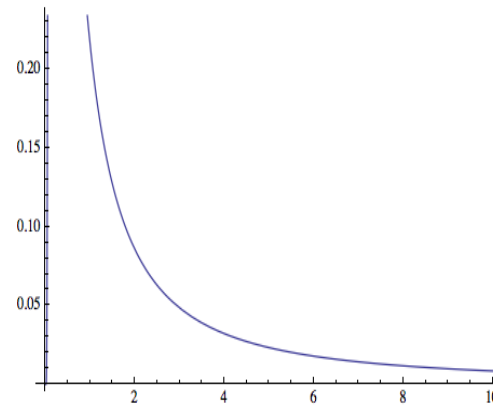
## (II)

effect on delta function  
=> differentiate step  
function and its  
transformed shape  
after the cable

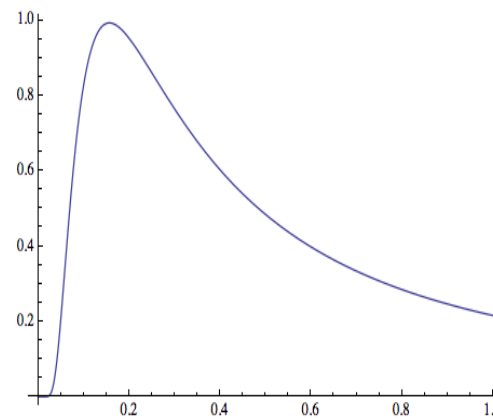
```
Pulse[x_] = D[Cable[x], x]
```

$$0.272296 e^{-0.232935/x} \left(\frac{1}{x}\right)^{3/2}$$

```
Plot[Pulse[x], {x, 0., 10.}]
```

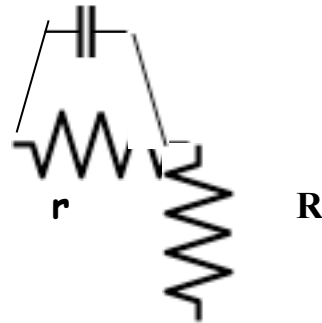


```
Plot[Pulse[x], {x, 0., 1.}]
```



# Correction with filter

- Again old story! (196?) use a pole zero filter  $rC + R$
- 



- $r/(R+r) = \text{frac} = \text{attenuation of low frequency}$
- $rc \approx \text{time constant}$
- A delta function becomes  $\delta - (\text{Frac}/rc) * e^{-(t/rc)}$  optimum values are  $\text{frac}=0.16$   $rc=16\text{ns}$

# Integrated results for delta function after cable and correction

Time after $\delta$	integrated ( $\delta$ +cable)	integrated( $\delta$ +cable +cor)	$\Delta$ of last col
1ns	0.494	0.492	
10 ns	0.829	0.732	
25 ns	0.891	0.779	
50 ns	0.923	0.777	-0.002
75 ns	0.937	0.781	+0.004
100 ns	0.946	0.786	+0.005
1000 ns	0.983	0.815	+0.029/40

# Impact

- **Two main consequences**
  - **Contribution to following samples**
  - **Modification of flat top over  $\pm 2ns$  (quantitative value? This should be calculated using measured pulse + calculated filter +integral) Calculation on this is the main priority in my opinion!**

# Pro and Con of possible corrections(I)

- **Subtraction of following pulse in FPGA and improvement of flat top with R on integrator**
  - **Pro simplest for hardware**
  - **Con: mixes pedestal subtraction and 2-integrator calibration with cable correction**
  - **Con: Heavy in firmware**
- **Filter using an added stage before the integrator in ASIC**
  - **Pro: probably the cleanest method**
  - **Con: added work and complication for ASIC**
  - **Con: Asic becomes set-up specific :depends on cable (Is this a real problem???)**

## Pro and Con (II)

- Put the compensation filter on the PCB (needs 4 components to conserve 50 ohms adaptation)
- A possible solution is  $r=8$  ohms  $C = 2$ nanofarad (2.2?)
- $R= 50$  ohms of the Asic in// with (a resistance of 262 ohms + an inductance of 5microHenry in series)
- This allows to have an input impedance of 50 ohms at all frequencies (until the inductance stops working= max frequency=150MHz in radiospare)
  - **Pro: rather flexible and optimised at PCB time (later than ASIC)**
  - **Con: 4X32 components per cards including 32 5-microH inductances**
  - **Con: could impact the noise? (to be calculated or tested)**



# Conclusion

- **Need to define time scale (time for final decision)**
- **In my opinion a least a test of the PCB solution should be done and tested to insure the existence of the solution at least as back-up**