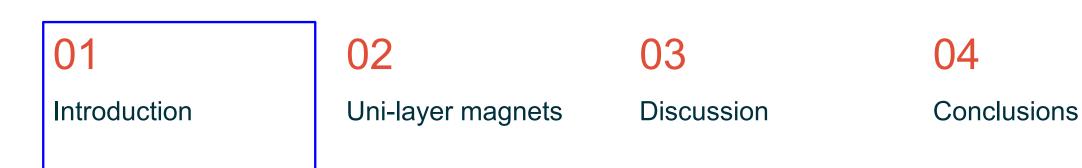




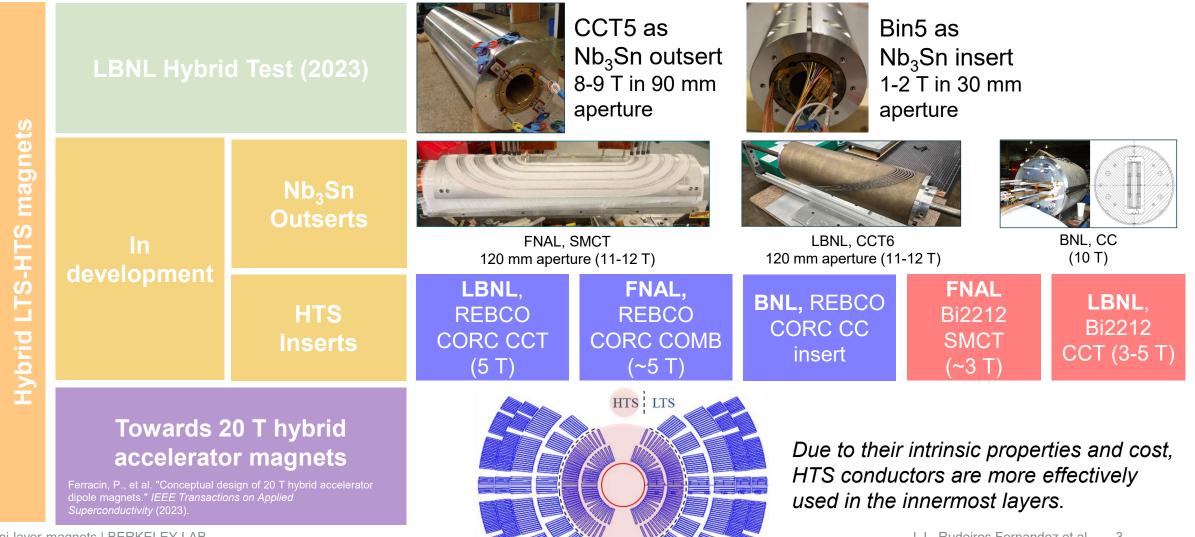
J. L. Rudeiros Fernandez, P. Ferracin

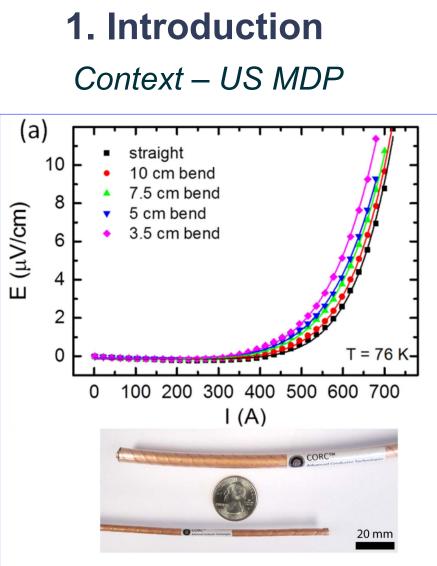
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1. Introduction

Context – US Magnet Development Program: Towards very high-field magnets





Weiss, Jeremy D., et al. "Introduction of CORC® wires: highly flexible, round high-temperature superconducting wires for magnet and power transmission applications." *Superconductor science and technology* 30.1 (2016): 014002.

A few examples of HTS insert magnet concepts



V. Kashikhin (Fermilab)

X. Wang

(LBNL)

A. Zlobin *(Fermilab)*

1. Introduction

Key challenges in the performance and cost of future high-field superconducting magnets:

- Magnet design's ability to deal with high Lorentz forces and resulting strain-stress in the strain-sensitive conductor (i.e. **stress management structures**).
- Efficient use of the conductor to create a certain magnetic field (i.e. required length of conductor for an equivalent integrated field along the beam path). In terms of efficiency, the ability to implement "grading" is also crucial.
- Magnet design's that enables the effective use of HTS superconductors (i.e. magnet designs that could leverage all the potential of HTS superconductors without subjecting it to degradation due to required geometrical or manufacturing conditions).
- Easiness, scalability, and cost-effective manufacturing of the coil and magnet.

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Two special cases of a system of infinitely long current lines

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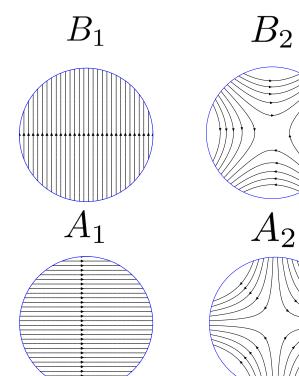
A few words on field-quality

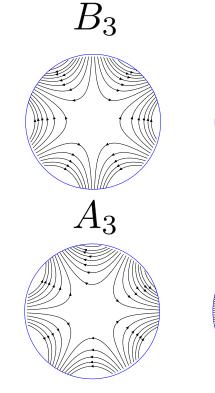
 $C_n = B_n + iA_n$

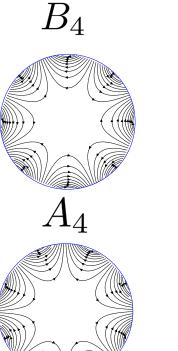
$$\mathbf{B}(z) = \sum_{n=1}^{\infty} C_n \left(\frac{z}{R_{ref}}\right)^{n-1} \quad \text{with} \quad z \in D$$

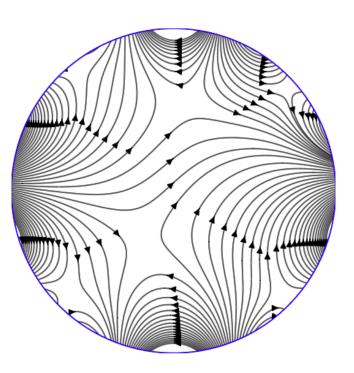
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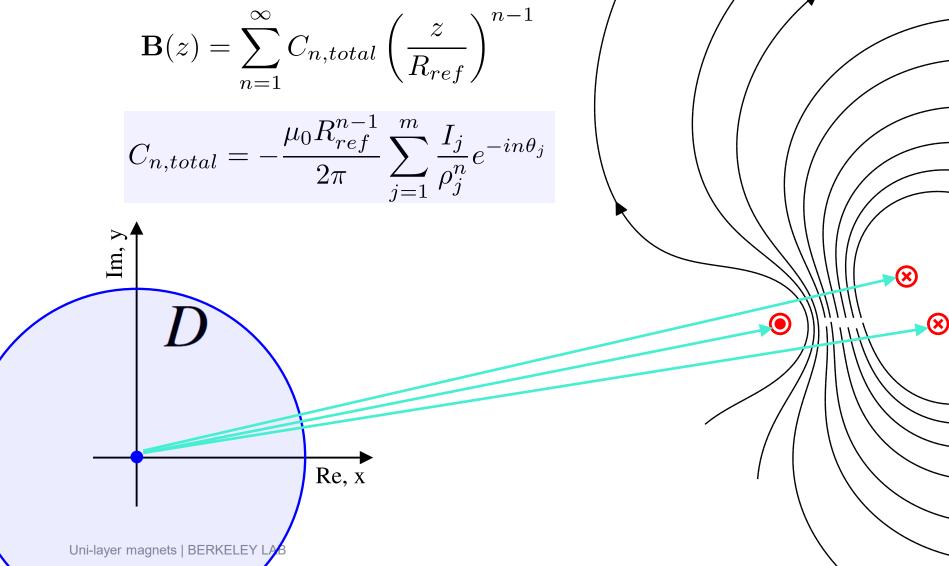




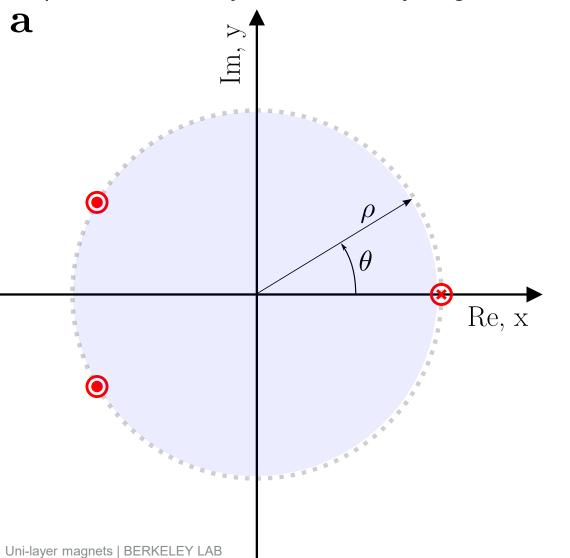




A few words on field-quality



A special case of a system of infinitely long current lines



$$\mathbf{B}(z) = \sum_{n=1}^{\infty} C_{n,total} \left(\frac{z}{R_{ref}}\right)^{n-1}$$

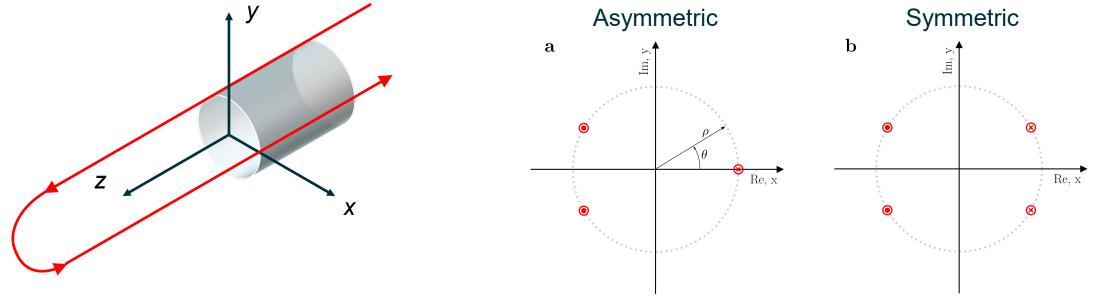
$$C_{n,total} = -\frac{\mu_0 R_{ref}^{n-1}}{2\pi} \sum_{j=1}^m \frac{I_j}{\rho_j^n} e^{-in\theta_j}$$

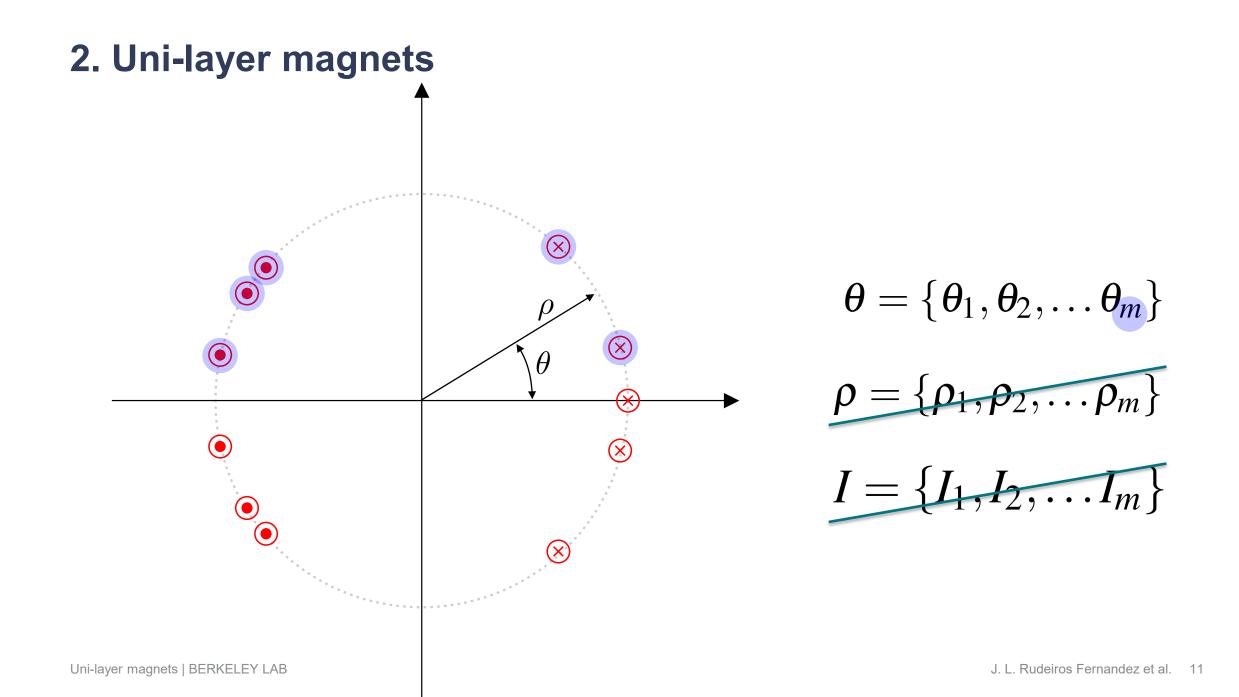
$$C_{n,total} = -\frac{\mu_0 I}{2\pi} \frac{R_{ref}^{n-1}}{\rho^n} \sum_{j=1}^m s_j e^{-in\theta_j}$$

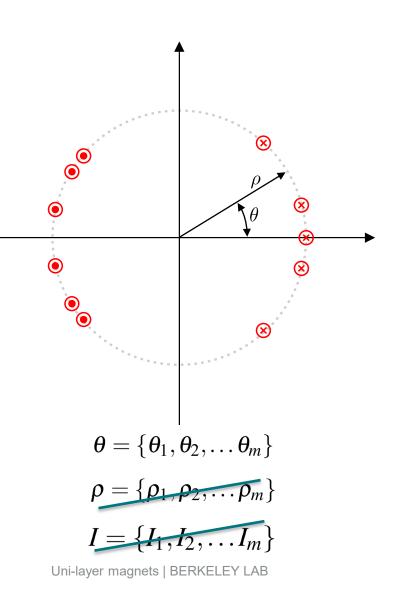
$$C_{n,t} = \frac{\mu_0 I}{2\pi} \frac{R_{ref}^{n-1}}{\rho^n} \left(-1 + 2\cos(n\theta_t)\right)$$

Definition

An idealized **Uni-layer** magnet can be defined as a magnet that generates a **B** field within a straight region of space by a system of current lines, all parallel to the *z-axis* (along the straight section of the magnet) of a Cartesian coordinate system (i.e. perpendicular to the *xy-plane*), that lay within a single continuous surface, and that are connected by a single continuous path that does not cross itself.







Harmonic of interest: **B**₁

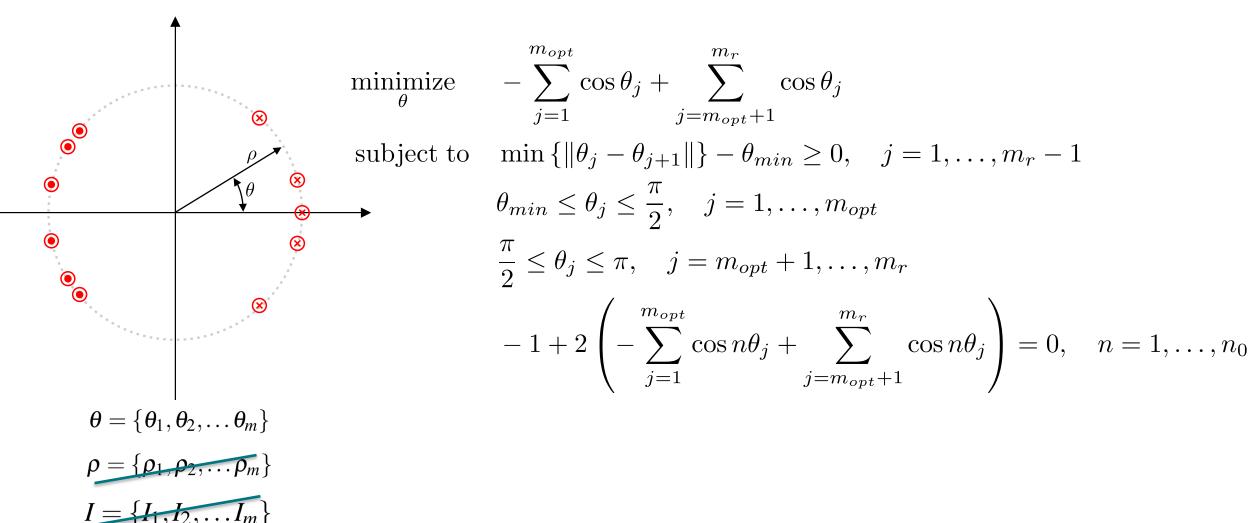
$$C_{n,total} = -\frac{\mu_0 I}{2\pi} \frac{R_{ref}^{n-1}}{\rho^n} \sum_{j=1}^m s_j e^{-in\theta_j}$$
$$-\sum_{j=1}^{m_{opt}} \cos\theta_j + \sum_{j=m_{opt}+1}^{m_r} \cos\theta_j$$

Geometrical constraint: Minimum distance between conductors

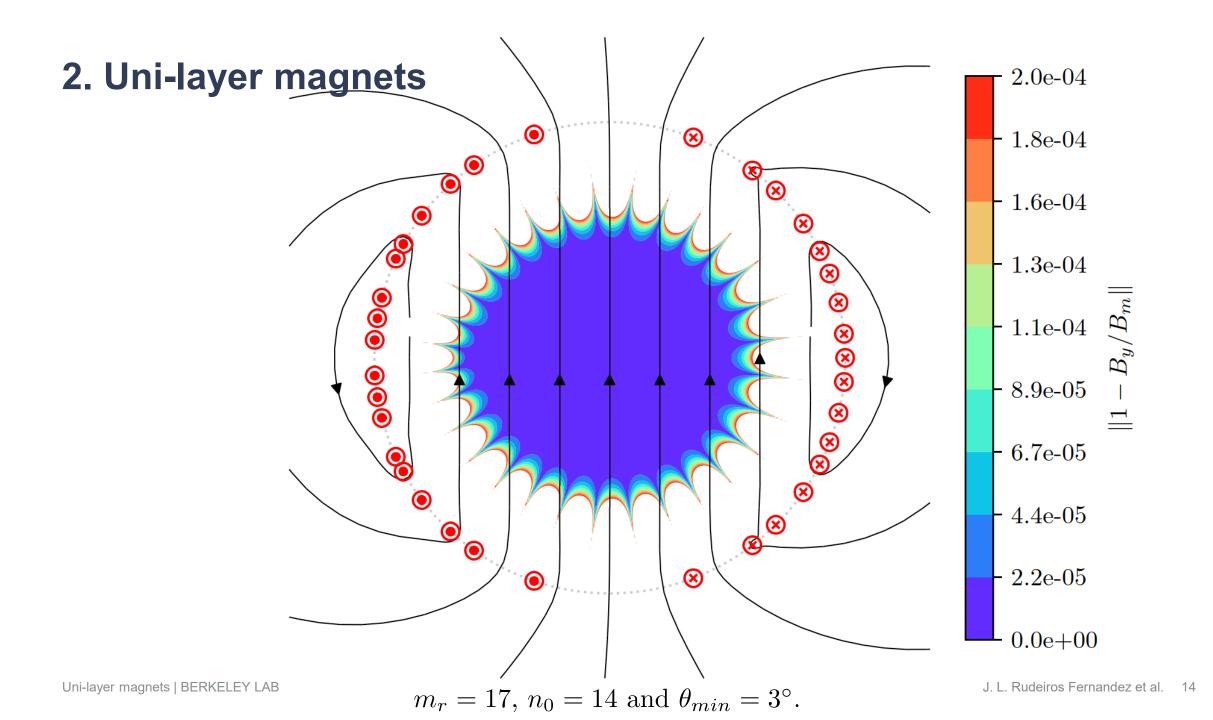
$$\min \{ \|\theta_j - \theta_{j+1}\| \} - \theta_{min} \ge 0, \quad j = 1, \dots, m_r - 1$$

Geometrical constraint: *Left-Right* position

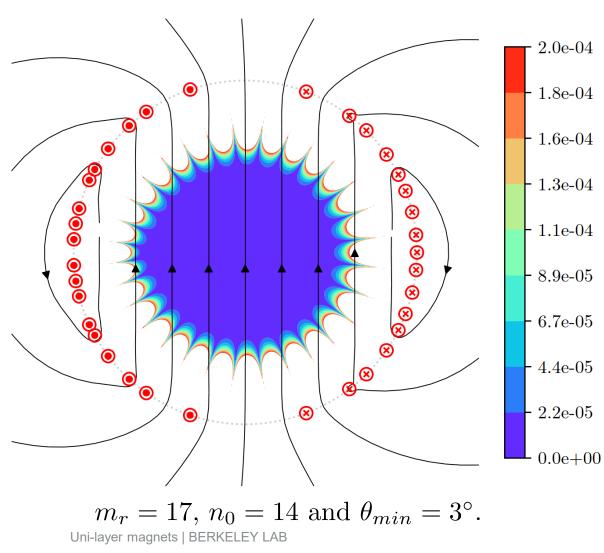
$$\begin{aligned} \theta_{min} &\leq \theta_j \leq \frac{\pi}{2}, \quad j = 1, \dots, m_{opt} \\ \frac{\pi}{2} &\leq \theta_j \leq \pi, \quad j = m_{opt} + 1, \dots, m_r \\ \text{Harmonics to be cancelled} \\ &-1 + 2 \left(-\sum_{j=1}^{m_{opt}} \cos n\theta_j + \sum_{j=m_{opt}+1}^{m_r} \cos n\theta_j \right) = 0, \quad n = 1, \dots, n_0 \\ & \text{J. U}_{\text{Rudeiros Fernandez et al. 12}} \end{aligned}$$



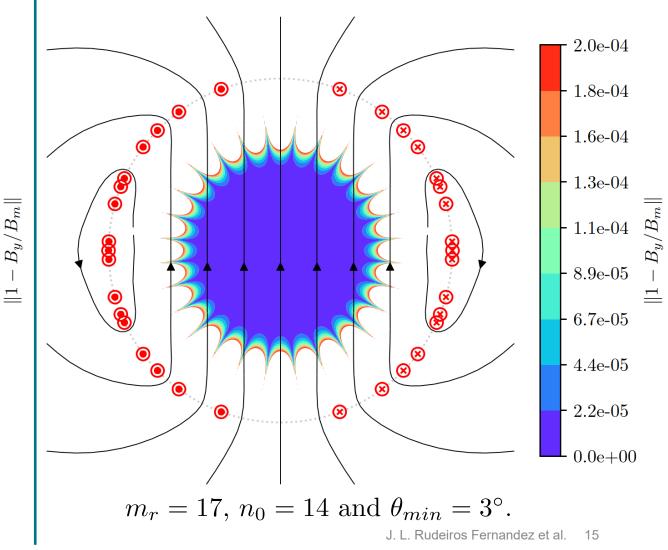
Uni-layer magnets | BERKELEY LAB



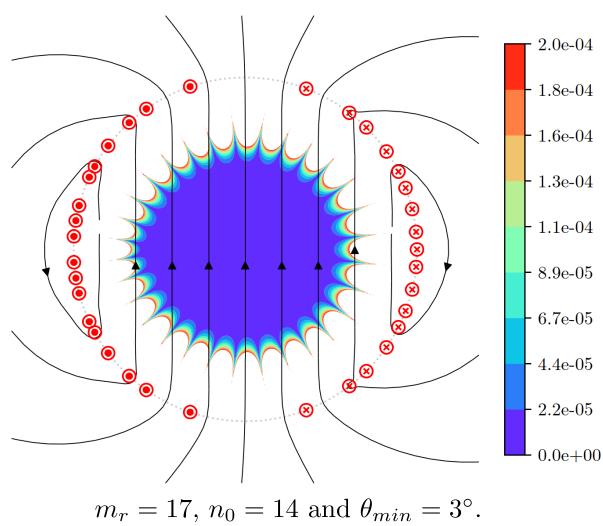
Example asymmetric solution







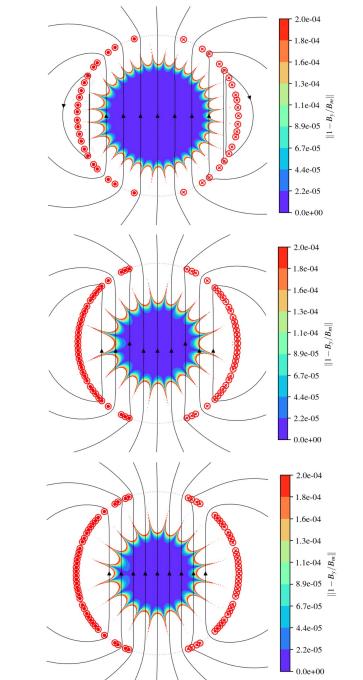
Example **asymmetric** solution



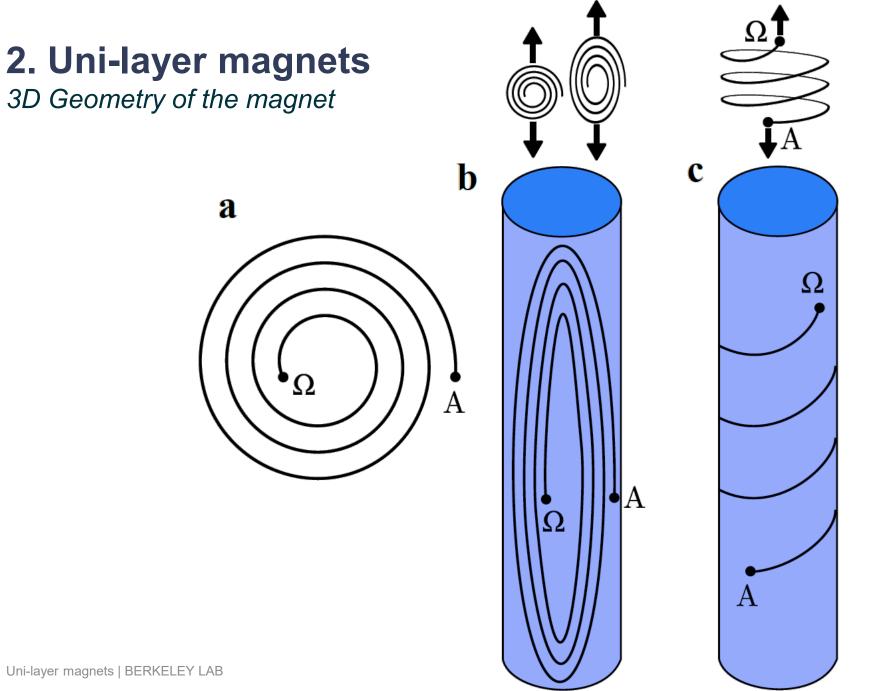
 B_m

g

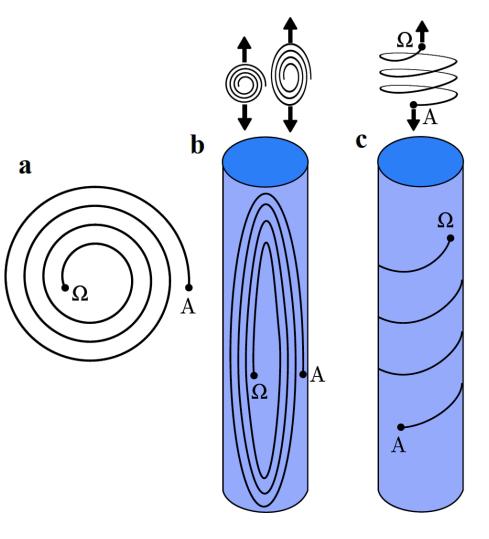
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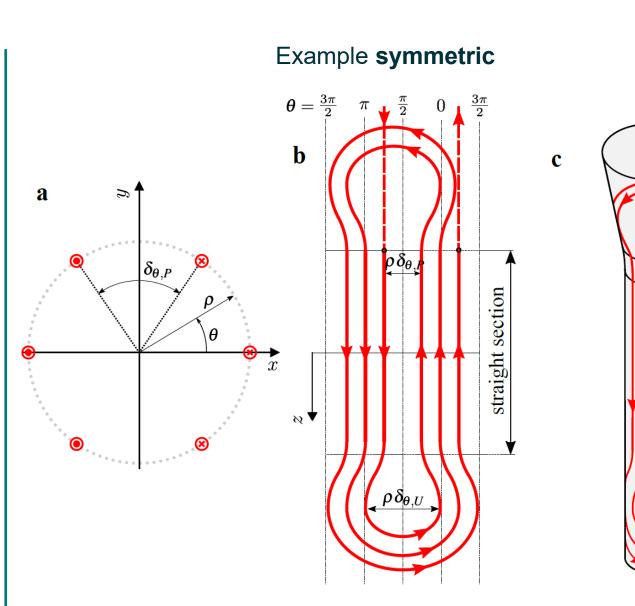


Uni-layer magnets | BERKELEY LAB



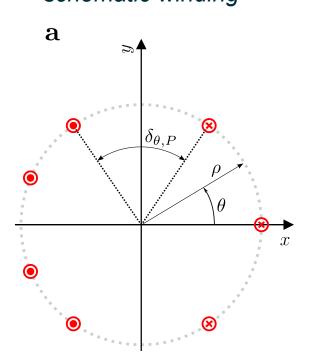
3D Geometry of the magnet





2. Uni-layer magnets – Surface and winding configuration

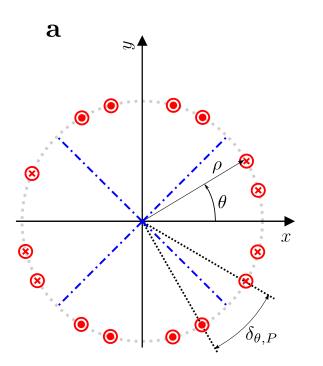
Asymmetric dipole schematic winding



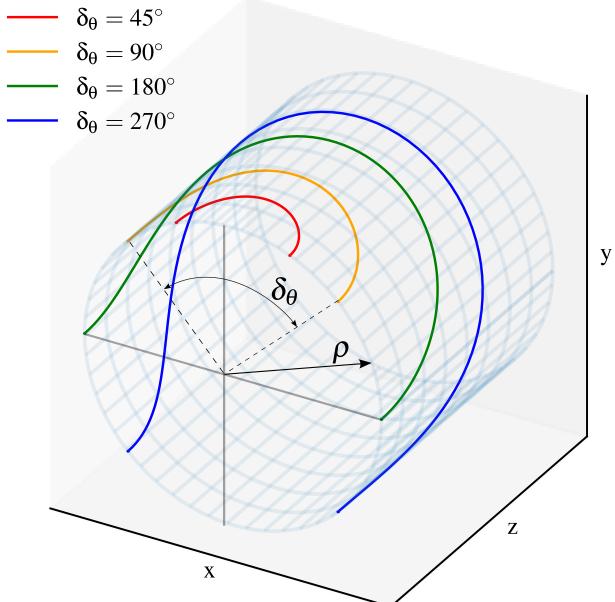
2. Uni-layer magnets – Surface and winding configuration

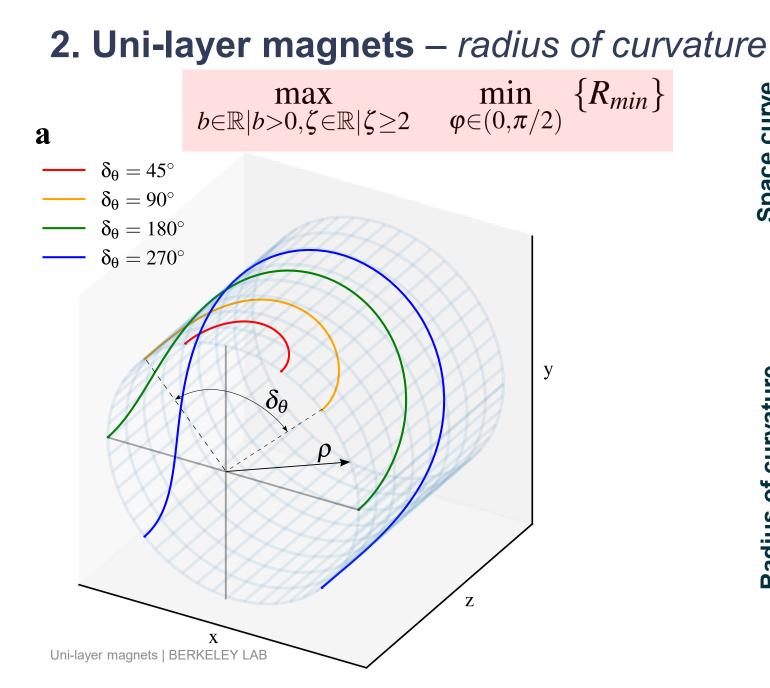
Asymmetric quadrupole

schematic winding



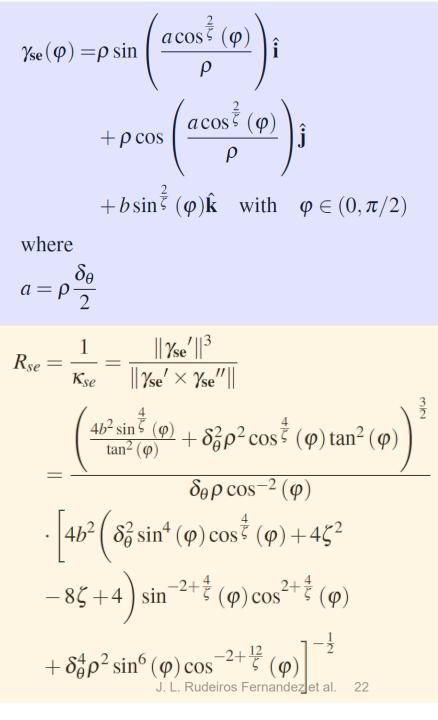
2. Uni-layer magnets – radius of curvature



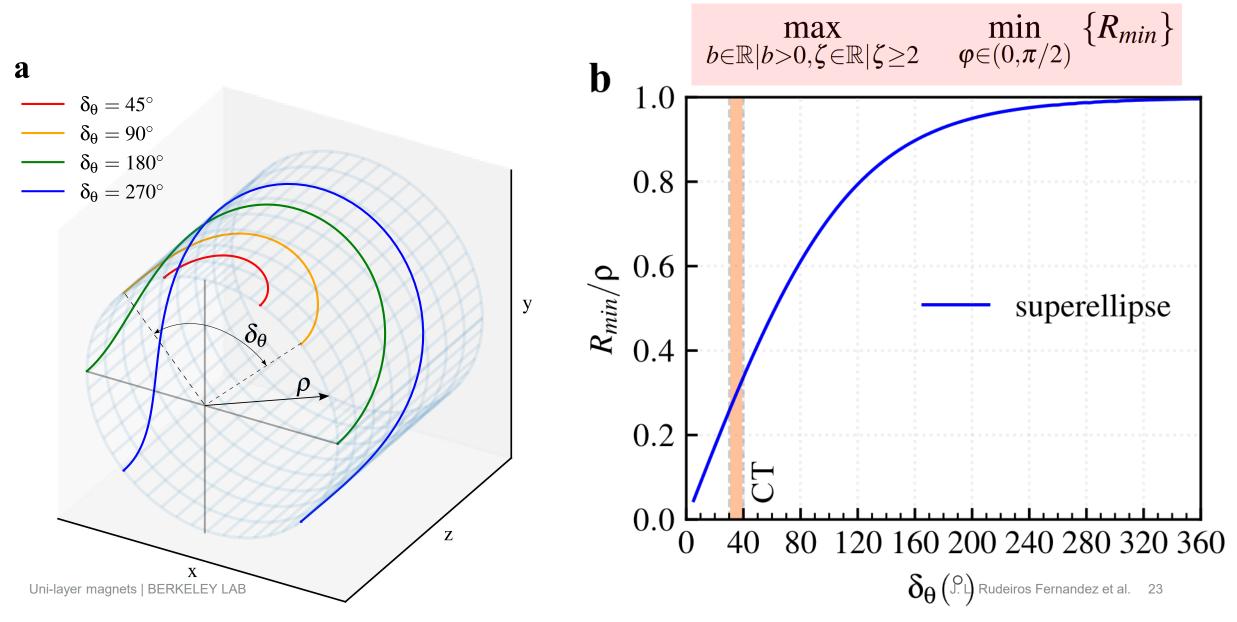




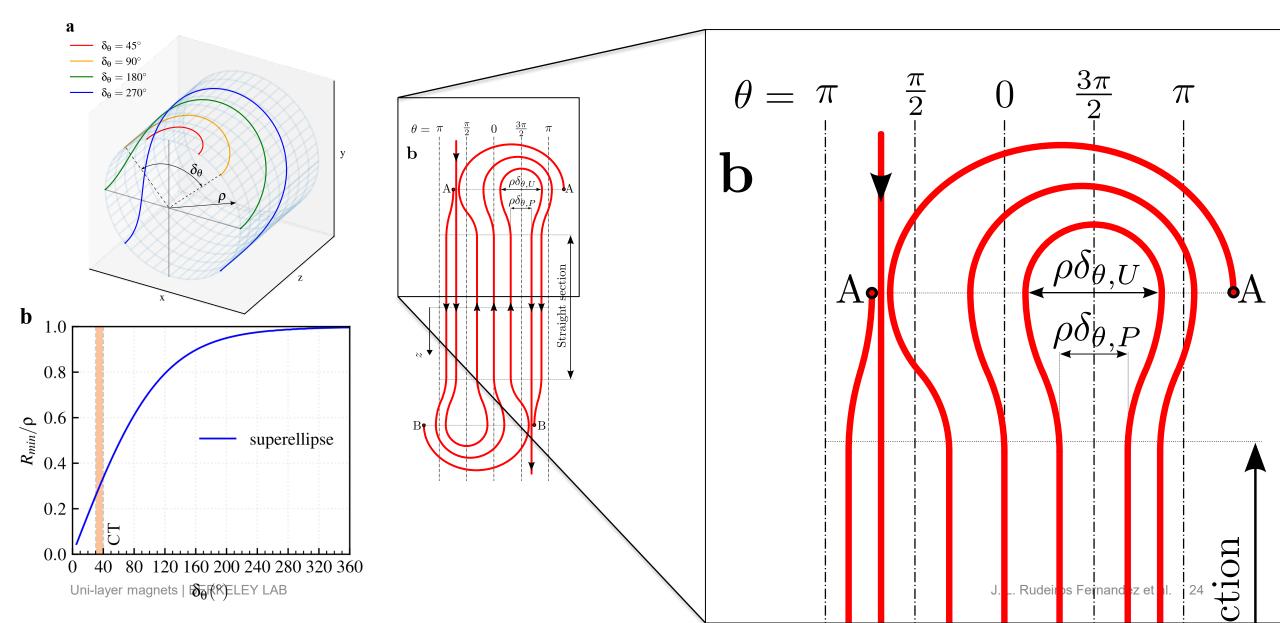
Radius of curvature



2. Uni-layer magnets – radius of curvature



2. Uni-layer magnets – radius of curvature



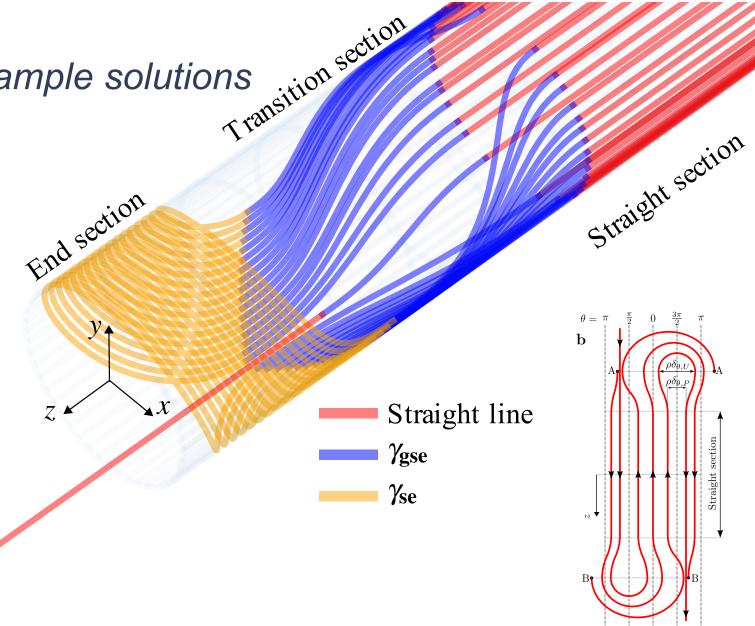
2. Uni-layer magnets – example solutions

$$\gamma_{gse}(\Phi) = \rho \cos\left(\theta_h - \delta_i \sin^{\frac{2}{\chi}}(\Phi)\right) \hat{\mathbf{i}} + \rho \sin\left(\theta_h - \delta_i \sin^{\frac{2}{\chi}}(\Phi)\right) \hat{\mathbf{j}} + \left(z_0 - l_t \cos^{\frac{2}{\zeta}}(\Phi)\right) \hat{\mathbf{k}} \quad \text{with } \Phi \in (0, \pi/2)$$

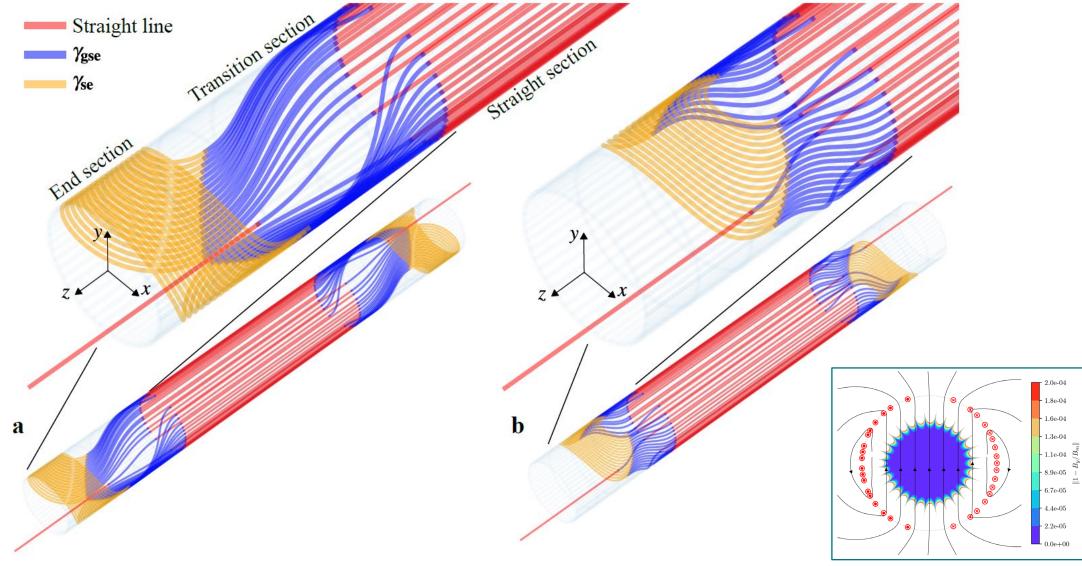
$$\gamma_{se}(\varphi) = \rho \sin\left(\frac{a\cos^{\frac{2}{\zeta}}(\varphi)}{\rho}\right)\hat{\mathbf{i}} + \rho \cos\left(\frac{a\cos^{\frac{2}{\zeta}}(\varphi)}{\rho}\right)\hat{\mathbf{j}} + b\sin^{\frac{2}{\zeta}}(\varphi)\hat{\mathbf{k}} \quad \text{with} \quad \varphi \in (0, \pi/2)$$

where

 $a = \rho \frac{\delta_{\theta}}{2}$



2. Uni-layer magnets – *example solutions*



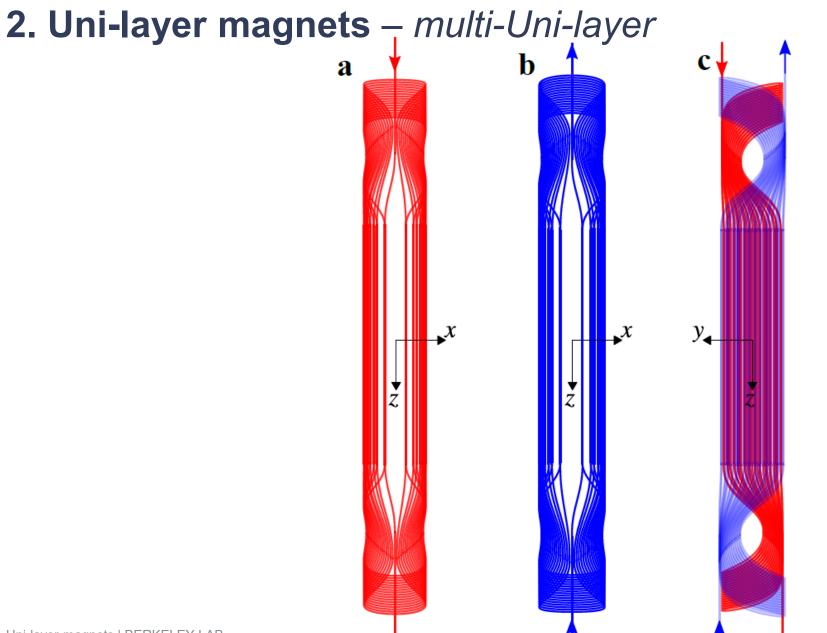


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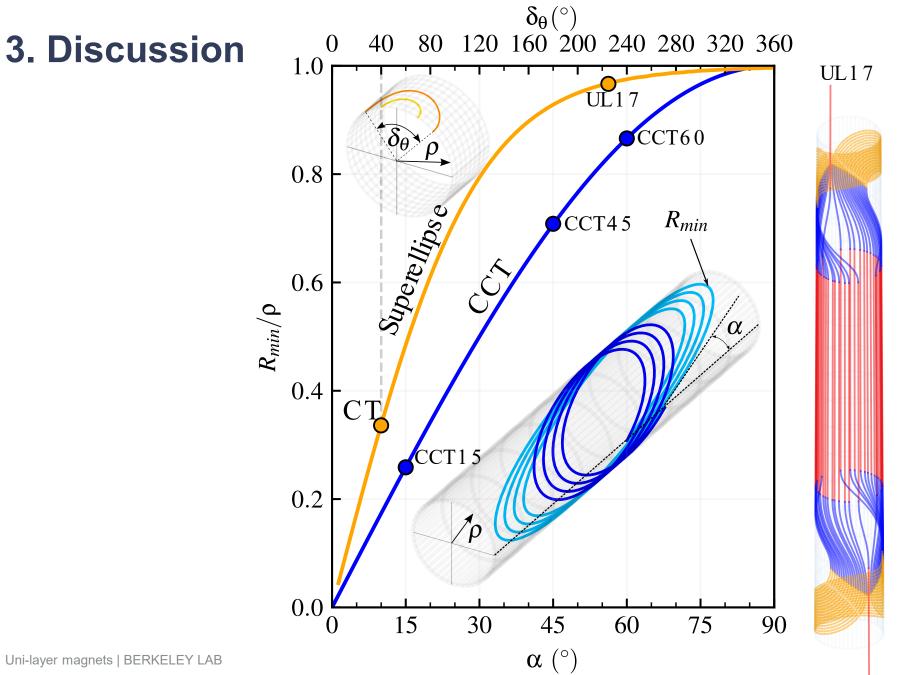
01

02

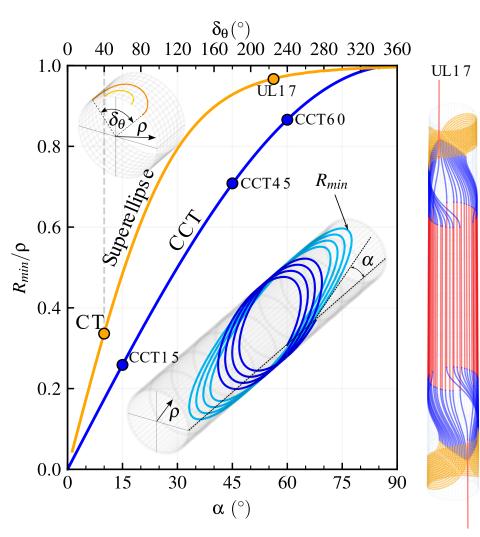
Two special cases of a system of infinitely long current lines Uni-layer magnets

03 Discussion 04

Conclusions



3. Discussion

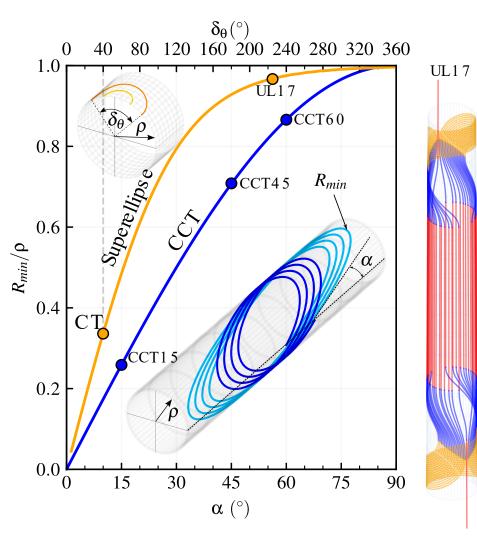


- $R_{min,c}$
- Same conductor.
 - Same length of the conductor.
 - Same current.

Smallest possible aperture diameter (i.e. 2ρ) that can be wound without degrading the conductor assuming a minimum bending radius of $R_{min.c}$ = 25 mm (i.e. similar to CORC_® wire).

Coil design	Smallest possible aperture (mm)	Dipole field transfer function for the same constant length of conductor across all coil designs (T/kA)	Central field ratio (B _m /B _{m,UL17})
UL (UL17)	51	0.218	1
$CT (\delta_{\theta} = 40^{\circ})$	147	0.075	0.34
CCT (α = 15°)	192	0.051	0.23
CCT ($\alpha = 45^{\circ}$)	71	0.081	0.37
CCT ($\alpha = 60^{\circ}$)	58	0.064	0.29

3. Discussion

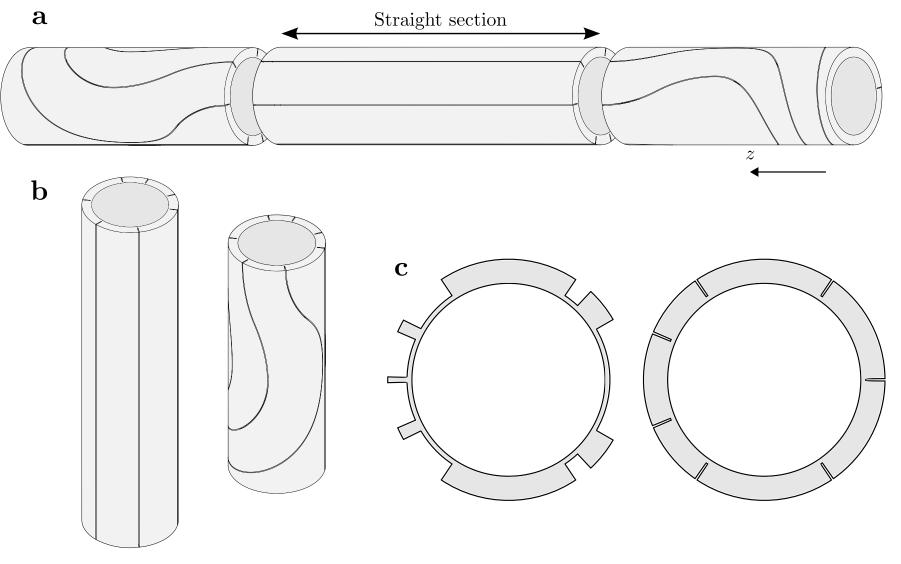


- Same conductor. Same length of the conductor. $\dot{R_{min,c}}$
 - Same current.

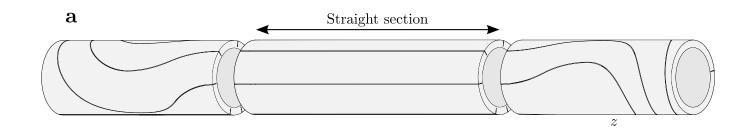
Minimum radius of curvature R_{min} required in the conductor to wind a coil with an aperture of 50 mm.

	Minimu	Dipole field transfer	
	m radius	function for the same	Central field
Coil design	of	constant length of	ratio
	curvatur	conductor across all coil	(B _m /B _{m,UL17})
	e (mm)	designs (T/kA)	,
UL (UL17)	24	0.222	1
$CT (\delta_{\theta} = 40^{\circ})$	8.5	0.222	1
$CCT (\alpha = 15^{\circ})$	6.5	0.197	0.89
CCT ($\alpha = 45^{\circ}$)	17	0.115	0.52
CCT ($\alpha = 60^{\circ}$)	21	0.075	0.34

3. Discussion - Fabrication



3. Discussion



Some thoughts:

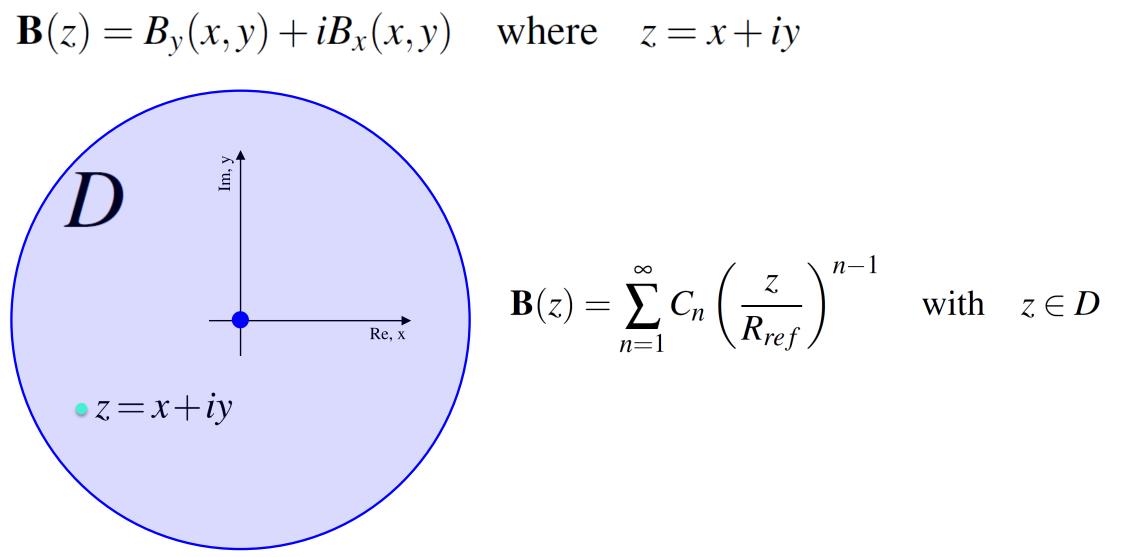
- Significant increase in the minimum bending radius required in the conductor, compatible with REBCO-based CORC wire.
- Minimization of overall required tooling for manufacturing (similar to CCT).
- Relative simplification of manufacturing of the mandrel in relation to other stressmanaged structures.
- Reduced required effort in metrology and magnet assembly (no half to close the aperture).
- Possibility of field quality measurements and cold testing of individual layers before magnet assembly.
- Elimination of internal layer jumps and enhanced potential for layer-to-layer grading.
- Compact combined function magnets.

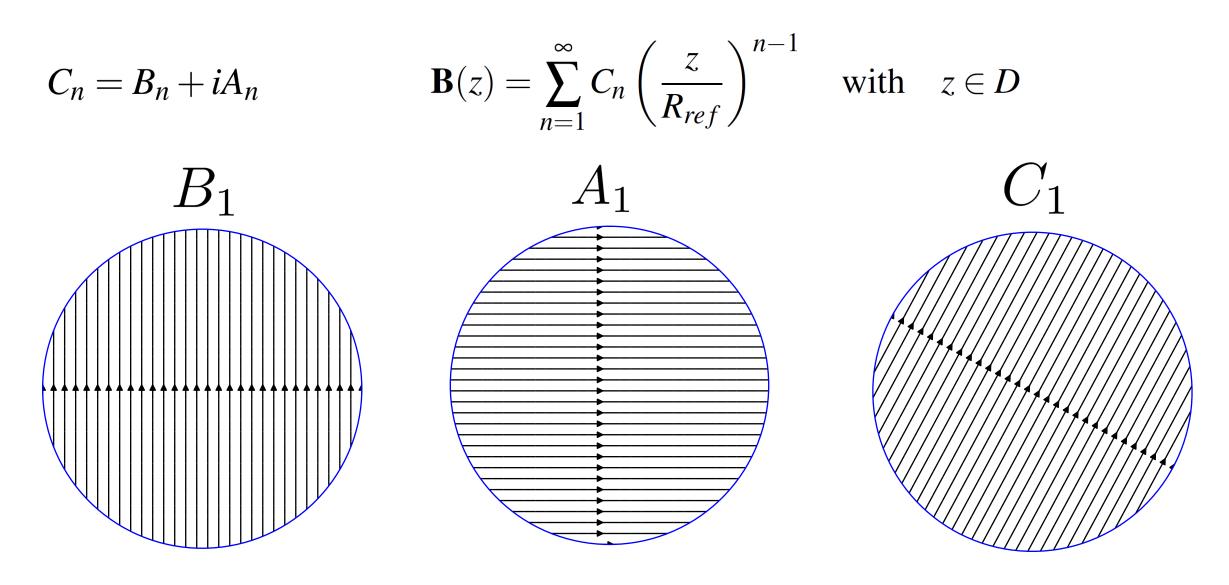
4. Conclusions

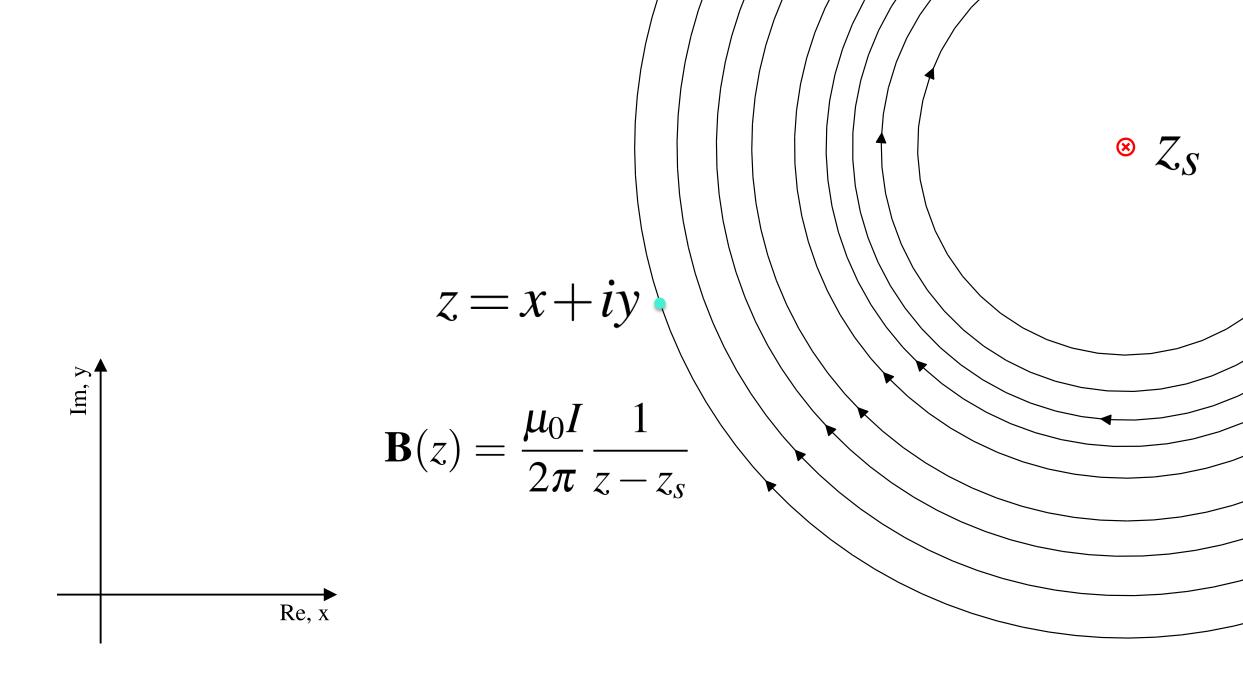
- The novel concept of **Uni-layer** magnets has been presented.
- The asymmetric Uni-layer magnet presents significant advantages over other concepts, merging the advantages of the costheta and CCT concepts, providing a high-quality field in terms of harmonics, within a single-layer (no internal layer jumps), using a single continuous conductor, and with a higher minimum radius of curvature required in the conductor during winding.
- This new concept is especially advantageous in the domain of stress-managed magnets for HTS conductors.

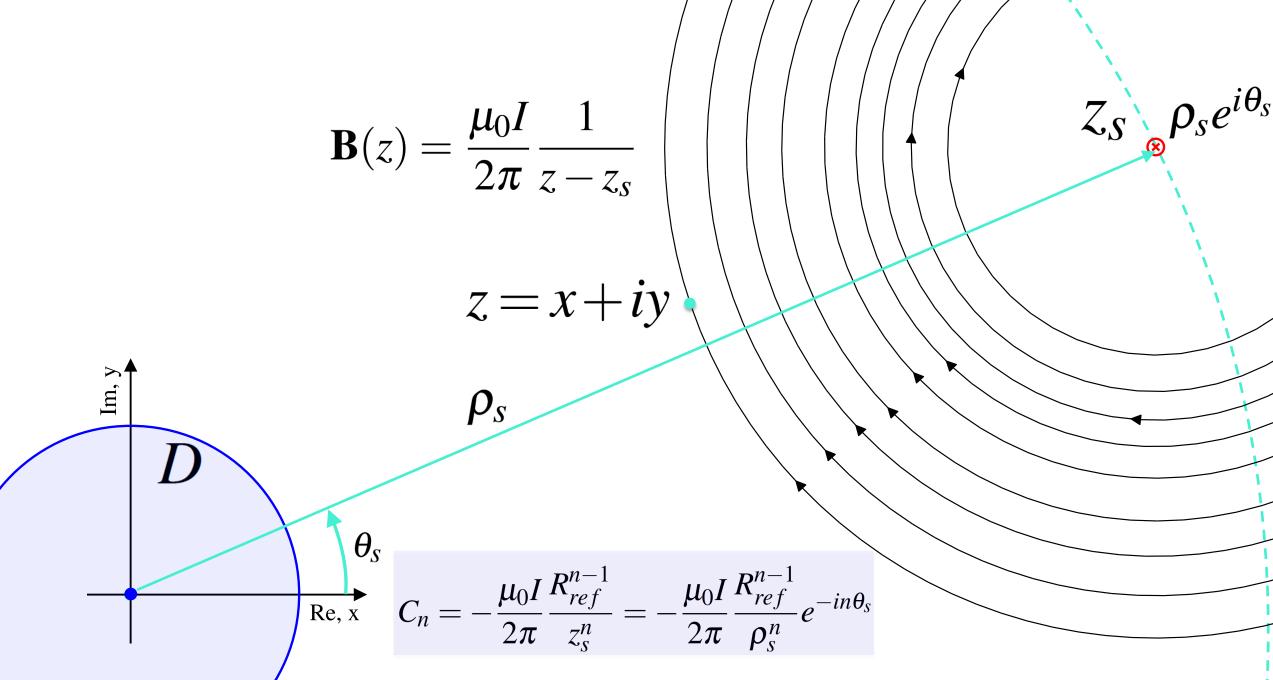
Thank you!!!

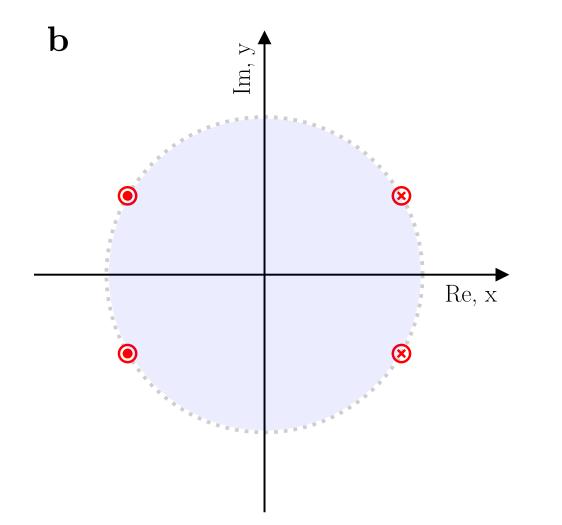
Rudeiros Fernández et al 2023 Supercond. Sci. Technol. https://doi.org/10.1088/1361-6668/acc281











$$\mathbf{B}(z) = \sum_{n=1}^{\infty} C_{n,total} \left(\frac{z}{R_{ref}}\right)^{n-1}$$

$$C_{n,total} = -\frac{\mu_0 R_{ref}^{n-1}}{2\pi} \sum_{j=1}^m \frac{I_j}{\rho_j^n} e^{-in\theta_j}$$

$$C_{n,total} = -\frac{\mu_0 I}{2\pi} \frac{R_{ref}^{n-1}}{\rho^n} \sum_{j=1}^m s_j e^{-in\theta_j}$$

$$C_{n,q} = \frac{\mu_0 I}{\pi} \frac{R_{ref}^{n-1}}{\rho^n} \cos\left(n\theta_q\right) \left(1 - e^{in\pi}\right)$$