

***Cold CMOS Electronics
for the readout of
very large LAr TPCs***

Craig Thorn
On behalf of the LBNE LAr Working Group

TIPP 2011
June 11, 2011

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Outline

- LAr40 overview
- The Cold Electronics System
- Analog Front End
- ADC
- Summary

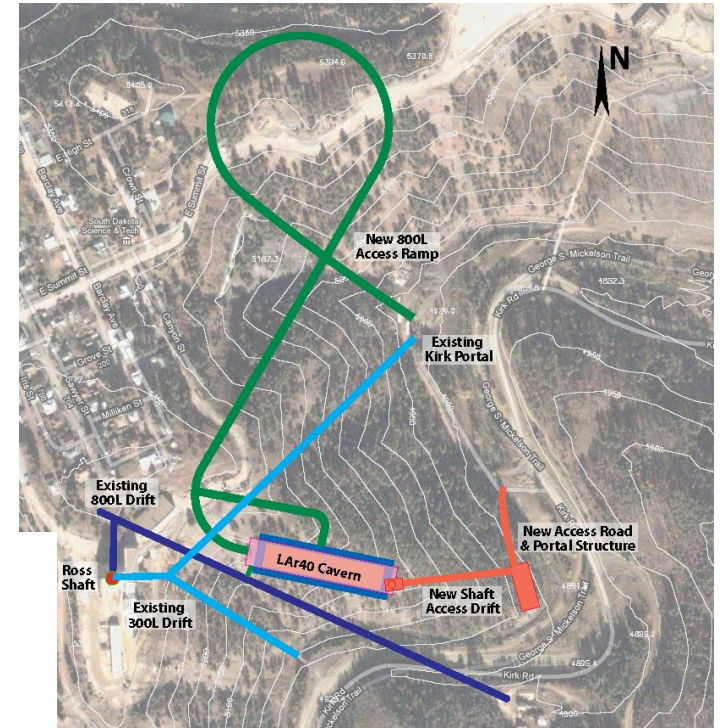
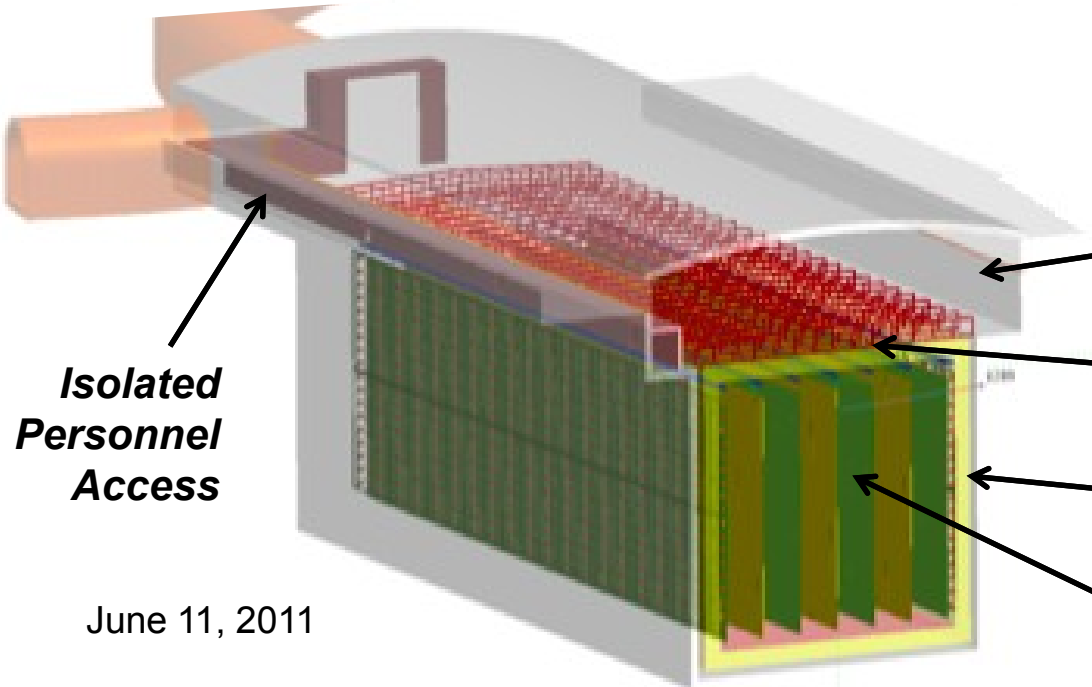
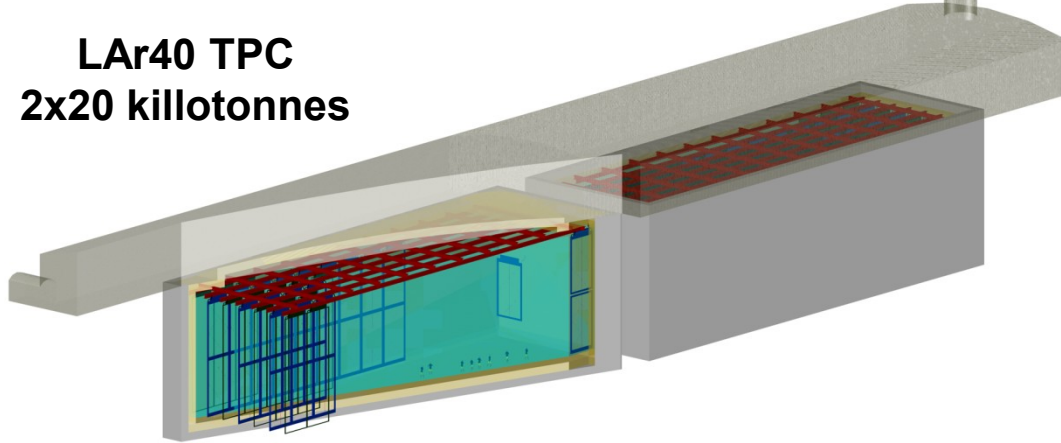
Other presentations at TIPP 2011 describing LBNE LAr40:

- *[412] Membrane cryostat technology and prototyping program towards kton scale Neutrino detectors, **Rucinski***
- *[422] Designs of Large Liquid Argon TPCs — from MicroBooNE to LBNE LAr40, **Yu***
- *[223] Front End Readout Electronics of the MicroBooNE Experiment, **Chen***

LBNE LArTPC inside Cryostats in Cavern

Deep Underground Science & Engineering Lab (DUSEL) in Homestake, South Dakota, USA

LAr40 TPC
2x20 kilotonnes



High Bay Access
Truss Lid for Cryostat

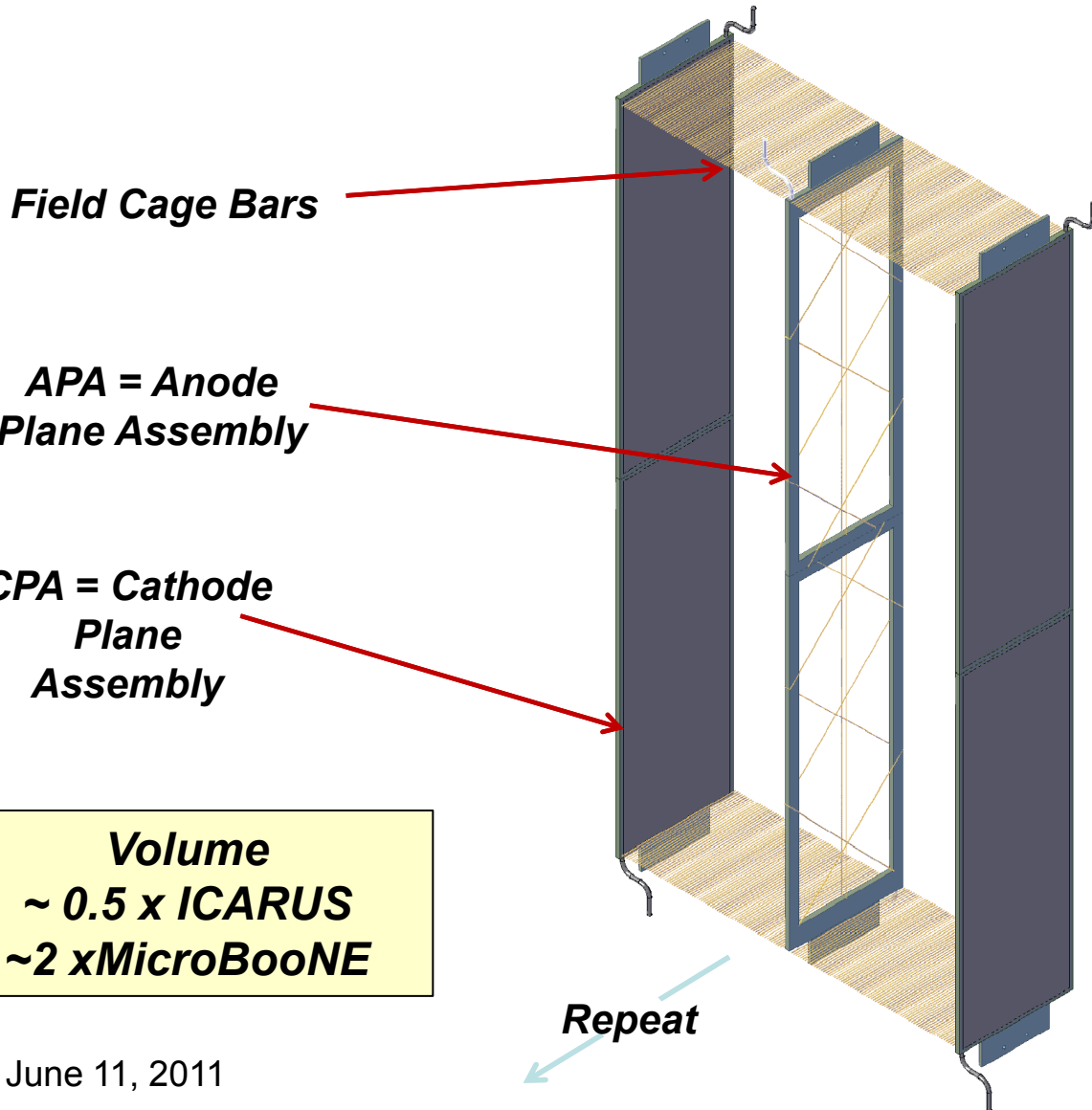
Membrane Cryostat

TPC Arrays

Membrane cryo program toward

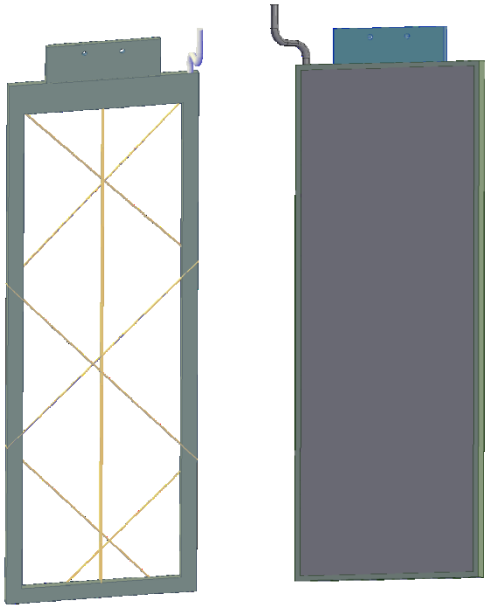
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APA + CPA Assemblies form TPC modules in the Cryostat



Unit Cell
TPM = TPC Module
= 1 APA + 1 CPA
(+1 "terminal" CPA)

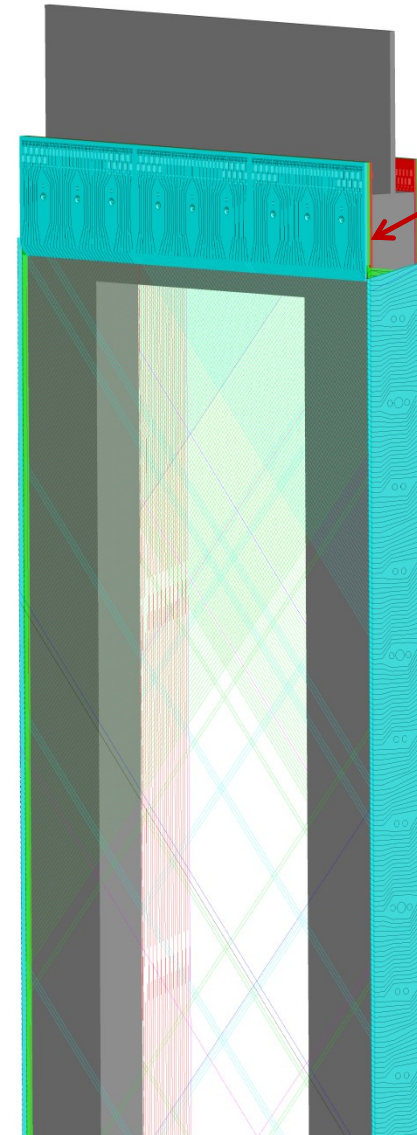
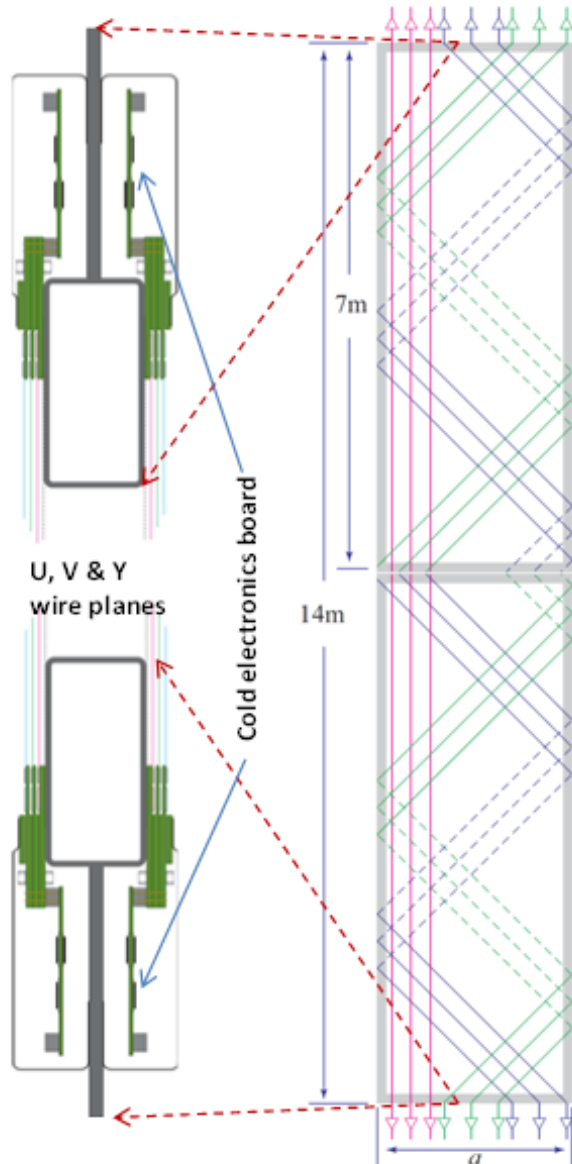
Volume
~ 0.5 x ICARUS
~ 2 x MicroBooNE



APA

CPA

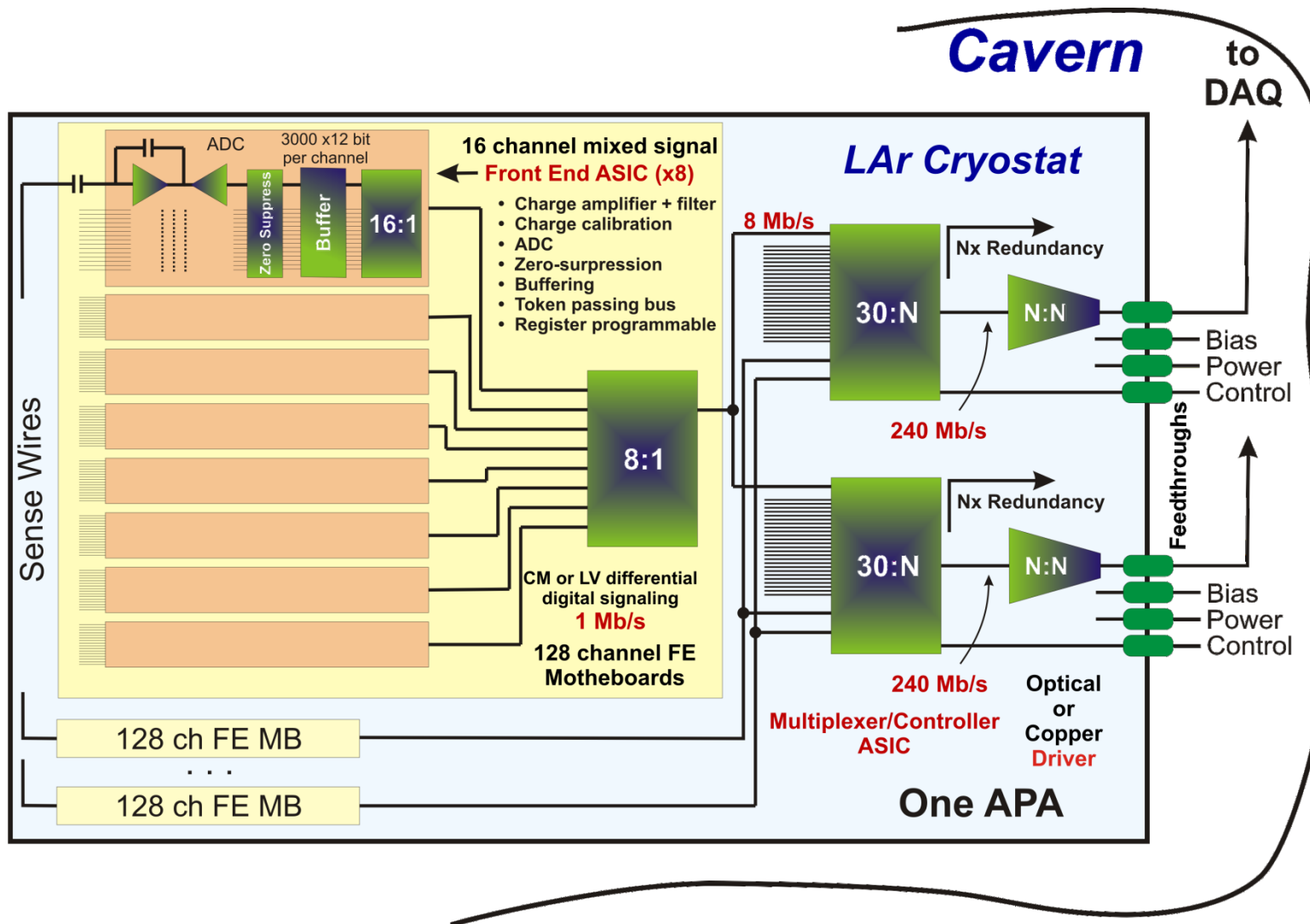
Anode Plane Assembly (APA): the core element of a TPC unit



Cold Electronics

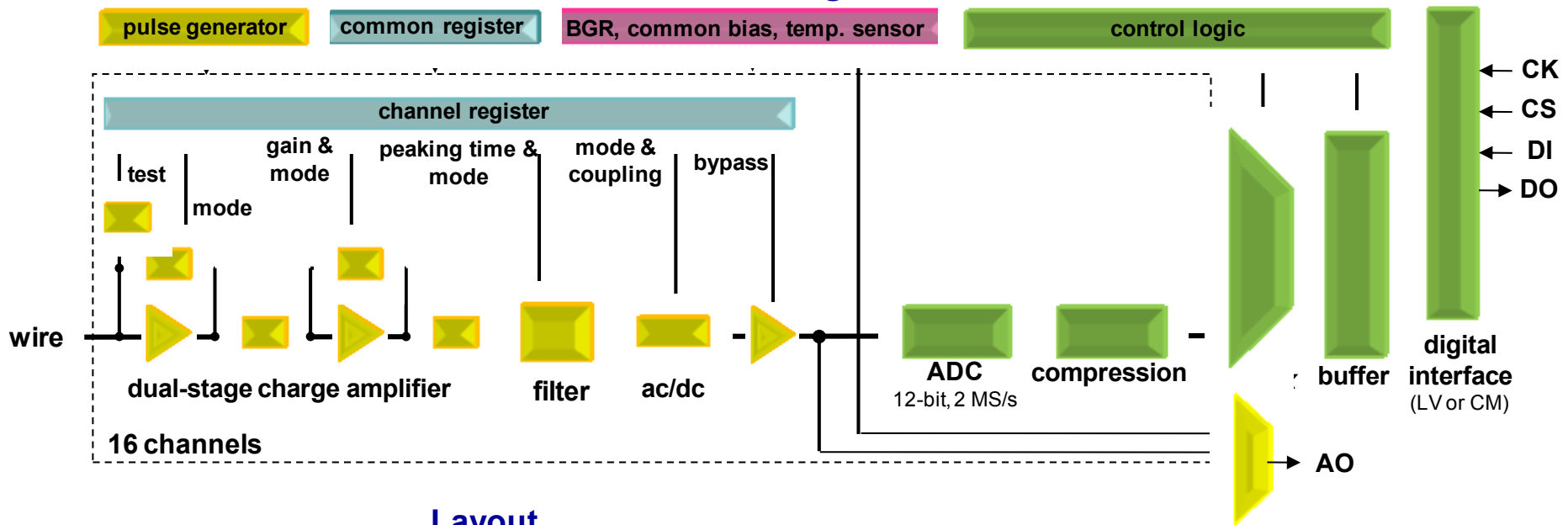
APAs are structural and electrical units containing all sense wires and readout electronics. They can be tested in LN2 and stored and transported in shipping containers.

LAr TPC - Cold CMOS Electronics Block Diagram – Reference Design

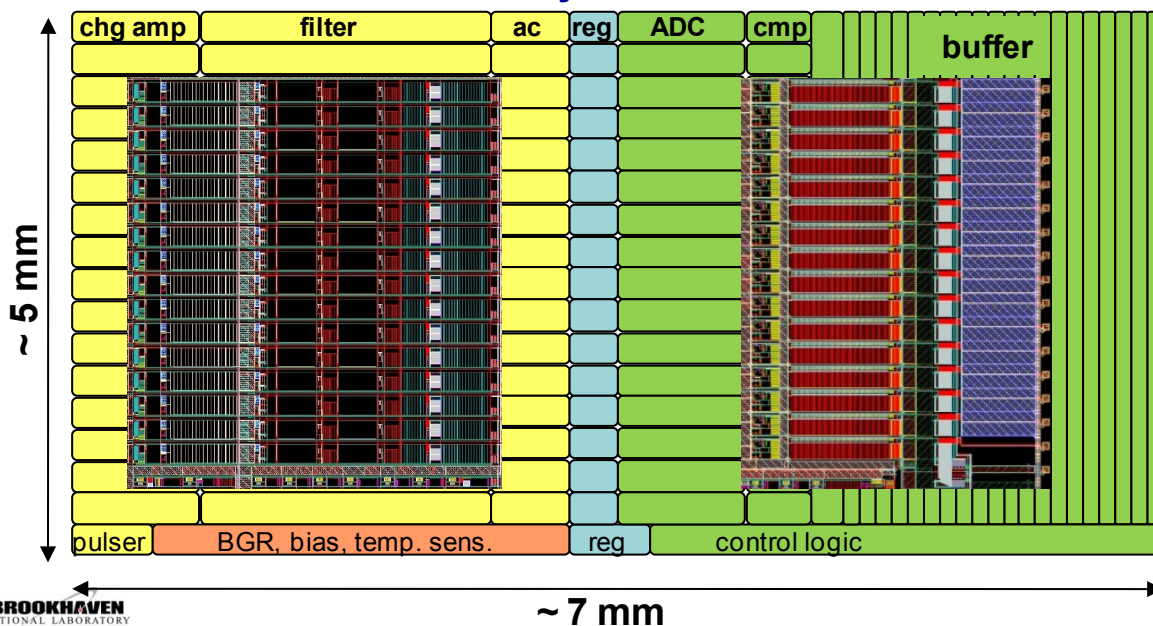


LAr TPC Front-End ASIC

Block Diagram



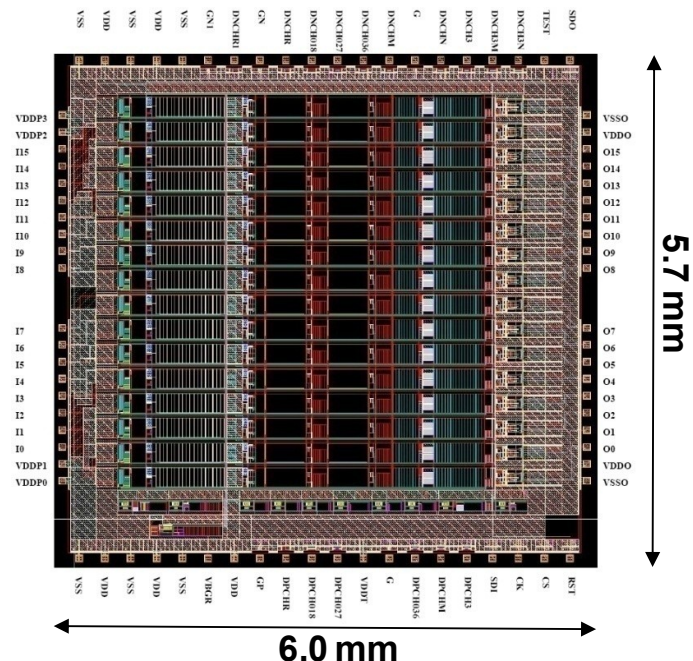
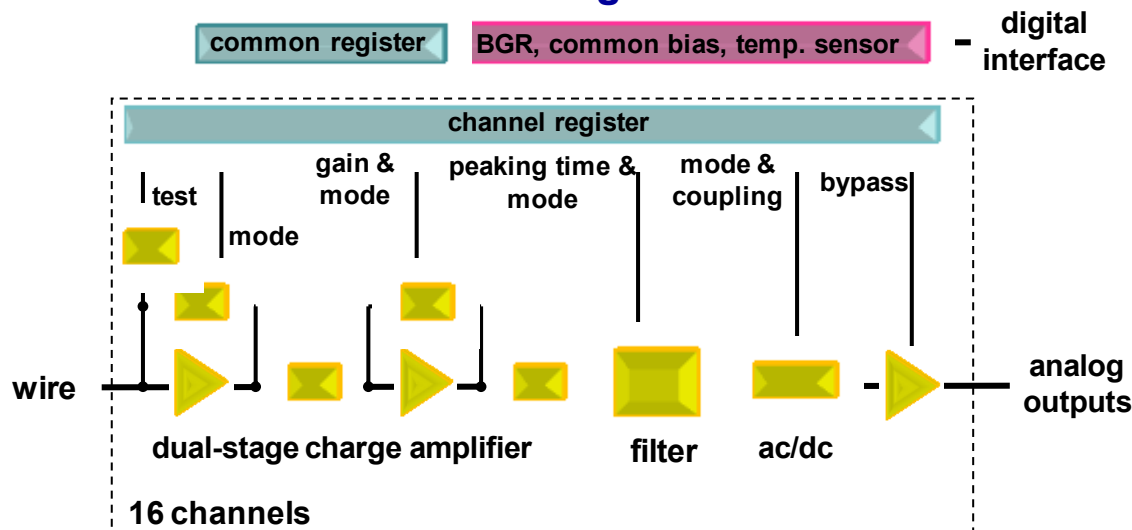
Layout



- 16 channels
- charge amplifier (adj. gain)
- high-order filter (adj. time constant)
- ac/dc, adjustable baseline
- test capacitor, channel mask
- ADC (12-bit, 2 MS/s)
- compression, discrimination
- multiplexing and digital buffering
- LV or CM digital interface
- pulse generator, analog monitor
- temperature sensor
- LAr environment (> 20 years at 88K)
- estimated total size ~ 6 x 8 mm²
- estimated power ~ 10 mW/channel

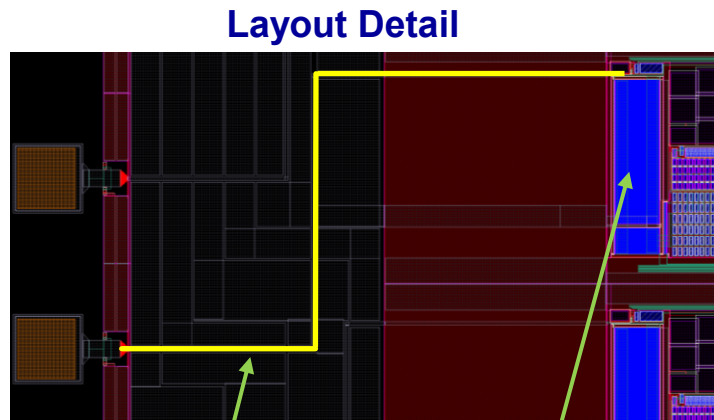
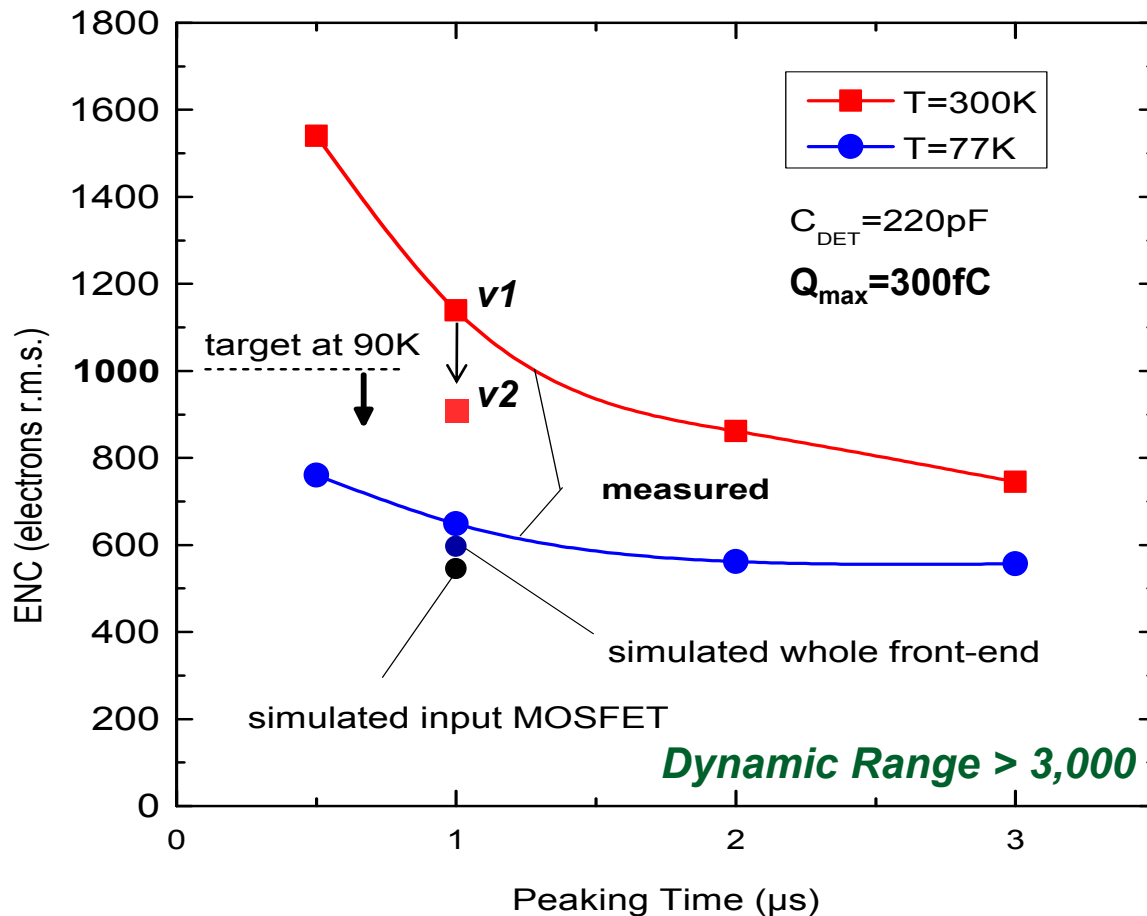
Analog ASIC

Block Diagram



- 16 channels
- charge amplifier, high-order filter
- adjustable gain: 4.7, 7.8, 14, 25 mV/fC (charge 55, 100, 180, 300 fC)
- adjustable filter time constant (peaking time 0.5, 1, 2, 3 μs)
- selectable collection/non-collection mode (baseline 200, 800 mV)
- selectable dc/ac coupling (100μs)
- rail-to-rail analog signal processing
- band-gap referenced biasing
- temperature sensor (~ 3mV/°C)
- 136 registers with digital interface
- 5.5 mW/channel (input MOSFET 3.9 mW)
- single MOSFET test structures
- ~ 15,000 MOSFETs
- designed for room (300K) and cryogenic (77K) operation
- technology CMOS 0.18 μm, 1.8 V

Noise Measurements



Input Line
 $L \approx 1\text{ mm}$
 $W = 3.5\ \mu\text{m}$
 (M3 + M4)
 $R_{77\text{K}} \approx 3\ \Omega$
 $R_{300\text{K}} \approx 12\ \Omega$

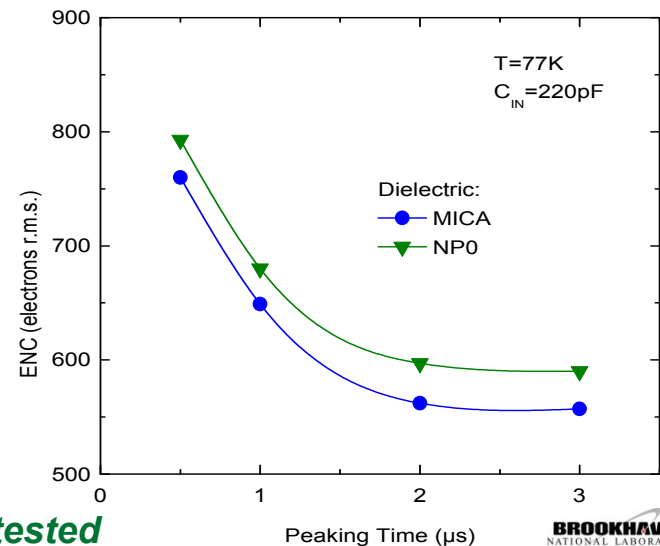
Input MOSFET
 $L = 270\text{ nm}$
 $W = 10\text{ mm}$
 (50 $\mu\text{m} \times 200$)
 $g_{m,77\text{K}} \approx 90\text{ mS}$ (11 Ω)
 $g_{m,300\text{K}} \approx 45\text{ mS}$ (22 Ω)

Measurements affected by:

- input line **parasitic resistance**
 - $\sim 150\text{ e}^-$ at 77 K ($\sim 590\text{ e}^-$ at 300K)
 - addressed in next revision
- C_{IN} **dielectric noise (not present in wire)**
 - $\sim 60\text{ e}^-$ at 77 K

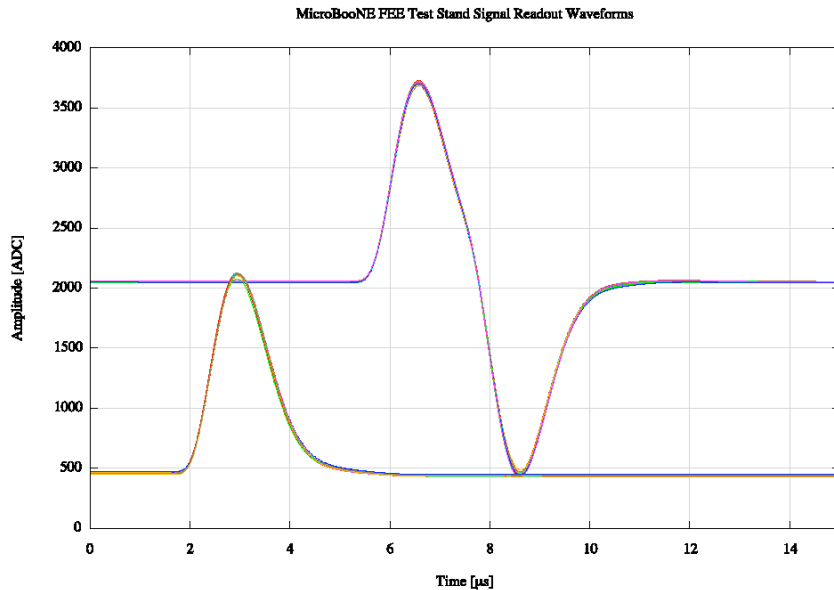
$$d\text{ENC} \approx \sqrt{2kTC_{\text{IN}} \text{tg } \delta}$$

$$\approx \begin{cases} 200\text{ e}^- & \text{for NPO} \\ 60\text{ e}^- & \text{for MICA} \end{cases}$$

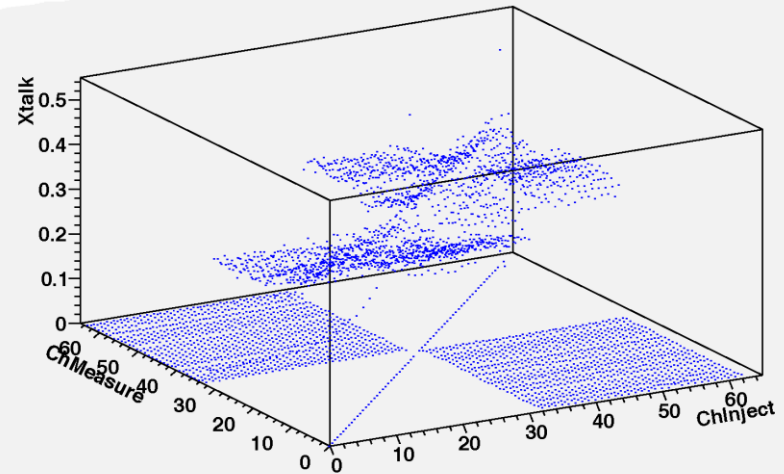


ASIC revision 2 designed and fabricated, currently being tested

FEE ASIC Evaluation



Crosstalk Measurement



FEE Test Stand for MicroBooNE is operational

- Full front end electronics chain, from CMOS ASIC to Receiver/ADC board, data is acquired to PC through FPGA board and Gigabit Ethernet
- One temporary 32 channel cold cable is available which has one broken channel
- Without detector capacitance, noise is $\sim 200e^-$ with 1us peaking time
- Nonlinearity and crosstalk are less than 0.5%
- FEE test stand will be upgraded with the second version of ASICs and two prototype cold cables, more tests to follow

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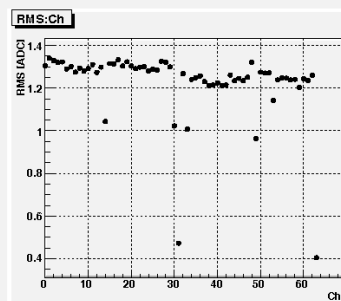
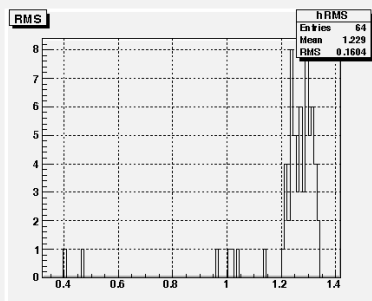
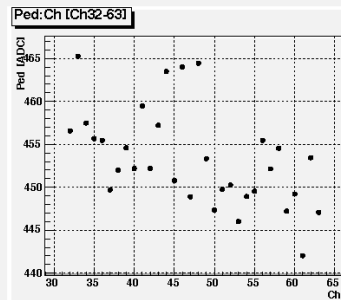
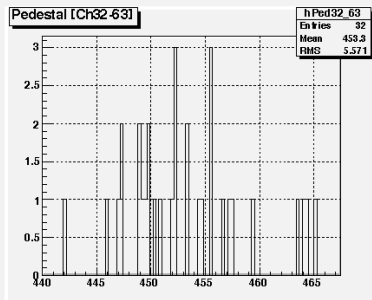
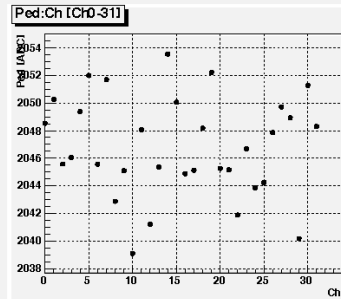
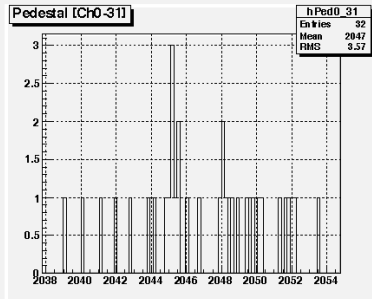
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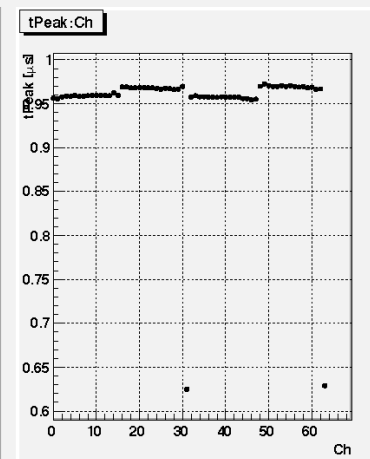
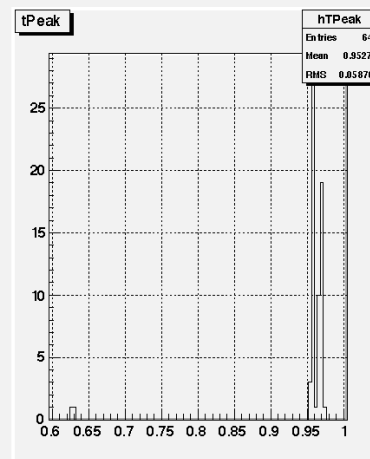
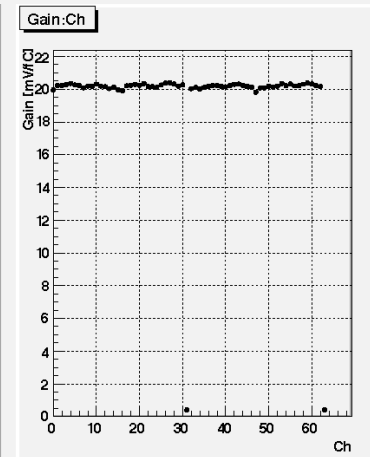
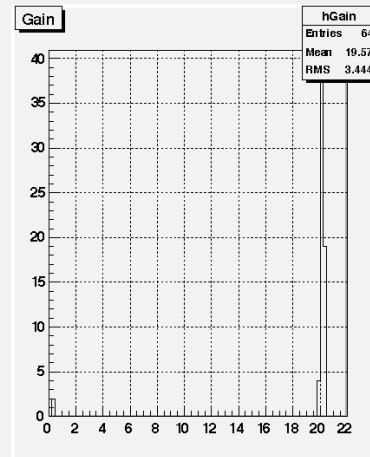
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FEE Analog ASIC Evaluation Noise, Gain & Shaping Time

Noise Measurement

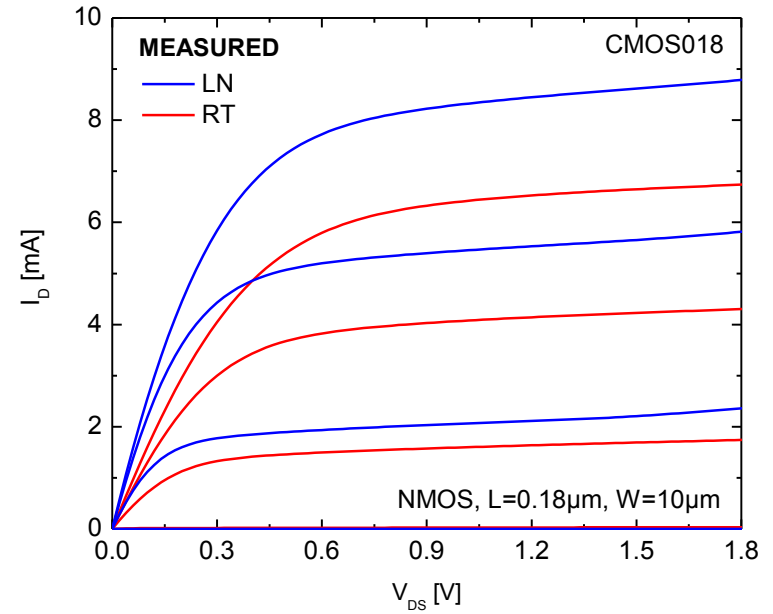
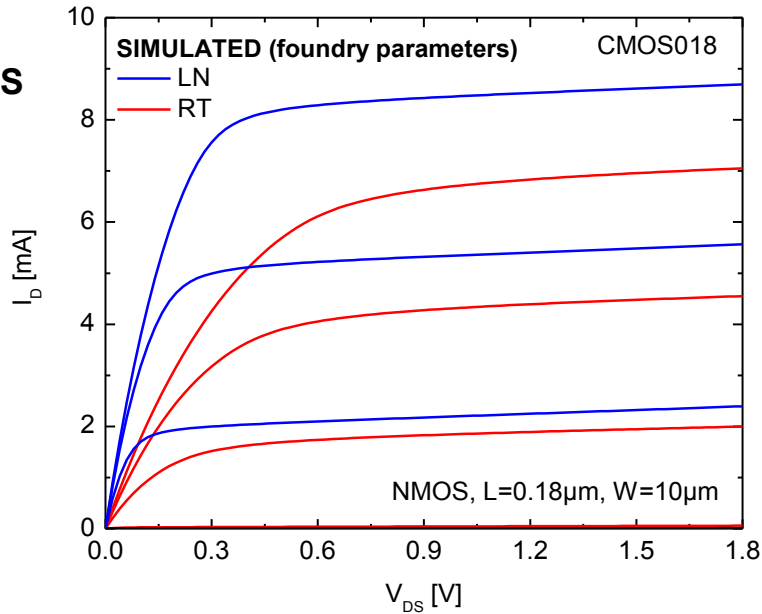


Gain & Peaking Time Measurement

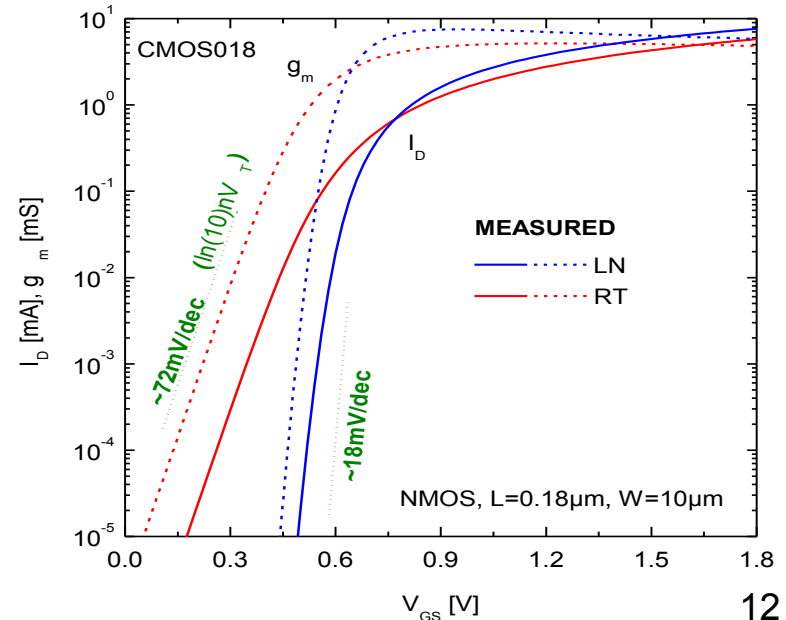
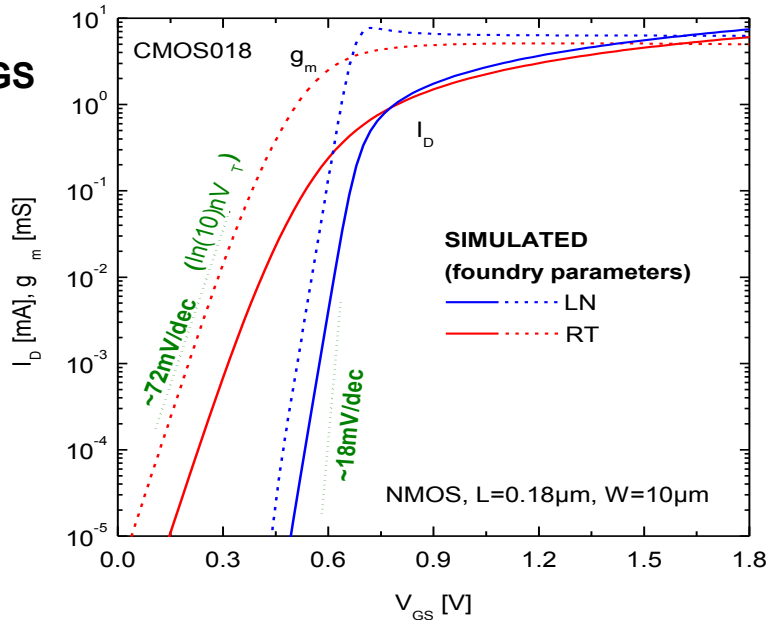


MOS Static Model

I_D vs V_{DS}

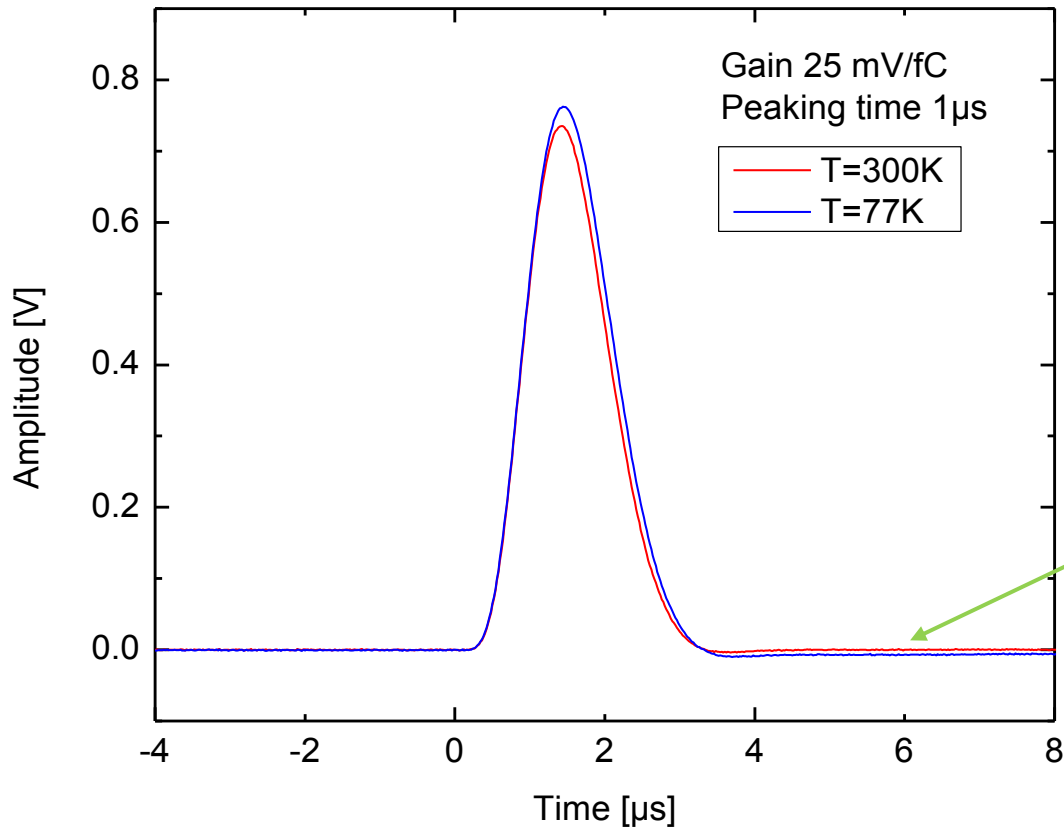


I_D vs V_{GS}



Some differences in saturation voltage, sub-threshold slope, transconductance

Signal Measurements



Bandgap Reference

$$V_{\text{BGR}} \approx \begin{cases} 1.185 \text{ V} & \text{at } 300 \text{ }^\circ\text{K} \\ 1.164 \text{ V} & \text{at } 77 \text{ }^\circ\text{K} \end{cases}$$

variation $\approx 1.8 \%$

Temperature Sensor

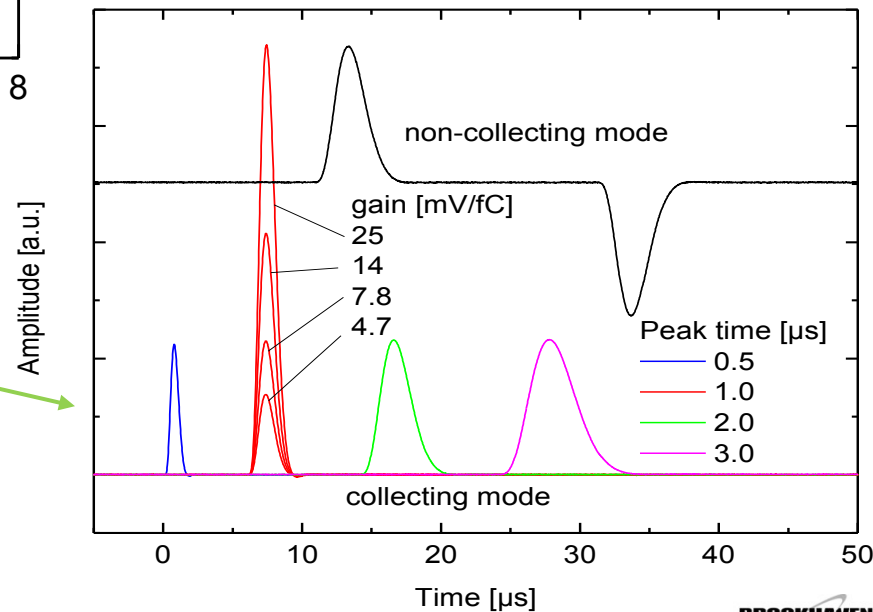
$$V_{\text{TMP}} \approx \begin{cases} 867.0 \text{ mV} & \text{at } 300 \text{ }^\circ\text{K} \\ 259.3 \text{ mV} & \text{at } 77 \text{ }^\circ\text{K} \end{cases}$$

$\sim 2.86 \text{ mV} / \text{ }^\circ\text{K}$

**Pole-zero cancellation at 77K
to be addressed in next revision**

Adjustable **gain**, peaking time and baseline

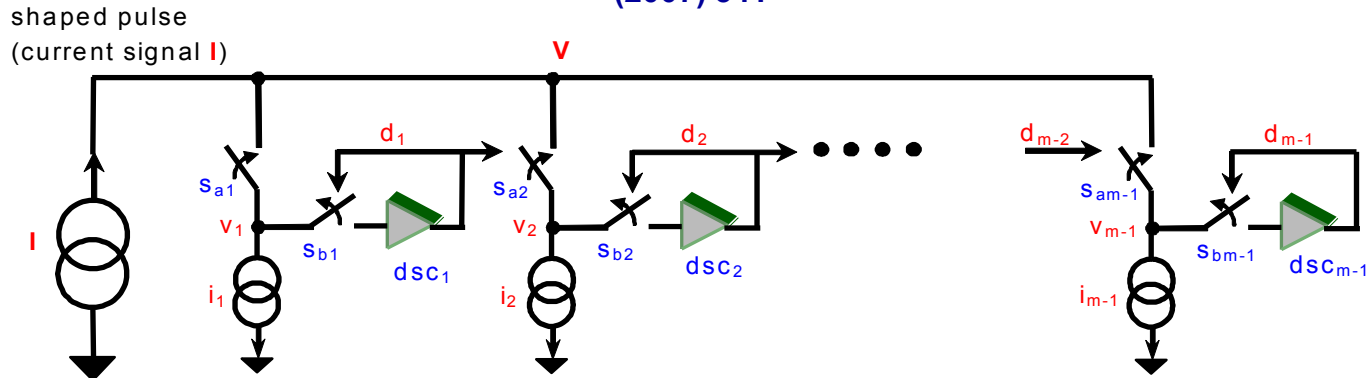
maximum charge 55, 100, 180, 300 fC



ADC - Architecture

Clockless low power ADC stage

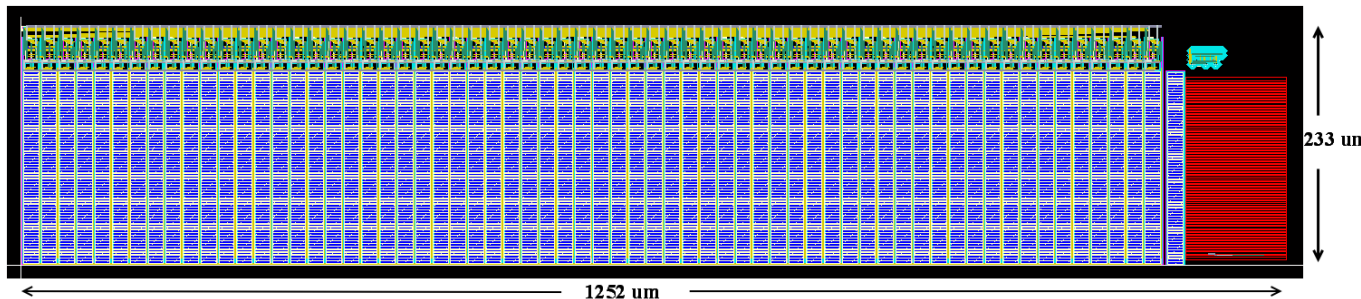
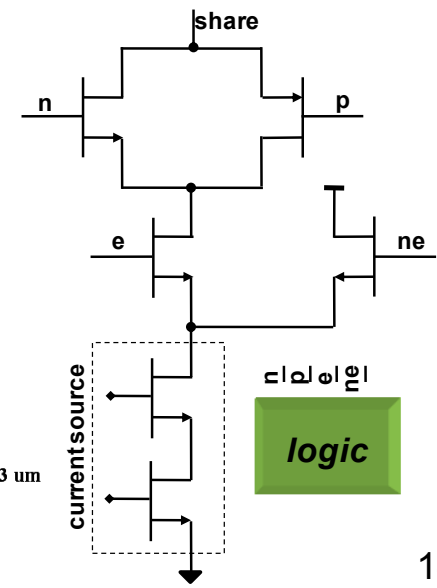
Demonstrated in ASIC for SNS, see De Geronimo, *et al.*, IEEE Trans NSS, 54 (2007) 541



Current mode ADC

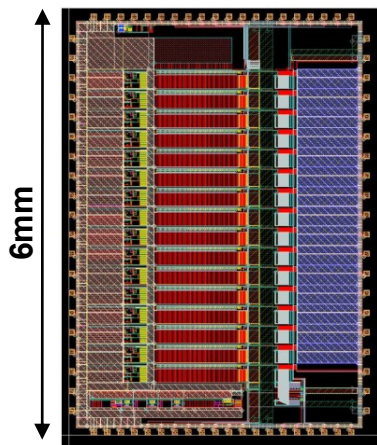
- **dual stage** 6-MSBs in 150ns, 6-LSBs in 250ns
- single trigger conversion per stage
- **12-bit resolution**
- **2 MS/s** conversion rate
- power dissipation **3.6 mW at 2 MS/s**
- **power-down** option for low rate applications
 - wake up in few tens of ns
- layout size: 0.23 mm x 1.25 mm

ADC cell



ADC - Preliminary Results

16-channel
ADC+buffer

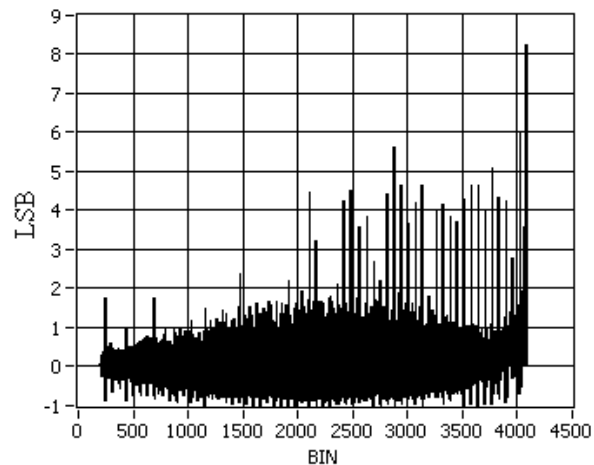


4.3 mm

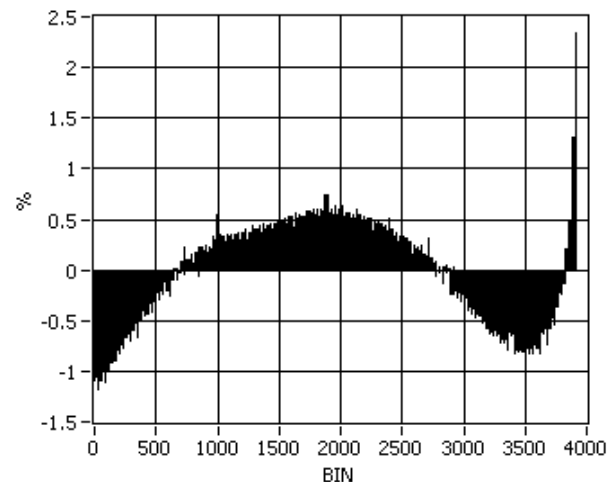
77 K

300 K

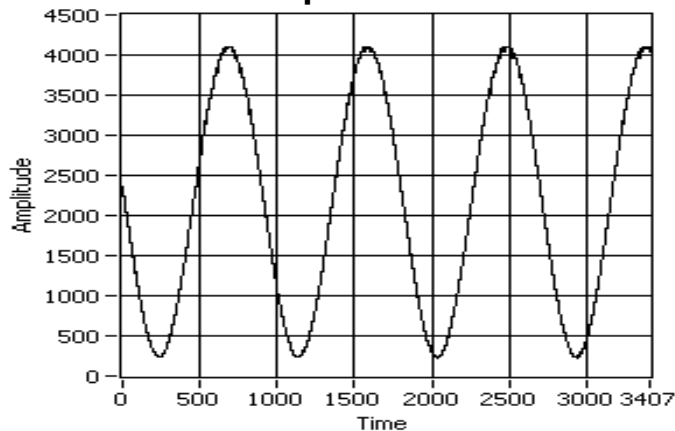
Differential Nonlinearity (DNL)



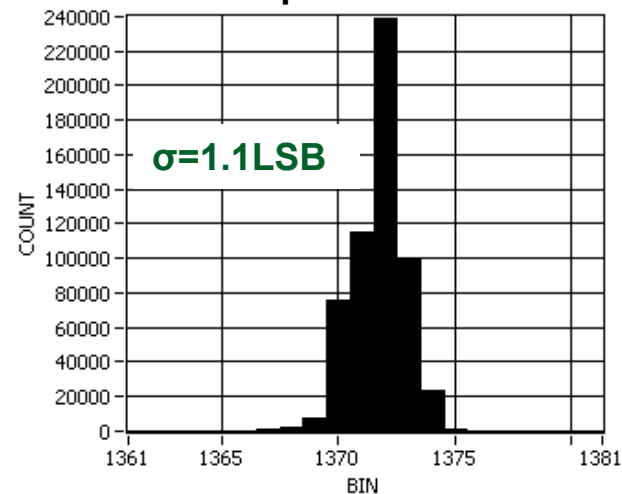
Integral Nonlinearity (INL) Percent



ADC output - 1.4 V sine



ADC output - 500mV dc



- operation verified at room and cryogenic temperatures
- differential non-linearity limited by timing design error in control circuit
- integral non-linearity limited by mismatch (linear → common centroid)

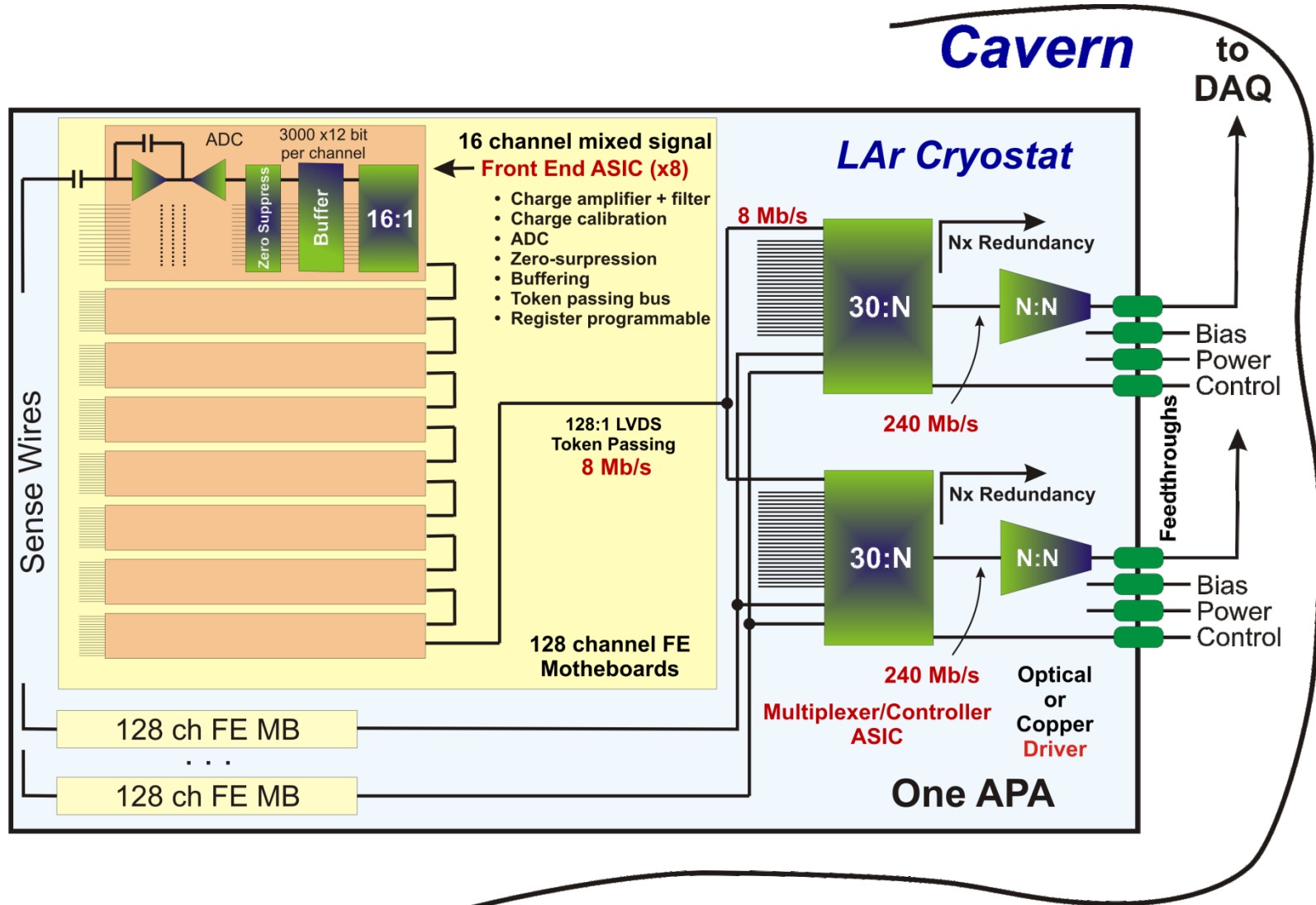
ASIC revision being designed, to be fabricated in July

Conclusions and Future Work

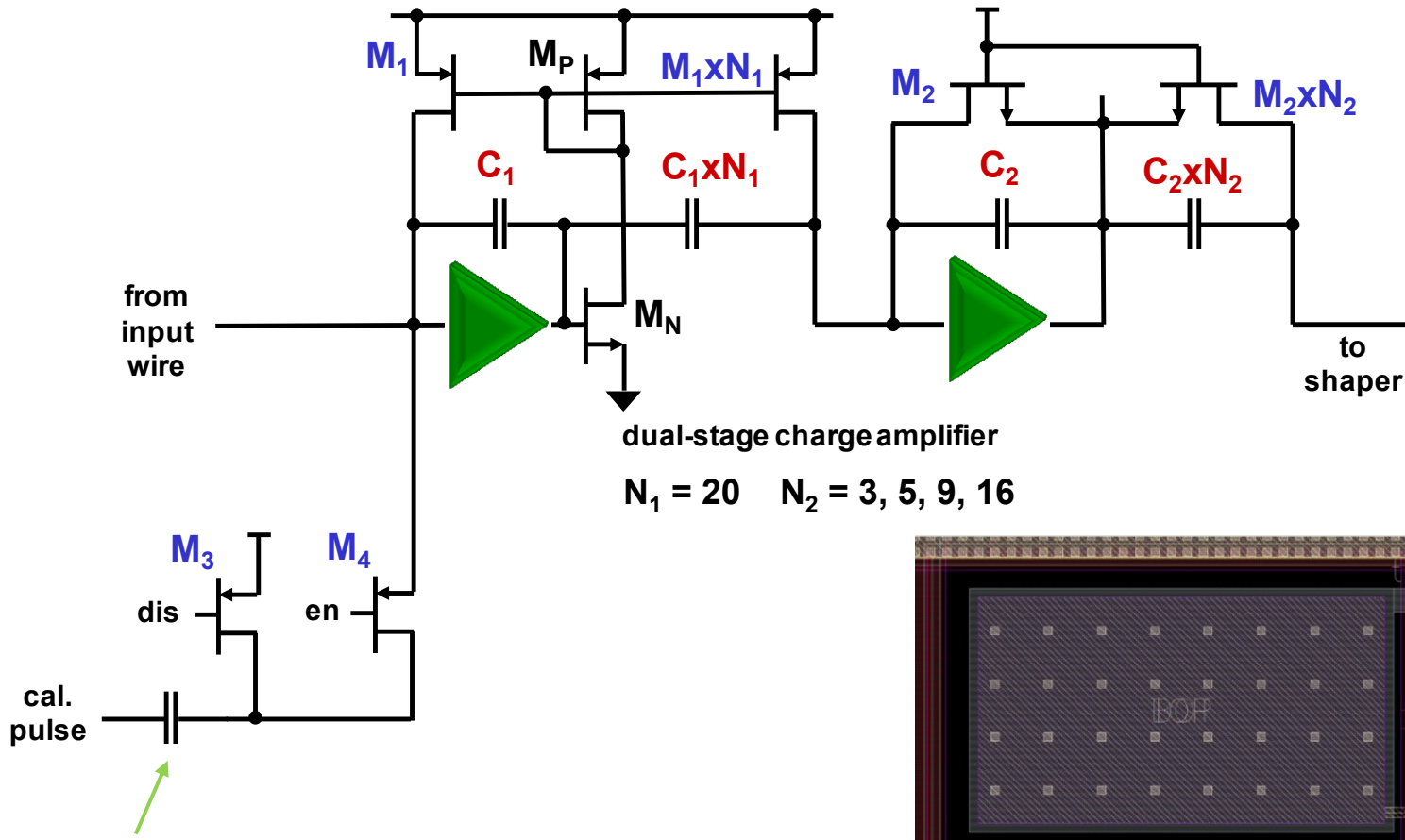
- CMOS **performs better** at cryogenic temperatures
- **Defined and predictable** design for cryogenic T is possible
- **Low-noise** at cryogenic T demonstrated
 - ENC < 1,000 e⁻ at 200pF ~5mW/ch.
 - characterization and modeling of CMOS 180nm
- **Long lifetime** at cryogenic T possible with guidelines
- Critical **building blocks** - front-end & ADC - developed
- Future work
 - Improve cryogenic static models
 - Optimize ADC
 - Merge, add zero-suppression & buffering, and finalize

Backup Slides

LAr TPC - Cold CMOS Electronics Block Diagram – Alternate Design



Calibration Scheme

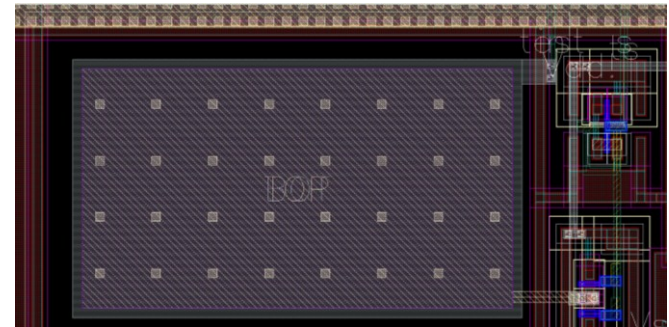


$C_{INJ} = 180$ fF nominal

Integrated injection capacitance ($10 \times 18 \mu\text{m}^2$)

Disabled (grounded) when unused

Measured with high-precision external capacitance



$$C_{INJ} \approx \begin{cases} 184 \text{ fF} & \text{at } 300 \text{ K} \\ 183 \text{ fF} & \text{at } 77 \text{ K} \end{cases}$$

change ~ 0.5%

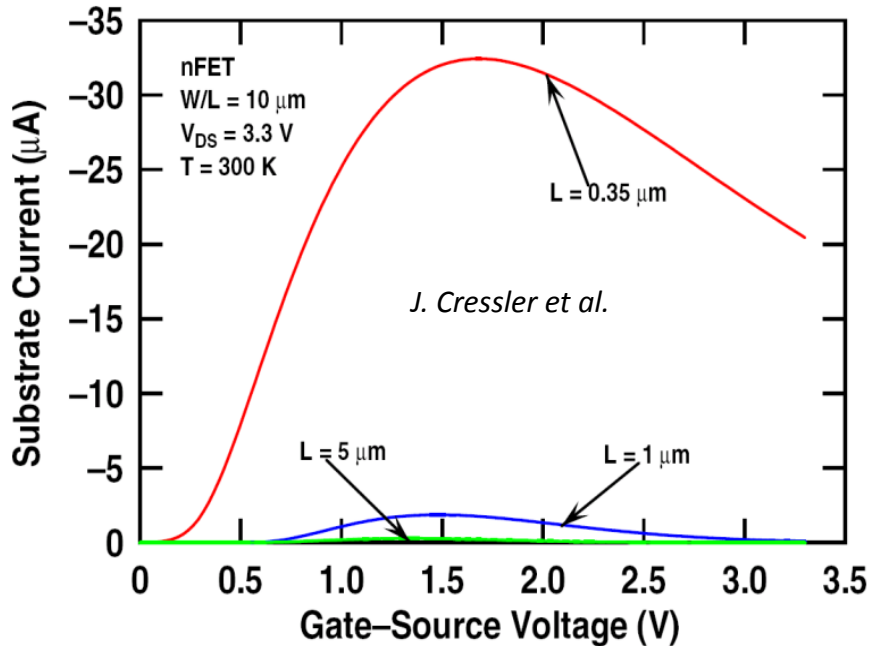
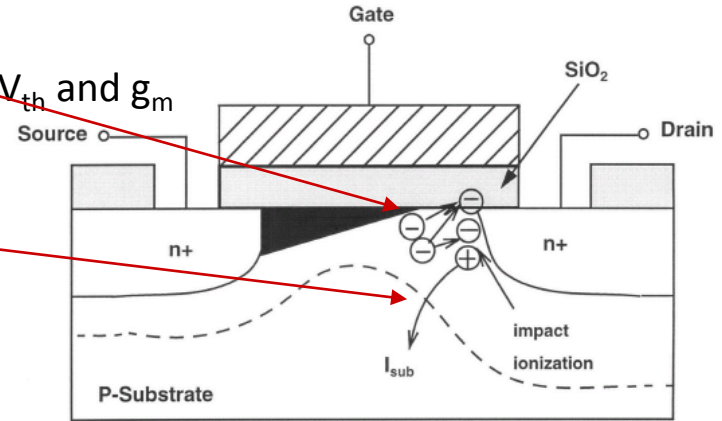
Lifetime - Basic Mechanism

- **Degradation is due to impact ionization**

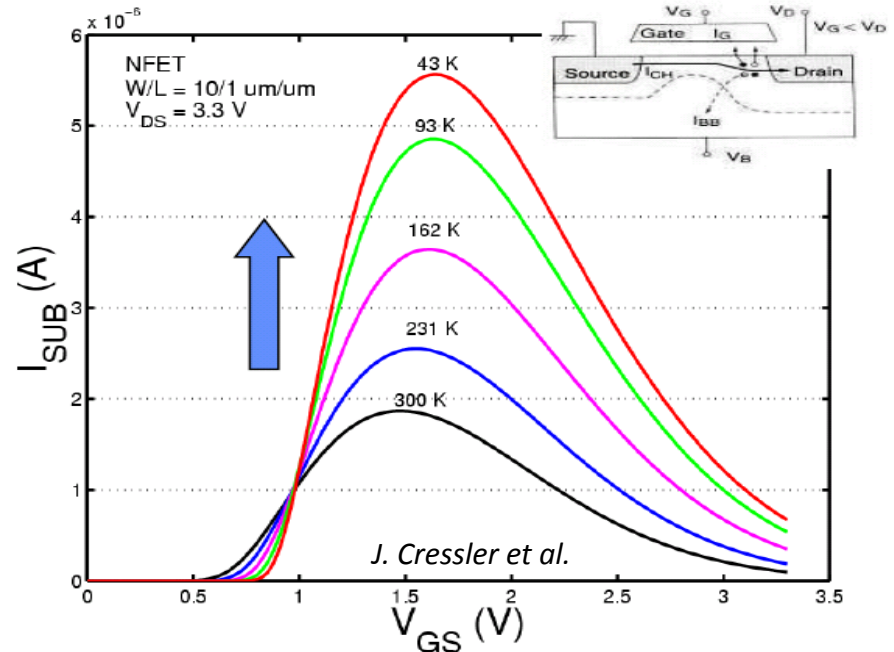
- charge trap in oxide, interface generation \rightarrow shift in V_{th} and g_m

- **Substrate current is a monitor of impact ionization**

- increases with **drain voltage**
- is higher in **short channel** devices
- has a **maximum** at $V_{GS} \approx V_{DS}/2$



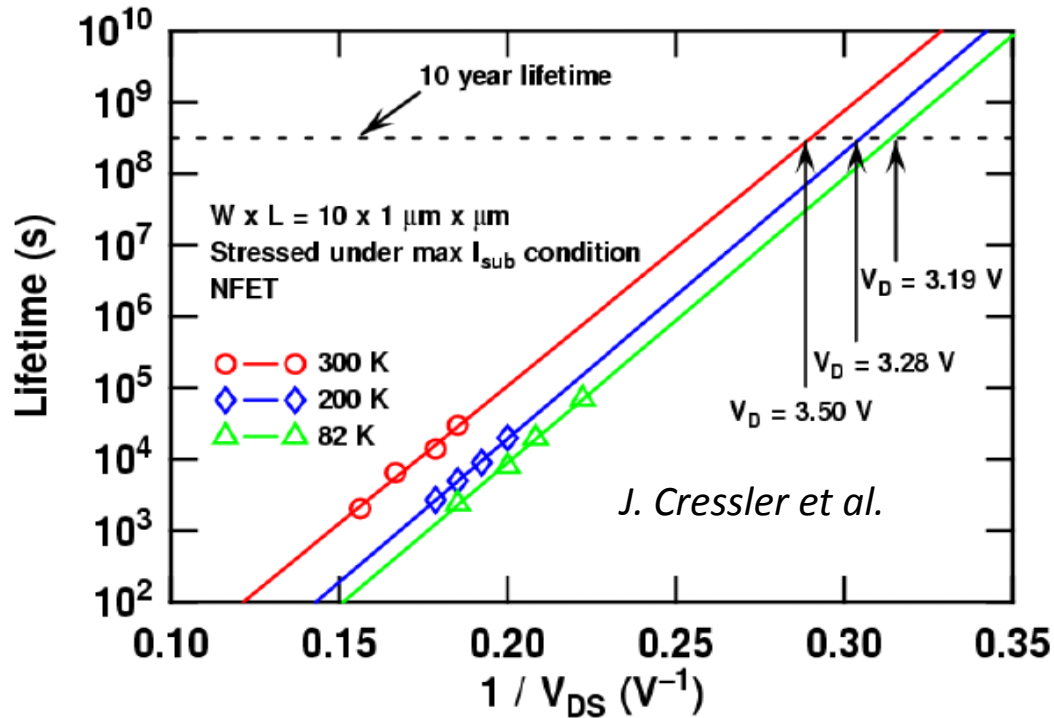
- **increases as the temperature decreases**



Commercial technologies are rated 10 years lifetime (10% g_m shift) in continuous ring oscillator operation: $T = 300$ K, $L = L_{min}$, $V_{ds} = \text{nominal } V_{DD} + 5\%$, $V_{GS} \approx V_{DS}/2$

Lifetime - Design Guidelines

Accelerated tests at increased V_{DS} allow extrapolation of lifetime



Desired lifetime at low temperature can be achieved by:

1. **decreasing V_{DS}** (e.g. decreasing the supply voltage)
2. **decreasing J_D** (i.e. decreasing the drain current density)
3. **increasing L** (i.e. non-minimum channel length devices)

Design guidelines can be obtained for:

analog circuits

- operate devices at **low current density**
- use **non-minimum channel length L**

digital circuits

- operate devices at **-10% of nom. V_{DD}**
- use **non-minimum channel length L**
- operate at **low clock frequency**