



LLRF for EUCARD Crab Cavities

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Crab Parameters







Tolerance for Gaussian Amplitude Errors







To minimise required cavity kick R12 needs to be large (25 metres suggested) Vertical kicks from unwanted cavity modes are bad one needs R34 to be small. For 20 mrad crossing and using as 12 GHz structure

$$V_{\text{crab}} = \frac{\theta_{\text{c}} E_{\text{o}} c}{2R_{12}\omega} = \frac{2 \times 10^{-2} \times 1.5 \times 10^{12} \times 3 \times 10^{8}}{2 \times 25 \times 2\pi \times 12 \times 10^{12}} = 2.4 \text{ MV}$$

Error in kick tilts effective collision from head on.

Luminosity Reduction Factor

$$S \approx \frac{1}{\sqrt{1 + \left\{\frac{\sigma_z \theta_c}{4\sigma_x} \frac{\left(\left|\delta V_1\right| + \left|\delta V_2\right|\right)}{V_{crab}\right\}^2}} \qquad \text{gives}$$

amplitude error on each cavity	1.0%	1.5%	2.0%	2.5%	3.0%
luminosity reduction	0.9953	0.9914	0.9814	0.9714	0.9596













CLIC Solution



Wakefields

Phase and amplitude control

Beamloading compensation

Phase synchronisation (4.4 fs)

Phase reference

Phase measurement calibration

Phase stability

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Large irises Small number of cells Strong damping

Passive during 156 ns bunch train High energy flow through cavity * small number of cells * high group velocity * low efficiency

Same klystron drives both cavities Temperature stabilized waveguide

Optical interferometer

DBM and Down conversion to ~ 1 GHz, Digital phase detection Staggered sample and hold

Thick irises Strong cavity cooling





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Fill Time and Beamloading



For each cell solve energy equation

$$\frac{dU_n}{dt} = \frac{(U_{n-1} - U_n)}{L_{cell}} v_g - U_n \frac{\omega}{Q} - q f_{rep} \delta x \omega \sqrt{\frac{\omega}{c} \frac{R}{Q} U_n}$$

convection - dissipation - beamloading

(n > 1)

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Cavity Parameters as on last slide and Q = 6381

Beam offset (mm)	-0.4	0.0	0.4
Power entering cell 1 (MW)	6.388	6.388	6.388
Power leaving cell 16 (MW)	5.619	5.341	5.063
Ohmic power loss (MW)	1.071	1.047	1.023
Beamload power loss (MW)	-0.302	0.000	0.302
E max for cell 1 (MV/m)	51.1	51.1	51.1
Efficiency	12.04%	16.39%	20.74%
Kick (MV)	2.428	2.400	2.372

A short inefficient cavity with a high power flow achieves adequate amplitude stability

Can we make the gradient?

Is pulse heating OK (consider low temperature operation)?





Cavity variables



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Number of cells for 2.4 MV transverse kick not exceeding surface field 110 MV m⁻¹.











Cavity power requirement for a beam offset of 0.375 mm







Beam Loading

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- As the electric field in dipole cavities vary with offset the beamloading changes with time and cannot be predicted.
- The beam is too short for feedback.
- Need to design to minimise the effect (increase convection or dissipation)









Luminosity loss



Increasing dissipation increases heating, this is not recommended

We can increase power convection by increasing the structure group velocity.

The group velocity is dependant on the iris radius.

However we have limited power (20 MW) so we can only increase the convection so much.



Luminosity loss for a beam offset of 0.125 mm







Wakefactor ~ R/Q times cells



R/Q multiplied by cell number plotted against iris radius to give figure of merit for wakefields





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Assume 40 m waveguide run from Klystron to each Crab cavity

For copper s =5.8e7 S/m and at 11.994 GHz	Attenuation	Transmission	Over moded
Rectangular TE10 EIA90 (22.9 x 10.2 mm)	0.098 dB/m	40.6%	no
Rectangular TE10 special (24 x 14 mm)	0.073 dB/m	51.3%	no
Circular TE11 (r = 9.3 mm)	0.119 dB/m	33.3%	no
Circular TE11 (r = 12 mm)	0.055 dB/m	60.4%	TM10
Circular TE01 (r = 40 mm)	0.010 dB/m	91.2%	extremely

Available klystron has nominal output of 50 MW Divide output for two beam lines = 25 MW For standard rectangular waveguide we have 10.2 MW available (OK for 15 cells) For special rectangular waveguide we have 12.8 MW available (OK for 11 cells) For circular 12mm TE11 waveguide we have 15.1 MW available (OK for 11 cells)

(note that mode conversion from circular TE11 to circular TM10 is vanishingly small for properly designed bends)



CLIC LLRF Timing

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Crab Cavity RF



- Beamloading constrains us to high power pulsed operation
- Intra bunch phase control looks impossible for a 140 ns bunch

SOLUTION

- One Klystron (~ 20 MW pulsed) with output phase and amplitude control
- Intra bunch delay line adjustment for phase control (i.e. between bunch trains)
- Very stable cavities



RF Layout and Procedure



Once the main beam arrives at the crab cavity there is insufficient time to correct beam to cavity errors. These errors are recorded and used as a correction for the next pulse.

- 0. Send pre-pulse to cavities and use interferometer to measure difference in RF path length (option1)
- 1. Perform waveguide length adjustment at micron scale (option 2 use measurements from last pulse)
- 2. Measure phase difference between oscillator and outward going main beam
- 3. Adjust phase shifter in anticipation of round trip time and add offset for main beam departure time
- 4. Klystron output is controlled for constant amplitude and phase
- 5. Record phase difference between returning main beam and cavity
- 6. Alter correction table for next pulse

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Phase measurement

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- r.m.s. cavity to cavity synchronisation requirement is 4.4 fs
- hence r.m.s klystron to cavity stability requirement is 3.1 fs (two paths)
- phase velocity of light in the waveguide will be just over 3.0×10⁸
- hence waveguide length must be steady at the precision of 10⁻⁶ metres

(Expansion example without expansion joints) Control waveguide temperature to say 0.3° C. Copper expansivity 17 x 10^{-6} K⁻¹. 40 metre waveguide could vary in length by 200 μ m. Waveguide wavelength ~ 25 mm Expansion ~ 200 μ m Phase shift ~ 2.9 degrees which is 150 times the allowance!

It is probably that the waveguide will have expansion joints and so the real question is about the lateral stability of the cavity and the klystron.









Double balanced mixer Mini-circuits SIM-24MH+ RF 7.3 GHz to 20 GHz, IF DC to 7.2 GHz Need about 24 dB of amplification to resolve 1 milli-degree at 12 GHz, signal to noise before amplifier for 40 MHz bandwidth ~ 6 dB Measurement looks feasible.

But need digital phase detection as before for calibration









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Board Development



The calibration of the DBM is performed using a digital phase detector after down-conversion to 1.3 MHz.

This hardware is underdevelopment but is so far not meeting the resolution specification.



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