



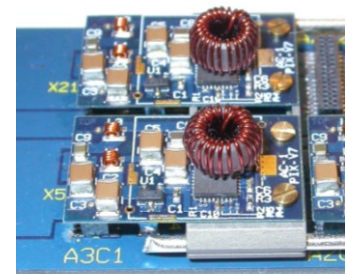
Powering for Future Detectors: DC-DC Conversion for the CMS Tracker Upgrade

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1. Physik. Institut B, RWTH Aachen University

Vertex 2011, Rust, Austria
June 23rd, 2011





CMS strip tracker end cap



CMS strip tracker ready for installation



- **Trackers need kilowatts of power:** e.g. CMS strips $\sim 33\text{kW}$
 - power consumption will increase for SLHC: higher granularity, more functionality
- Due to long (50m) cables, power losses are (already today) similar to detector power
- Routing of services complex and nested, cable channels full and total current limited
- Cabling inside tracker volume adds to material budget

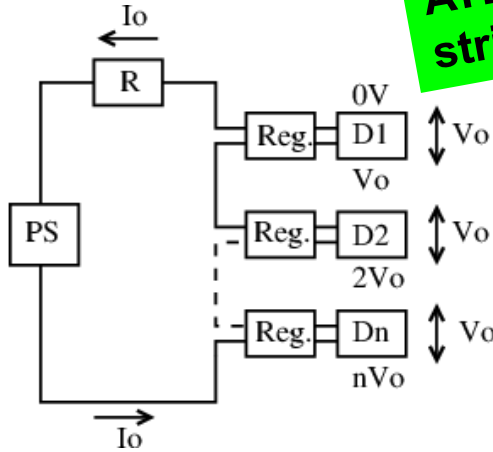
→ **Novel powering schemes need to be exploited**

Powering Schemes



Serial Powering

ATLAS pixels and strips upgrades?



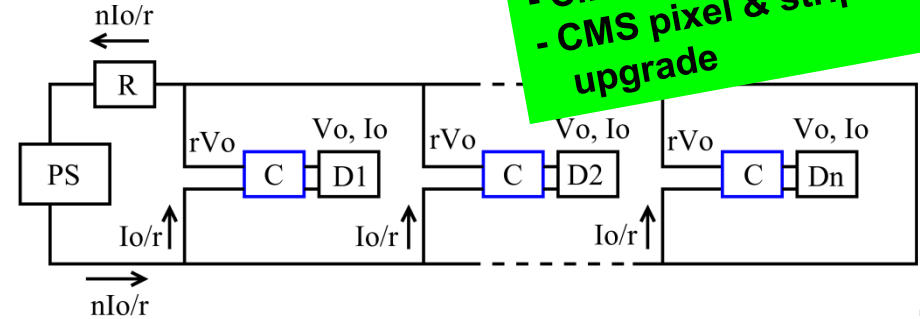
$$V_{\text{drop}} = R \cdot I_0$$

$$P_{\text{drop}} = R \cdot I_0^2$$

- Powered from constant current source
- Shunt regulator and transistor to take excess current and stabilize voltage
- + Number of modules in chain can be large
- + Adds very little extra material
- No solid system ground → biasing, AC-coupled communication etc.
- Inefficient if different current consumptions (e.g. end caps)

DC-DC conversion

ATLAS pixels and strips upgrades?
CMS HCAL upgrade
CMS pixel & strips upgrade



$$P = U \cdot I = (r \cdot U) \cdot (I/r)$$

$r = \text{conversion ratio}$

$$P_{\text{drop}} = R \cdot (I/r)^2$$

- Need radiation-hard magnetic field tolerant DC-DC converter
- + Standard grounding, biasing, control & communication scheme
- + Fine for very different current consumption
- Conversion ratio limited by technology and efficiency
- Switching devices → switching noise
- Output current per converter limited



The CMS Tracker Upgrade



As a result of a review process, the CMS tracker has chosen **DC-DC conversion as baseline solution**, and maintains Serial Powering as back-up (January 2009).

Around 2016: Exchange of the CMS pixel detector

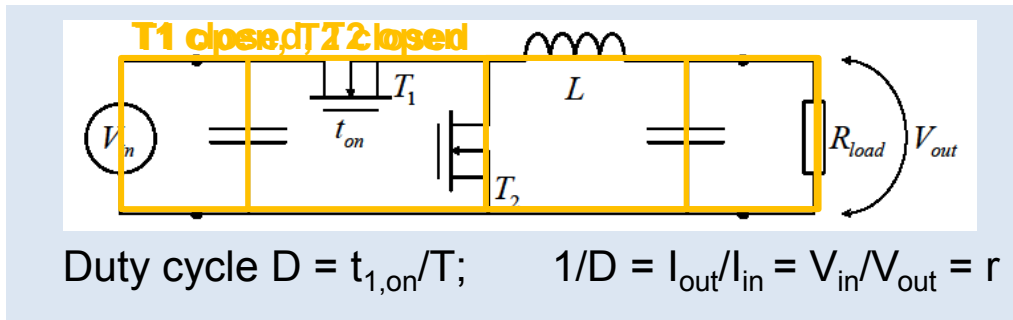
- Similar to today's detector, but less material, reduced data losses, CO₂ cooling
 - 3 Barrel layers → 4 barrel layers; 2 disks → 3 disks
 - Number of readout chips (ROCs) increases by factor 1.9
 - Unacceptable power losses in cable trays
- **DC-DC buck converters with conversion ratio of 3-4**
(Semi-conductor technology limits input voltage to < 12V, and V_{out} = 2.5 and 3.3V)

Around 2022: Exchange of the whole CMS tracker

- Higher granularity → more readout channels
 - Tracker is supposed to contribute to Level 1 trigger → higher power consumption
- **DC-DC converters with conversion ratio of 8-10**



DC-DC converters can be based on many different principles and layouts
 → concentrate here on so-called **buck converters**



Why buck converters?

- High currents with high efficiency
- Comparably simple & compact
- Output voltage regulation by Pulse Width Modulation (not shown)

Challenges

- Radiation tolerance of high voltage (15V) power transistors
- Switching with MHz frequencies → “switching noise“ through cables (conductive)
- Saturation of inductor ferrite cores in magnetic field → air-core inductor
 → radiated noise emissions
- Maximization of efficiency & minimization of material and size



Buck Converter ASICs



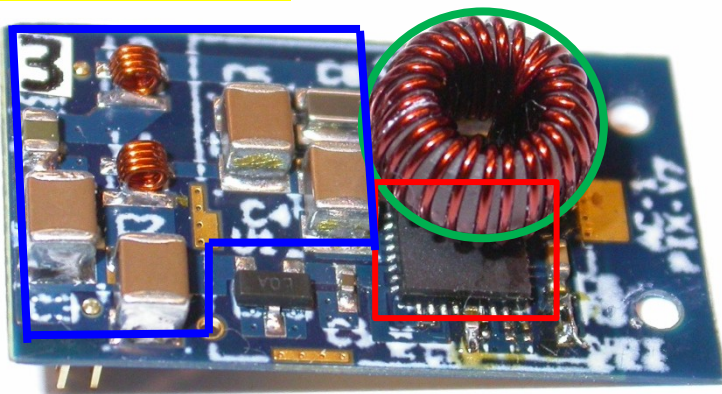
- ASIC includes transistors and voltage regulation circuit
- ASIC is being developed within **CERN electronics group (F. Faccio et al.)**
- Radiation tolerance of many semi-conductor technologies evaluated
 - **AMIS I3T80** 0.35 μ m (ON Semiconductor, US)
 - functional up to dose of 300Mrad & fluence of $5 \cdot 10^{15}$ p/cm²
 - no Single Event Burnout effect
- **AMIS prototypes:**
 - AMIS1 (2008) → AMIS2 (2009) → AMIS3 (problems)
 - AMIS4 with full functionality (submitted in January 11)
- Work with second supplier (IHP, Germany) to improve radiation tolerance
 - two prototypes in 2010, but ASIC development on-hold due to issues

SEB = Single Event Burnout

= ionizing particle in source turns parasitic npn transistor on → destructive current



“PIX_V7“:



ASIC: AMIS2 by CERN

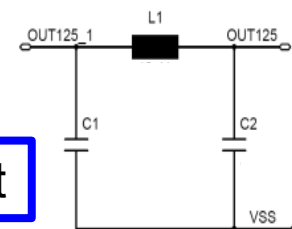
$I_{out} < 3A$
 $V_{in} < 12V$
 f_s configurable, e.g. 1.3MHz

PCB:

2 copper layers a 35 μ m
 0.3mm thick
 Large ground area on bottom for cooling

Toroidal inductor:

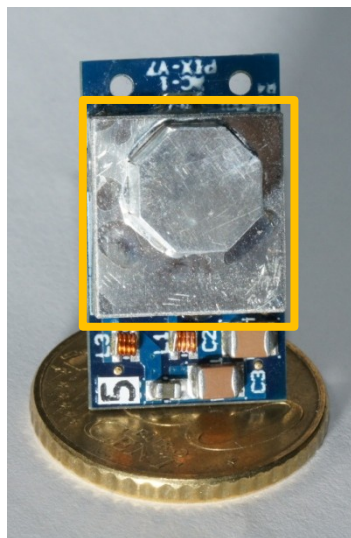
$L = 450nH$
 $R_{DC} = 40m\Omega$
 Plastic core



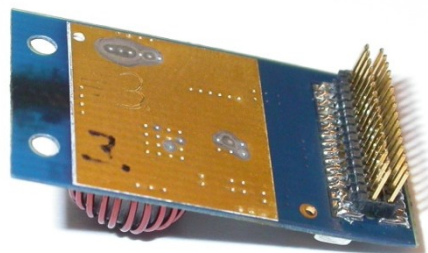
Pi-filters at in- and output

Shield

Design guidelines from CERN group have been implemented.



$A = 28 \times 16 \text{ mm}^2$
 $M \approx 2.5g$
 3.8% of a radiation length



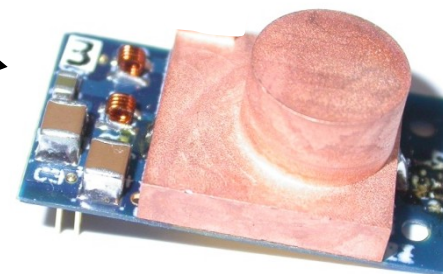
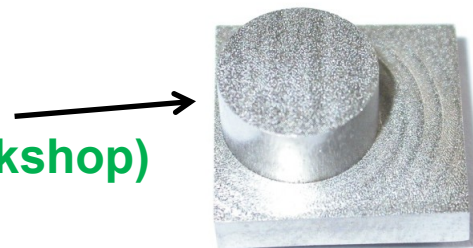


The shield has three functions:

- 1) to shield radiated emissions from inductor
- 2) to reduce conducted noise by means of segregation between noisy and quiet parts of board (less coupling)
- 3) to provide cooling contact for coil through its solder connection to PCB, since cooling through contact wires not sufficient

Several technologies are under evaluation:

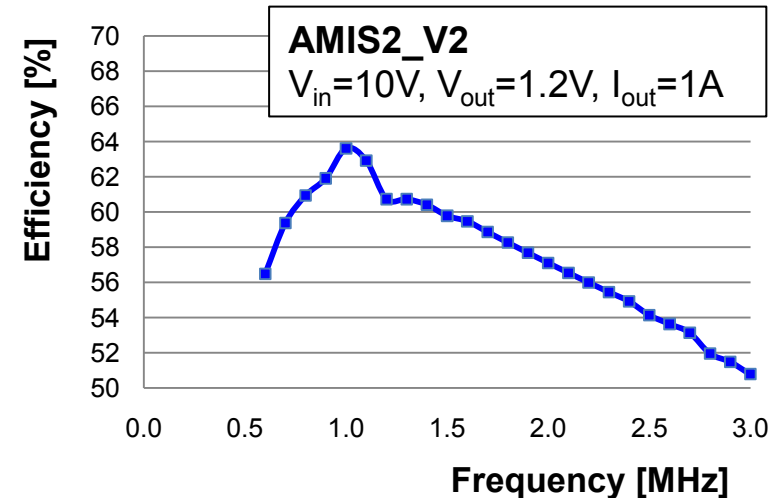
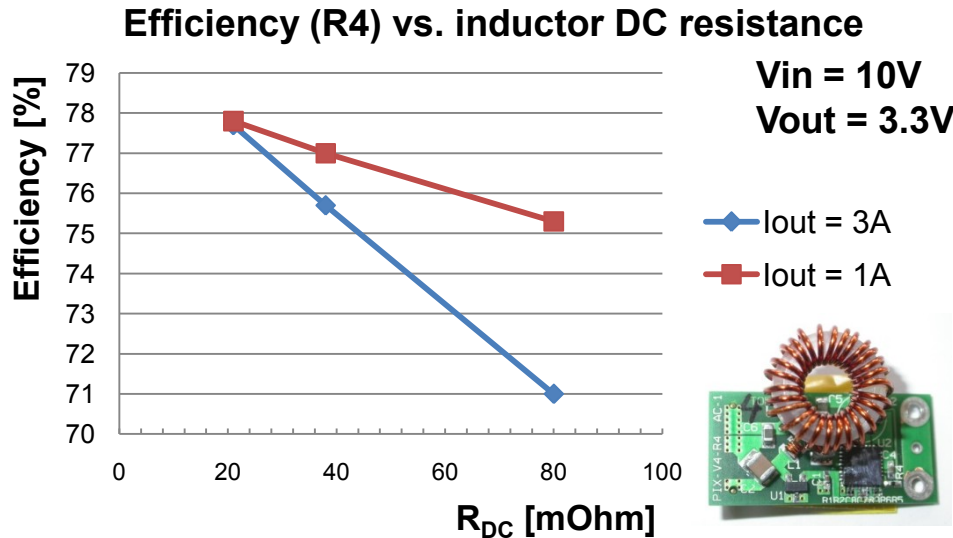
- **Aluminium shields of 90 μ m thickness (milled in our Workshop)**
- **Plastic shields (PEEK) coated with a metal layer**
e.g. galvanic deposition of copper (30 μ m – 60 μ m)

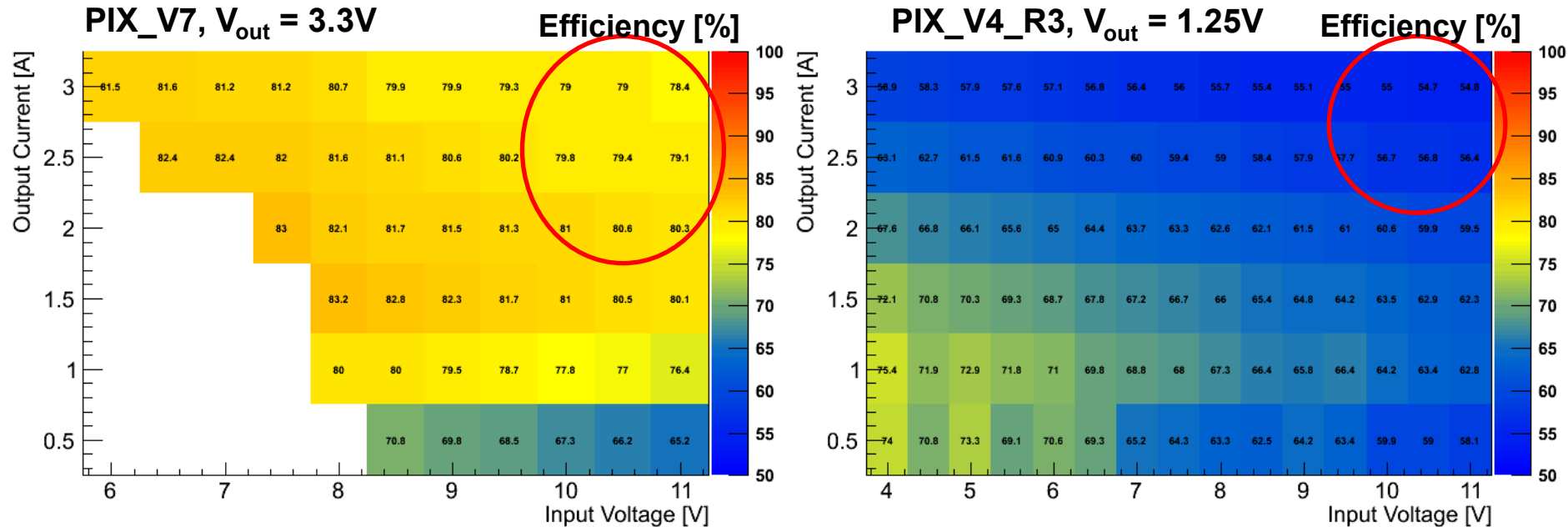


Shape driven by geometrical constraints



- Efficiency = P_{out} / P_{in}
- Resistive losses from
 - chip (R_{on} of transistors)
 - wire bonds
 - inductor
- Resistive losses $\sim 1/f_s$; switching & driving losses $\sim f_s$
- Need to balance efficiency vs. mass, volume & EMC





[White regions: regulation not working properly, V_{out} too low]

- **Phase 1 conditions:** $V_{out} = 3.3V$ or $2.5V$, $I_{out} < 2.8A$, conversion ratio of 3-4
 → **75% - 80% efficiency:** ok

- **Phase 2 conditions:** $V_{out} = 1.25V$, $I_{out} = 3A$, conversion ratio of 8-10
 → about **55% efficiency:** too low

Possible solution: combine with a on-chip “switched capacitor” converter with $r = 2$

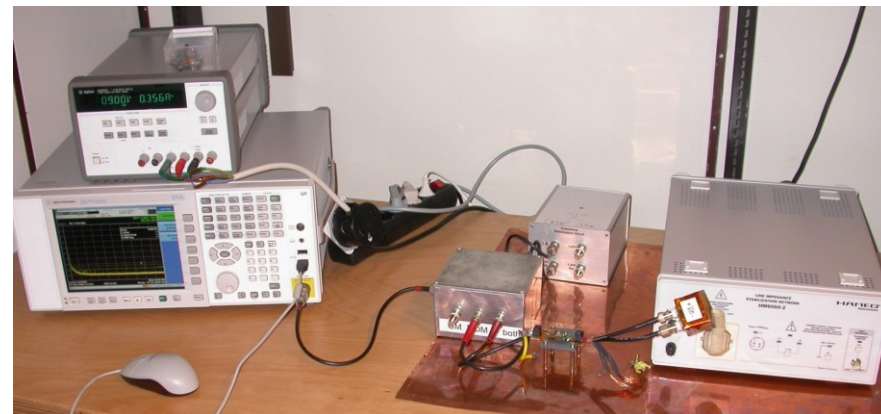
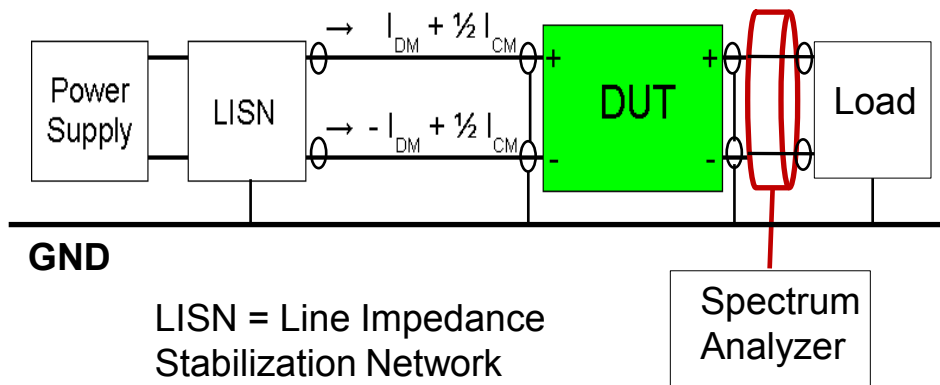


Conductive Noise

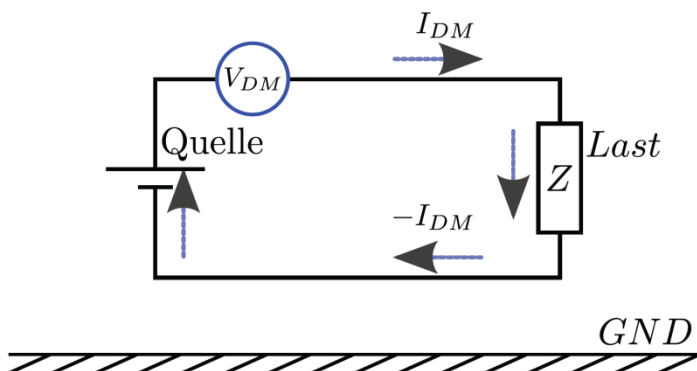


Noise through cables (conductive noise) was studied with EMC set-up

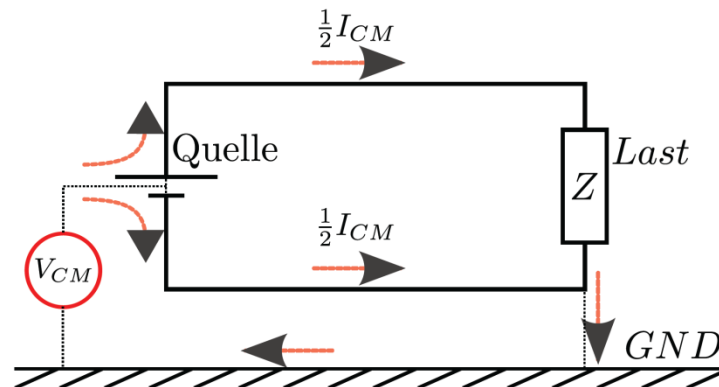
EMC = electromagnetic compatibility



Differential Mode (DM), “ripple“



Common Mode (CM)

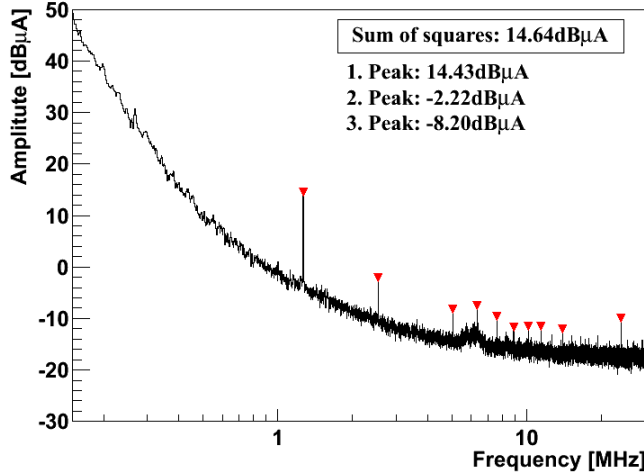




Conductive Noise

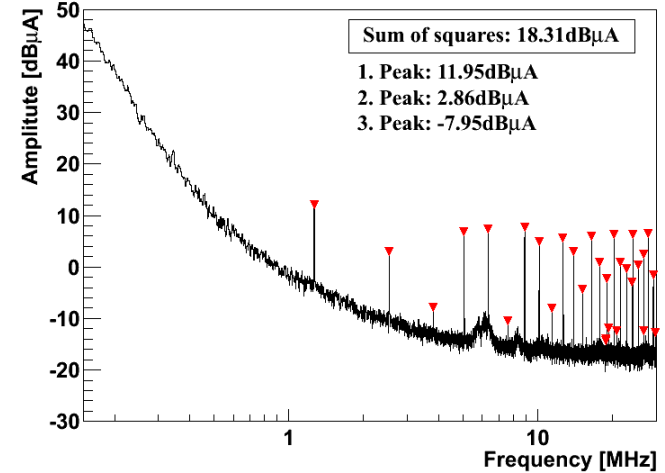


Differential Mode, no shield

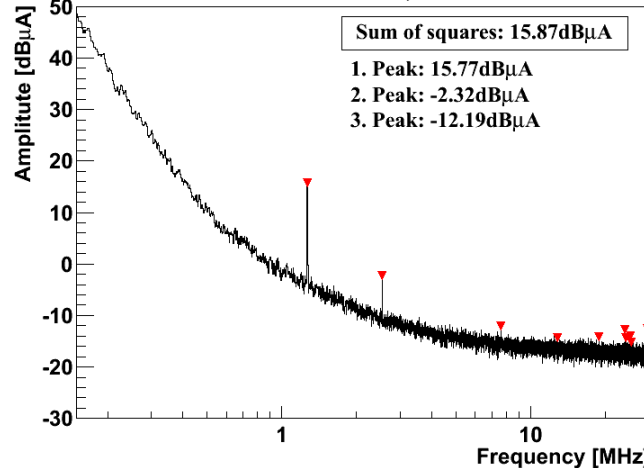


PIX_V7
output noise
 $V_{out} = 3.3V$
 $V_{in} = 10V$
 $f_s = 1.3MHz$
 $L = 450nH$

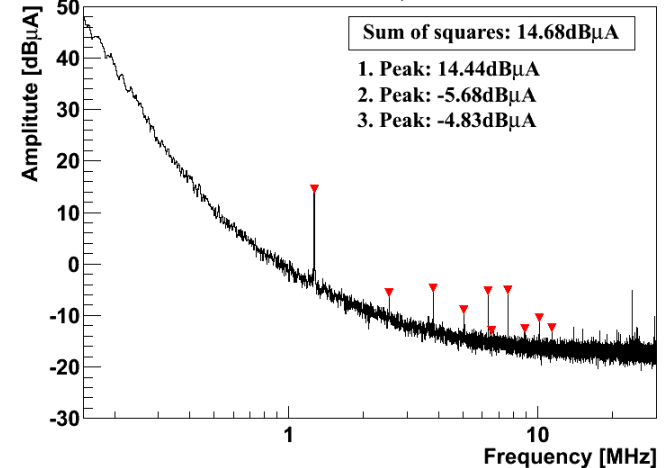
Common Mode, no shield



Differential Mode, with shield



Common Mode, with shield



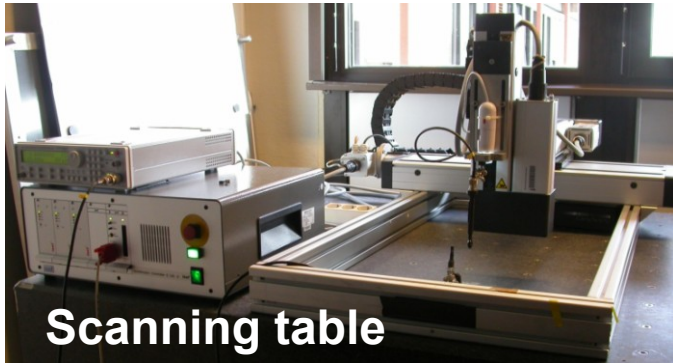
→ Large reduction of CM above 2 MHz due to shield



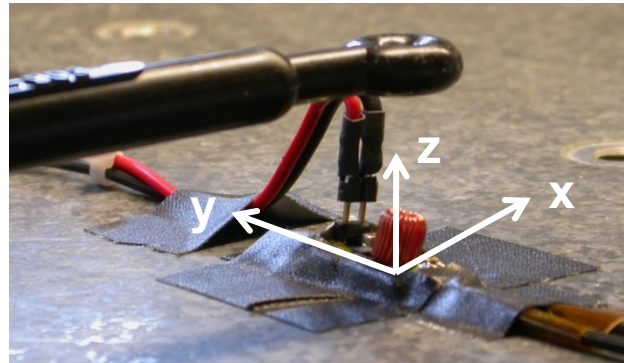
Radiated Noise Emissions



- Large fast changing currents through inductor \rightarrow magn. near field can induce noise
- Field of air-core **toroid** has been measured and inductor shape optimized

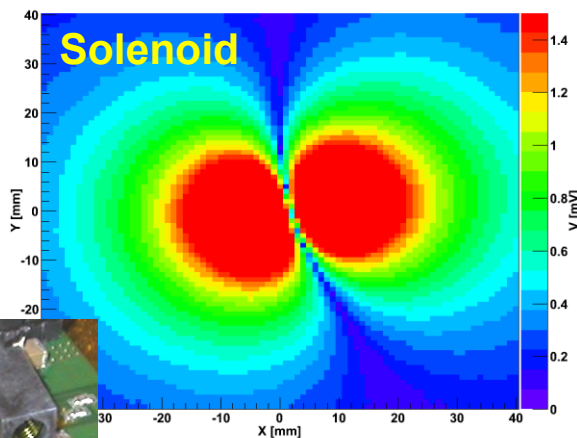


Scanning table



Emitted field is measured with a pick-up probe and spectrum analyzer [height of 1. peak]

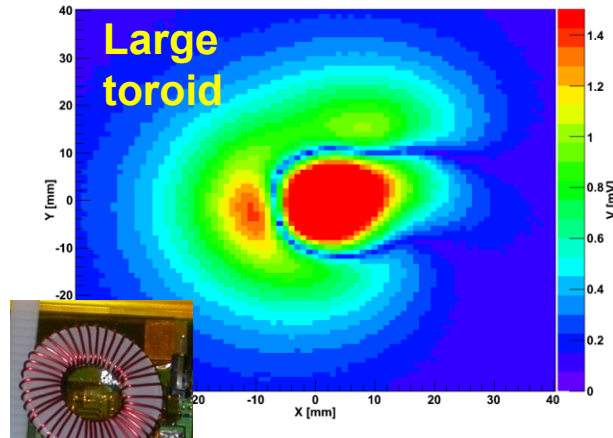
B_z measured in x-y-plane, 1.5 mm above coil:



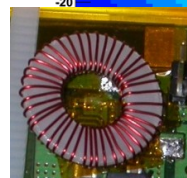
538nH, 90m Ω , 500mg



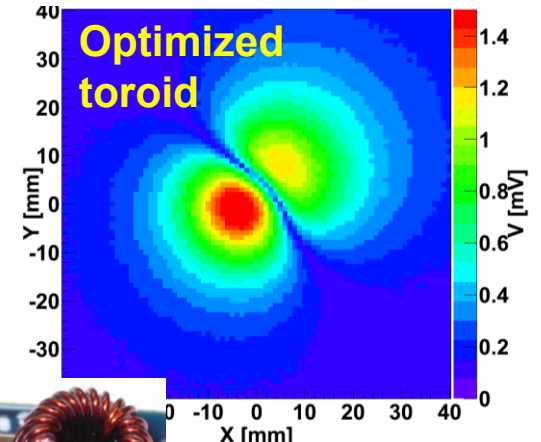
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618nH, 104m Ω , 783mg



Powering for Future Detectors



450nH, 40m Ω , 650mg



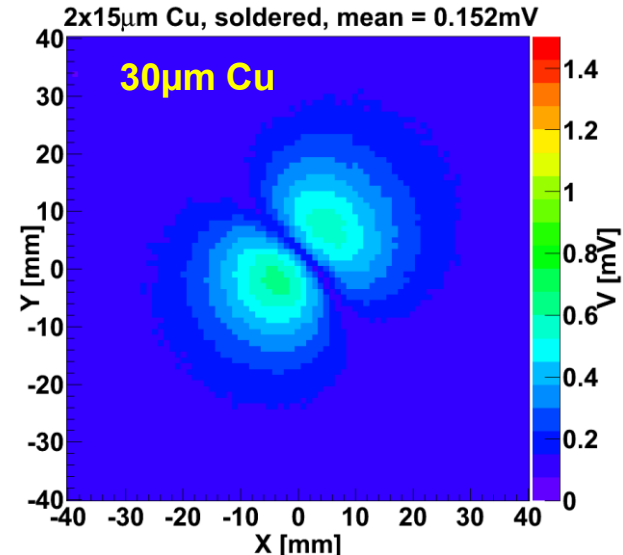
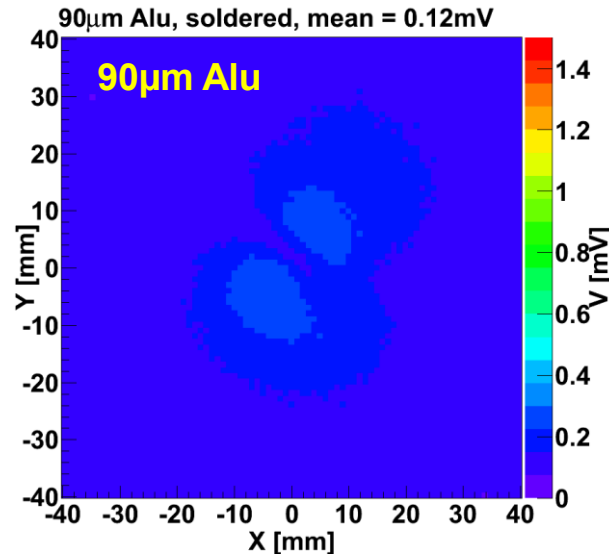
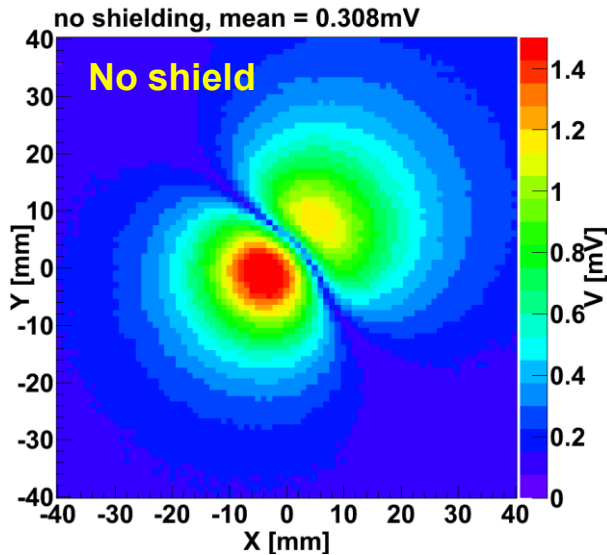


Shielding from Radiated Noise



Shielding of magnetic field: Eddy currents in metallic shield

- 90 μm milled Aluminium shield works fine
- Plastic shield coated with 30 μm Cu worse and adds $\sim 40\%$ more material (but probably cheaper)





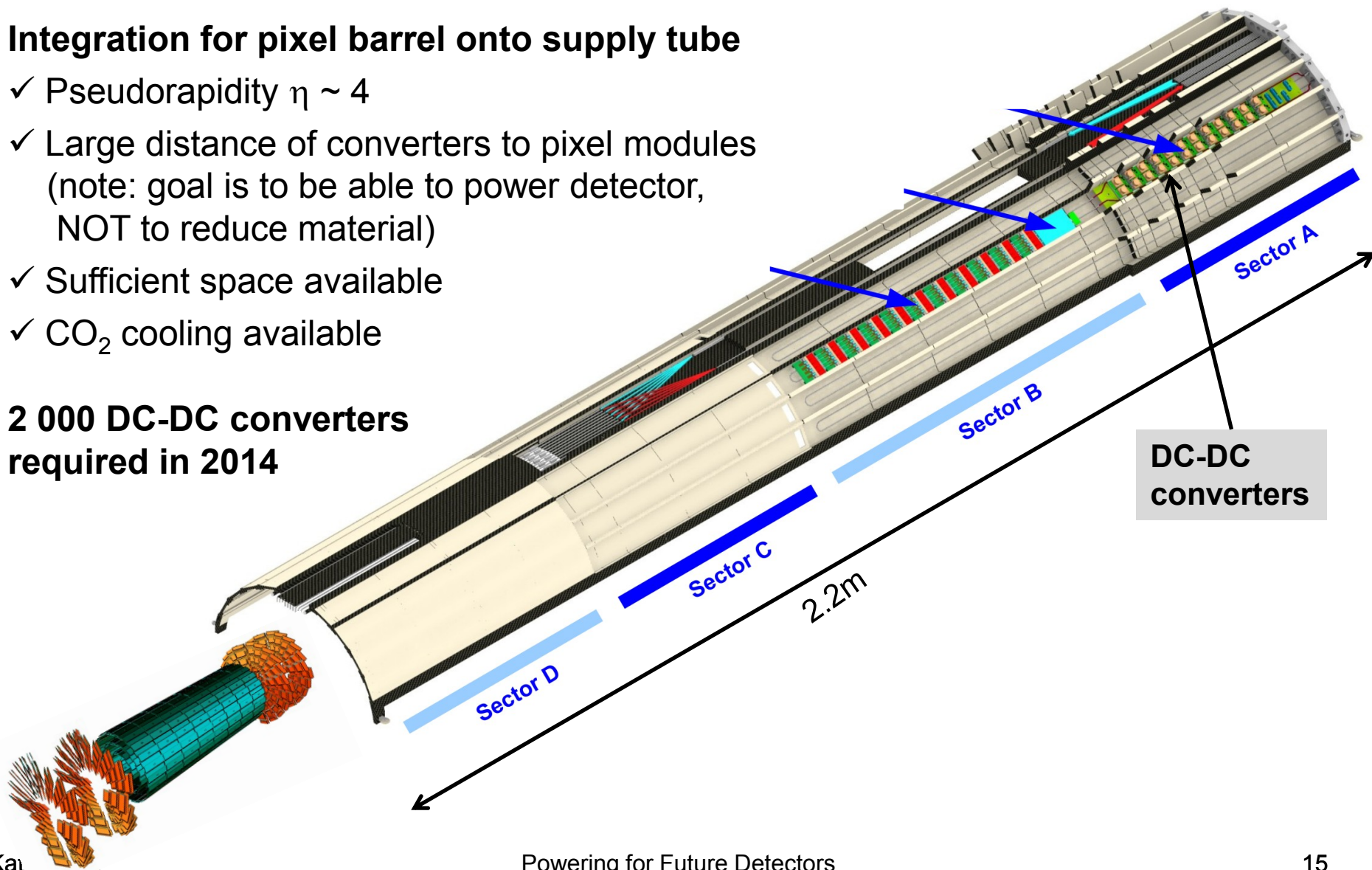
Integration into Phase-1 Pixel Detector



Integration for pixel barrel onto supply tube

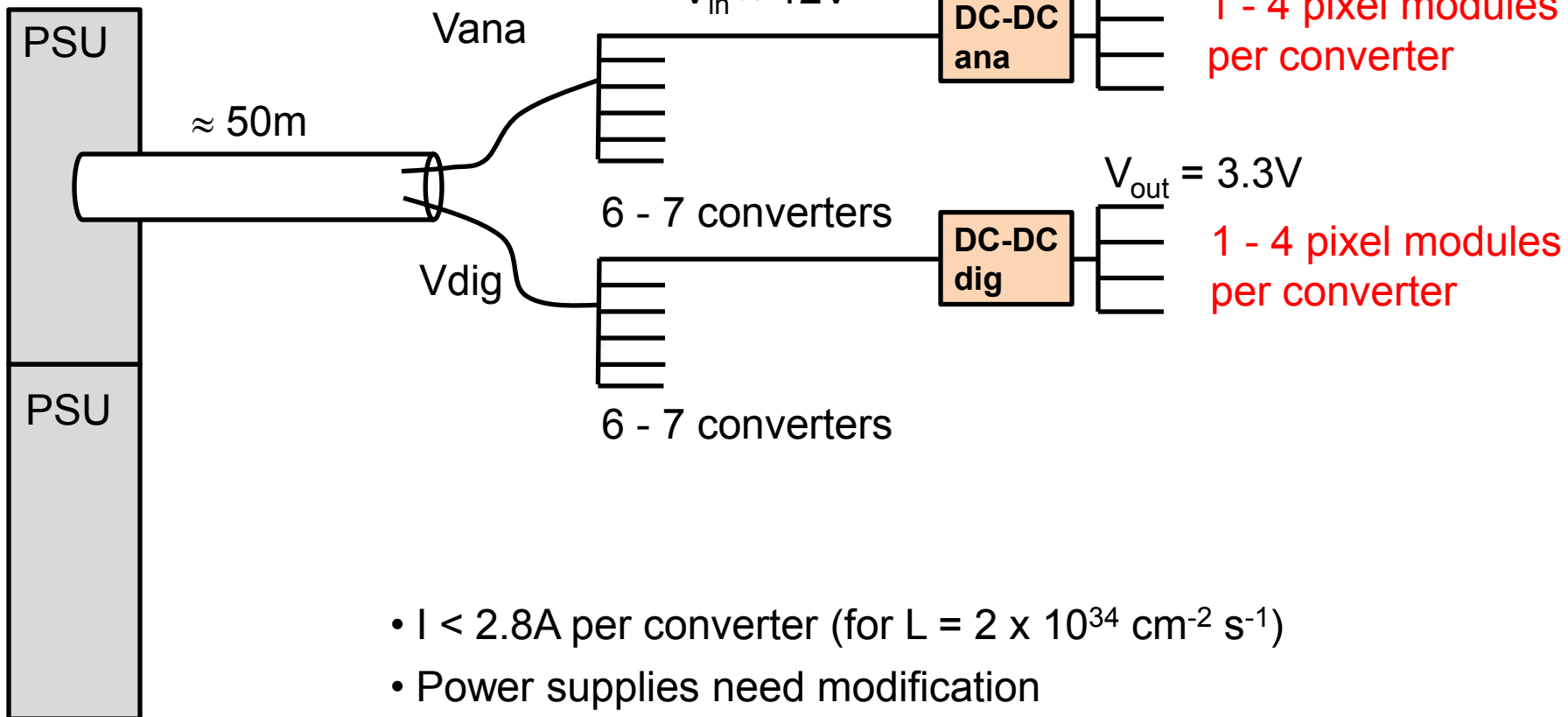
- ✓ Pseudorapidity $\eta \sim 4$
- ✓ Large distance of converters to pixel modules (note: goal is to be able to power detector, NOT to reduce material)
- ✓ Sufficient space available
- ✓ CO₂ cooling available

2 000 DC-DC converters required in 2014





CAEN
A4603



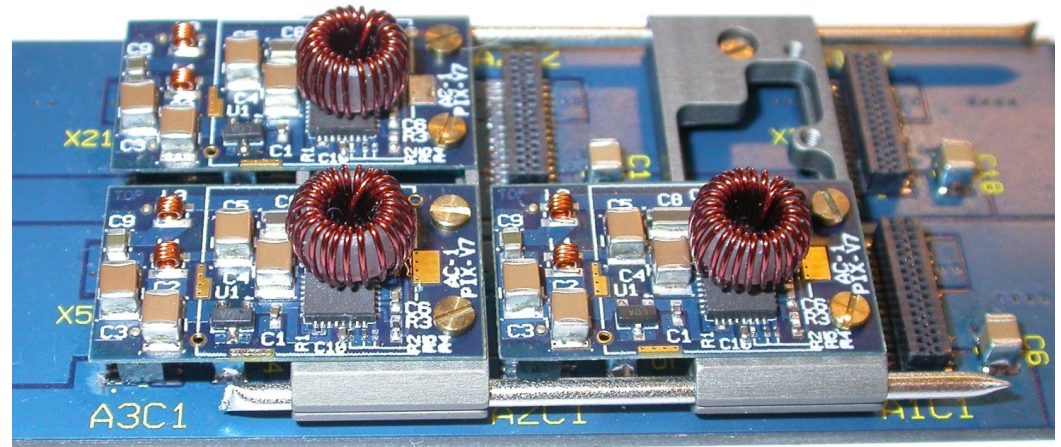
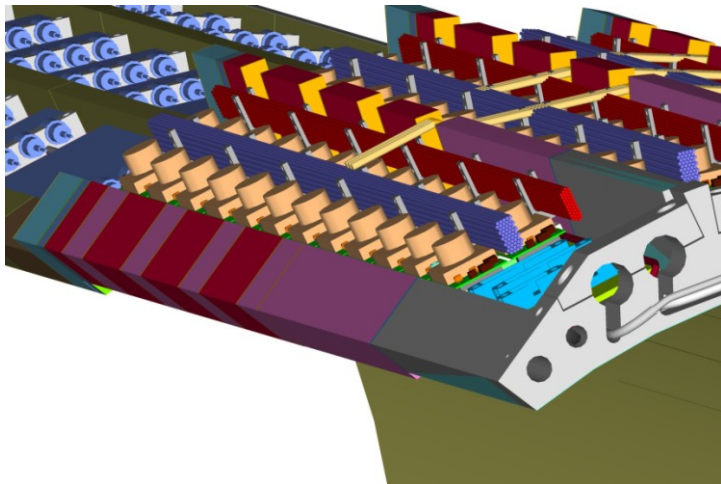
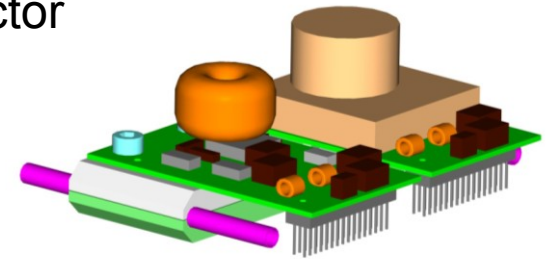
- $I < 2.8A$ per converter (for $L = 2 \times 10^{34} \text{ cm}^{-2} \text{ s}^{-1}$)
- Power supplies need modification
- No remote sensing

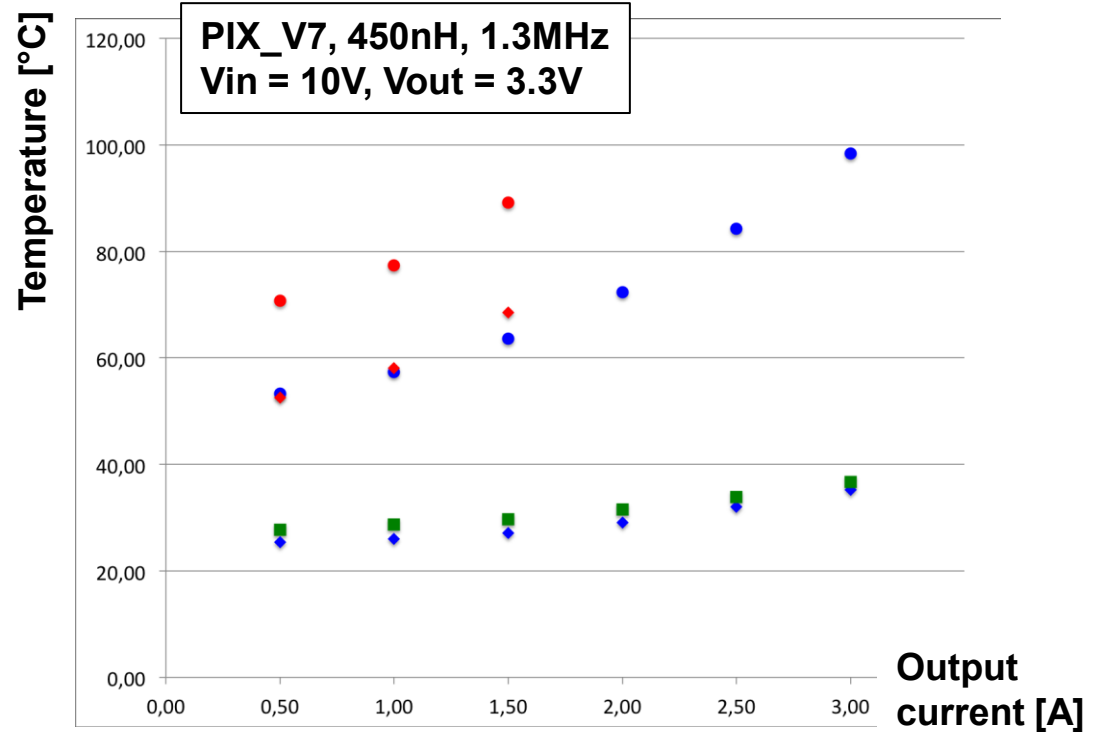
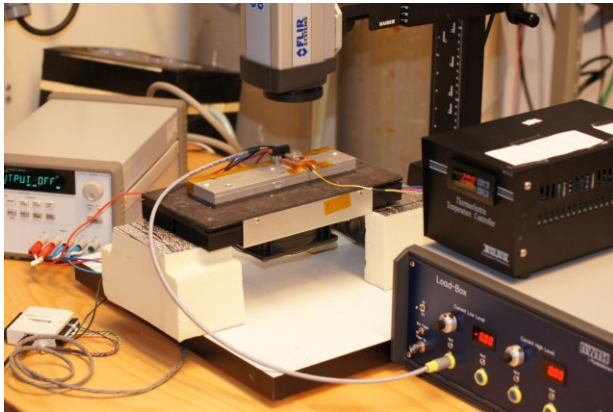


Integration into Phase-1 Pixel Detector



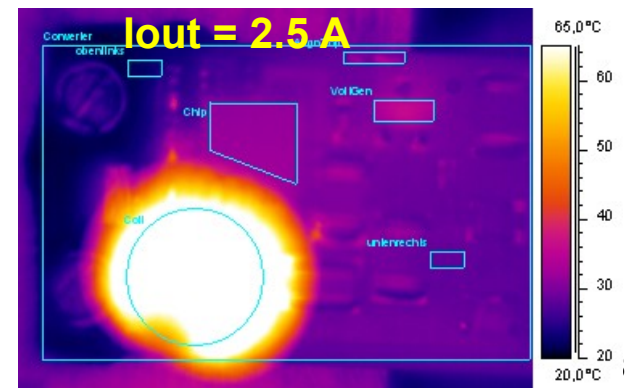
- 26 DC-DC converters per channel
- Power dissipation $\sim 50\text{W}$ per channel
- Cooling bridges clamp around CO_2 pipes
- Chip cooled through PCB backside
- Shield (soldered to PCB) acts as cooling contact for inductor





- Coil without cooling
- ◆ Chip without cooling
- Coil with cooling, no shield
- ◆ Chips with cooling, no shield
- Shield temperature

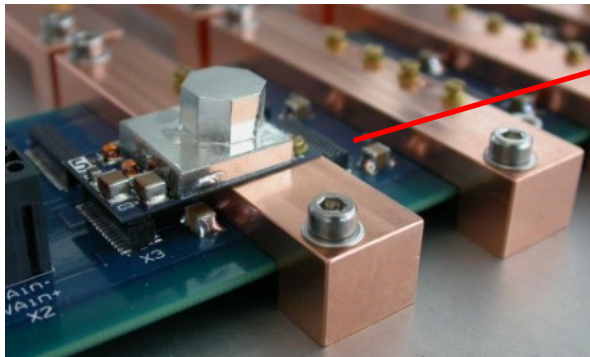
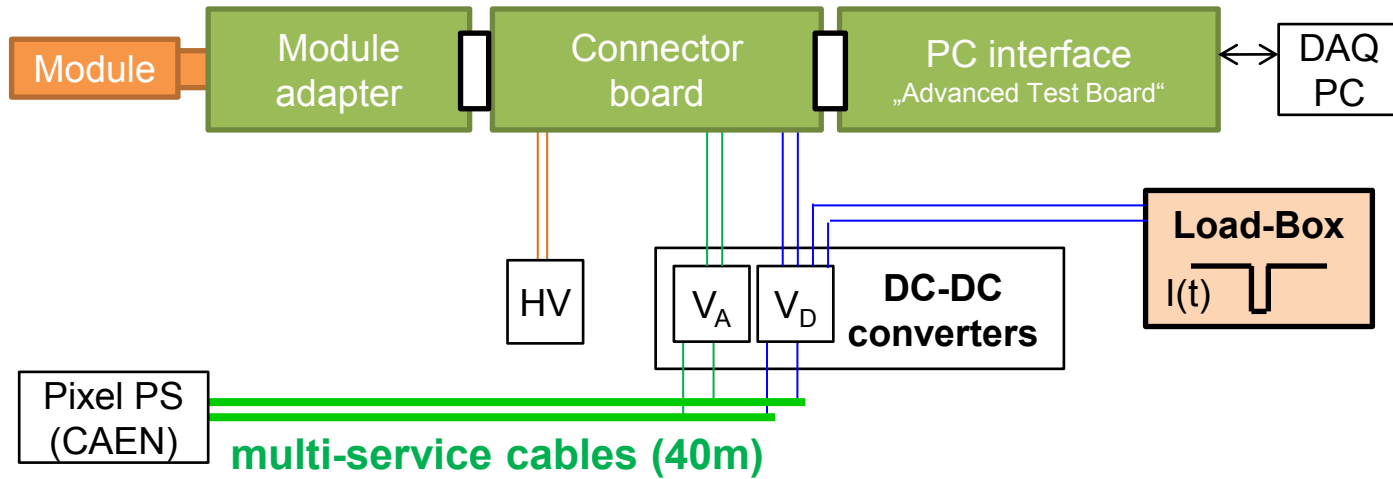
- Measurements with Flir infrared camera
- Peltier element set to +20°C
- Cooling of chips via backside of PCB is very effective
- Coil needs to be connected to cooling contact (shield)
- Good agreement with Finite Element simulations



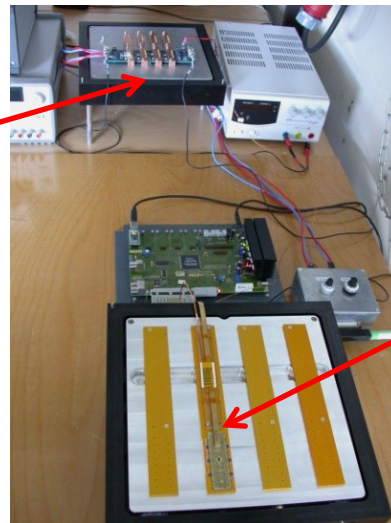
System Tests with Pixel Modules



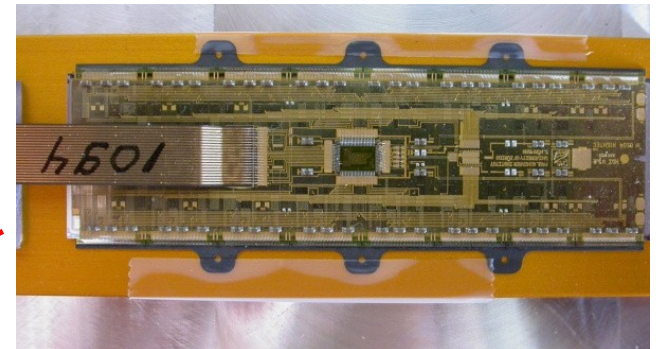
The effect of buck converters on the noise of today's pixel modules has been studied:



DC-DC converter on bus board



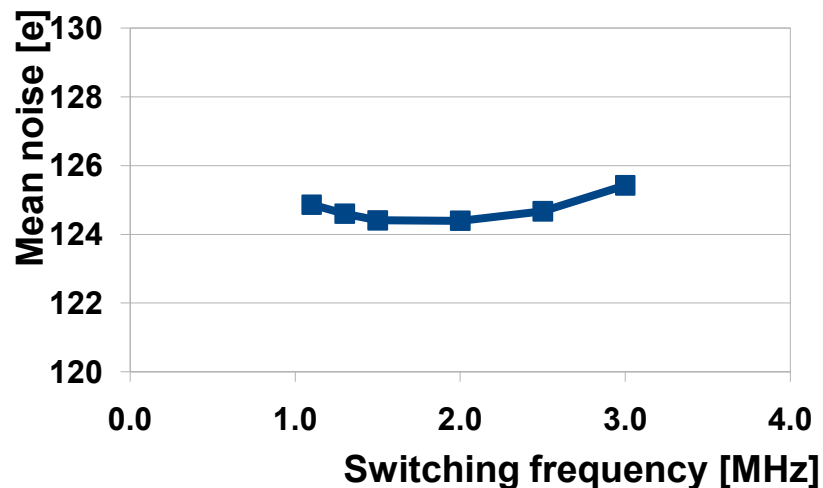
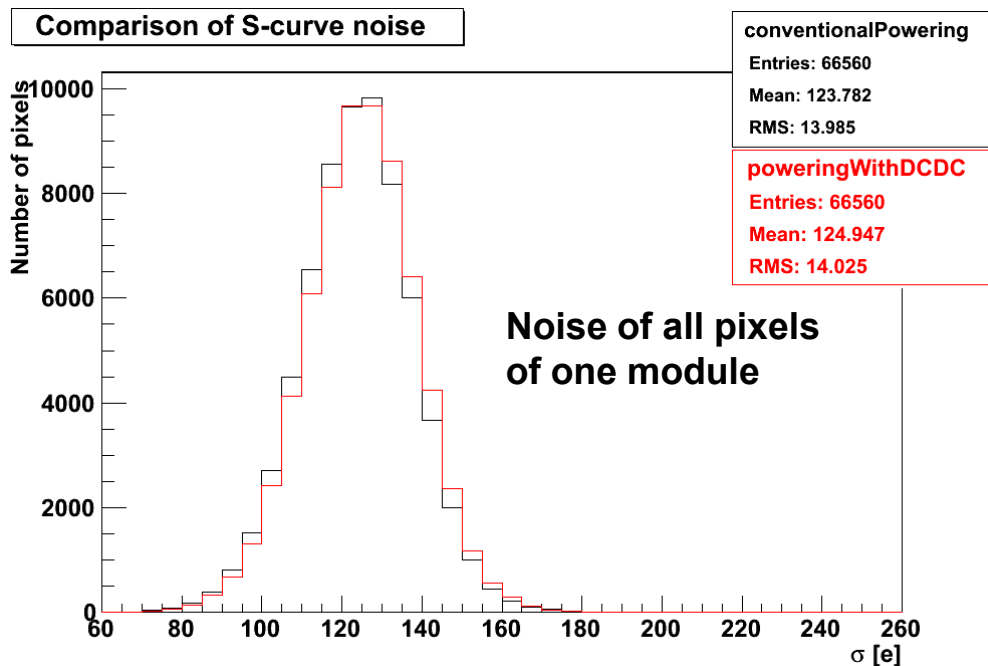
Powering for Future Detectors



Pixel module



- **Threshold scan:** efficiency for internal calibration pulse vs its amplitude
- Fit “s-curve“ with error function → width corresponds to noise

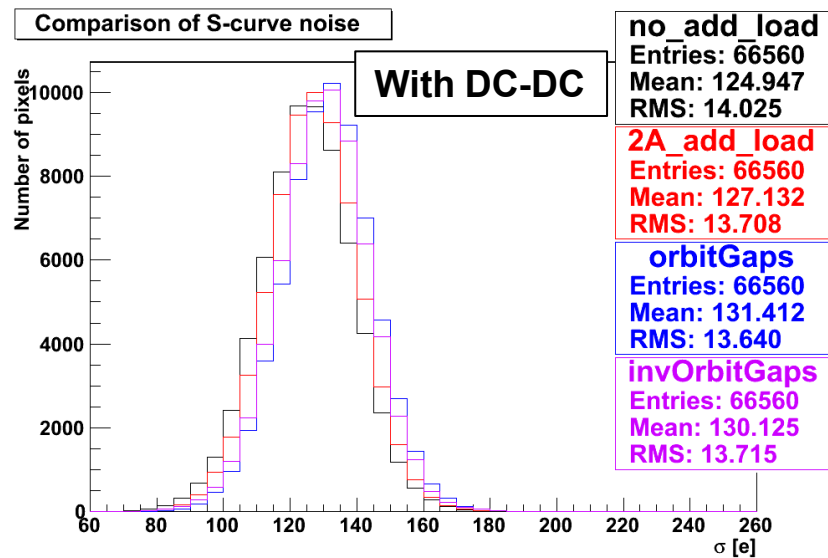
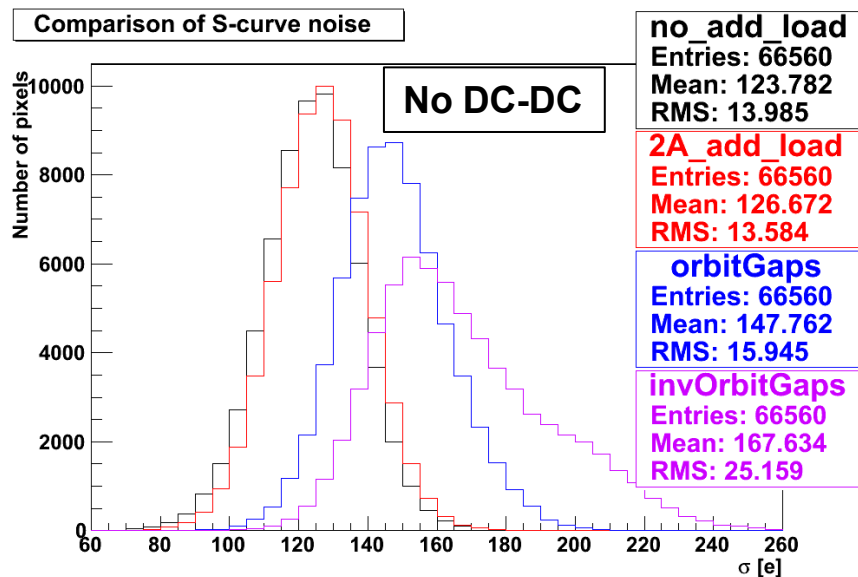


- Change in noise due to DC-DC converter is below 1%
- Noise is flat over considered switching frequency range (1-3 MHz)



- Sparsified readout → digital power consumption depends on particle fluence
- LHC bunches are not equally distributed: **3μs “abort gap” every 89μs is not filled**
- Digital current per converter drops within ~ 50ns from **2.7A to 1.0A** ($2 \cdot 10^{34} \text{cm}^{-2}\text{s}^{-1}$)
 → stability of power supply chain for large load variations to be checked

Result: Sensitivity to load changes with DC-DC converters much reduced





Summary



- **Novel powering schemes have to be exploited for the LHC upgrades**
- **CMS tracker has opted for a DC-DC conversion powering scheme**
- **Prototypes with sufficient efficiency and low noise in hands**

- **Next big step: AMIS4 ASIC (expected in summer)**

- **Many more things to be done:**
 - **More realistic system tests**
 - **Controls**
 - **Mass reduction for phase-2 (e.g. aluminium coil)**
 - **Establish efficient scheme for larger conversion ratios (e.g. 2 stages)**