RF Systems I

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CERN

The CERN Accelerator School

Introduction to Accelerator Physics

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Outline

- Introduction
- Choice of parameters
  - Frequency and voltage
- RF cavity parameters
  - Shunt impedance, beam loading, power coupling
- Power amplifiers
  - Tube or solid state
  - Local feedbacks
- Longitudinal beam control system
  - Building blocks: RF source and receiver
  - Phase, radial and synchronization loops
- Summary

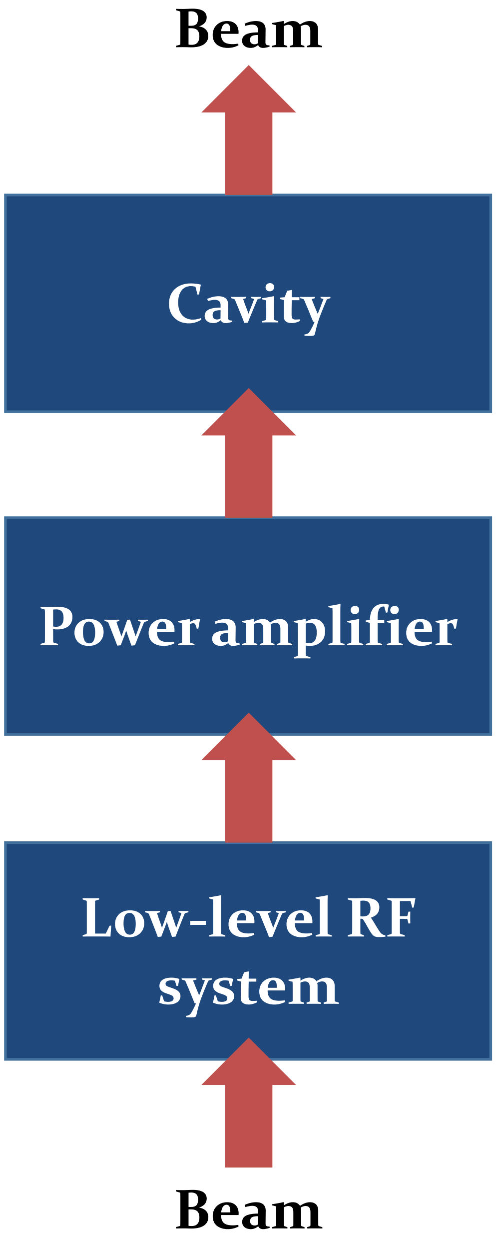
Introduction

Introduction

- The radiofrequency (RF) system transforms a string of magnets into an accelerator
- Cavity most is the most visible part of an RF system
  - → On top of the RF system food chain
  - → Interacts directly with beam

→ What is below?

→ How are RF signals generated which make the beam feel comfortable?

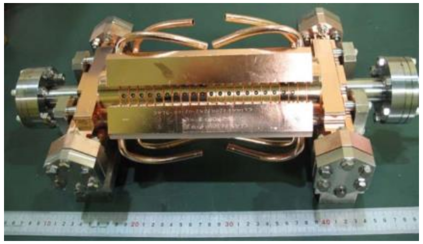


Frequency and wavelength ranges

PS longitudinal damper

PS main RF system

SPS 200 MHz



CLIC 12 GHz

100 kHz
3 km

1 MHz
300 m

10 MHz
30 m

100 MHz
3 m

1 GHz
30 cm

10 GHz
3 cm

100 GHz
3 mm

Long wave

Medium/ short wave

VHF

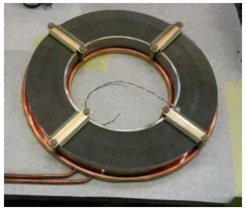
Microwave links

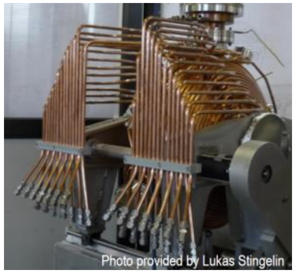
Amplitude ranges

Signals from beam pick-ups

LLRF systems

Low/Medium energy hadron RF





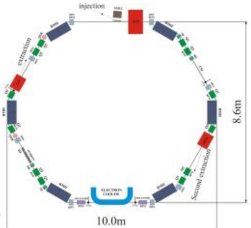
SLS

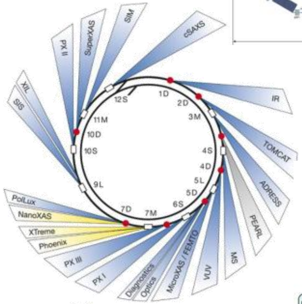
LHC: 16 MV

LEP: 3.6 GV total

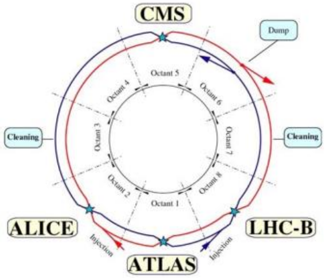
- 1 µV
- 1 mV
- 1 V
- 1 kV
- 1 MV
- 1 GV

Cooled hadron beams (ELENA)

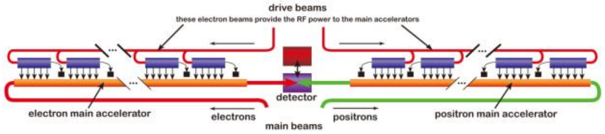




Electron light sources



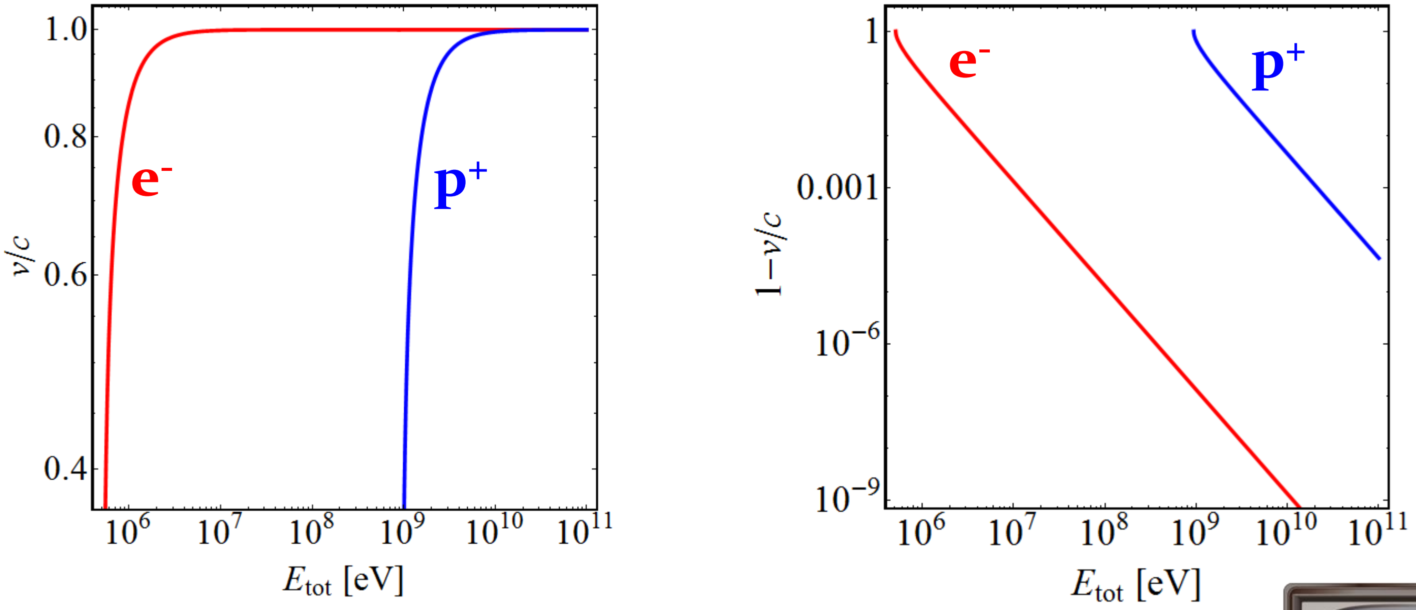
LHC

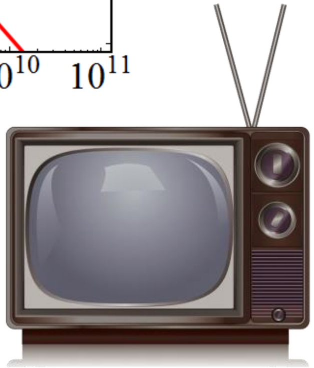


ILC and CLIC: several TV

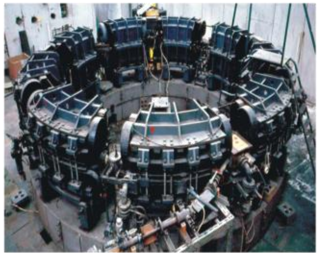
Particle velocity

- **Particle velocity depends on its type:** $\beta = v/c = \sqrt{1-(E_0/E)^2}$





- **Old television set (30 kV):** **Electrons at 30% of c_0**
  Protons just at 0.7%
- **Small synchrotron (500 MeV):** **Electrons at 99.99995%**
  Protons at 75.8%



→ Most electron accelerators at ‘fixed’ frequency

Parameter choices

RF system for high-energy accelerators

Accelerator type

- **Linear (single pass)**
  - **Electron**
    - Maximum RF voltage, constant velocity
  - **Hadrons**
    - Maximum RF voltage, variable velocity
- **Circular (multi-pass)**
  - **Electron**
    - Compensate synchrotron radiation losses
  - **Hadrons**
    - Sweep frequency with beam

$$f_{\mathrm{RF}} = n \cdot f_{\mathrm{rev}}{}^{*}$$

*Exceptions (rare) exist

Choice of frequency (range)

Why choose a low RF frequency?

Advantages	Disadvantages
• Large beam aperture • Long RF buckets, large acceptance • Wide-band or wide range tunable cavities possible • Power amplification and transmission straightforward	• Bulky cavities, size scales $\propto 1/f$, volume $\propto 1/f^3$ • Lossy material to downsize cavities • Moderate or low acceleration gradient • Short particle bunches difficult to generate
RF frequencies below ~200 MHz for	→ Some hadron linear accelerators → Cyclotrons → Low- and medium energy hadron synchrotrons

Why choose a high RF frequency?

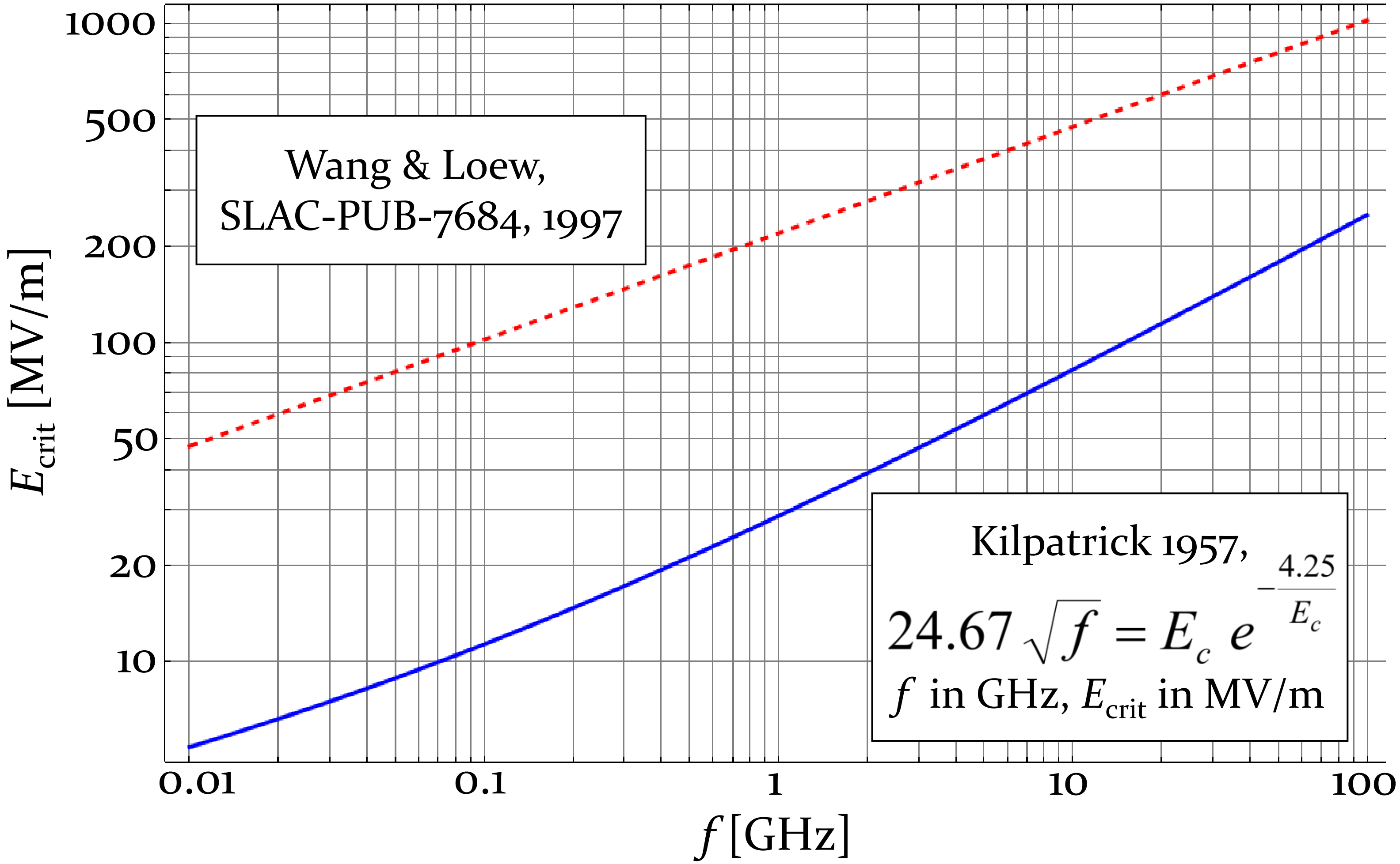
Advantages	Disadvantages
Cavity size scales $\propto 1/f$, volume $\propto 1/f^3$	Maximum beam available aperture scales $\propto 1/f$
Break down voltage increases	No technology for wide-band or tunable cavities
High gradient per length	Power amplifiers more difficult
Particle bunches are short	Power transmission losses

RF frequencies above ~200 MHz used for

→ Linear accelerators
→ Electron storage rings
→ High energy hadron storage rings

Limits to maximum gradient

- **Surface electric field in vacuum**



$$24.67\sqrt{f} = E_c\, e^{-\frac{4.25}{E_c}}$$

f in GHz, E_{crit} in MV/m

→ High frequencies preferred for large gradient

E. Jensen

Some standard frequencies

If exact RF frequency not critical, choose standard value

Accelerator	Frequency
Hadron synchrotrons (PSB, PS, JPARC RCS, MR)	<10 MHz
Hadron accelerators and storage rings (RHIC, SPS)	~200 MHz
Electron storage rings (LEP, ESRF, Soleil)	352 MHz
Electron storage rings (DORIS, BESSY, SLS,…)	499.6…499.8 MHz
Superconducting electron linacs and FELs (X-FEL, ILC)	1300 MHz
Normal conducting electron linacs (SLAC)	2856 MHz
High-gradient electron linac (CLIC)	11.99 GHz

→ **Off-the-shelf RF components easily available in frequency ranges used by industry**

→ **Exchange of developments and equipment amongst research laboratories**

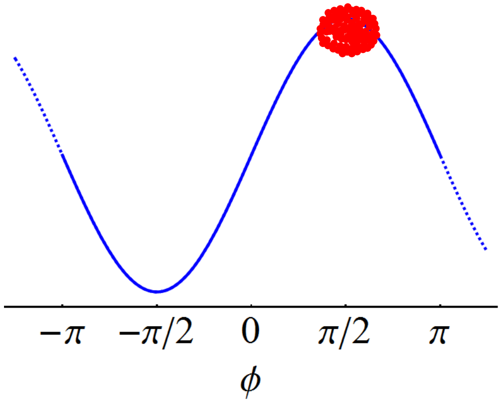
RF voltage

Minimum voltage requirement

- **RF system expected to provide given energy gain**

$$qV = \Delta E$$

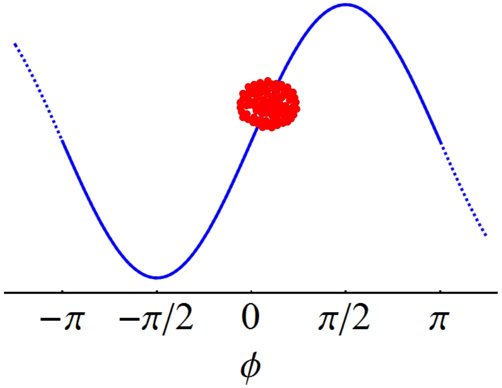
→ On-crest acceleration
→ Used in some linear accelerators
→ Insufficient in a circular accelerator



- **More voltage provided to avoid on-crest acceleration**

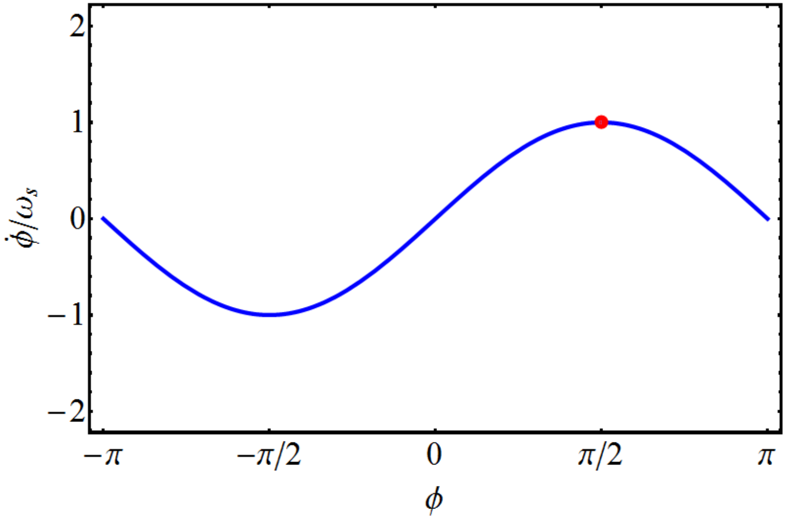
$$qV > \Delta E \rightarrow qV \sin(\phi_S) = \Delta E$$

→ Off-crest acceleration
→ Needed for circular accelerator
→ Higher voltage for given energy gain

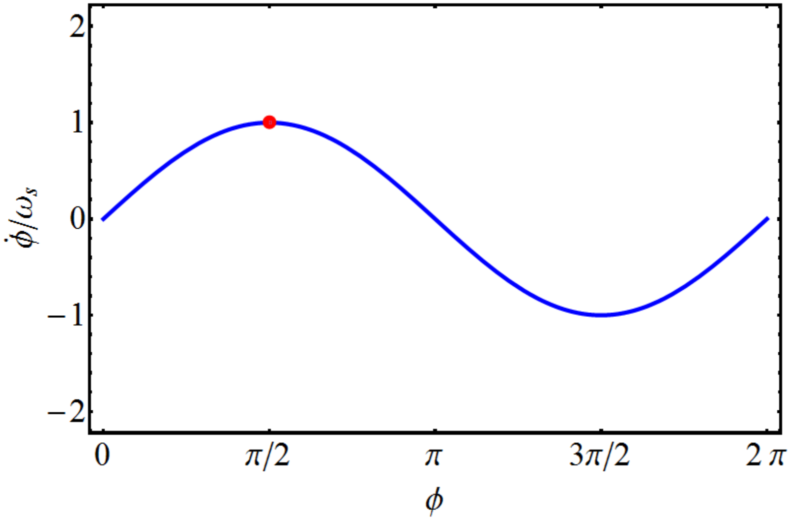


Bucket area dependence on stable phase

- In a circular accelerator the area in energy-time phase space (bucket area) depends on the stable phase



Below transition, ϕ_S = 0…90°



Above transition, ϕ_S = 90°…180°

- Typical synchronous phase with respect to 0° or 180°
  - Hadron accelerators: < 40°
  - Electron storage rings: ~ 20°

Minimum voltage requirement (circular)

The RF system must compensate

1. **Energy gain per turn due to changing magnetic field**

$$F_Z = F_L \quad \rightarrow \quad \frac{p}{q} = \rho B \quad \rightarrow \quad \dot{p} = q\rho\dot{B}$$

$$\dot{p} = \frac{\Delta p}{\Delta t} = \frac{m_0 c^2 \beta}{2\pi R}(\beta\Delta\gamma + \gamma\Delta\beta) = \frac{\Delta E_{\mathrm{turn}}}{2\pi R}$$

$$\boxed{\Delta E_{\mathrm{turn}} = 2\pi q\rho R\dot{B}}$$

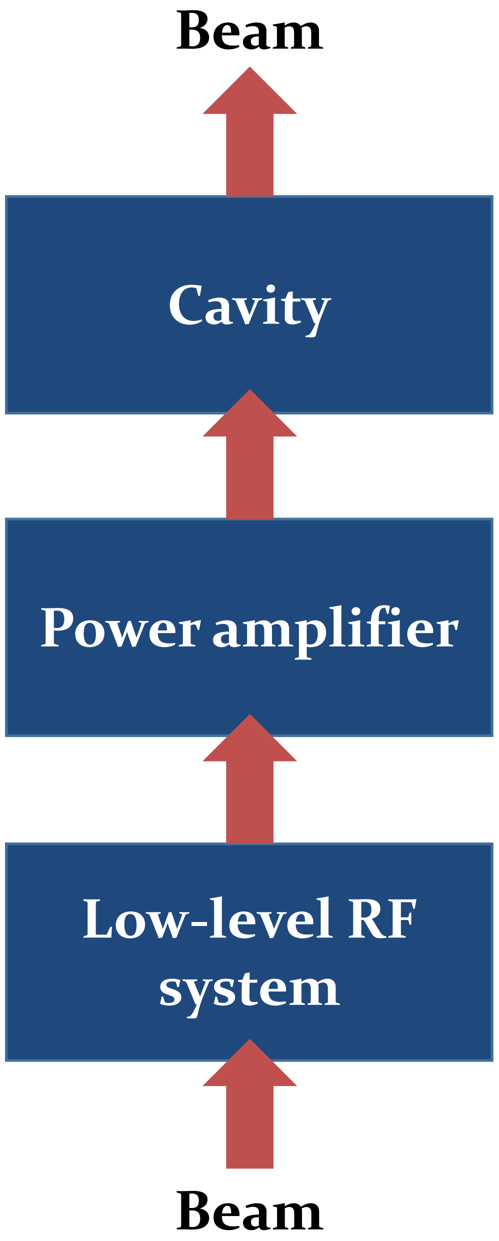
2. **Energy loss, e.g., due to synchrotron radiation (electrons)**

$$\boxed{\Delta E_{\mathrm{turn}} = \frac{e^2}{3\epsilon_0 (m_0 c^2)^4}\frac{E^4}{\rho}}$$

$$\Delta E_{\mathrm{turn}}[\mathrm{keV}] = 88.5 \cdot \frac{E^4[\mathrm{GeV}]^4}{\rho[\mathrm{m}]} \qquad \Delta P_{\mathrm{loss}}[\mathrm{kW}] = 88.5 \cdot \frac{E^4[\mathrm{GeV}]^4}{\rho[\mathrm{m}]} \cdot I_{\mathrm{B}}[\mathrm{A}]$$

→ $(m_p/m_e)^4 = 1836^4 \sim 1.1 \cdot 10^{13}$ times less for protons

RF system overview

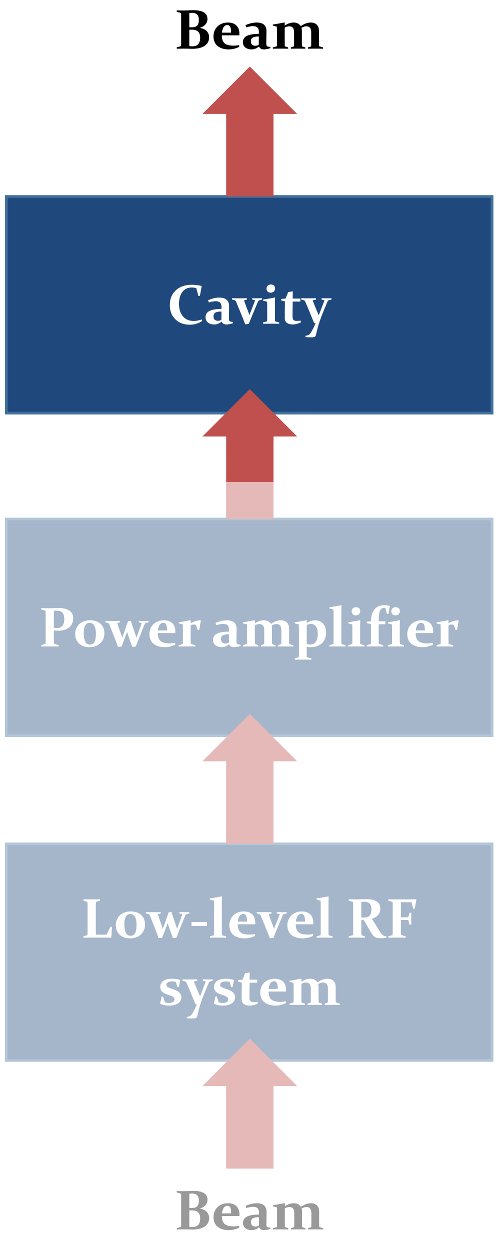


→ Convert RF power into longitudinal electric field

→ Amplify low-power signal from beam control to kW, MW or GW

→ Provide RF signals with correct frequency, amplitude and phase

RF system overview



Beam

Cavity

Power amplifier

Low-level RF system

Beam

→ **Convert RF power into longitudinal electric field**

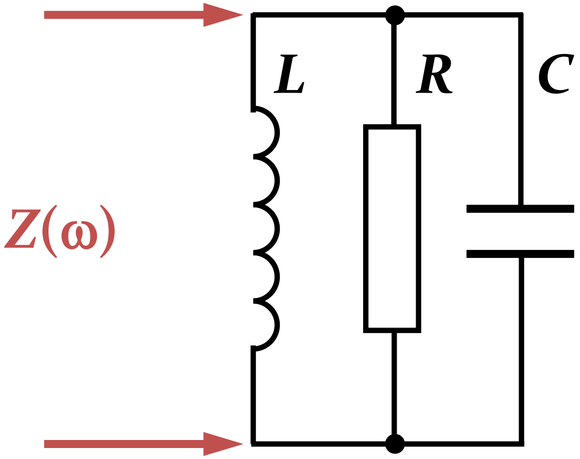
→ **Amplify low-power signal from beam control to kW, MW or GW**

→ **Provide RF signals with correct frequency, amplitude and phase**

RF cavity

Cavity parameters

- **The resonance of a cavity can be understood as simple parallel resonant circuit described by *R*, *L*, *C***



$$\frac{1}{Z(\omega)} = \frac{1}{R} + \frac{1}{i\omega L} + i\omega C$$

with $$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$E = \frac{1}{2}CV^2 = \frac{1}{2}LI^2$$

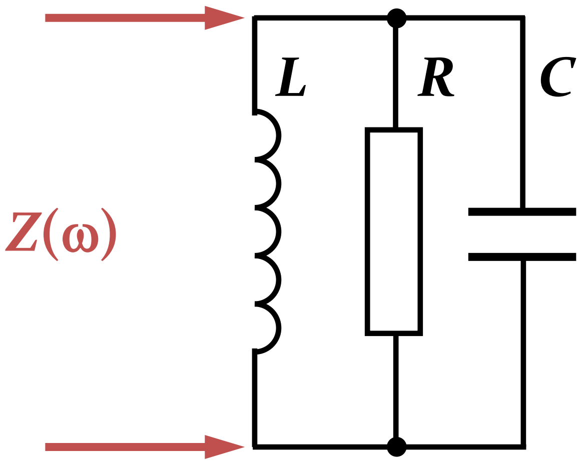
$$Q = \omega_0 \frac{\text{Stored energy}}{\text{Average power loss}} = \frac{\omega_0 E}{P}$$

$$P = \frac{1}{2}\frac{U^2}{R} = \frac{1}{2}I^2 R$$

$$Q = \omega_0 RC = \frac{R}{\omega_0 L}$$

Cavity parameters

- **The resonance of a cavity can be understood as simple parallel resonant circuit described by *R, L, C***



$$\frac{1}{Z(\omega)} = \frac{1}{R} + \frac{1}{i\omega L} + i\omega C$$

with $$\omega_0 = \frac{1}{\sqrt{LC}}$$

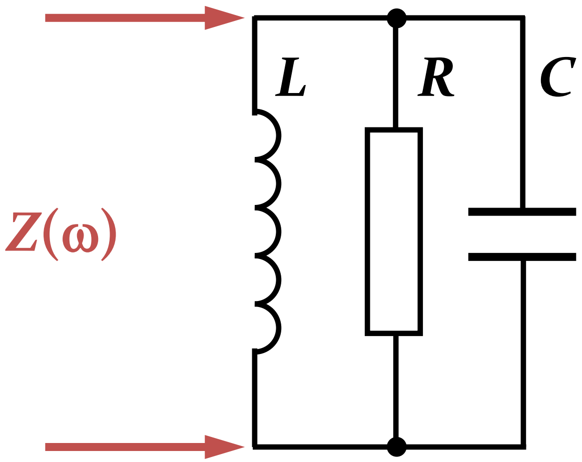
$$Q = \omega_0 RC = \frac{R}{\omega_0 L}$$

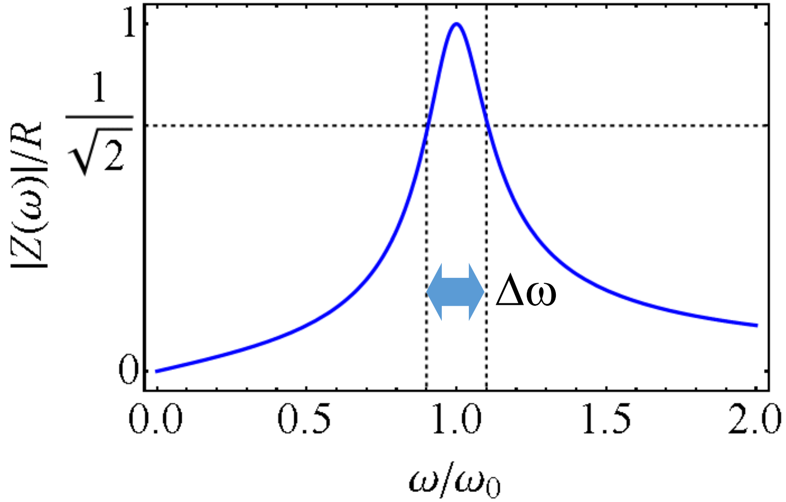
$$Z(\omega) = \frac{R}{1 + iQ\left(\frac{\omega^2 - \omega_0^2}{\omega\omega_0}\right)} \simeq \frac{R}{1 + 2iQ\frac{\Delta\omega}{\omega_0}}$$

→ **Resonant circuit can also be described by *R, R/Q,* ω_0 or any other set of three parameters**

Cavity parameters

- **The resonance of a cavity can be understood as simple parallel resonant circuit described by *R, L, C***





$$Q = \omega_0 RC = \frac{R}{\omega_0 L} \qquad Z(\omega) = \frac{R}{1 + iQ\left(\frac{\omega^2 - \omega_0^2}{\omega\omega_0}\right)} \simeq \frac{R}{1 + 2iQ\frac{\Delta\omega}{\omega_0}}$$

→ **Resonant circuit can also be described by *R, R/Q, ω_o* or any other set of three parameters**

Cavity parameters

- Most common choice by cavity designers ω_o, R, R/Q – why?

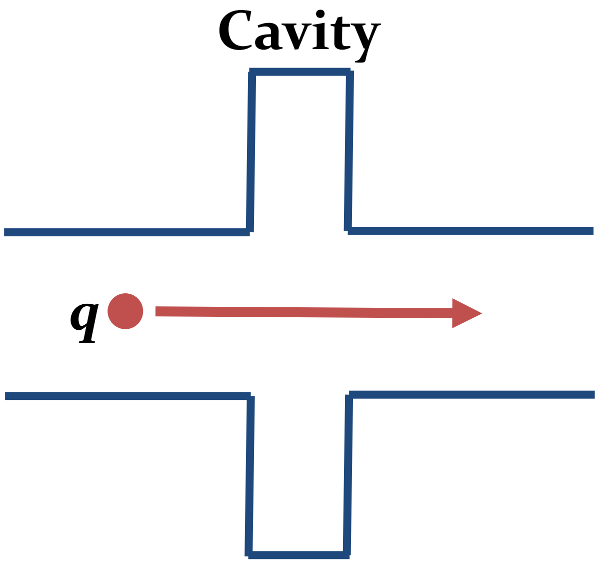
- Resonance frequency, ω_o
  - → Exactly defined for given application, e.g. hf_{rev}

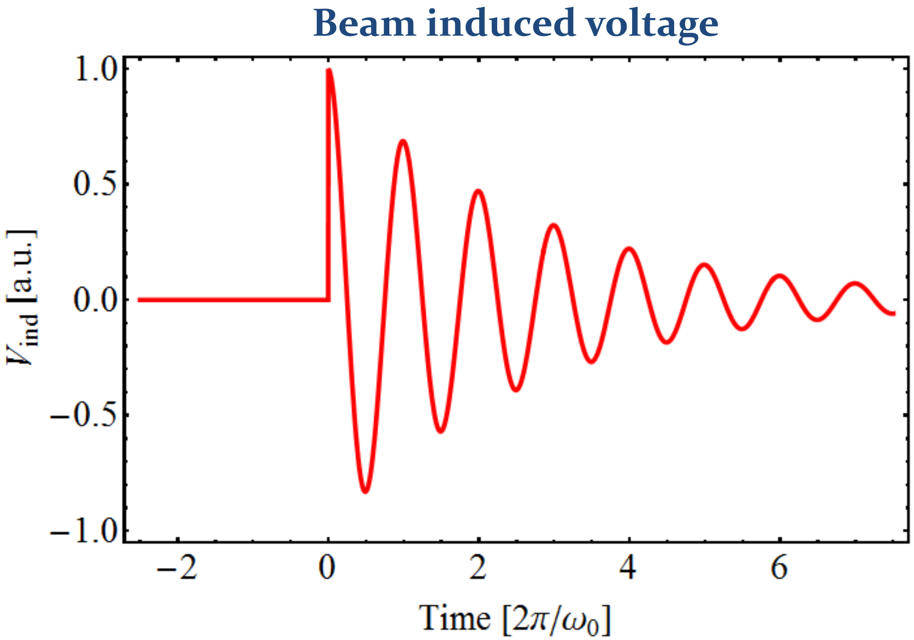
- Shunt impedance, R
  - → Power required to produce a given voltage without beam

- "R-upon-Q", R/Q
  - → Defined only by the cavity geometry
  - → Criterion to optimize a geometry
  - → Detuning with beam proportional to R/Q

Why R/Q?

→ **Charged particle experiences cavity gap as capacitor**





$$q = V_{\text{ind}} C$$

$$Q = \omega_0 R C \quad \rightarrow \quad \frac{1}{C} = \left(\frac{R}{Q}\right)\omega_0$$

$$\boxed{V_{\text{ind}}} = \frac{q}{C} \boxed{\propto \frac{R}{Q}}$$

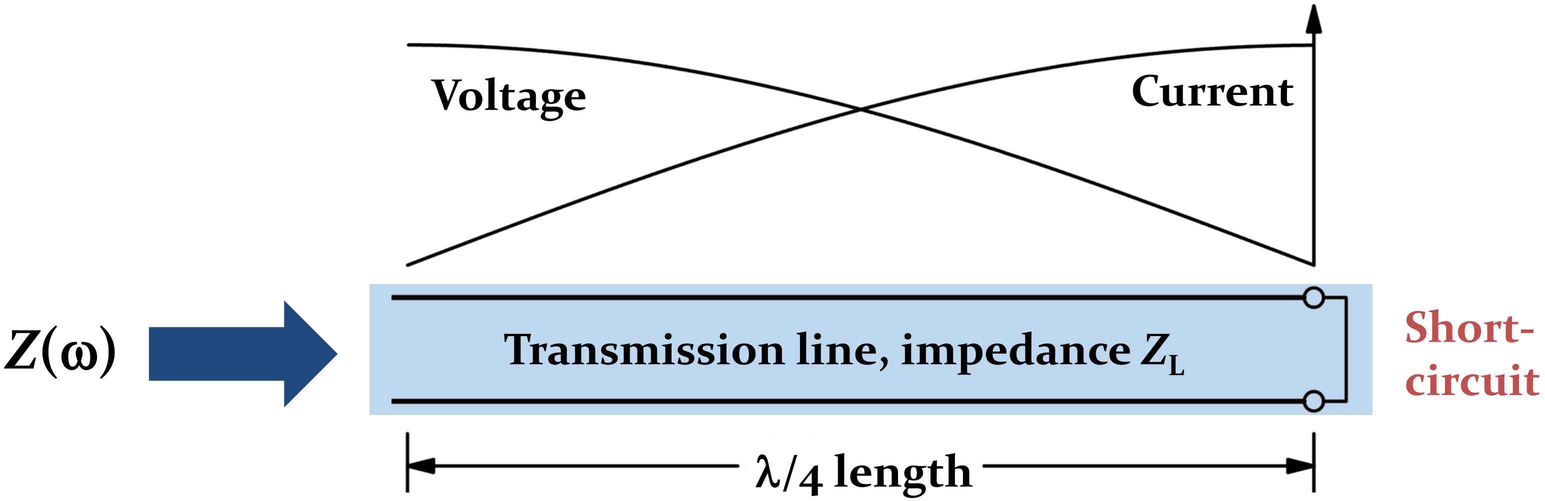
→ **Cavity geometry with small *R*/*Q* to reduce beam loading**

RF cavities in low frequency range

- RF wavelength large below ~10 MHz: >30 m

→ Would need huge cavities → too large for accelerators

→ Line resonators: λ/4 resonator



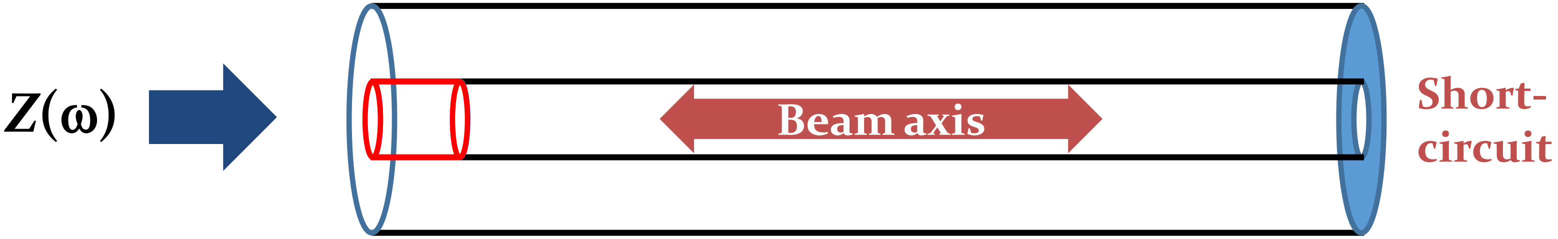
→ Short circuit on one side → Voltage is zero

→ Open end on other → No current but voltage

Why is this resonator so common in particle accelerators?

RF cavities in low frequency range

- Coaxial structure with inner conductor as beam pipe



Short-circuit

Accelerating gap, isolated

→ Still rather long geometry, 7.5 m at 10 MHz

→ Add capacitive or inductive shortening

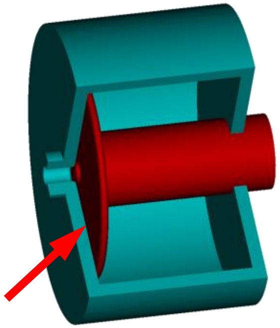
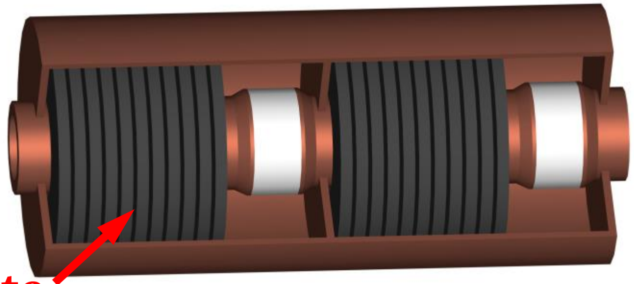


Plate capacitor

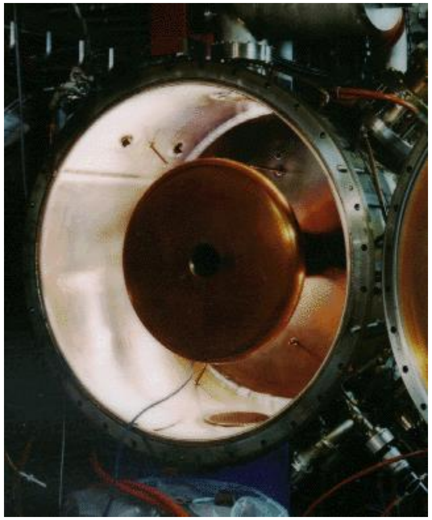


Ferrite inductivity

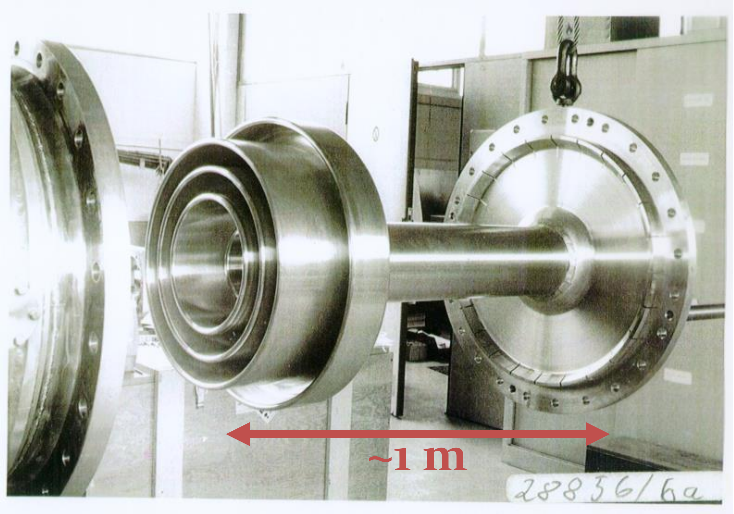
Capacitive loading

→ **Add capacitor at gap of cavity to shorten the resonator**

NSLS, 52.88 MHz



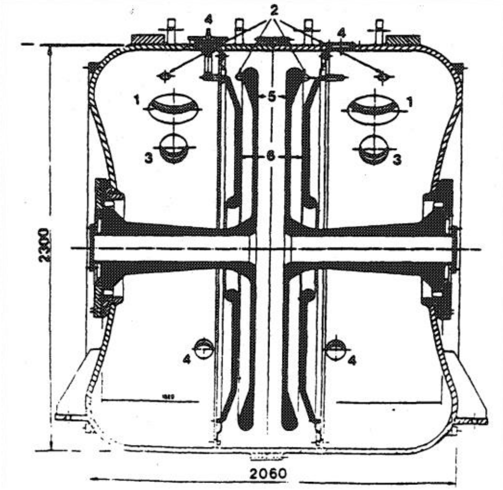
DESY PIA, 10.4 MHz, inner cond.



Outer cond.

M. Nagl

ACOL, 9.53 MHz

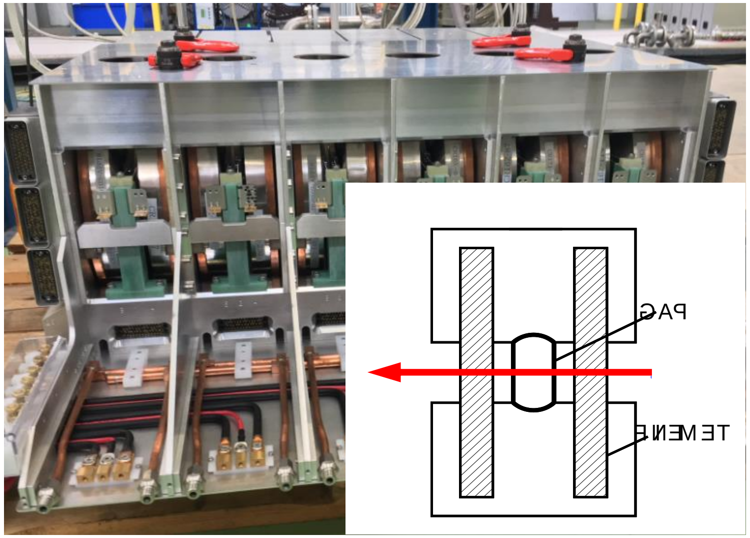


- → **Significantly reduces cavity size**
- → **Fixed frequency only**
- → **Small losses due to capacitor**
- → **Cavity in vacuum**

Inductive loading

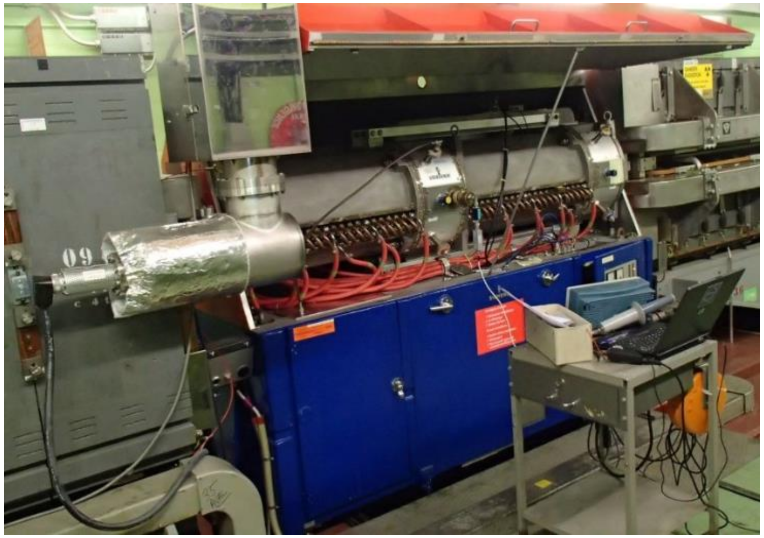
→ **Inductive loading with magnetic material shortens resonator from tens of meters to a device, lossy though**

CERN PSB Finemet cav., 0.6-18 MHz



M. Paoluzzi

CERN PS, double gap, 2.8-10 MHz

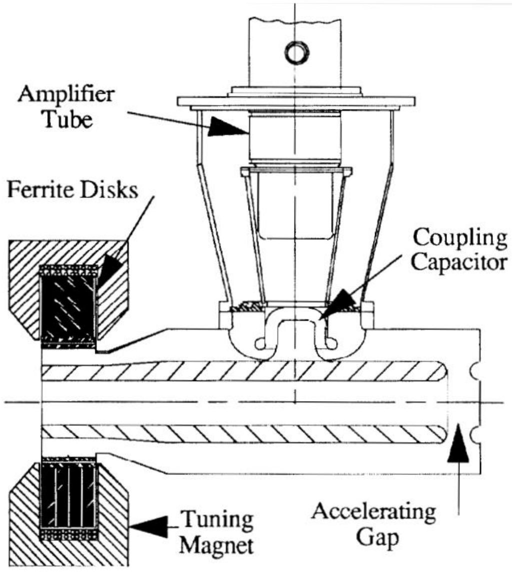


- **Additional advantage: permeability of ferrite can be controlled by DC bias current → variable inductivity**
  - → Cavity with programmable resonance frequency
  - → Essential for hadron acceleration in low-energy accelerators

Tunable cavities at higher frequencies

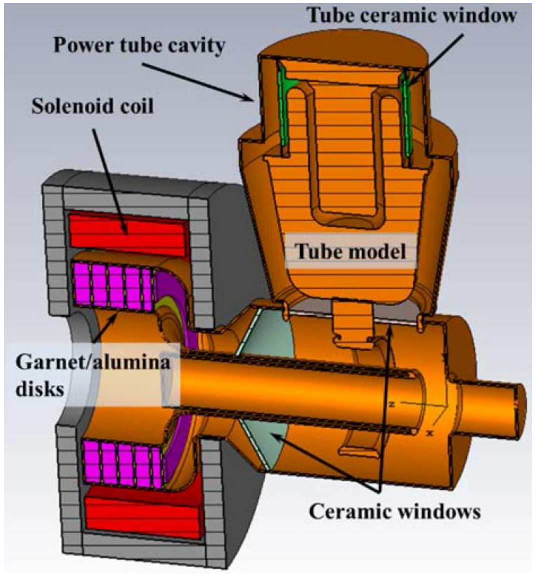
→ **Remove inductive or capacitive loading**

SSC Low Energy Booster,
~47 MHz to 60 MHz



C. C. Friedrichs et al., PAC91, p. 1020

FNAL Booster 2nd harmonic,
76 MHz – 106 MHz, 100 kV



R. L. Madrak, IPAC16, p. 130

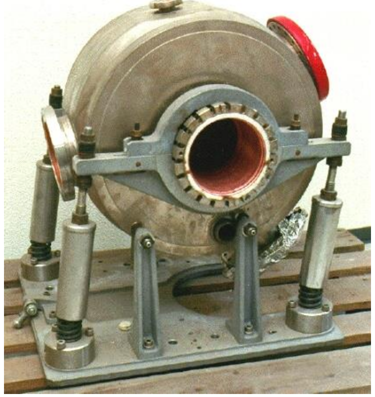
→ **Upper frequency limit for cavities with large tuning range**

Further increase frequency

→ **Remove inner conductor from coaxial set-up**

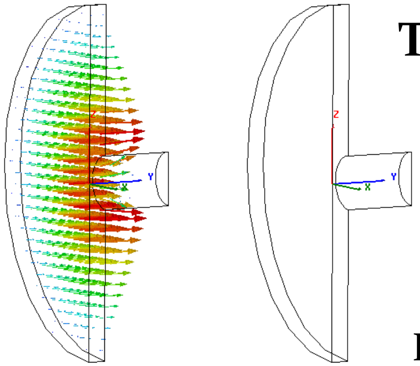
→ **The resonator becomes a pill-box cavity**

DORIS cavity



Electric field, TM$_{010}$-mode

Magnetic field, TM$_{010}$-mode



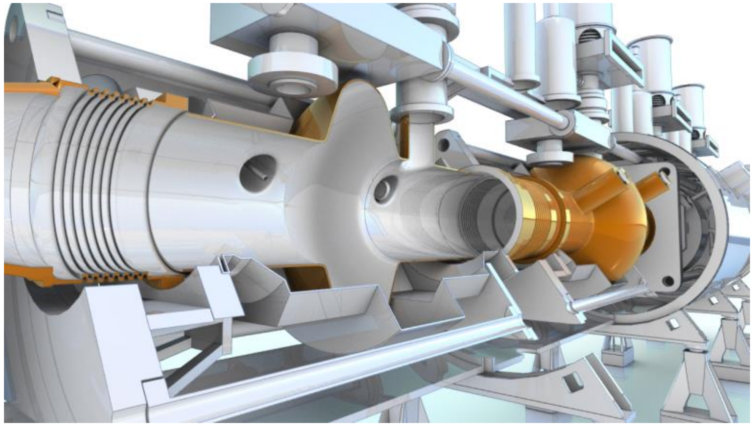
E. Jensen

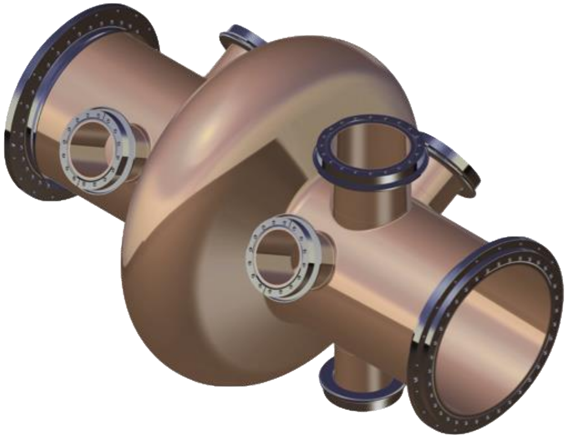
→ **The basis for cavity resonators**

Example: 400 MHz cavities in LHC

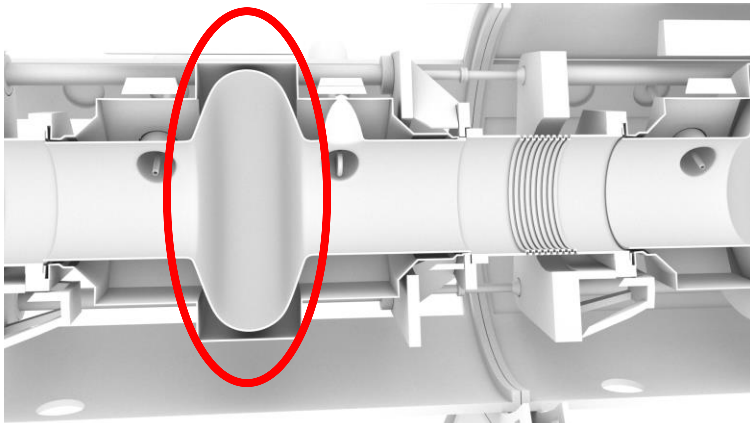
→ Reduce beam loading in RF cavities

→ Shunt impedance, *R*, low for small *R*/*Q* with normal conducting cavities → superconducting cavities in LHC





Bell shape: *R*/*Q* ~ 44 Ω, 400 MHz



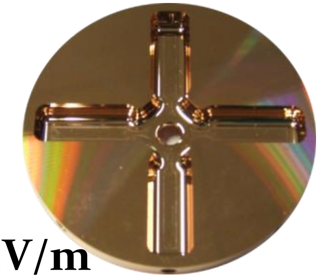
→ 2×8 cavities, 5.3 MV/m

$$\frac{1}{Q} = \cancelto{\sim 0}{\frac{1}{Q_{\text{cav}}}} + \frac{1}{Q_{\text{ext}}}$$

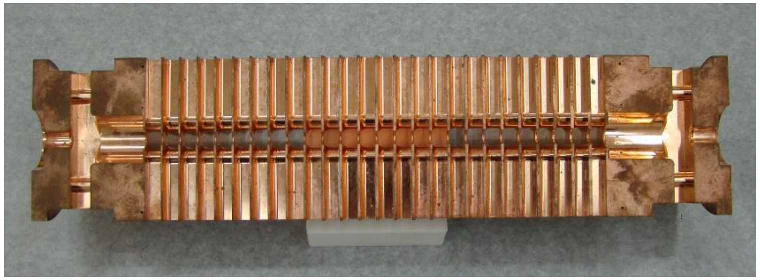
RF cavities in linear accelerators

- Beam only passes once → Maximize gradient
- Many accelerating cells to best reuse RF voltage

SuperHILAC, ~70 MHz, Berkley



CLIC, 12 GHz, ~100 MV/m



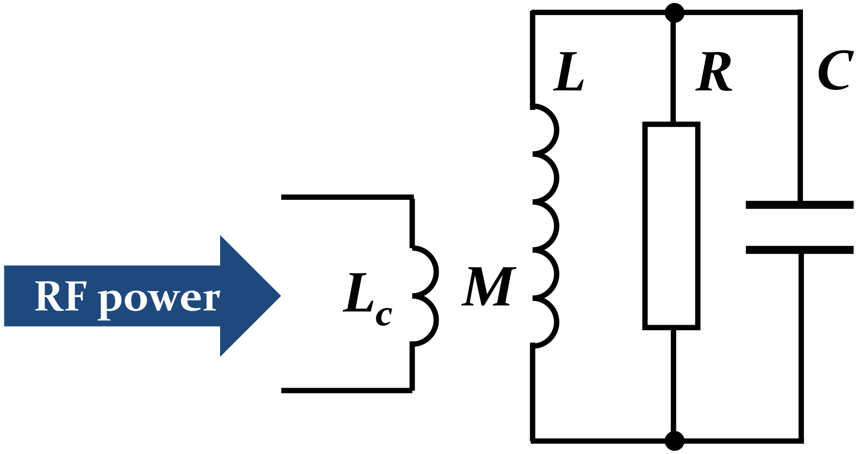
→ Cavity is the contrary to ‘one size fits all’

→ Many, many more variants

Coupling power into a cavity

Coupling power into a cavity

- Attack inductivity or capacitance of resonator, or combined



$$\frac{1}{Q} = \frac{1}{Q_{\text{cav}}} + \frac{1}{Q_{\text{ext}}}$$

→ Coupling loop forms transformer with resonator inductivity

L. Stigelin

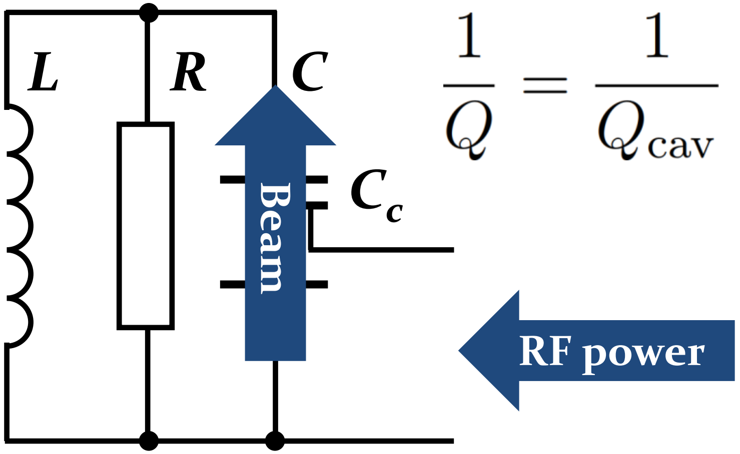
- Main coupler
  PSI cyclotron

→ ~1 MW at 50 MHz

A. Rodchenko

Coupling power into a cavity

- Attack inductivity or capacitance of resonator, or combined



$$\frac{1}{Q} = \frac{1}{Q_{\mathrm{cav}}} + \frac{1}{Q_{\mathrm{ext}}}$$

→ Capacitive divider to gap to transform generator impedance to cavity shunt impedance

→ Beam also couples capacitively via the gap

Coupler of CERN PS 40 MHz

→ Coupler forms one half of capacitor with the gap

Capacitive (electric) coupling

- **Coupling through an electric antenna**

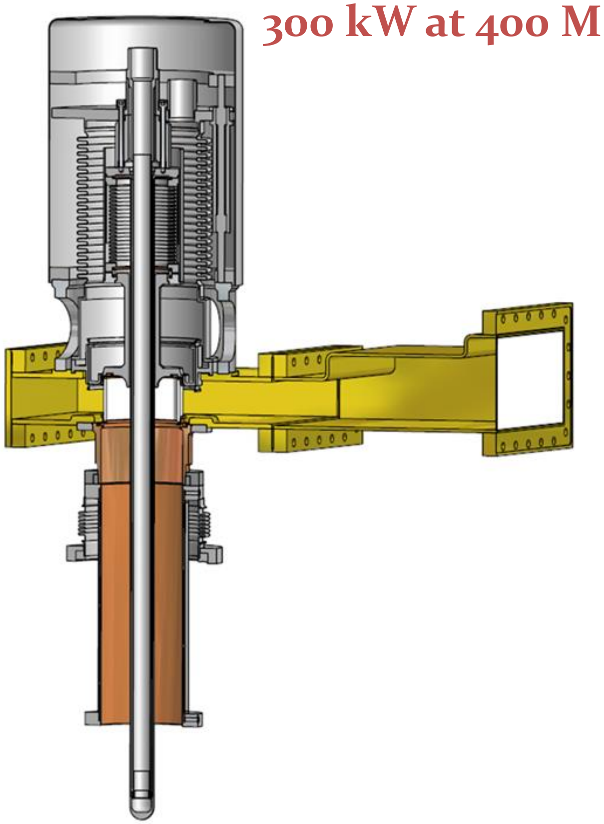
Electrical coupler to space

https://en.wikipedia.org/wiki/Transmitter_Solt

→ 2 MW at 540 kHz

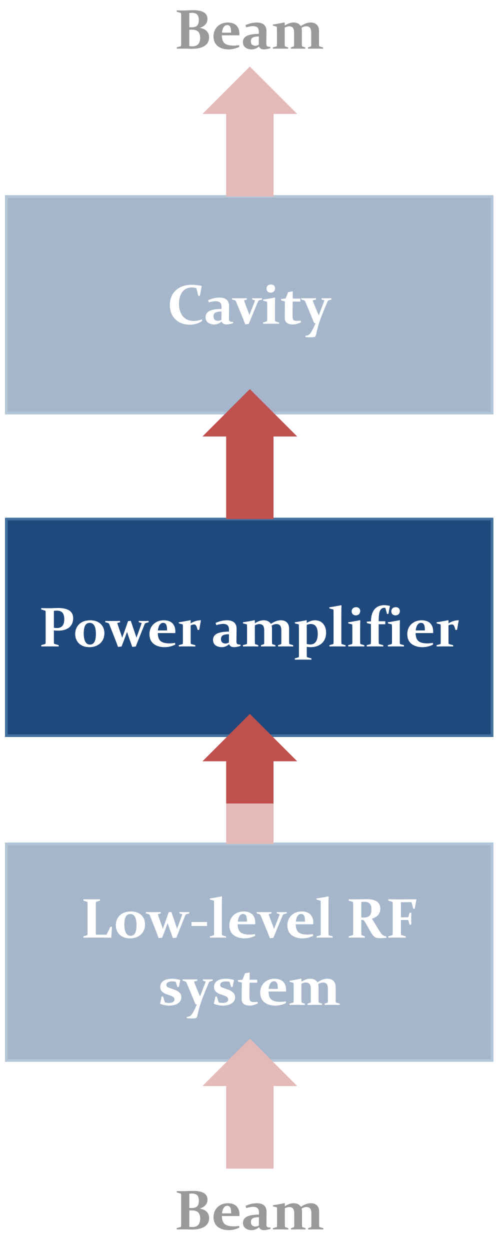
Power coupler of LHC cavities

300 kW at 400 MHz



→ Coupler antenna transmits directly into the cavity

RF system overview



→ **Convert RF power into longitudinal electric field**

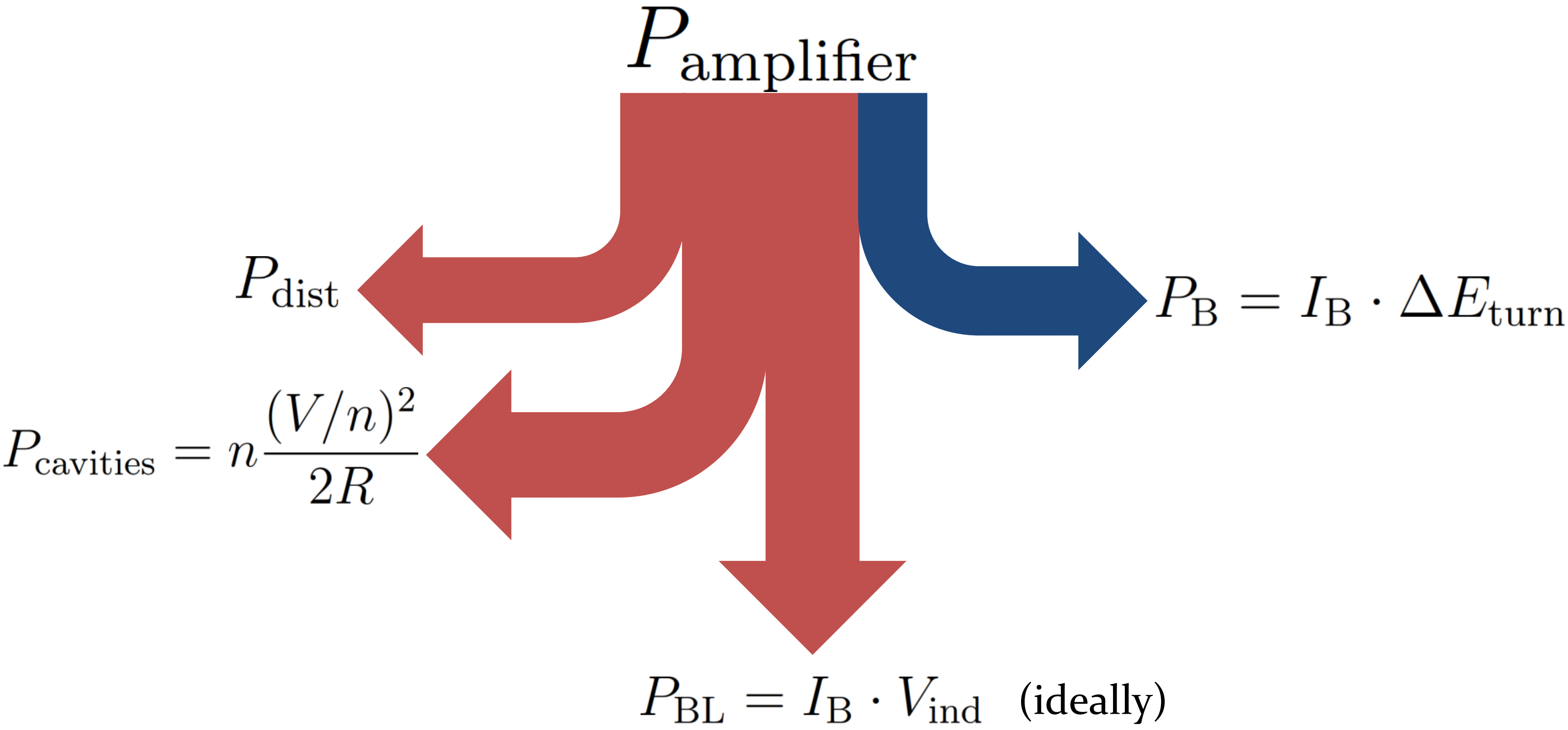
→ **Amplify low-power signal from beam control to kW, MW or GW**

→ **Provide RF signals with correct frequency, amplitude and phase**

Power amplifiers

How much power is required?

1. **Power to accelerate beam** → **Wanted**
2. **Compensate beam-induced voltage** → **Refl. *P***
3. **Compensate electrical losses in cavity** → **Heat**
4. **Compensate electrical losses in distribution** → **Heat**



$$P_{\text{amplifier}}$$

$$P_{\text{dist}}$$

$$P_{\text{B}} = I_{\text{B}} \cdot \Delta E_{\text{turn}}$$

$$P_{\text{cavities}} = n\frac{(V/n)^2}{2R}$$

$$P_{\text{BL}} = I_{\text{B}} \cdot V_{\text{ind}} \quad \text{(ideally)}$$

Power amplifiers

- Basically

$$\boxed{P_{\text{out}} = g \cdot P_{\text{in}} \text{ or } V_{\text{out}} = \sqrt{g} \cdot V_{\text{in}}}$$

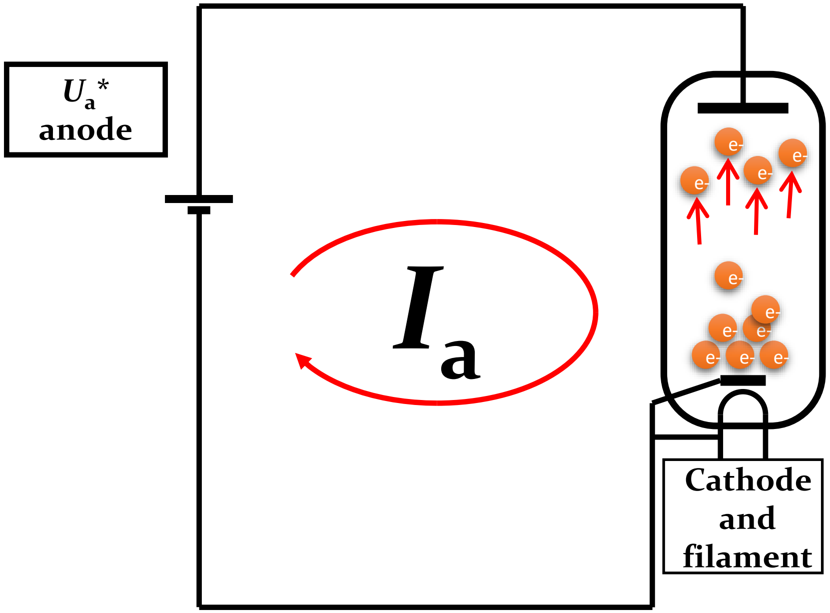
- The ideal power amplifier
  - → Large bandwidth: amplifies all frequencies equally
  - → No saturation, infinite power
  - → Zero delay
  - → No added noise
  - → Unconditionally stable and resistant to reverse power
  - → Radiation-hard

→ Unfortunately such a device has not been invented yet

→ Let us have a look at some real amplifiers

Basics of grid tube

- **From diode to tetrode amplifier**



- **Vacuum tube**
- **Heater + Cathode**
  - **Heated cathode**
    - **Coated metal, carbides, borides,…**
  - **thermionic emission**
  - **Electron cloud**
- **Anode**

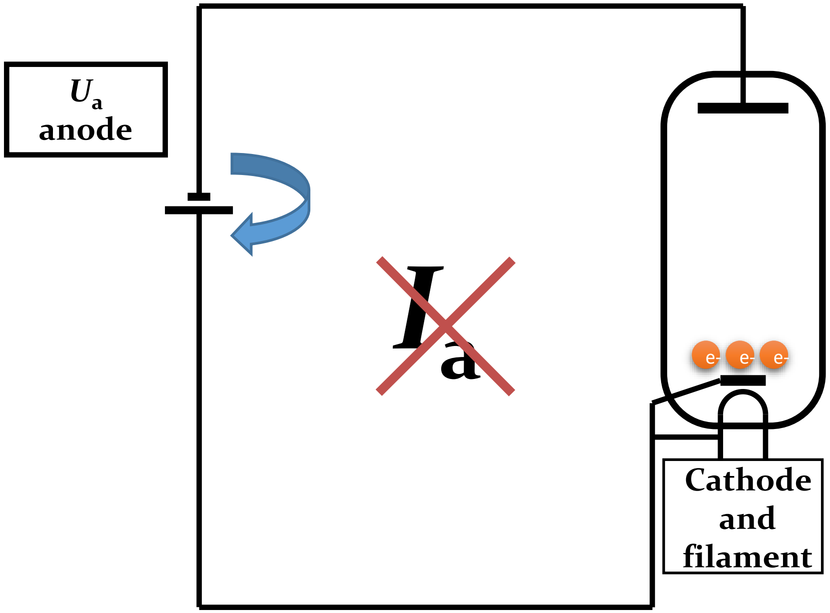
→**Diode**

*For tube amplifier designs voltages are named U instead of V

E. Montesinos

Basics of grid tube

- From diode to tetrode amplifier



- Vacuum tube
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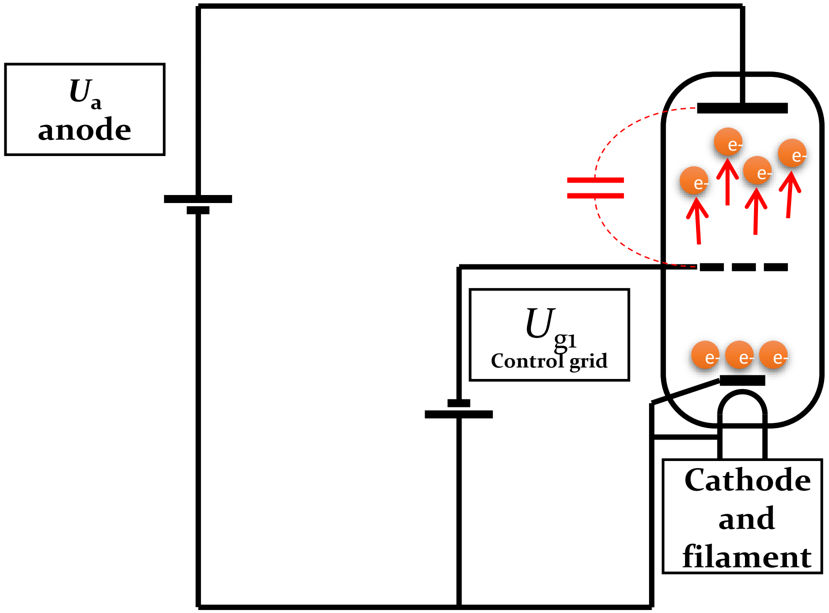
E. Montesinos

Basics of grid tube

- From diode to tetrode amplifier



→Triode

- Modulating the grid voltage proportionally modulates the anode current
- Transconductance
  - Voltage at grid
  - → Current at anode
- Limitations
  - Parasitic capacitor from anode to control grid (g_1)
  - Tendency to oscillate

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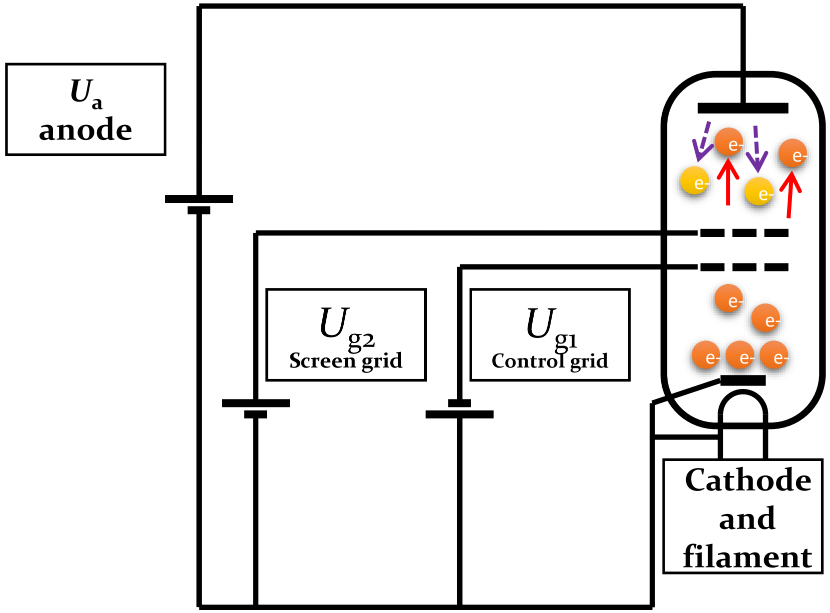
Basics of grid tube

- **From diode to tetrode amplifier**



→Tetrode

- Screen grid
  - Positive (lower anode)
  - Decouple anode and g1
  - Higher gain
- Limitations
  - Secondary electrons
  - Anode treated to reduce secondary emission

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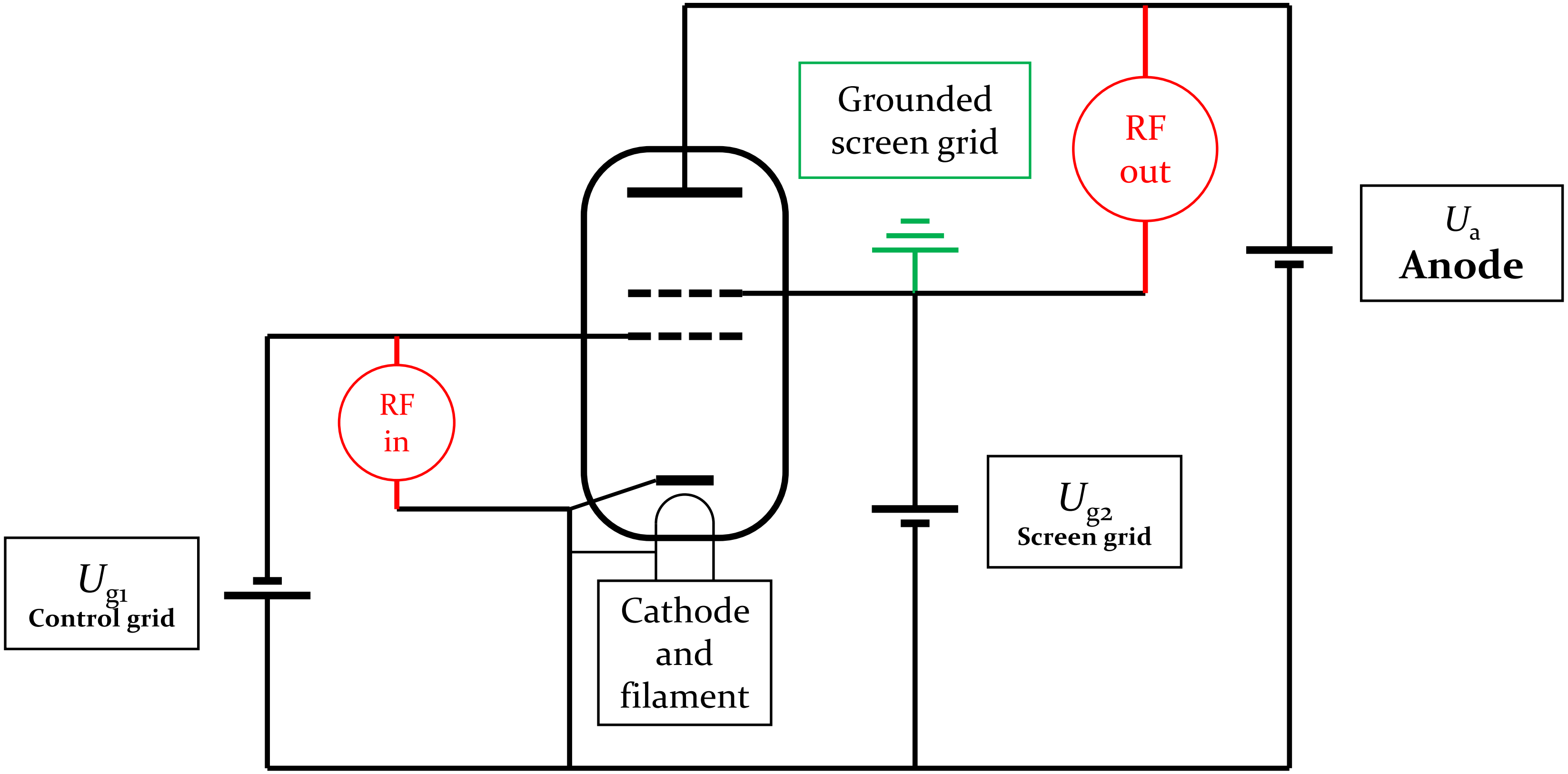
Tetrode based power amplifier

- **Example of SPS 200 MHz amplifier, tetrode RS2004**



→ Very simplified block diagram

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Example: Tetrode amplifier driving SPS RF

- Two transmitters, 2 × 1 MW at 200 MHz (almost continuous)
- Eight tetrodes per amplifier

RS2004 tetrode

Amplifier trolley

Complete transmitter

→ In operation since 1976

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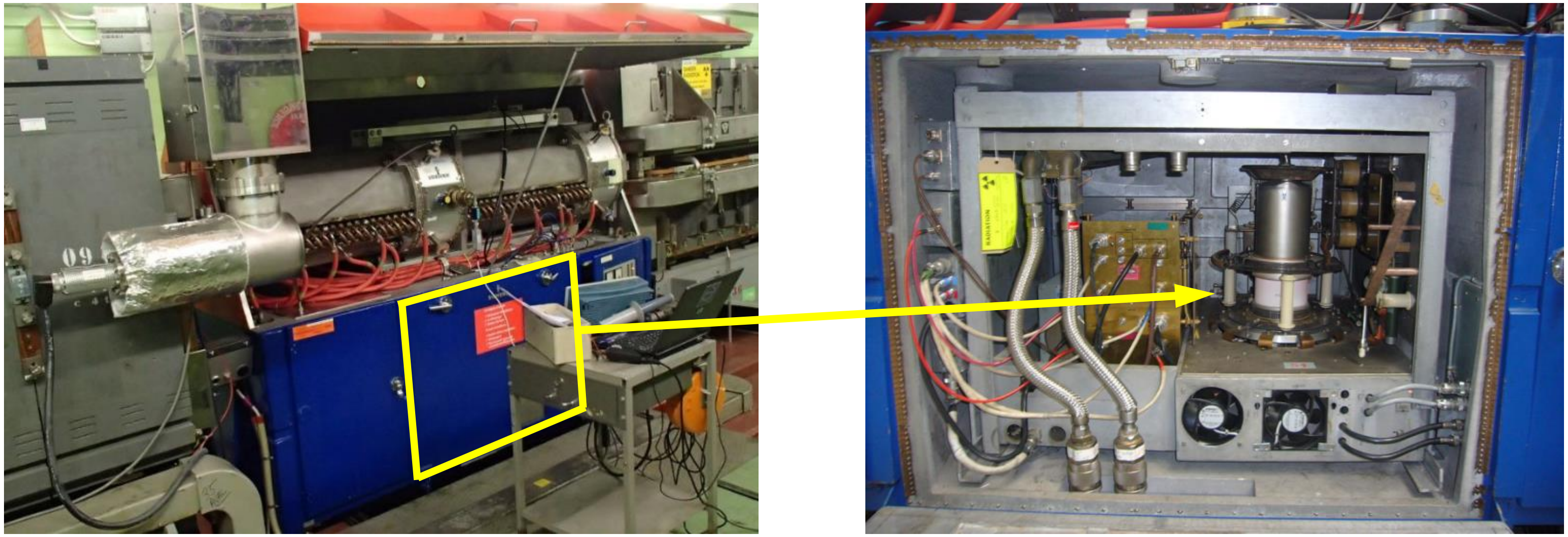
Tetrode amplifier driving PS RF

- → **Frequency range 2.8…10 MHz, ~60 kW per cavity, 11 units**
- → **Space constraints to have amplifier installed below cavity**

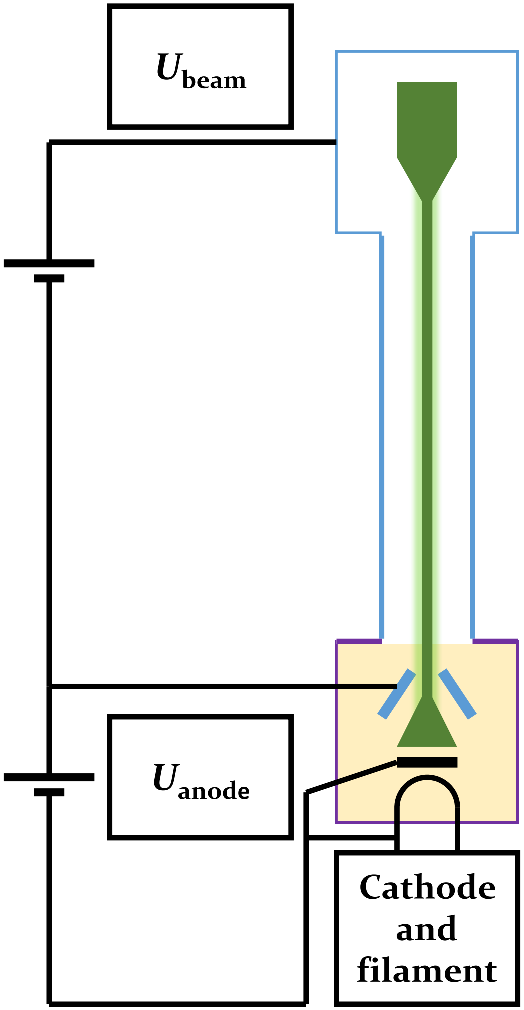


Amplifier trolley

- → **Tetrode is obvious choice**
  - → High power in small volume
  - → Operates in radioactive environment

Basics of linear beam tube

- Klystron: a complete mini-accelerator



- Klystrons velocity modulation
  - Converts the kinetic energy into RF power
- Vacuum tube
- Electron gun
  - Thermionic cathode
  - Anode
- Electron beam
- Drift space
- Collector
- e- constant speed until the collector

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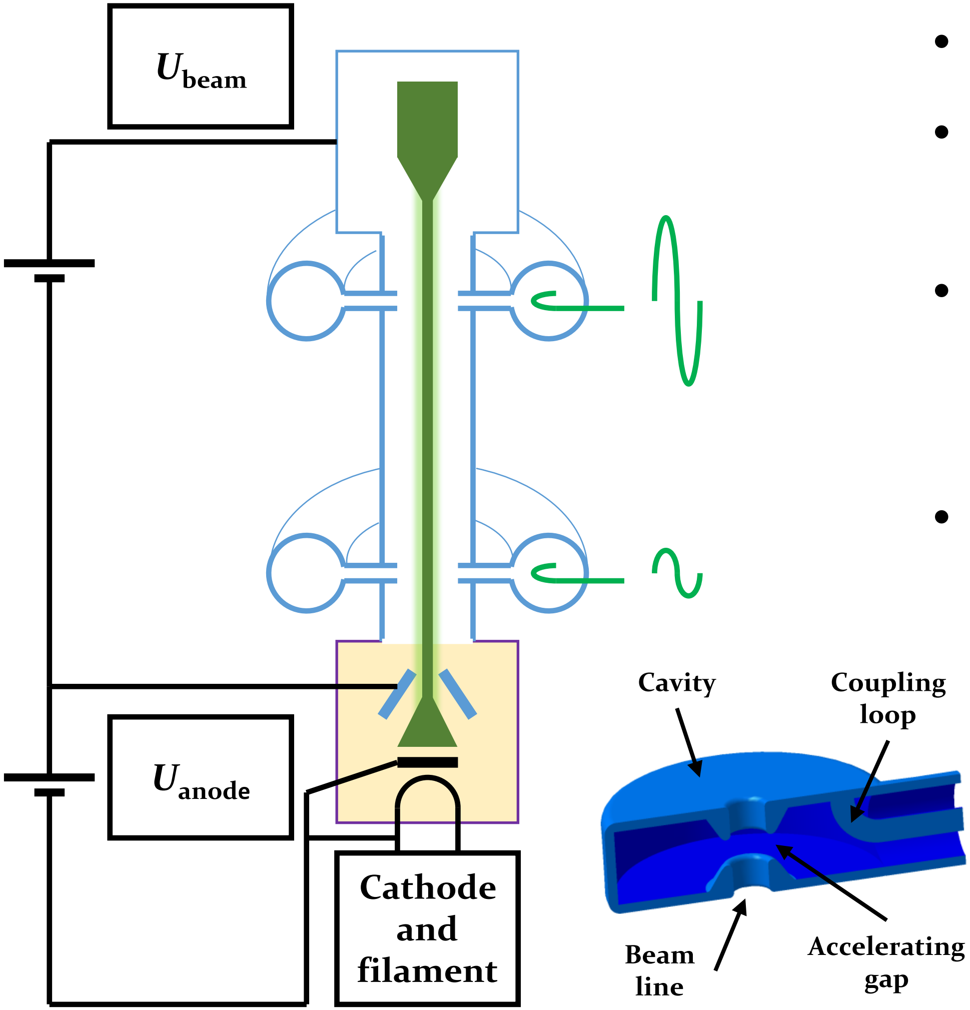
Basics of linear beam tube

- Klystron: a complete mini-accelerator

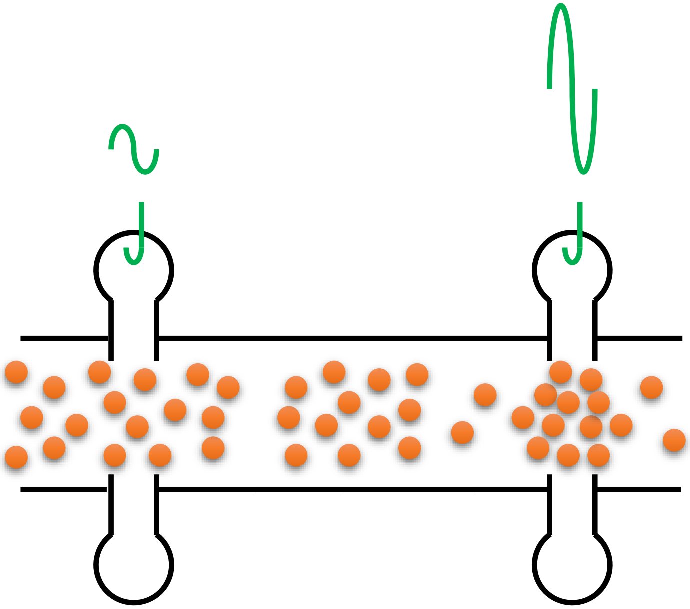


- Cavity resonators and drift
- RF input cavity (Buncher)
  → Modulates electron velocity
- Drift space
  → Faster electrons catch up
  → Slower electrons fall behind
- RF output cavity (Catcher)
  - Resonating atsame frequency as input cavity
  - At place where electrons are maximally bunched
  - Kinetic energy converted into voltage and extracted

E. Montesinos

Basics of linear beam tube

- **Klystron:** a complete mini-accelerator



- Cavity resonators and drift
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  - Resonating atsame frequency as input cavity
  - At place where electrons are maximally bunched
  - Kinetic energy converted into voltage and extracted

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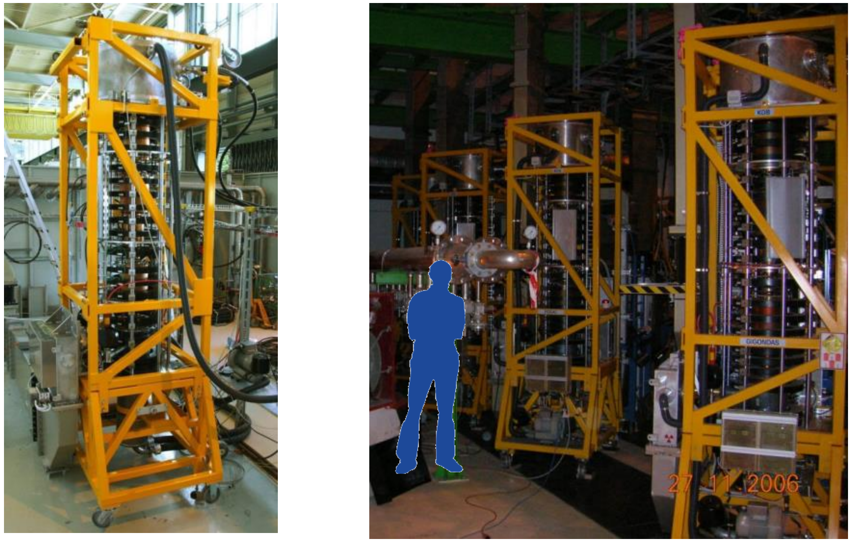
Example: Klystrons driving accelerators

- 2 × 8 cavities, each driven by separate 400 MHz klystron, 330 kW

→ First klystron amplifiers powering a hadron collider



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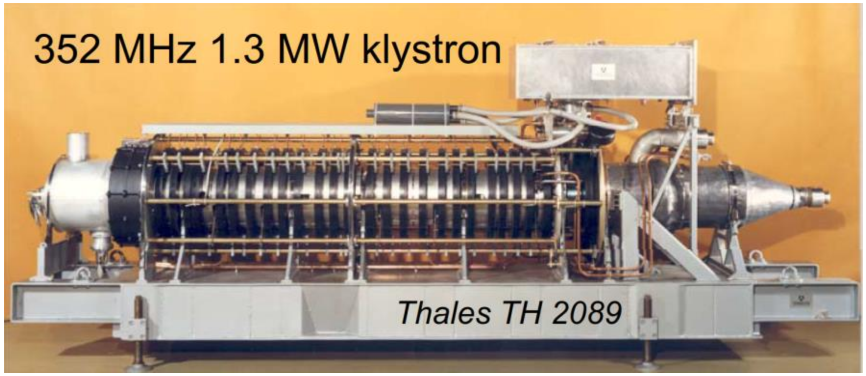
- 12 GHz pulsed klystron for CLIC

→ 50 MW in 1.5 μs

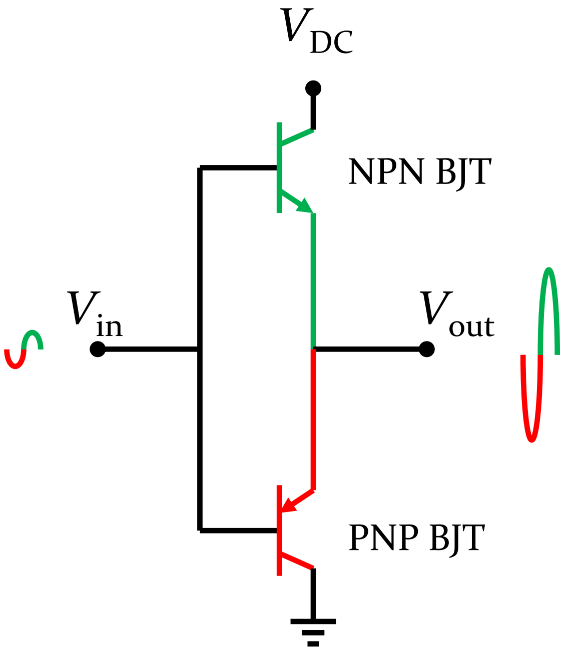
G. McMonagle

- Significantly more power was required to feed LEP (until 2000)

→ About 50 MW CW was installed at 352 MHz



Basics of RF solid state amplifiers



BJT: Bipolar Junction Transistor

- **In a push-pull circuit the RF signal is applied to two devices**
  - **One of the devices is active on the positive voltage swing and off during the negative voltage swing**
  - **The other device works in the opposite manner so that the two devices conduct half the time**

  →The full RF signal is then amplified

→Needs two different type of devices

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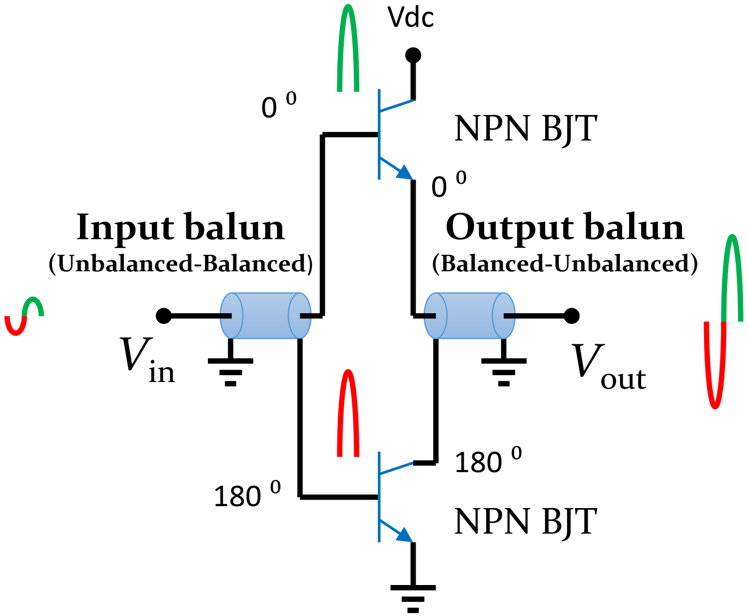
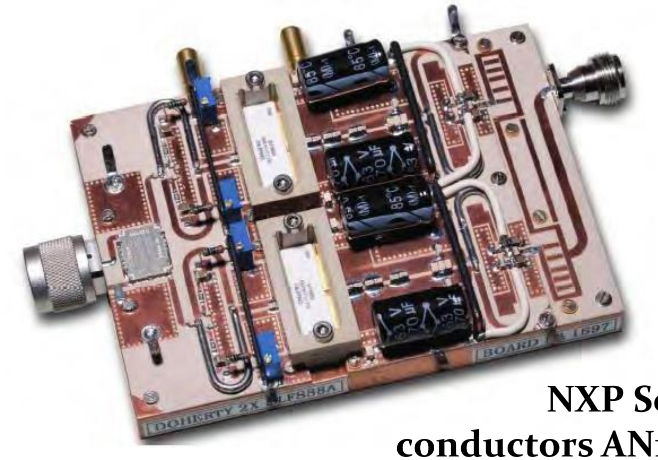
Basics of RF solid state amplifiers





NXP Semi-conductors AN11325
2-way Doherty amplifier with BLF888A

- **Another push-pull configuration is to use a balun (balanced-unbalanced)**
  - **Power splitter, equally dividing the input power between the two transistors**
  - **Balun keeps one port in phase and inverts the second port in phase**
- **Since the signals are out of phase only one device is On at a time**

→This configuration is easier to manufacture since only one type of device is required

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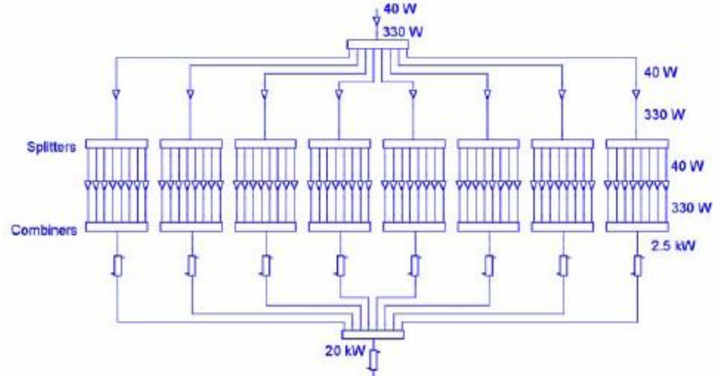
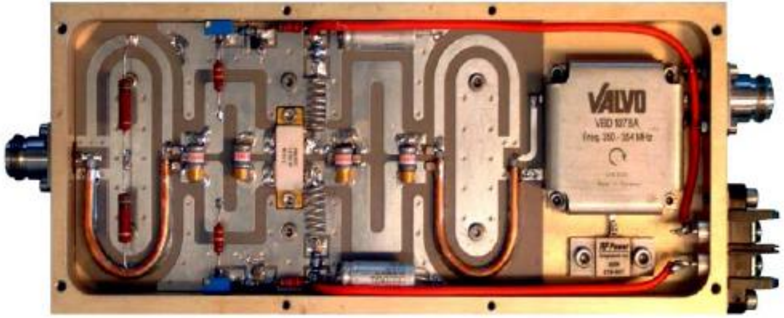
Example: Soleil 45 kW, 352 MHz

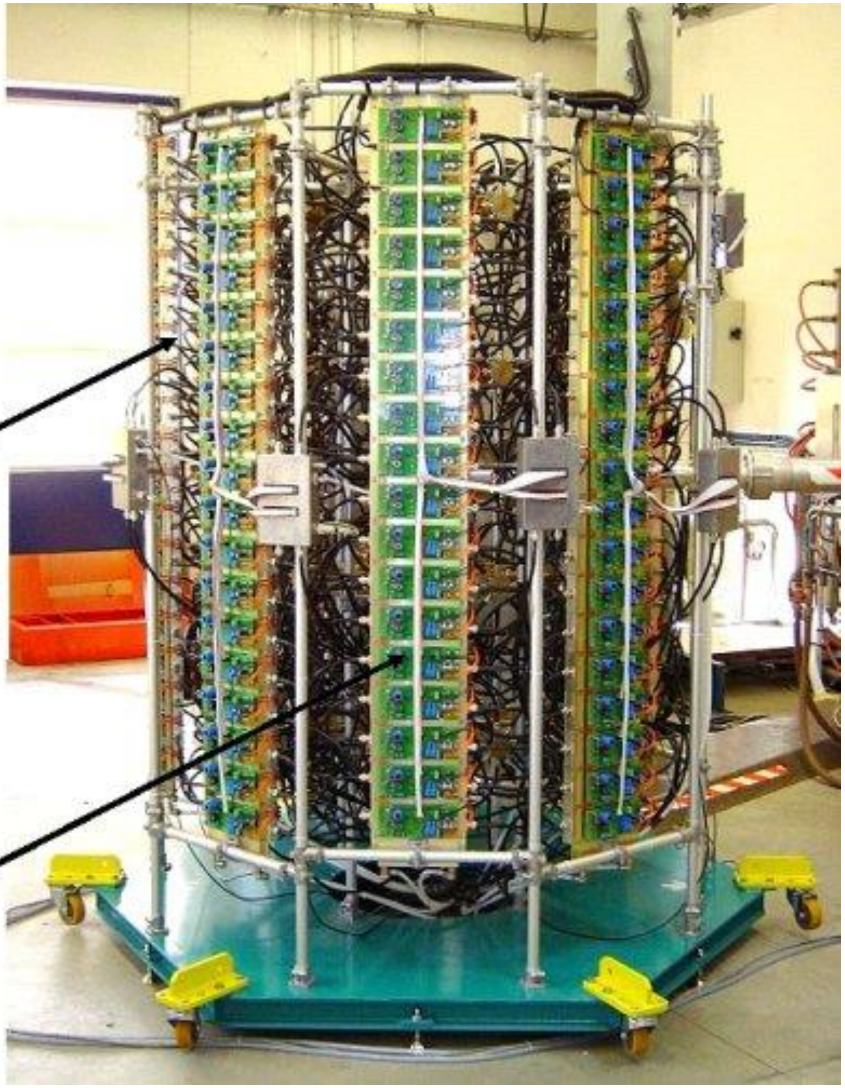
Electron storage ring running at 352 MHz



330 W amplifier module



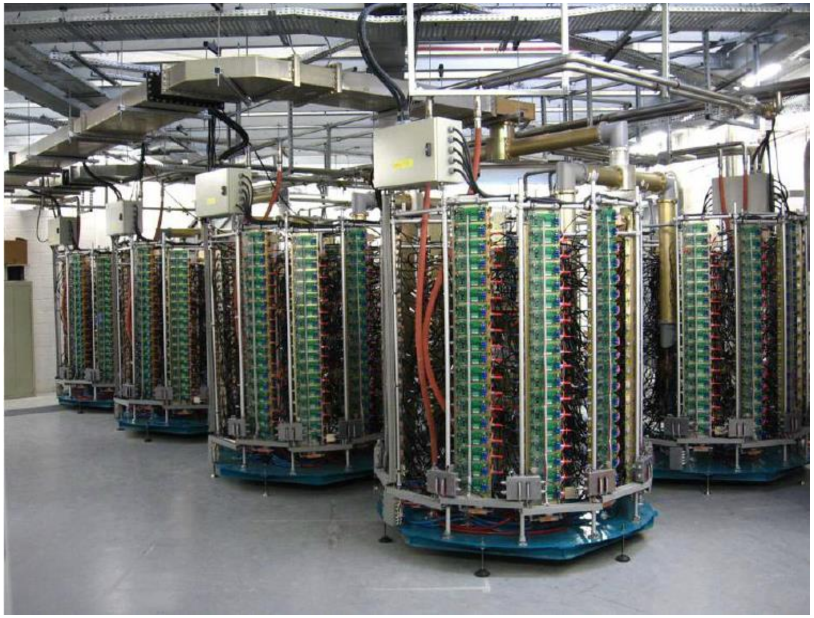
600 W, 300 V_{DC}/30 V_{DC} converter



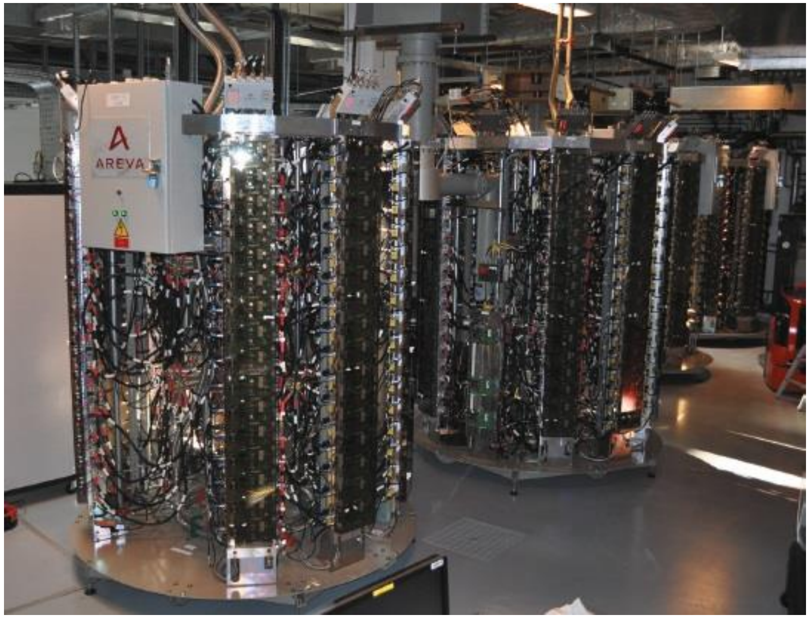
Example: Soleil 45 kW, 352 MHz

Large scale solid state amplifier installations

45 kW per tower (2004 and 2007)



150 kW per tower (2012)

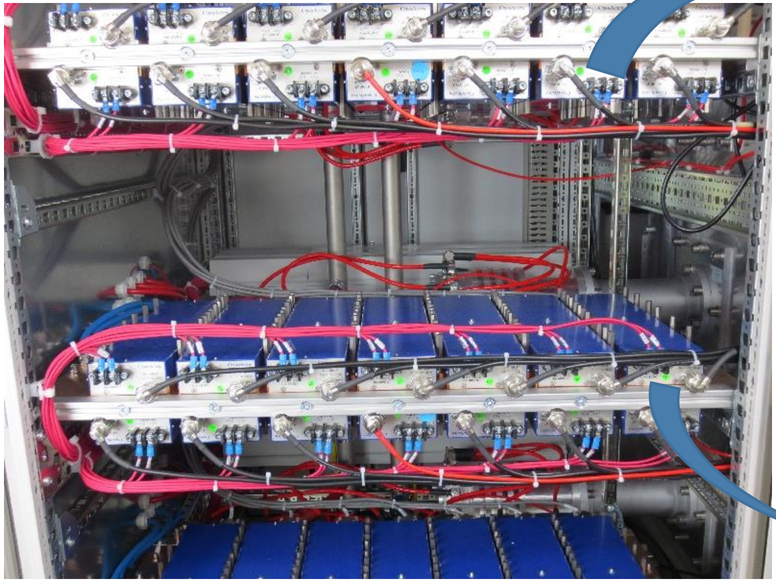


→ Requires a series of power combiners to moderate power per amplifier module to several tens of kilowatts

E. Montesinos

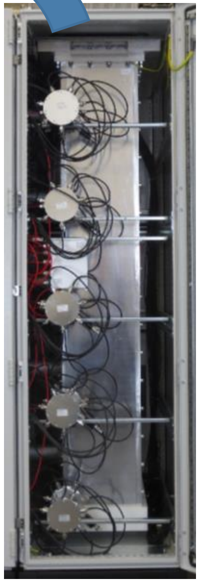
Example: BESSY II

500 MHz solid state amplifiers: 4 × 80 kW for storage ring,
40 kW for booster synchrotron



Amplifier modules

80 kW unit



Combiner

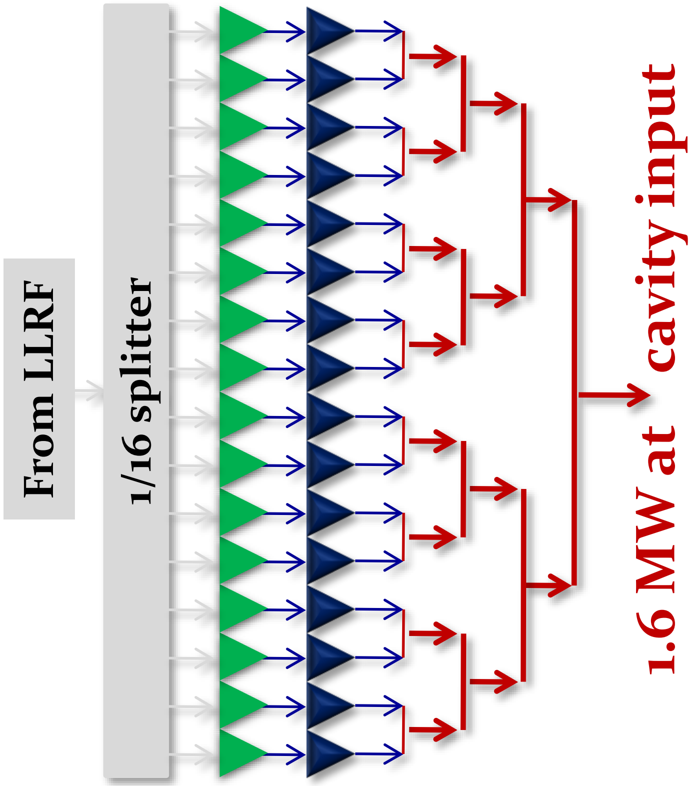
→ Power per module limited by RF transistors

→ Increasing with modern semiconductor devices

B. Schriefer

Example: SPS

200 MHz solid state amplifiers: 2 × 1.6 MW peak power,
2 × 16 towers per amplifier



From LLRF

1/16 splitter

1.6 MW at cavity input

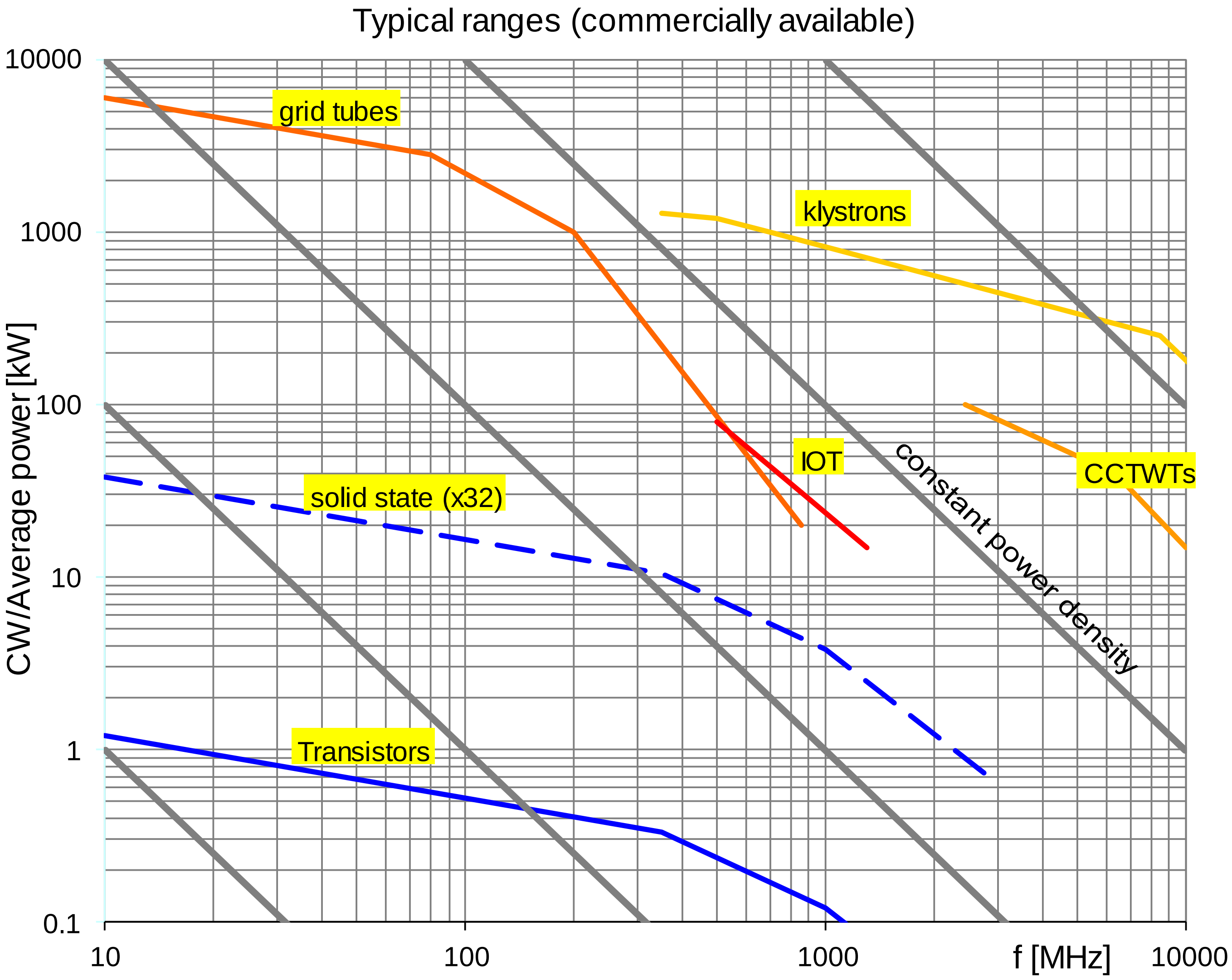
→ **80 modules per tower, 1280 modules with 5120 transistors per amplifier**

→ **Presently the largest RF installation in a particle accelerator**

E. Montesinos

RF power amplifier

Power capability of commercially available amplifier types



E. Jensen

How to choose the right RF amplifier?

Prefer tube amplifier, when	Prefer solid-state amplifier, when
• Amplifier must be installed in the accelerator tunnel	• Amplifier can be located in non-radioactive environment
• Expecting important spikes from beam induced voltage	• Circulator can be installed to protect the amplifier
• Large output power of a single device is required, without combiners	• Delay due to unavoidable combiner stages is little issue
• Not much space is available	• Sufficient space can be made available
• High peak power in pulsed mode	• Continuous operation
• Amplifier must be compact and/or close to cavity	• Amplifier can be separate from the cavity

→ Mostly no hard criteria → decide on case by case basis

Summary

- **RF system parameters**
  - → Choose frequency and voltage wisely
- **Parameters of RF cavities**
  - → R, R/Q
  - → No 'one-size fits' all
- **Power amplifier**
  - → Ideal amplifier does not (yet) exist
  - → Tube or solid-state based
- **Feedbacks and longitudinal beam control**
  - → Make the beam feel comfortable in bucket
  - → Beam phase, radial and synchronization loops

RF Systems II

H. Damerau

CERN

Introduction to Accelerator Physics

3 October 2023

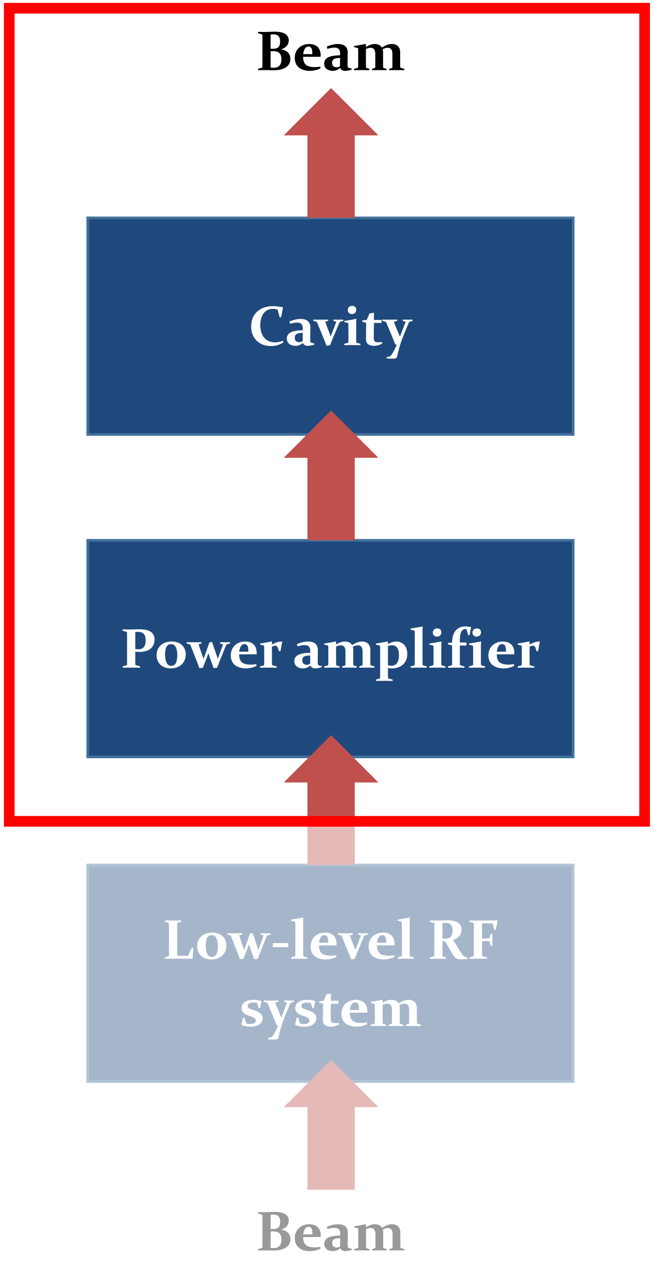
The CERN Accelerator School

Copyright statement and speaker's release for video publishing

Outline

- Introduction
- Choice of parameters
  - Frequency and voltage
- RF cavity parameters
  - Shunt impedance, beam loading, power coupling
- Power amplifiers
  - Tube or solid state
  - Local feedbacks
- Longitudinal beam control system
  - Building blocks: RF source and receiver
  - Phase, radial and synchronization loops
- Summary

RF system overview



Beam

Cavity

Power amplifier

Low-level RF system

Beam

→ **Convert RF power into longitudinal electric field**

→ **Amplify low-power signal from beam control to kW, MW or GW**

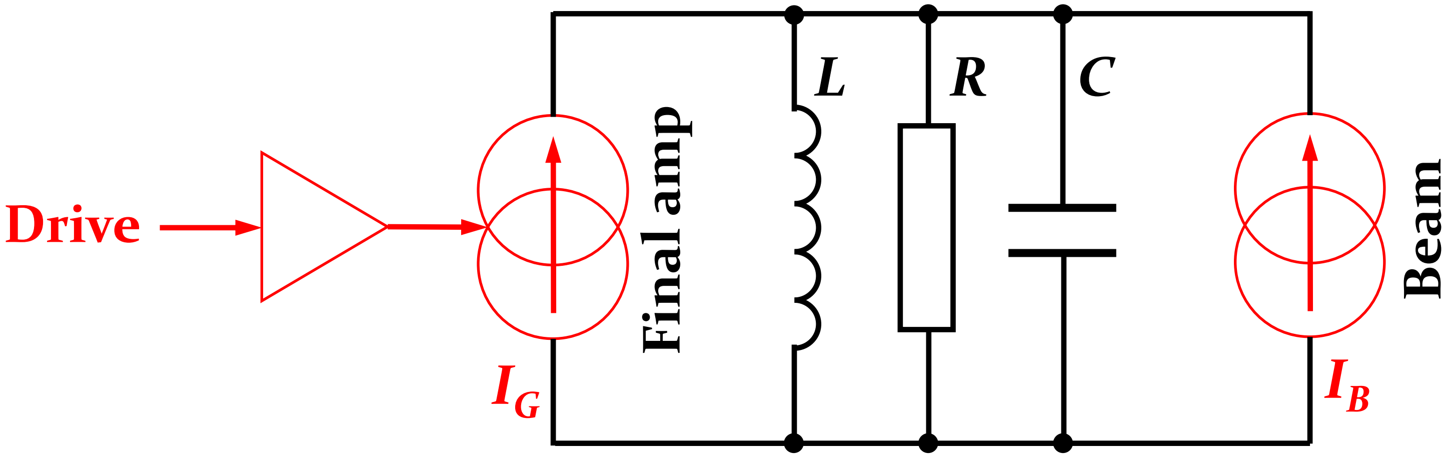
→ **Provide RF signals with correct frequency, amplitude and phase**

Local feedbacks

Reduction of cavity impedance

- Energy transfer from cavity to beam, but from beam to cavity

→ Both, RF generator and beam can induced voltage in cavity

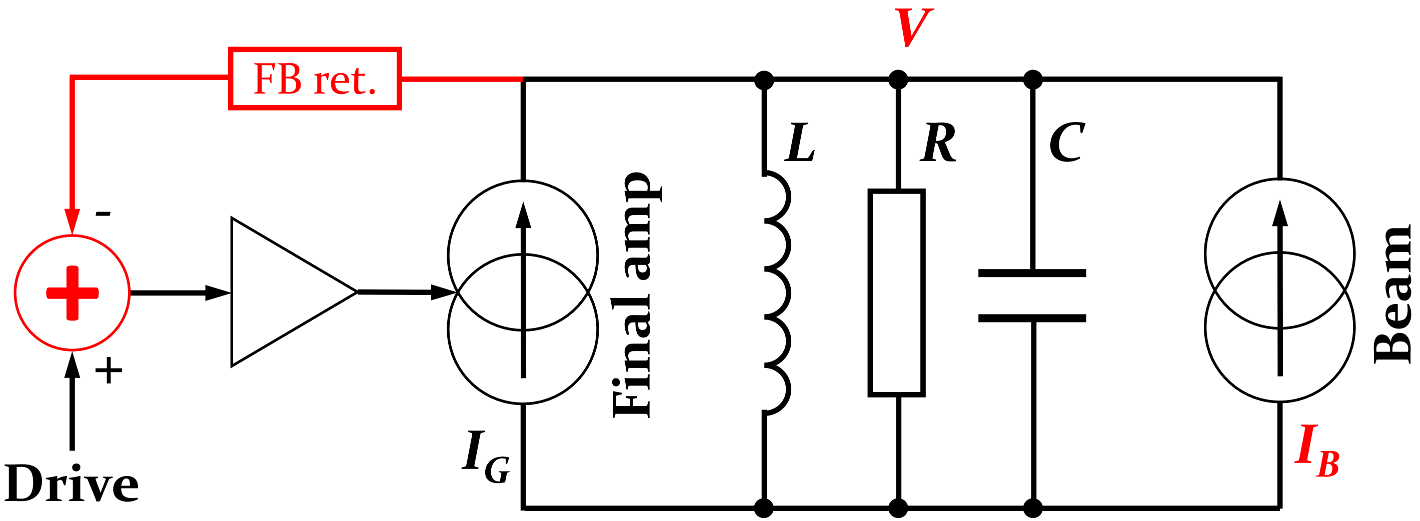


1. Reduce beam induced voltage by reducing *R*, **but not efficient**
   → Obviously needs more power → $$$
2. Feedback to decrease the apparent impedance for the beam
   → Use amplifier to counteract beam induced voltage

Reduction of cavity impedance

- **Energy transfer from cavity to beam, but from beam to cavity**

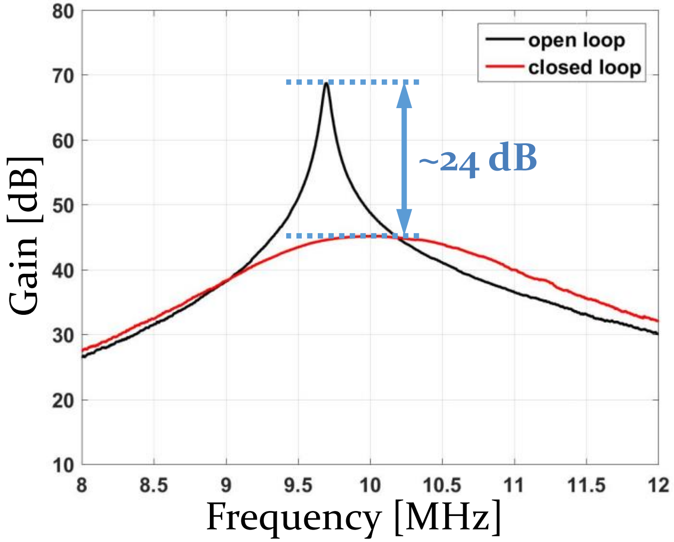
→ **Both, RF generator and beam can induced voltage in cavity**



1. **Compare drive signal (no beam) with gap (beam and generator)**
2. **Amplify inverted difference**

$$Z_{\mathrm{eq}}(\omega) = \frac{dV}{dI_B} = \frac{Z(\omega)}{1 + g_{\mathrm{OL}}}$$

Example: 10 MHz RF system in CERN PS



Transfer function with and without feedback

- Feedback gain of 24 dB

→ Equivalent impedance, $Z_{eq}(\omega)$ reduced

→ Impedance for amplifier remains unchanged, $Z(\omega)$

$Z(\omega)$ → Cavity, amplifier with feedback ← $Z_{eq}(\omega)$

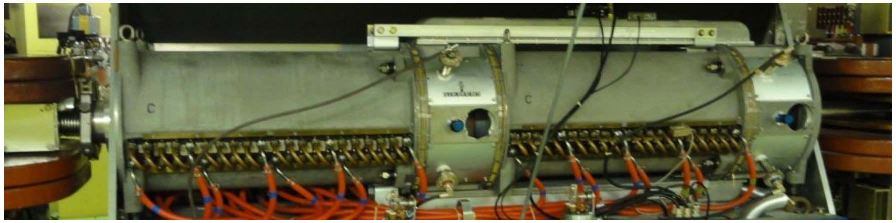
Why not further reduction with more gain?

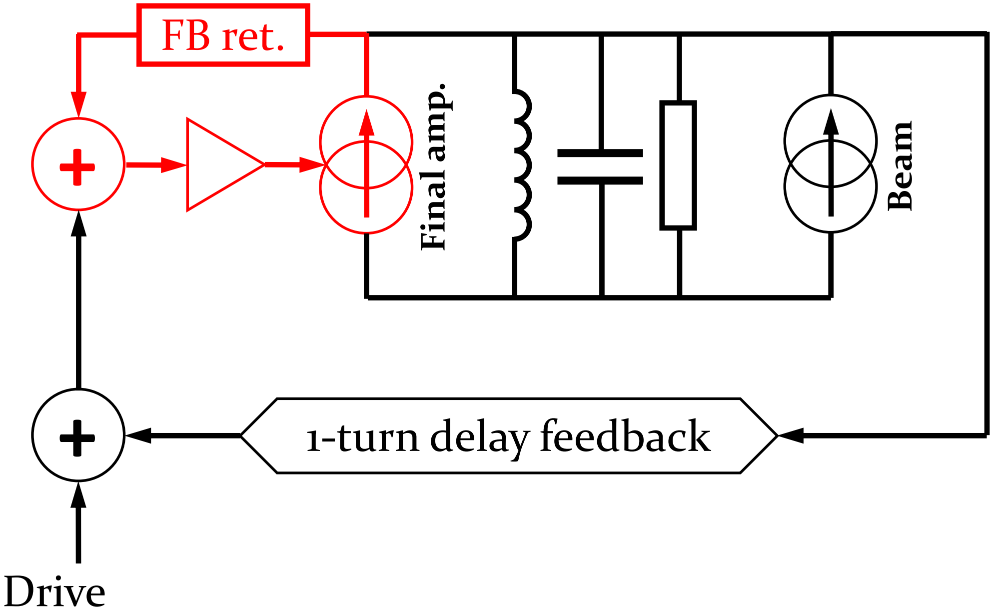
- Subtraction of gap voltage and drive signal imperfect due to
  1. Delay of cables and amplifier
  2. Parasitic resonances of amplifier and cavity system

Bandwidth ↑ ↔ Achievable gain ↓

Example: 10 MHz RF system in CERN PS

- 10 + 1 ferrite loaded cavities, tunable from 2.8…10 MHz

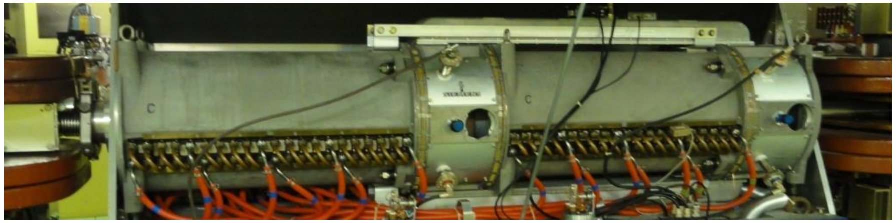


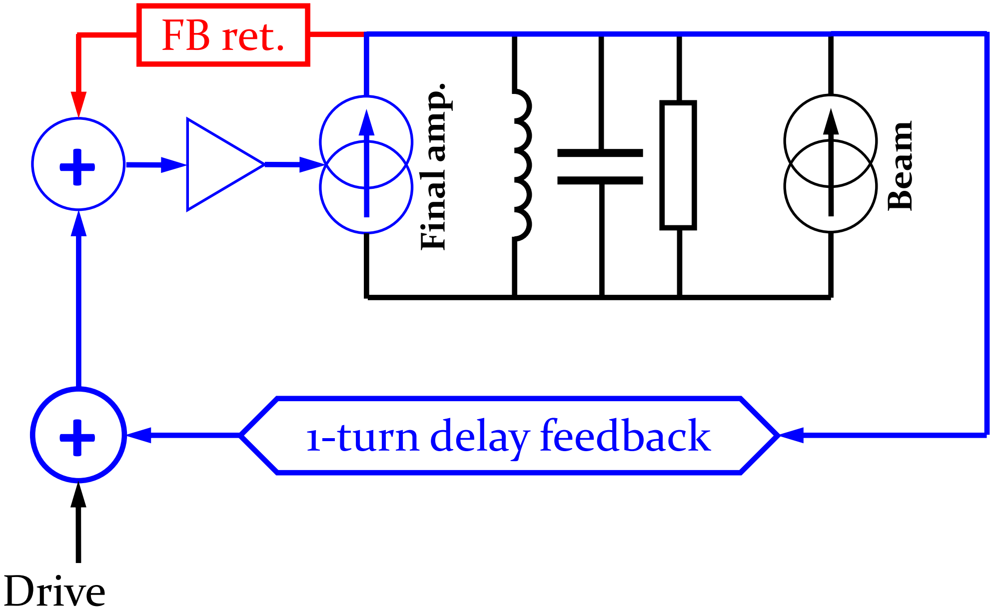


- Fast wide-band feedback around amplifier (internal)
  → Gain limited by delay

Example: RF feedback with 1-turn delay

- 10 + 1 ferrite loaded cavities, tunable from 2.8…10 MHz





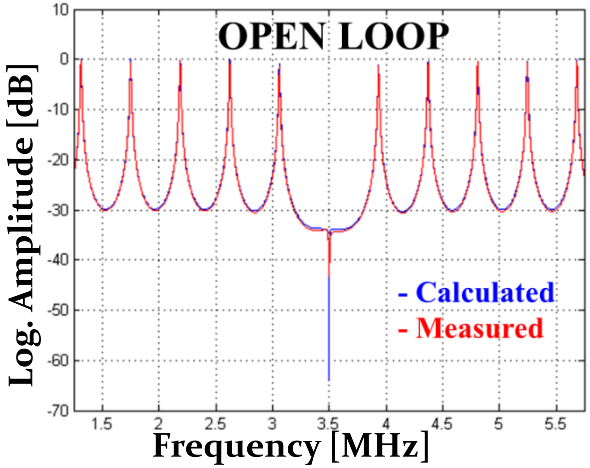
- **Fast wide-band feedback around amplifier (internal)**
  → Gain limited by delay

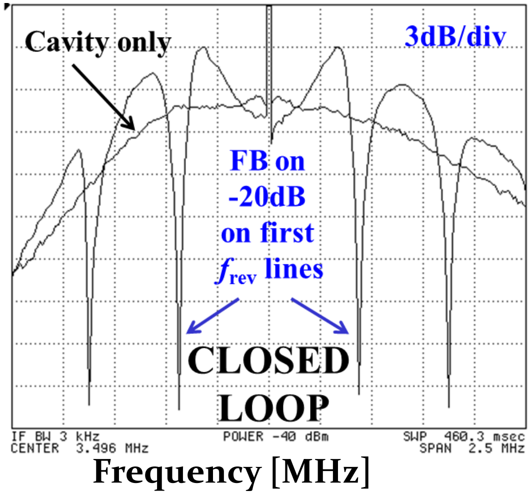
- **1-turn delay feedback**
  → High gain at $n \times f_{\mathrm{rev}}$

Example: RF feedback with 1-turn delay

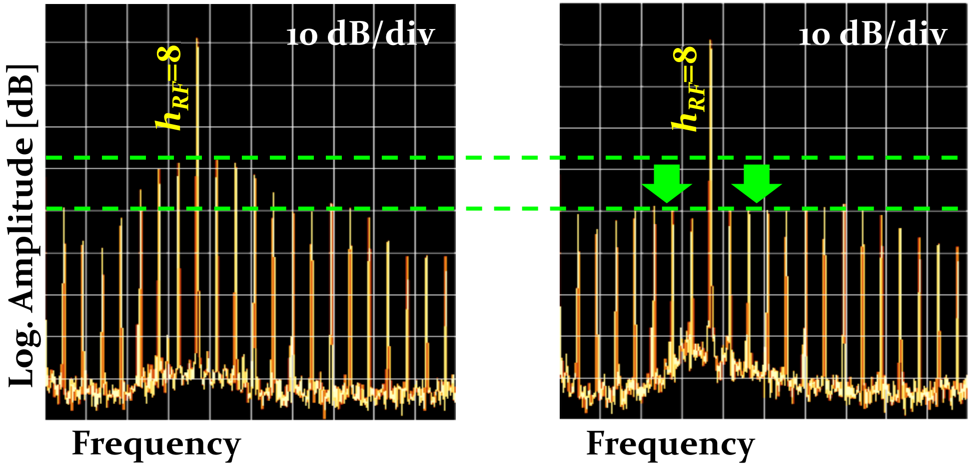
→ **Reduce cavity impedance beyond stability limit of wide-band FB**

Open/closed loop transfer functions





Spectrum at cavity gap return


Feedback off

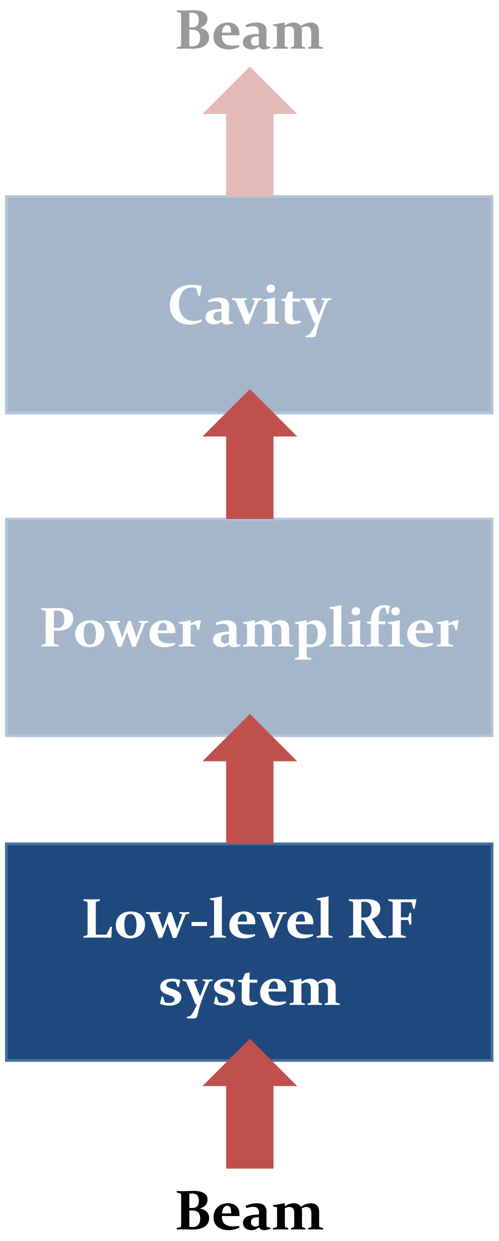
Feedback on

D. Perrelet

→ **Important additional impedance reduction**

→ **Clever usage of beam periodicity in circular accelerator**

RF system overview



→ **Convert RF power into longitudinal electric field**

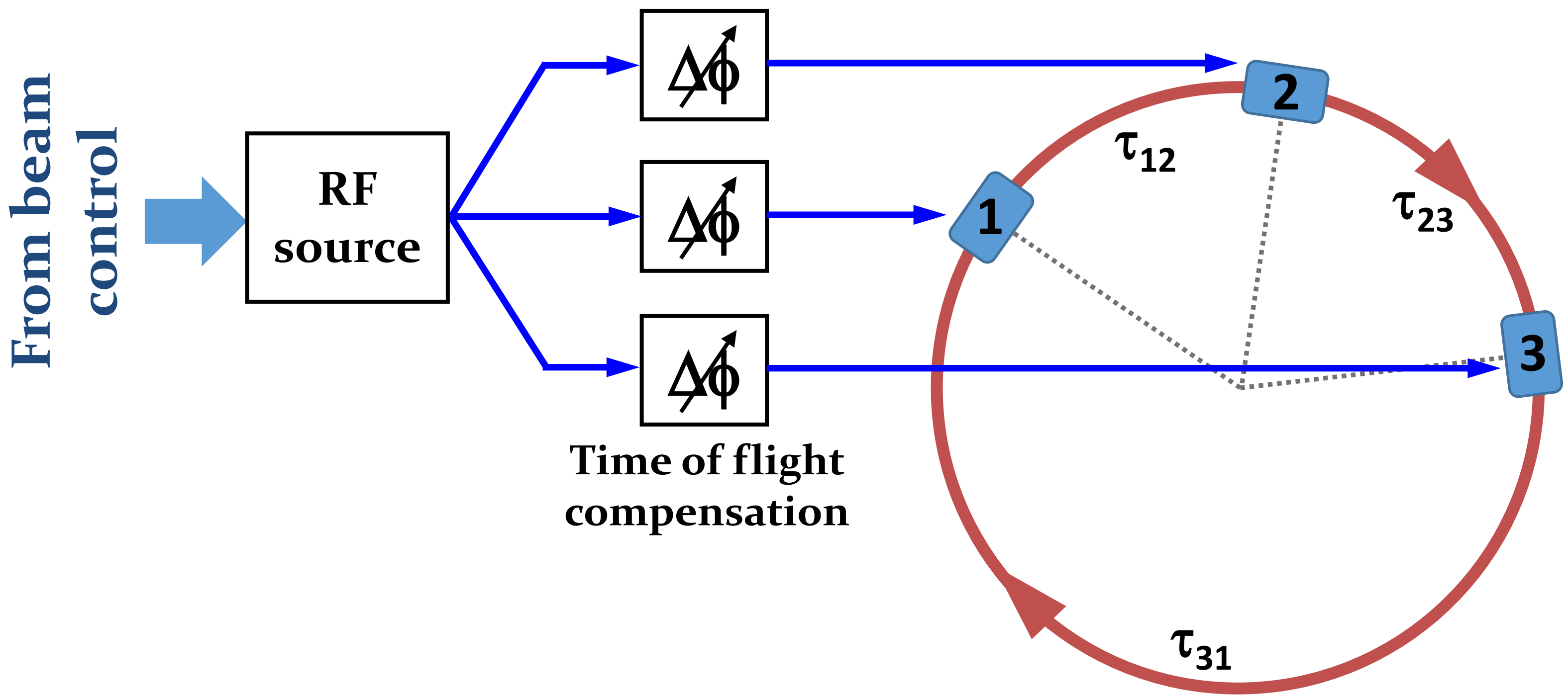
→ **Amplify low-power signal from beam control to kW, MW or GW**

→ **Provide RF signals with correct frequency, amplitude and phase**

Global feedbacks
Low-level RF beam control

Longitudinal beam control

- Local feedbacks → Act on individual RF stations
- Global feedbacks → Act on all RF stations simultaneously



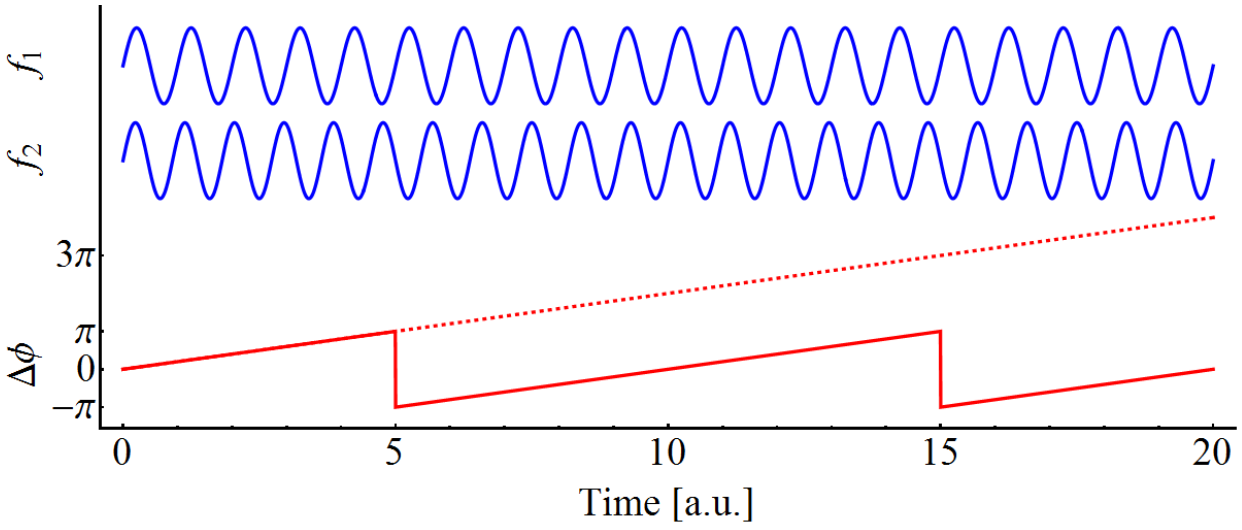
→ RF distribution to compensate time of flight between stations

→ Beam control drives all stations like a single one

Basic building blocks

Measure phase differences

- Two signals at different frequencies ω_1 and ω_2



→ Phase difference, $\Delta\phi$, between both signals changes linearly

→ Ambiguity to distinguish between $\Delta\phi = -\pi, \pi, -3\pi, 3\pi, \ldots$

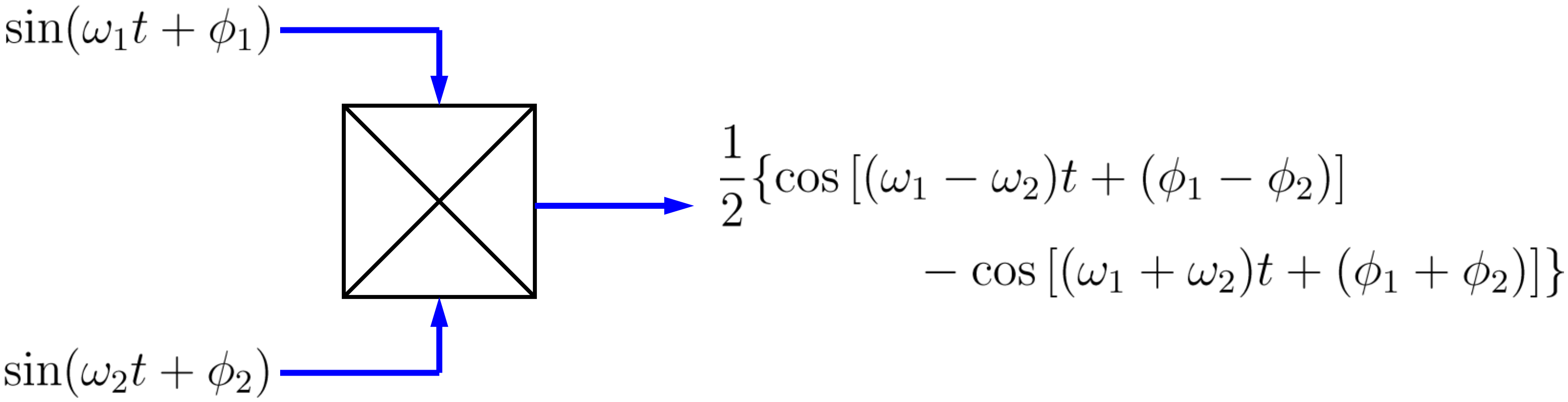
→ Saw-tooth in phase means constant frequency difference

→ Equivalence of frequency and phase

$$\omega = \frac{d\phi}{dt} \quad \leftrightarrow \quad \phi = \int \omega \, dt$$

Mixer or multiplier

- **Example: analogue 4 quadrant multiplier and low pass filter**

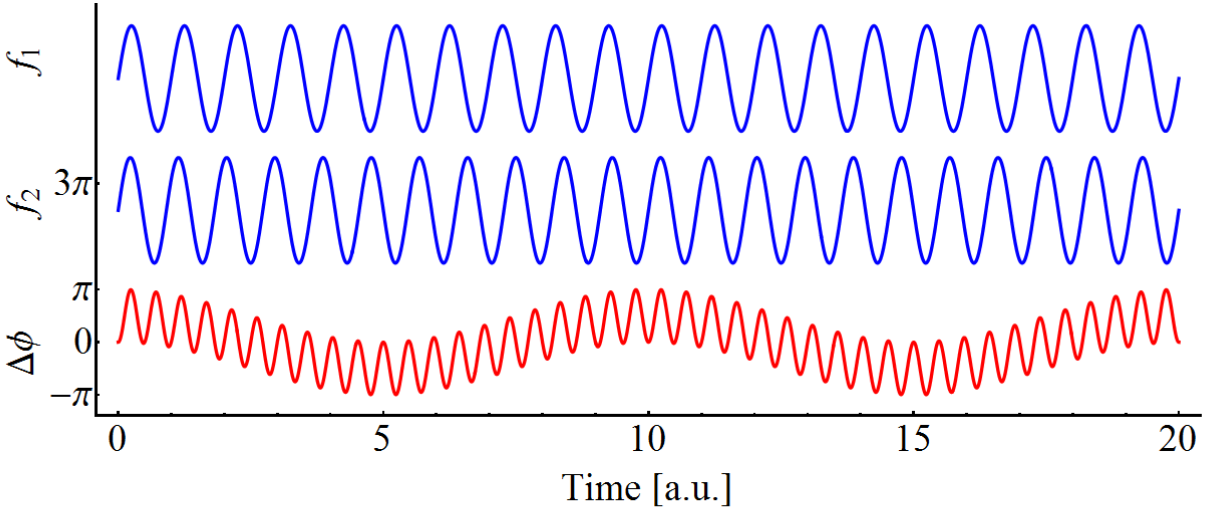


$\sin(\omega_1 t + \phi_1)$

$\sin(\omega_2 t + \phi_2)$

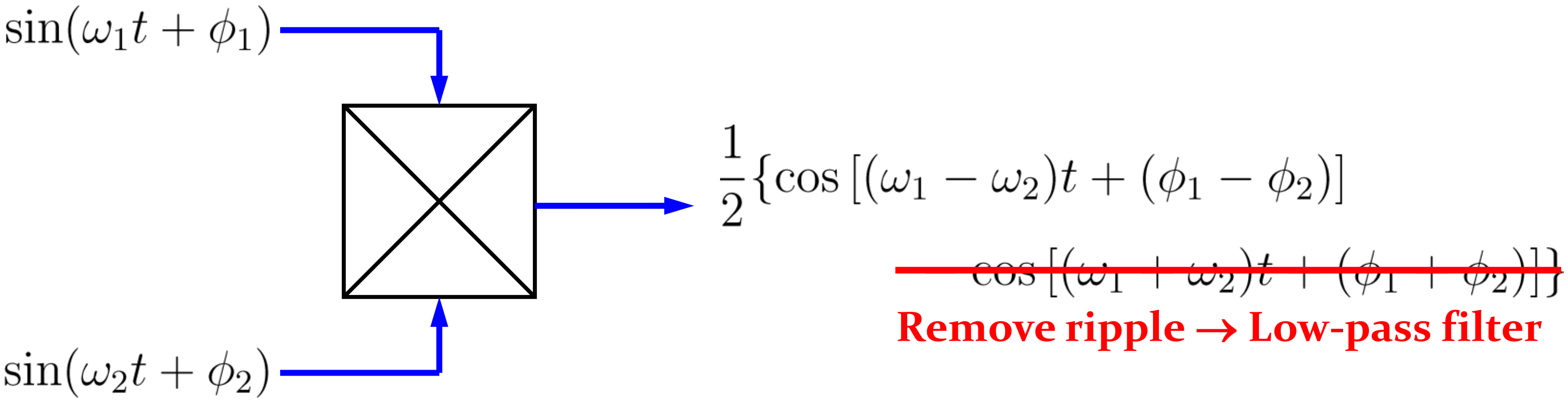
$$\frac{1}{2}\{\cos\left[(\omega_1 - \omega_2)t + (\phi_1 - \phi_2)\right] - \cos\left[(\omega_1 + \omega_2)t + (\phi_1 + \phi_2)\right]\}$$

- **Signals:**



Mixer or multiplier

- **Example: analogue 4 quadrant multiplier and low pass filter**



$\sin(\omega_1 t + \phi_1)$

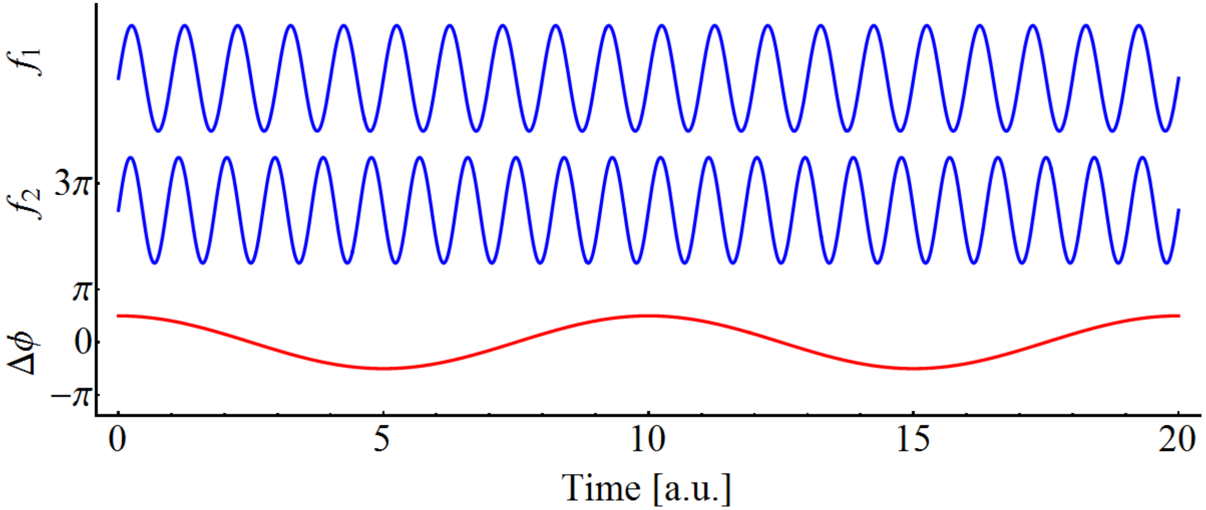
$\sin(\omega_2 t + \phi_2)$

$$\frac{1}{2}\{\cos\left[(\omega_1 - \omega_2)t + (\phi_1 - \phi_2)\right]$$

~~$\cos\left[(\omega_1 + \omega_2)t + (\phi_1 + \phi_2)\right]\}$~~

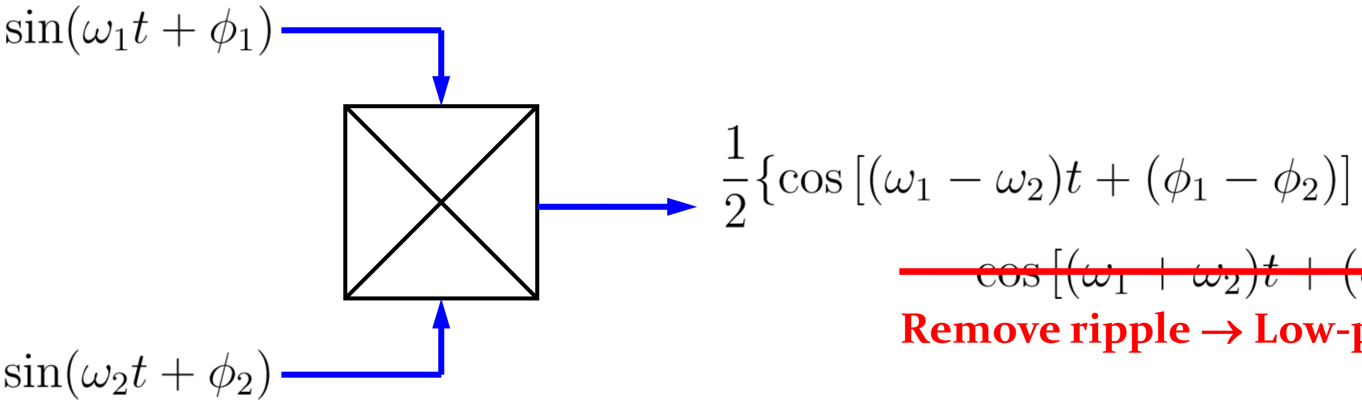
Remove ripple → Low-pass filter

- **Signals:**



How to detect phase differences?

- **Example: analogue 4 quadrant multiplier and low pass filter**



$\sin(\omega_1 t + \phi_1)$

$\sin(\omega_2 t + \phi_2)$

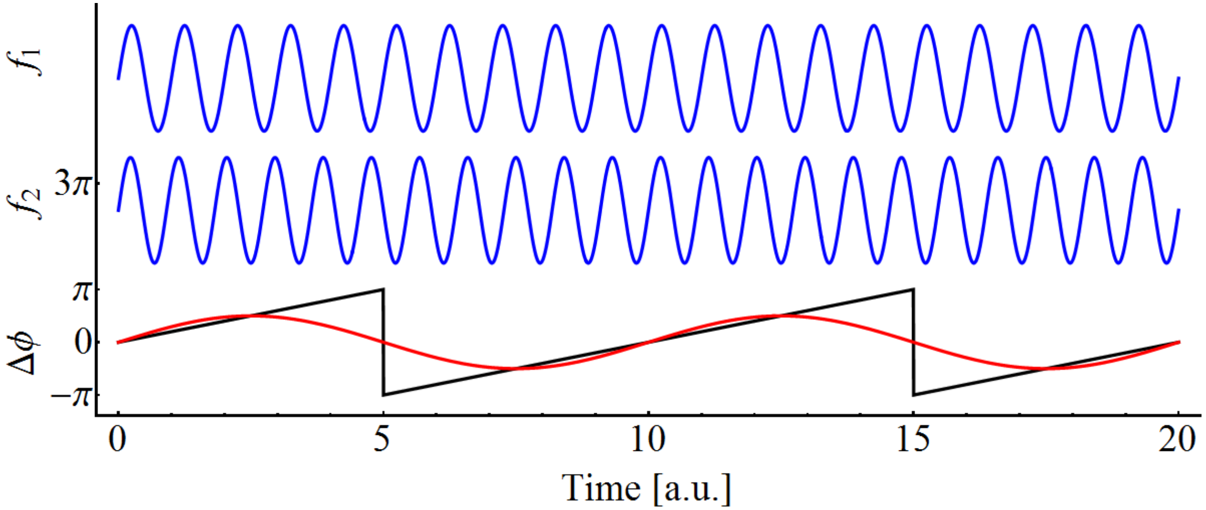
$$\frac{1}{2}\{\cos\left[(\omega_1 - \omega_2)t + (\phi_1 - \phi_2)\right]$$

~~$\cos\left[(\omega_1 + \omega_2)t + (\phi_1 + \phi_2)\right]\}$~~

Remove ripple → Low-pass filter

Relative: arbitrary shift by 90°

- **Signals:**



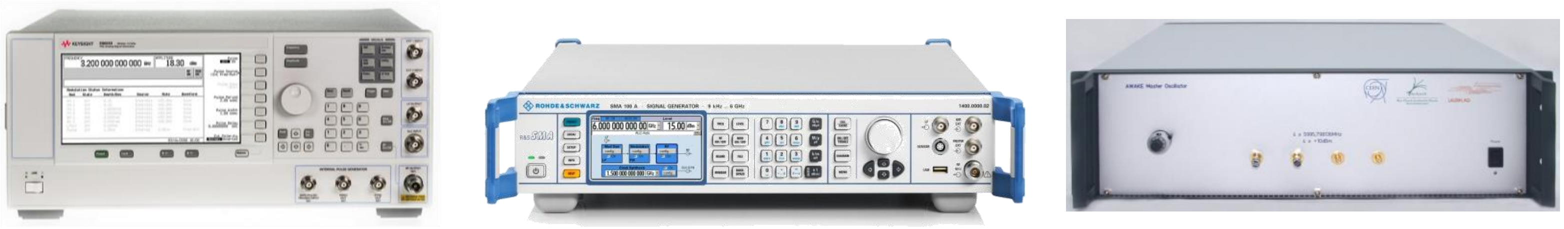
- **Phase discriminator in approximately +/-90° range**

RF sources

RF sources

What finally generates the RF signal to power amplifier and cavity?

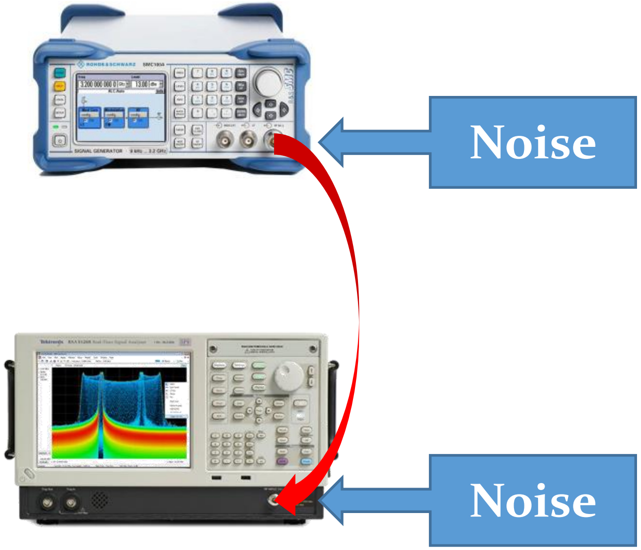
→ Need an RF source!



- Electron accelerators
  - Off-the-shelf high-performance laboratory generators as reference: BESSY SR, CERN CTF3
  - Dedicated commercial fixed-frequency sources with low phase noise: free electron lasers, CERN AWAKE
- Proton accelerators
  - Special sweeping RF sources, controlled by beam-based loops: mostly in-house developments

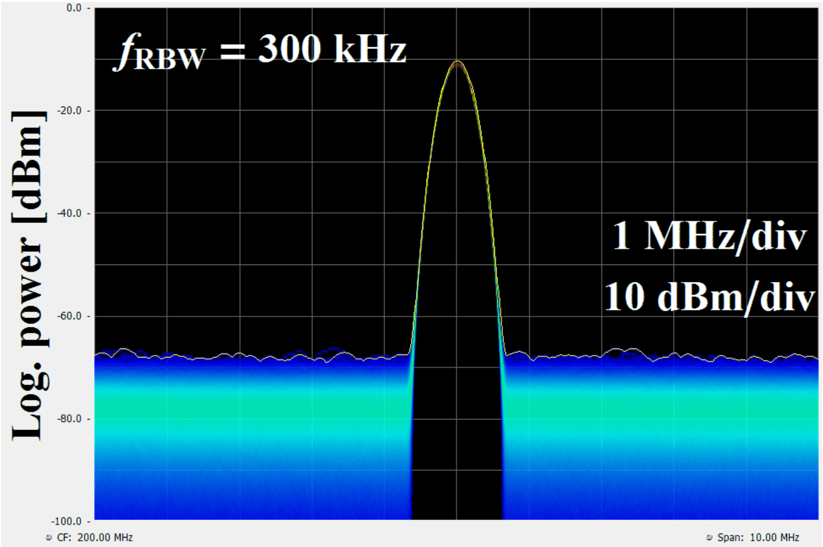
Noisy RF signals

- Degradation of signal quality due to noise
    - Amplitude and/or phase jitter
- What is the difference between a coherent signal and noise?



Noise

Noise



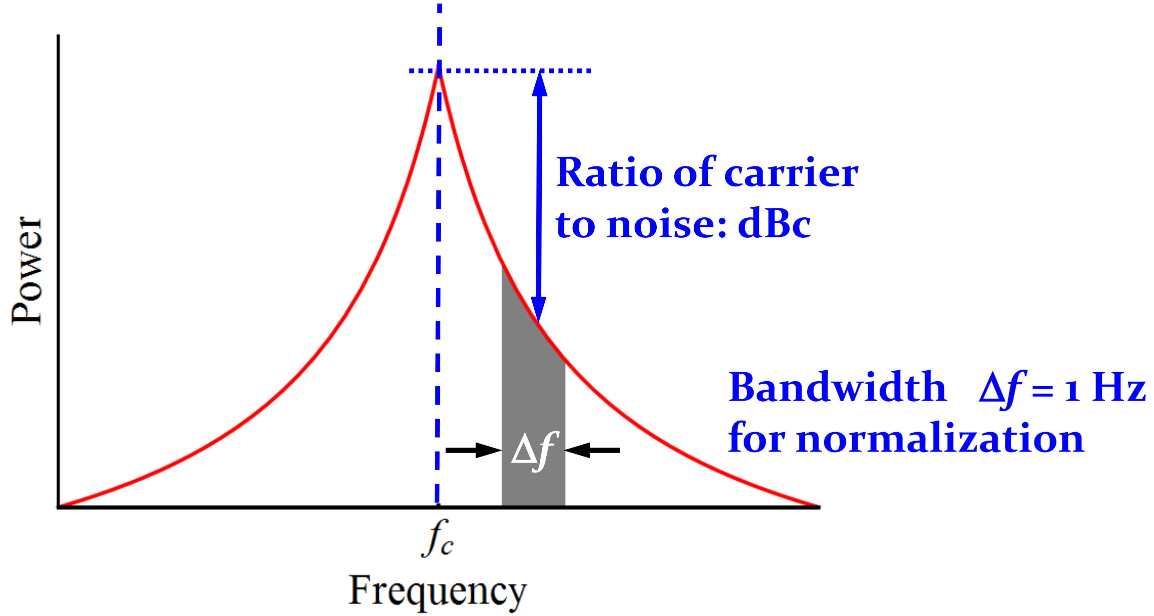
→ Amplitude of coherent, quasi monochromatic signal (at 200 MHz) is independent of observation bandwidth

→ Incoherent noise power (dominated by spectrum analyzer front-end amplifier/mixer) is proportional to bandwidth

→ Thermal noise power $\frac{P}{\Delta f} = k_{\mathrm{B}} T = 1.38 \cdot 10^{-23}\ \mathrm{J/K} \cdot 296\ \mathrm{K} \simeq -174\ \mathrm{dBm/Hz}$

Analysis of phase noise

- **Compare noise power with carrier power as reference**



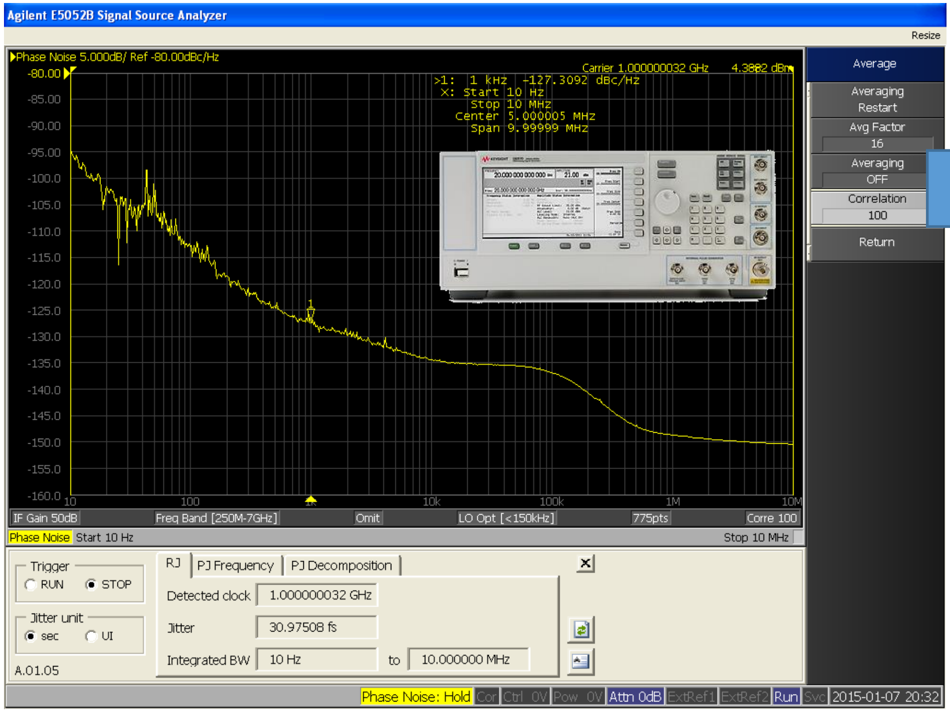
- **Noise power density** $\mathcal{L}(f) = \frac{\text{Power density}}{\text{Carrier power}} \left[\frac{\text{dBc}}{\text{Hz}}\right] = \frac{1}{2} S_\phi(f)$

→ **Its integral is the phase jitter and using** $\Delta t = \frac{\Delta \phi}{2\pi f_\mathrm{c}}$

the jitter in time becomes $\Delta t_\mathrm{rms} = \frac{1}{2\pi f_\mathrm{c}} \sqrt{\int_{f_1}^{f_2} S_\phi(f)\, df}$

Typical phase noise plots

- **Measure phase noise of a synthesized lab generator**



Agilent E5052B Signal Source Analyzer

Phase Noise 5.000dB/ Ref -80.00dBc/Hz

Carrier 1.000000032 GHz 4.3882 dBm

```
>1:  1 kHz  -127.3092 dBc/Hz
 X: Start 10 Hz
    Stop 10 MHz
  Center 5.000005 MHz
    Span 9.99999 MHz
```

Average

Averaging Restart

Avg Factor 16

Averaging OFF

Correlation 100

IF Gain 50dB | Freq Band [250M-7GHz] | Omit | LO Opt [<150kHz] | 775pts | Corre 100

Phase Noise Start 10 Hz Stop 10 MHz

Trigger ○ RUN ● STOP

Jitter unit ● sec ○ UI

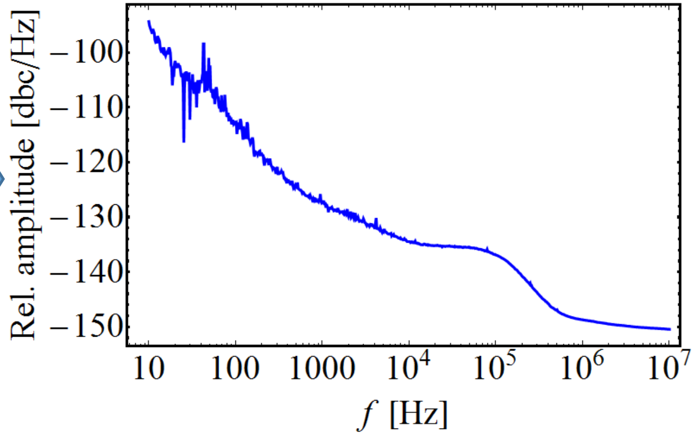
A.01.05

Detected clock 1.000000032 GHz

Jitter 30.97508 fs

Integrated BW 10 Hz to 10.000000 MHz

Phase Noise: Hold | Cor | Ctrl 0V | Pow 0V | Attn 0dB | ExtRef1 | ExtRef2 | Run | Svc | 2015-01-07 20:32



Frequency range	Δt_{rms} [fs]
10…100 Hz	12.4
100 Hz …1 kHz	5.4
1…10 kHz	5.4
10…100 kHz	11.1
100 kHz…1 MHz	13.0
Total	31.0

→ **Note: jitter values can be added as square root of quadratic sum**

$$\Delta t_{\mathrm{rms}} = \sqrt{\Delta t^2_{\mathrm{rms},1} + \Delta t^2_{\mathrm{rms},2} + \ldots}$$

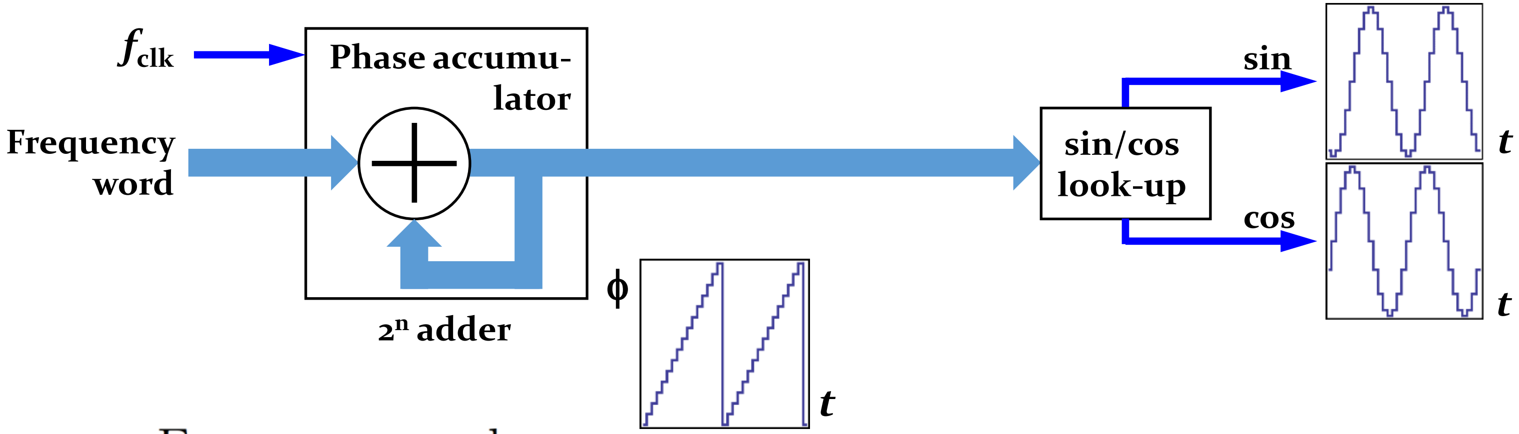
→ **Convenient split to relevant ranges**

Variable frequency: direct digital synthesis

- **Generate (almost) any frequency starting from a given clock frequency, f_{clk}**
- **Digitally programmable in frequency**

→ **RF signal**

➡ **Digital data**



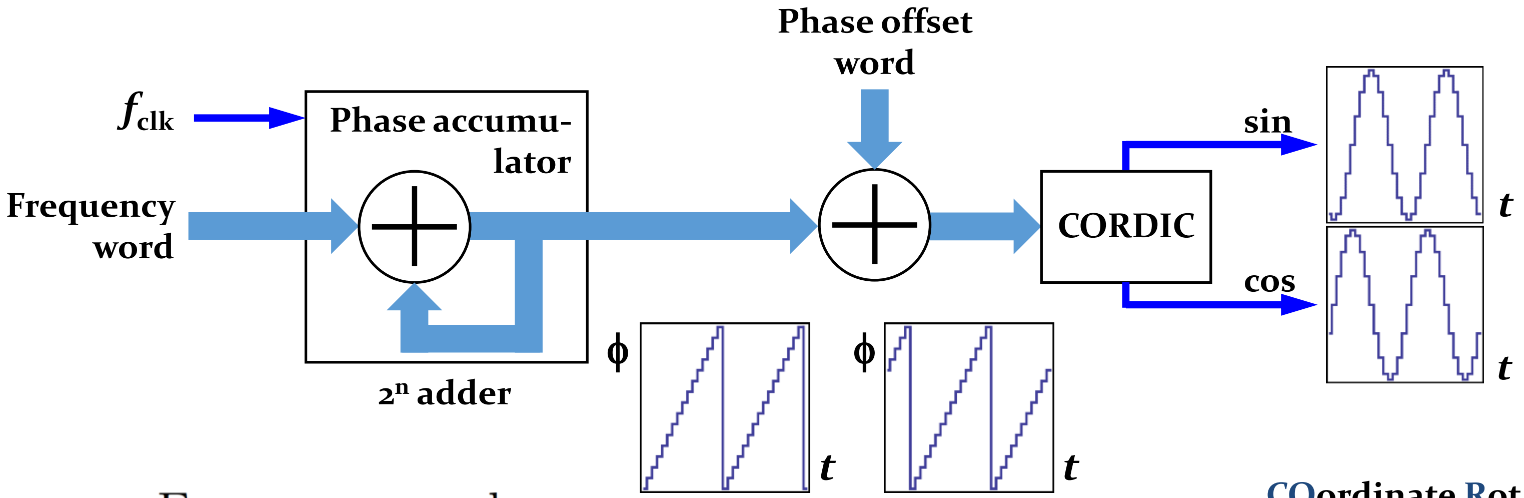
$$f_{\mathrm{out}} = \frac{\text{Frequency word}}{2^n} \cdot f_{\mathrm{clk}}$$

Variable frequency: direct digital synthesis

- Generate (almost) any frequency starting from a given clock frequency, f_{clk}
- Digitally programmable in frequency and phase

→ RF signal

⇒ Digital data



COordinate Rotation DIgital Computer

$$f_{\text{out}} = \frac{\text{Frequency word}}{2^n} \cdot f_{\text{clk}}$$

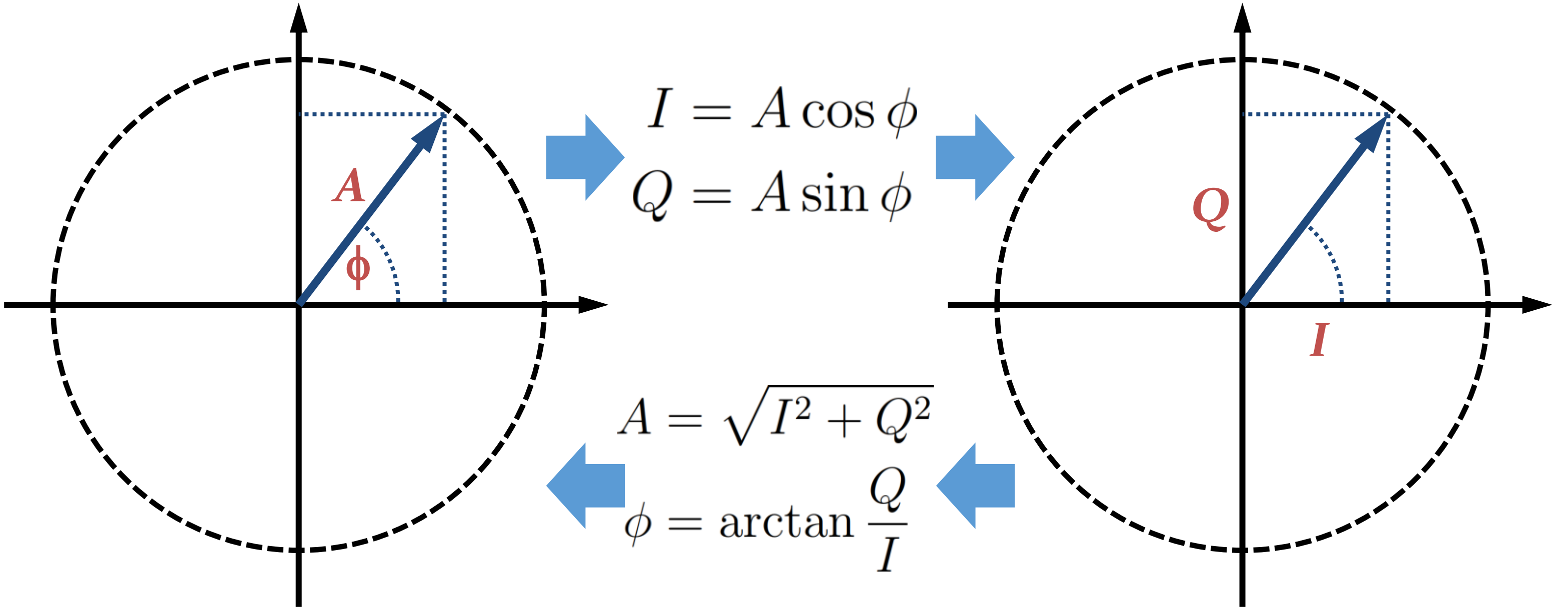
→ Two output signals with ideal 90° phase shift

→ Output signals are digital data streams

Receivers

I/Q representation of signals

- **Any signal can be represented by amplitude *A* and phase ϕ**



$$I = A\cos\phi$$
$$Q = A\sin\phi$$

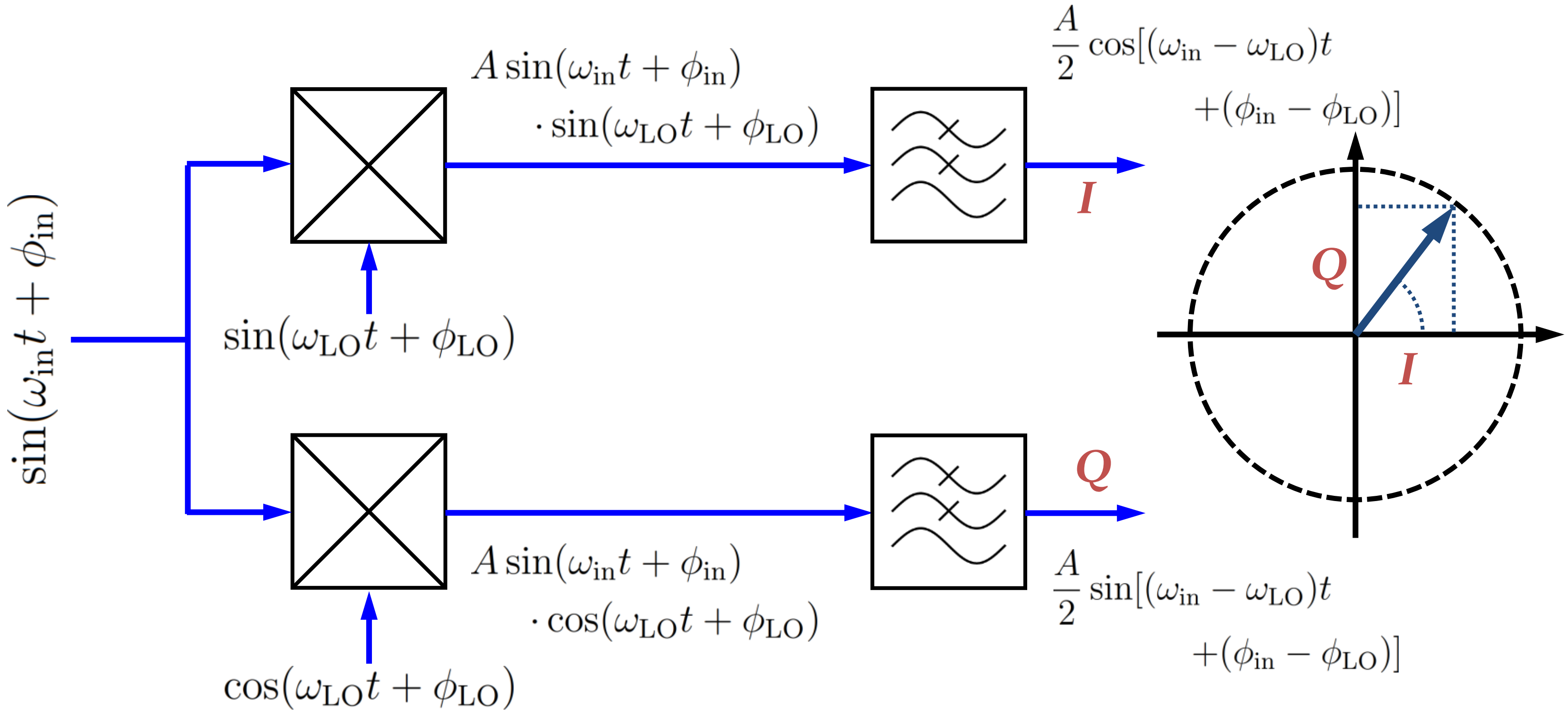
$$A = \sqrt{I^2 + Q^2}$$
$$\phi = \arctan\frac{Q}{I}$$

→ **In phase, *I* and quadrature, *Q* describe the same signal**

→ **Avoids phase discontinuities at 0, 2π, …**

Signal receivers

- **Radio with listens to beam or cavity signals**
- **Listens to amplitude and phase**



$\sin(\omega_{\text{in}} t + \phi_{\text{in}})$

$A \sin(\omega_{\text{in}} t + \phi_{\text{in}}) \cdot \sin(\omega_{\text{LO}} t + \phi_{\text{LO}})$

$\sin(\omega_{\text{LO}} t + \phi_{\text{LO}})$

$\frac{A}{2} \cos[(\omega_{\text{in}} - \omega_{\text{LO}})t + (\phi_{\text{in}} - \phi_{\text{LO}})]$

$A \sin(\omega_{\text{in}} t + \phi_{\text{in}}) \cdot \cos(\omega_{\text{LO}} t + \phi_{\text{LO}})$

$\cos(\omega_{\text{LO}} t + \phi_{\text{LO}})$

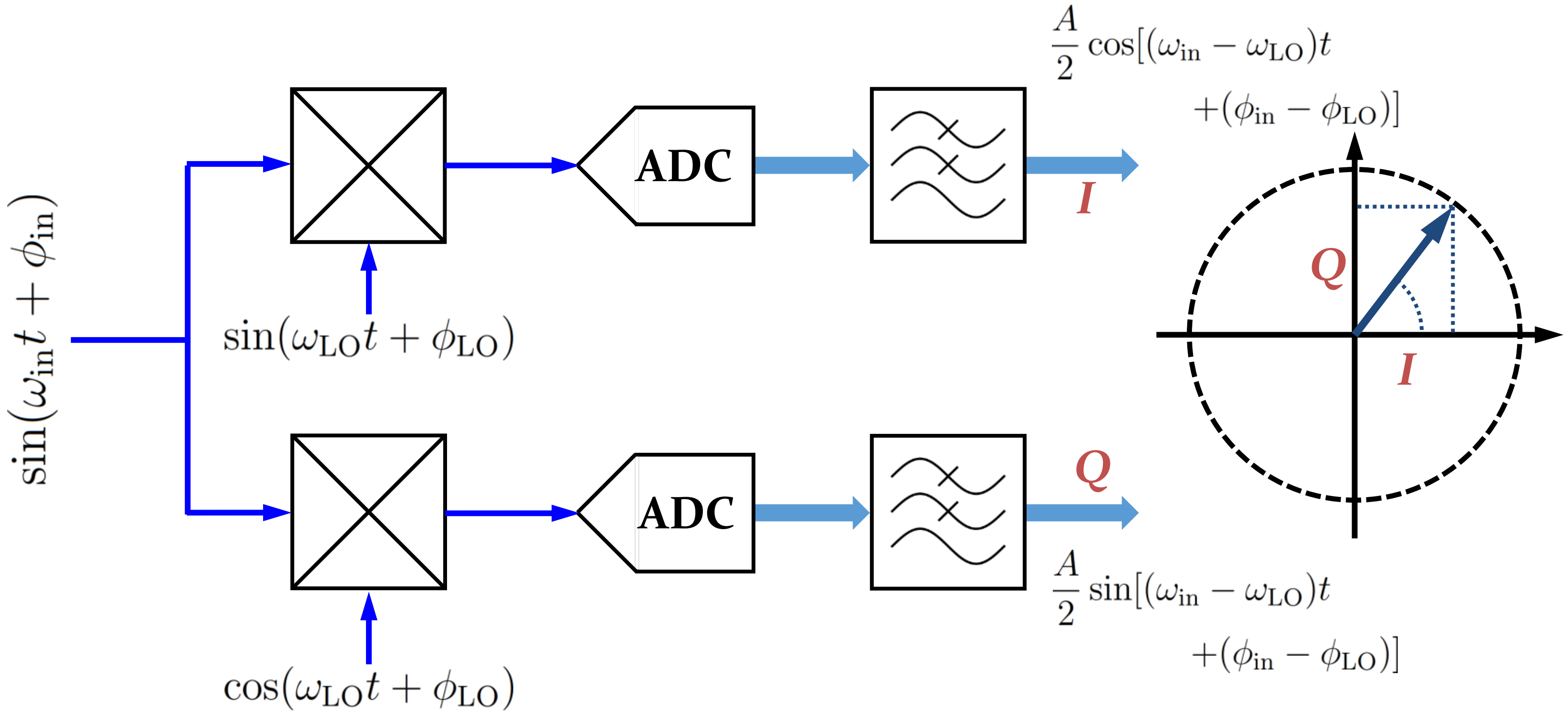
$\frac{A}{2} \sin[(\omega_{\text{in}} - \omega_{\text{LO}})t + (\phi_{\text{in}} - \phi_{\text{LO}})]$

→ **With $\omega_{\text{in}} \approx \omega_{\text{LO}}$ input signal is down-converted to base-band**

→ **Resulting I/Q vector rotates slowly with $\omega_{\text{in}} - \omega_{\text{LO}}$**

Digital receivers

- **No conceptual difference between analogue and digital**
- **Digitization can be performed at any level**

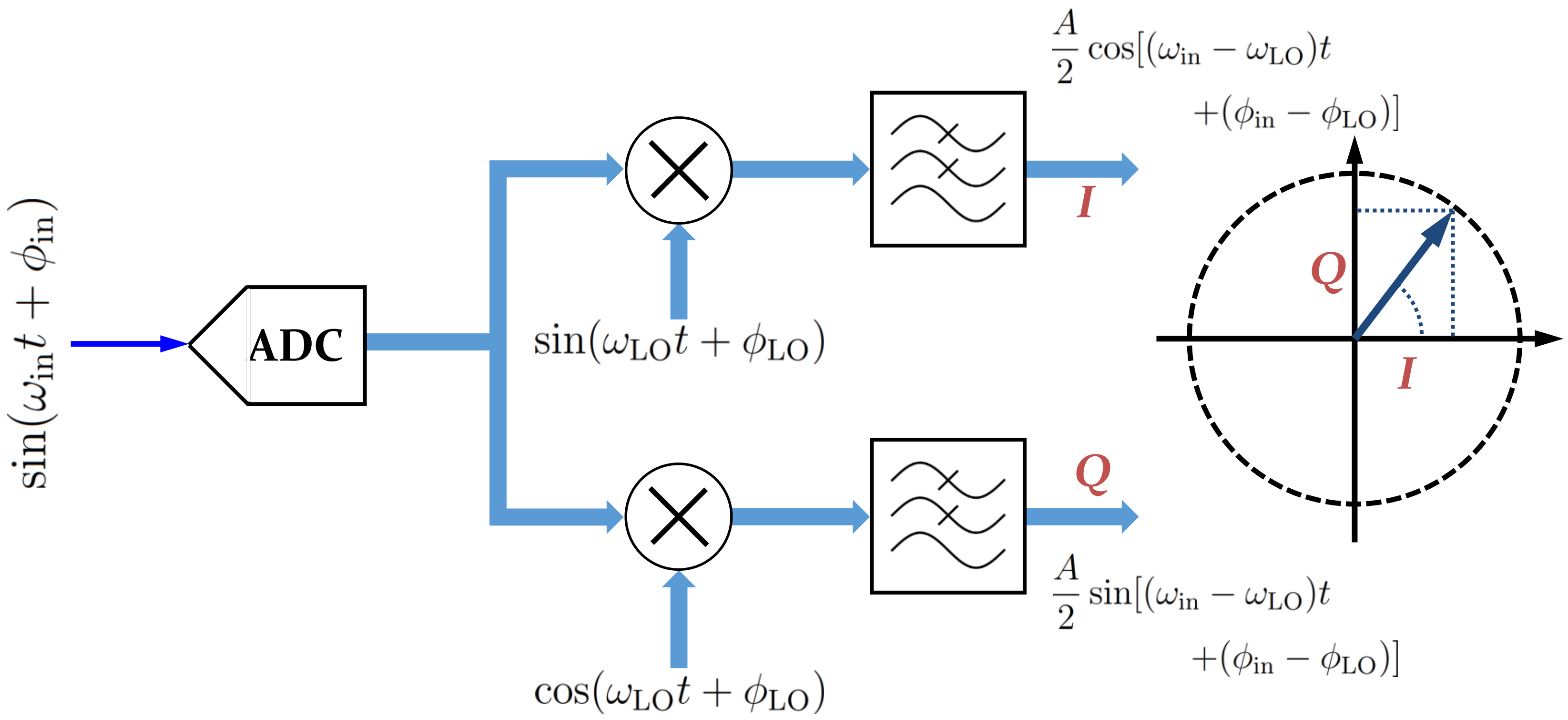


→ **Analog down-conversion of I and Q, then digital processing**

→ **High input frequencies beyond ADC sampling rates**

Digital receivers

- **No conceptual difference between analogue and digital**
- **Digitization can be performed at any level**



→ **Analogue mixers become digital multipliers**

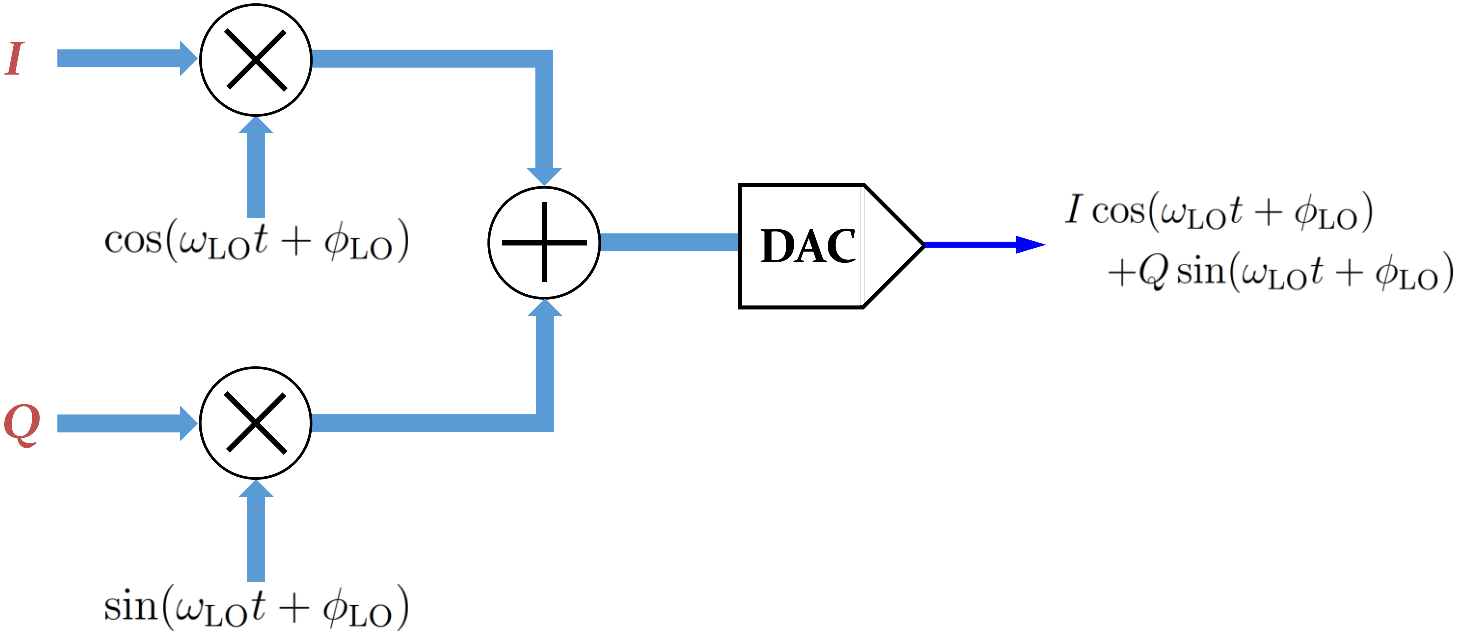
→ **All digital receiver**

→ **Theoretically perfect I/Q symmetry**

Vector modulator

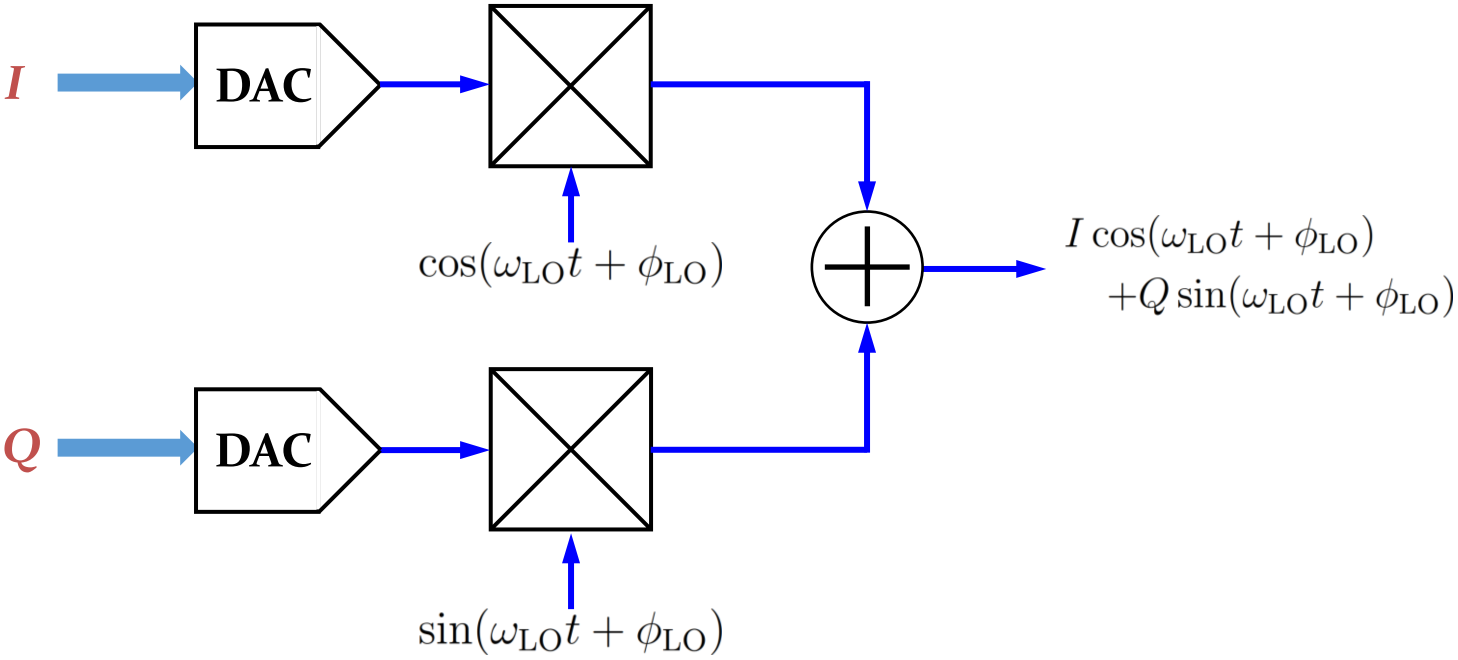
Invers receiver: vector modulator

- **Convert I/Q data into modulated RF signal**



Inverse receiver: vector modulator

- **Convert I/Q data into modulated RF signal**



→ **Perfect I/Q symmetry difficult to achieve**

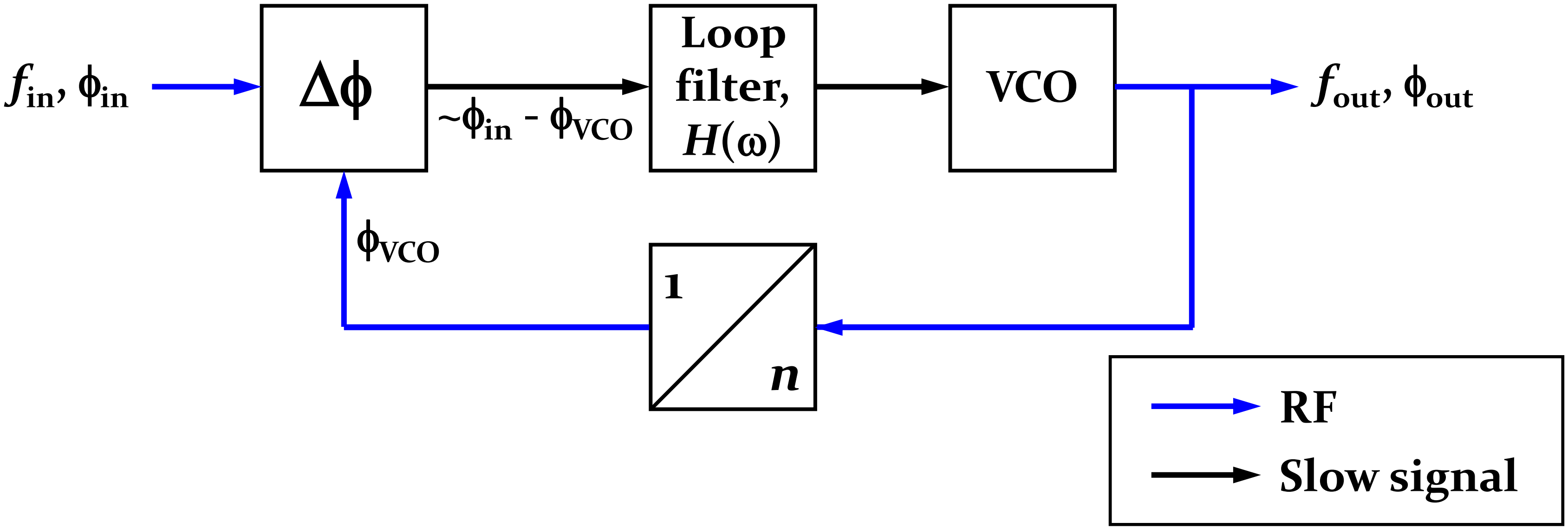
→ **Up-conversion of digital signal to a high RF frequency**

Beam phase loop

Electronic phase-locked loop

- **Frequency re-generation and multiplication**
- **Voltage controlled oscillator (VCO) locked in phase to input**

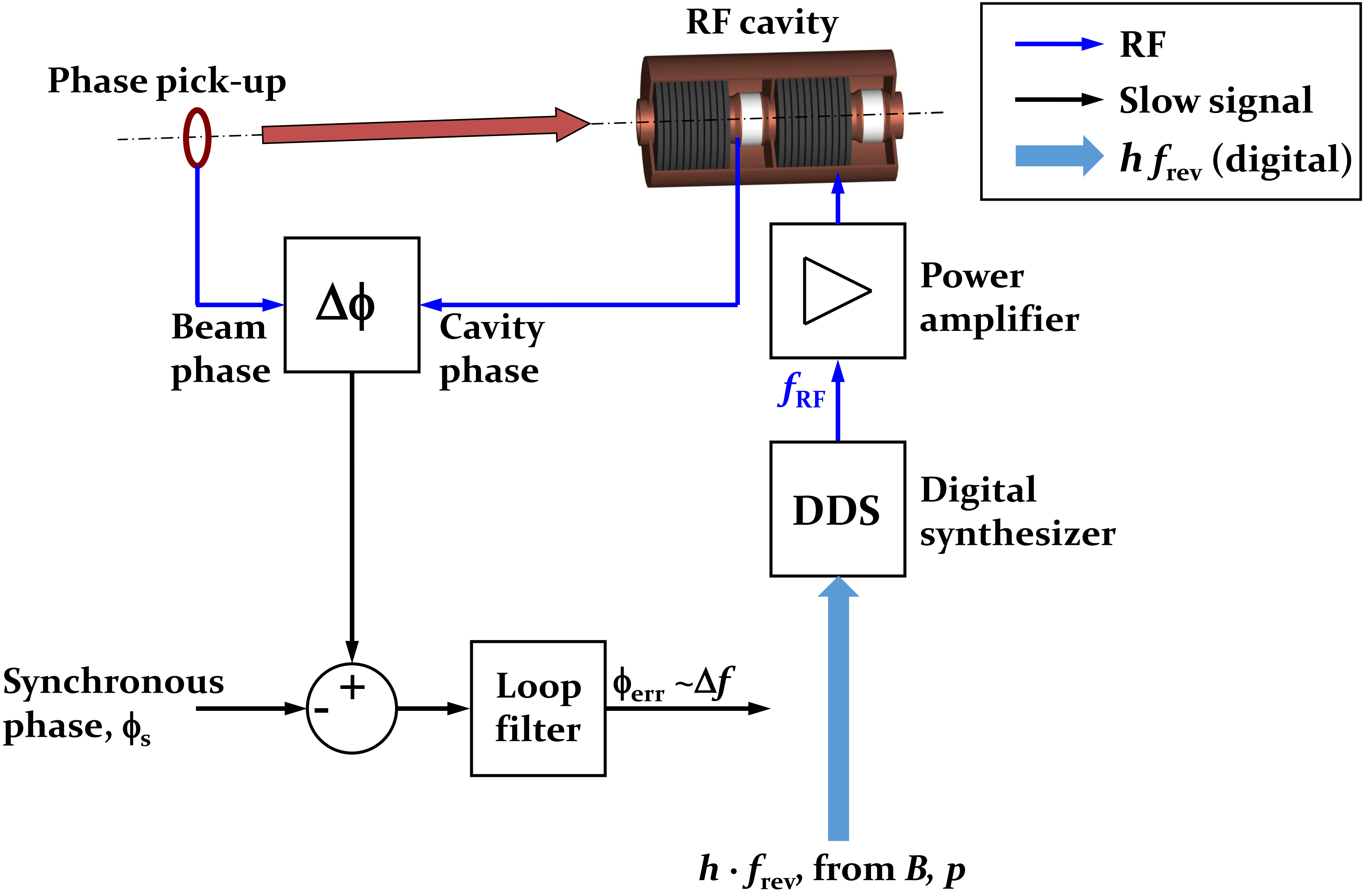
$$\omega_{\mathrm{VCO}} = 2\pi f_{\mathrm{VCO}} = \frac{d\phi}{dt} = K_{\mathrm{VCO}} V_{\mathrm{in}}$$



→ **Fixed phase relationship:** $\phi_{out}/n - \phi_{in} = \text{const.}$

→ **Optional divider:** $f_{out} = n \cdot f_{in}$

Beam phase loop



RF cavity

Phase pick-up

RF

Slow signal

$h\,f_{rev}$ (digital)

$\Delta\phi$

Beam phase

Cavity phase

Power amplifier

f_{RF}

DDS

Digital synthesizer

Synchronous phase, ϕ_s

+

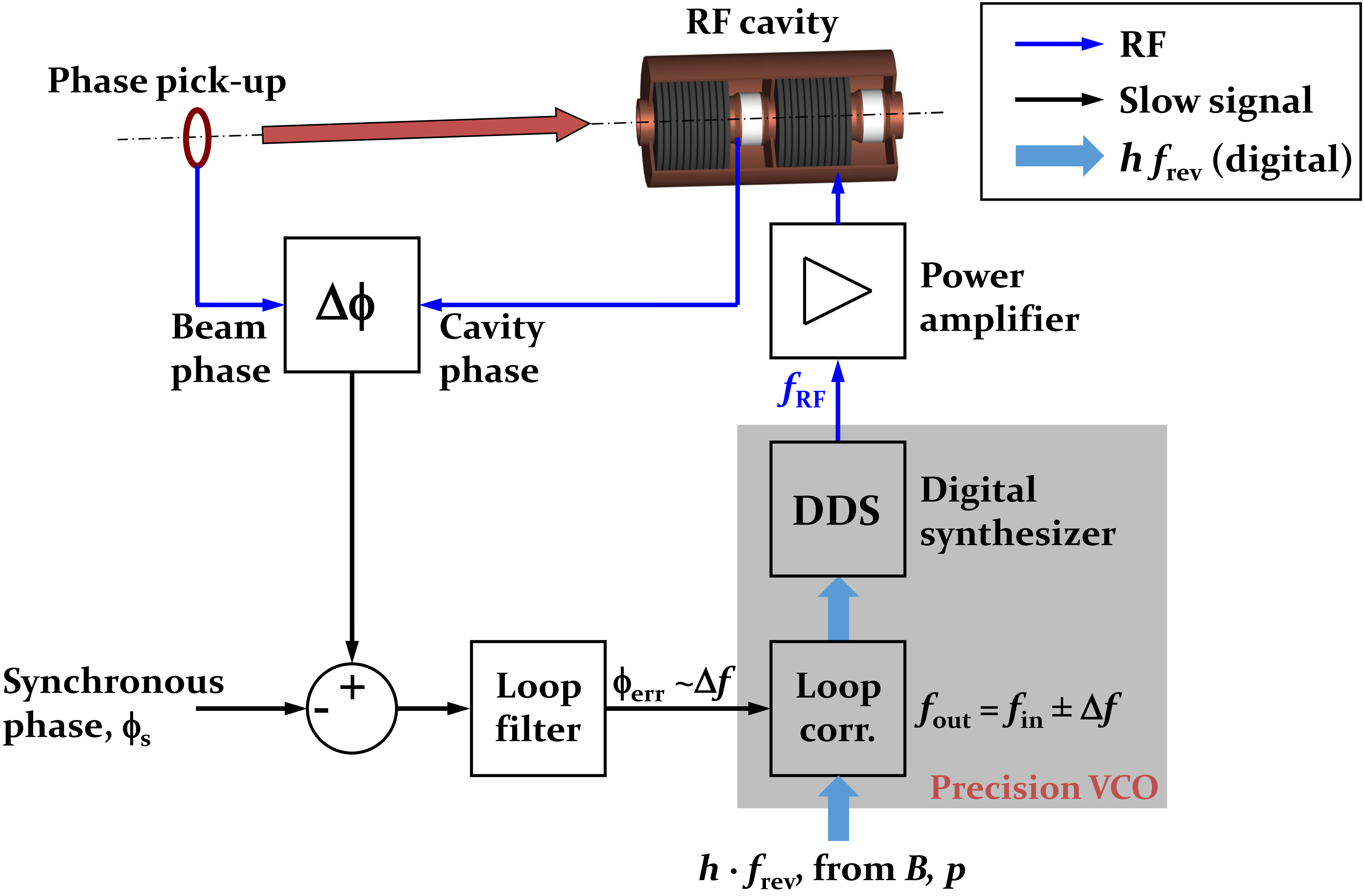
-

Loop filter

$\phi_{err} \sim \Delta f$

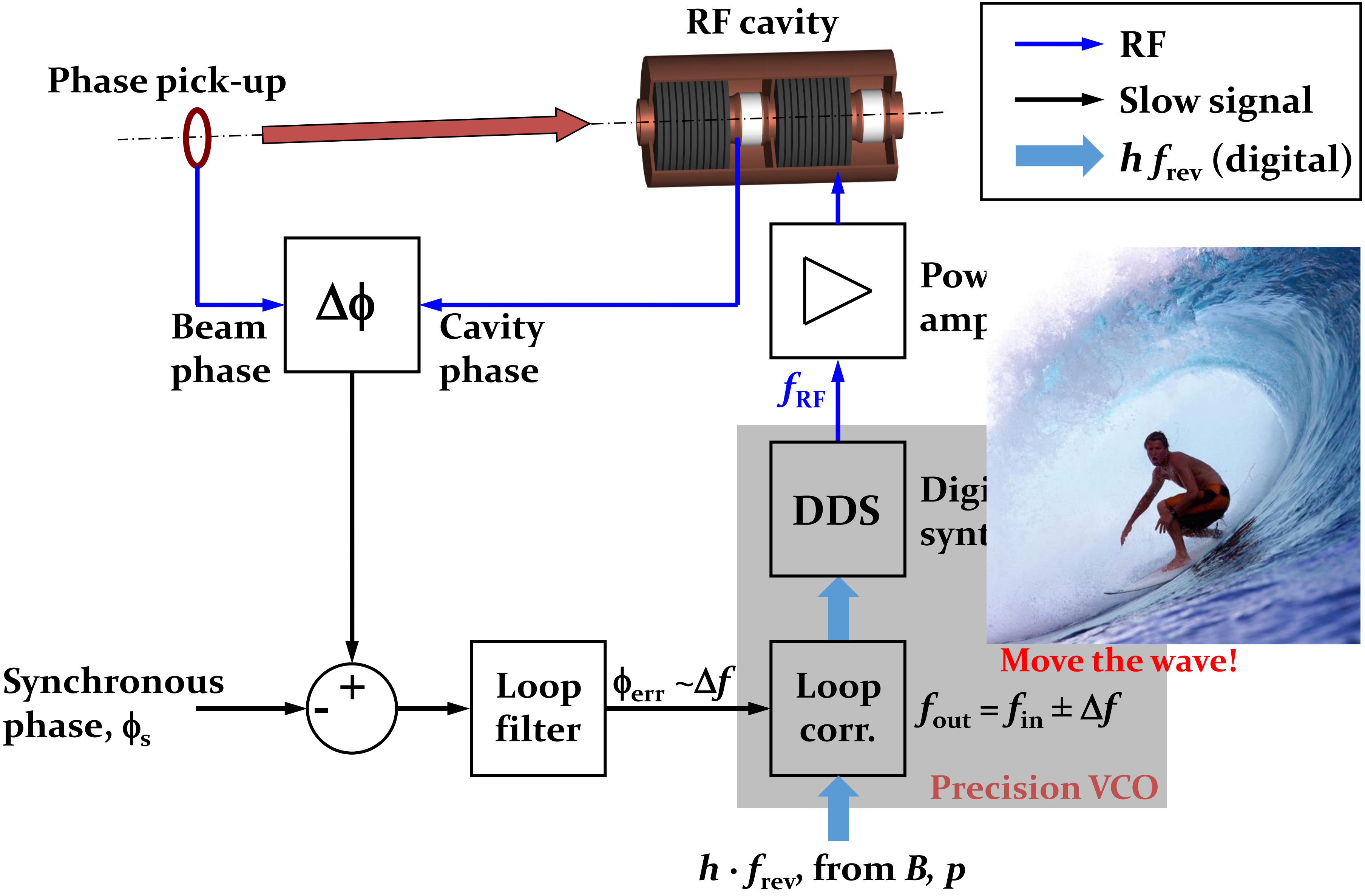
$h \cdot f_{rev}$, from B, p

Beam phase loop



→ **Phase-locked loop with beam phase as reference for RF system**

Beam phase loop



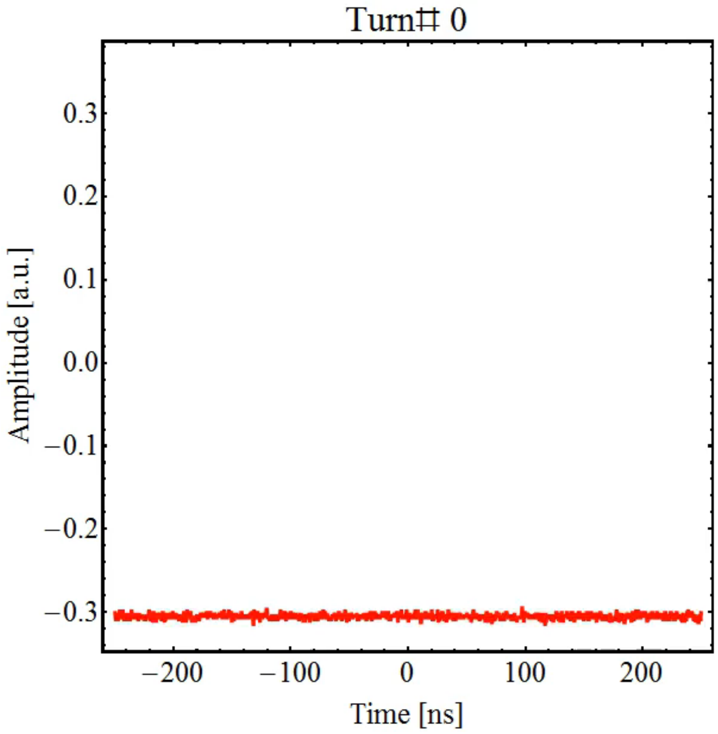
Move the wave!

→ Fast control of RF frequency to cavities, but no slow corrections

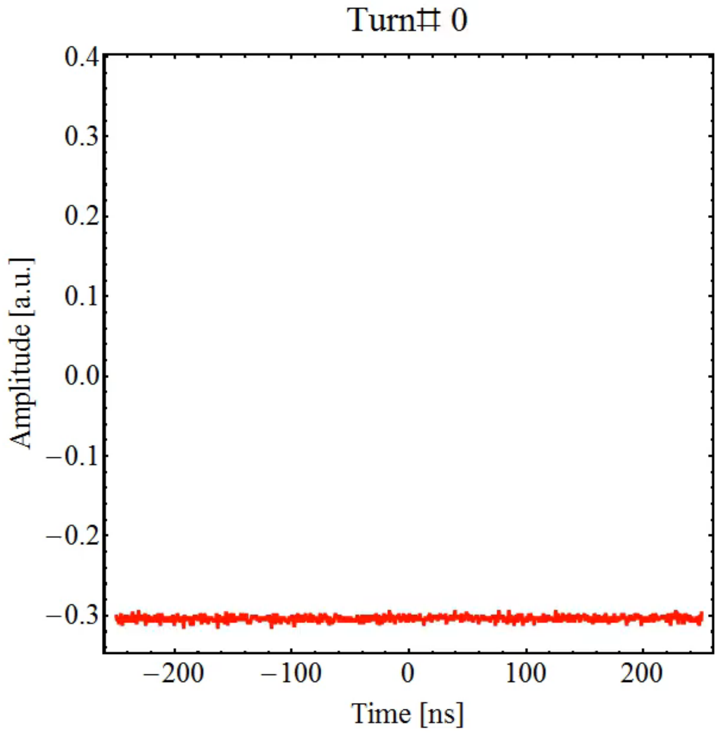
Effect of beam phase loop at injection

- **Example: Injection of a bunch from PS Booster into PS**

90° error, phase loop off



90° error, phase loop on

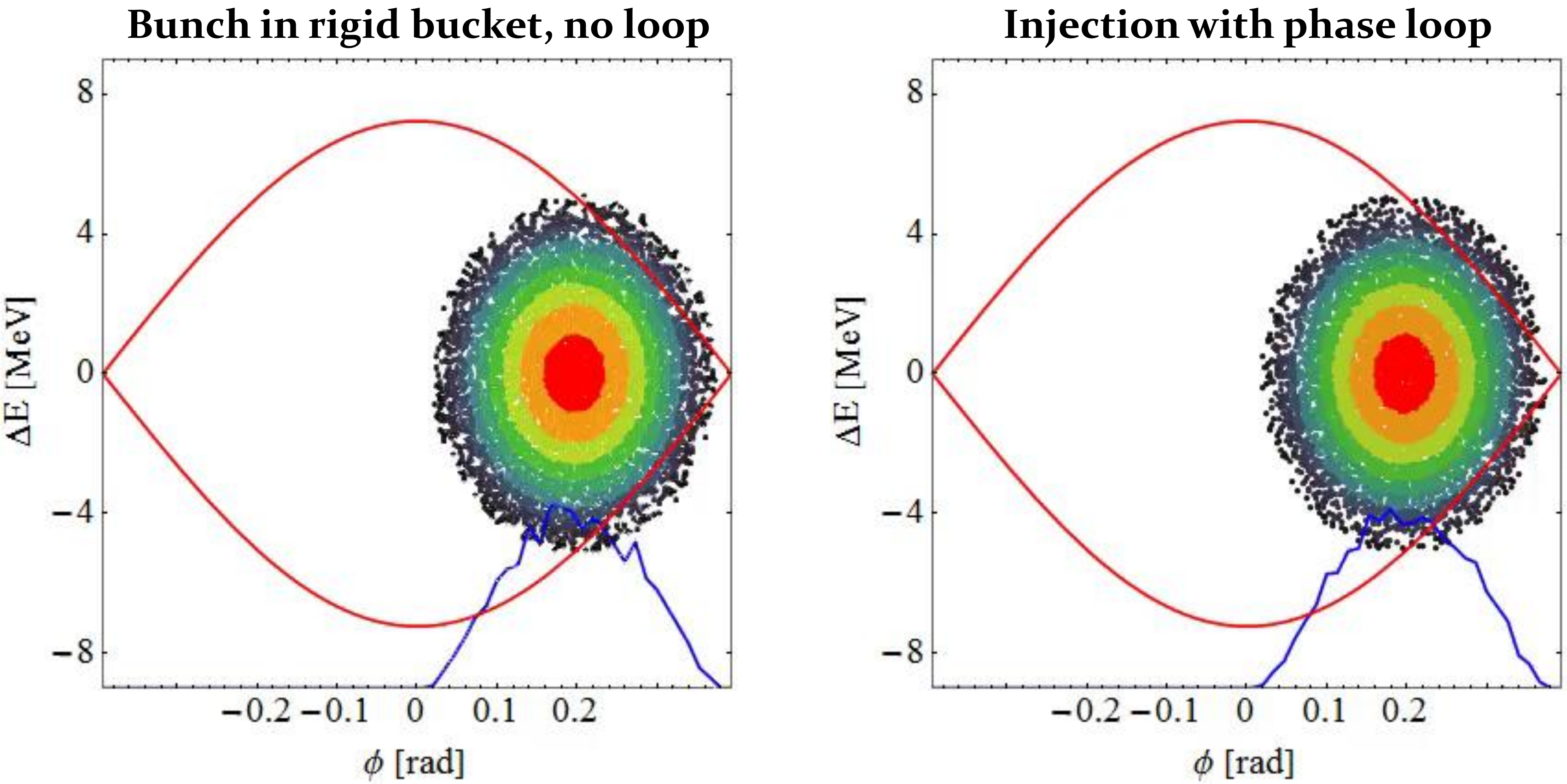


→ **Essential in hadron accelerators to keep RF locked to beam**

→ **How does this look like in longitudinal phase space?**

Effect of beam phase loop at injection

→ **Essential in hadron accelerators to keep RF locked to beam**


Bunch in rigid bucket, no loop

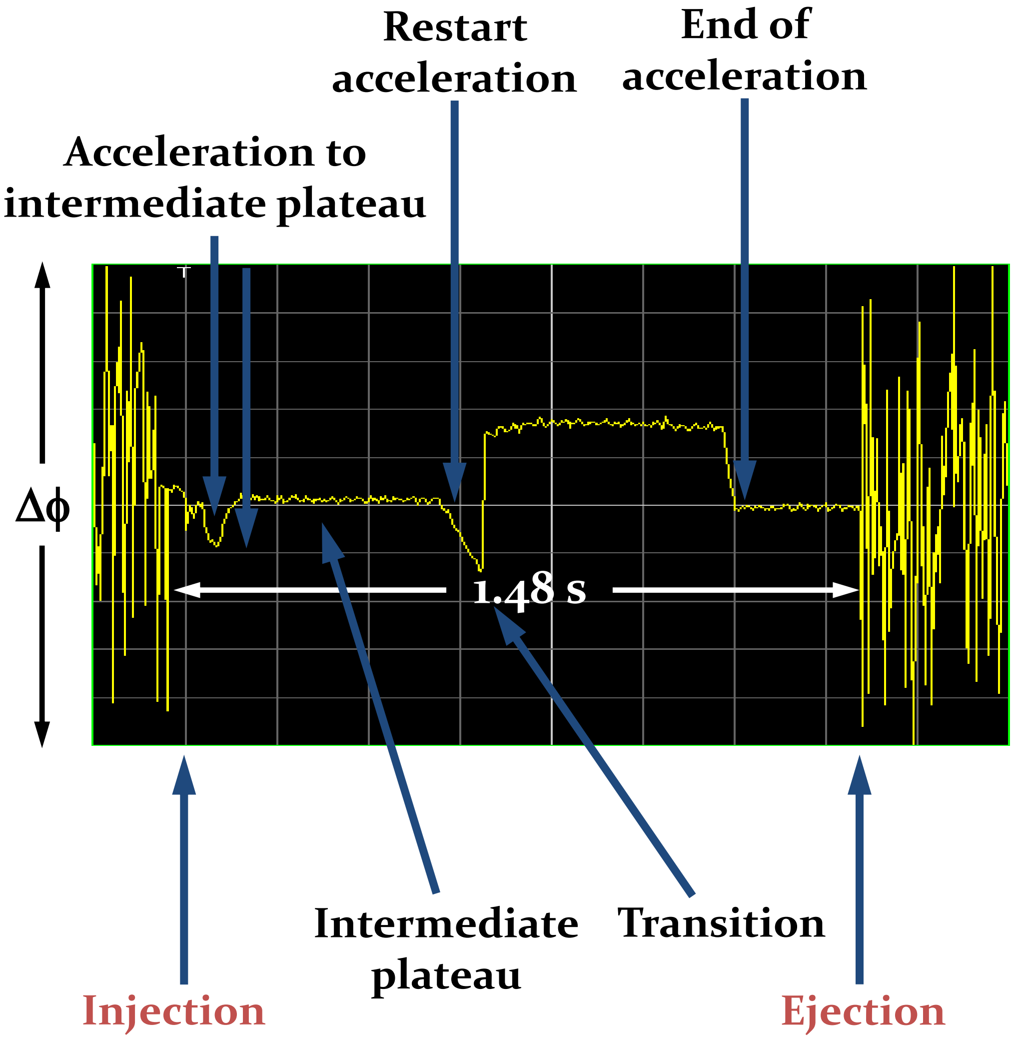
Injection with phase loop

→ **Even large transients (injection, transition) are controlled**

→ **Only minor longitudinal perturbation**

Beam phase loop during acceleration

- → What happens with phase loop during acceleration?
- → During plateaus the phase between RF and beam is either 0° or 180°
- → Fast phase changes well handled, but need slow frequency correction
- → Radial or synchronization loop



Restart acceleration

End of acceleration

Acceleration to intermediate plateau

Injection

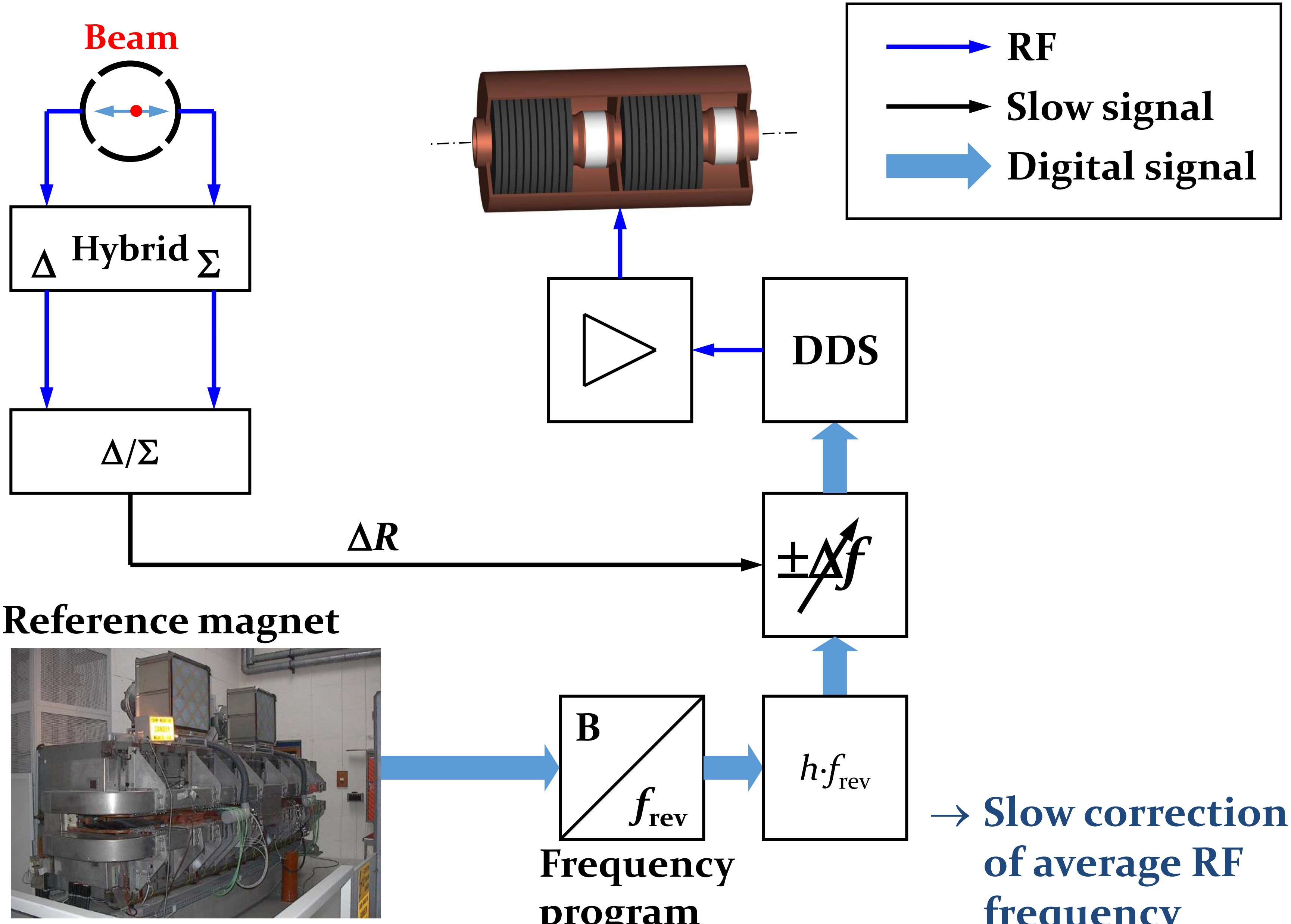
Intermediate plateau

Transition

Ejection

Radial loop

Radial loop



→ Slow correction of average RF frequency

Radial loop

- **Slow correction of RF frequency to keep beam centred**

Why needed at all with arbitrary precision synthesizers driving the RF system?

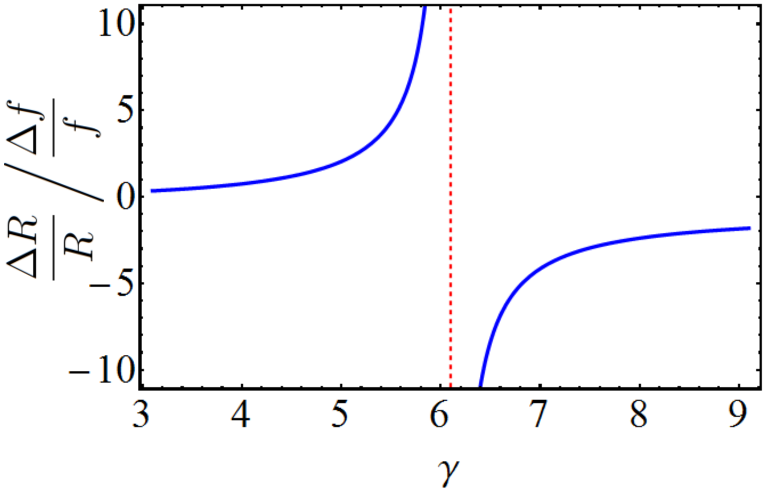
→ **At transition energy**

  → **Longer path of higher energy particle compensated by higher velocity**

  → **No revolution frequency change for energy offset**

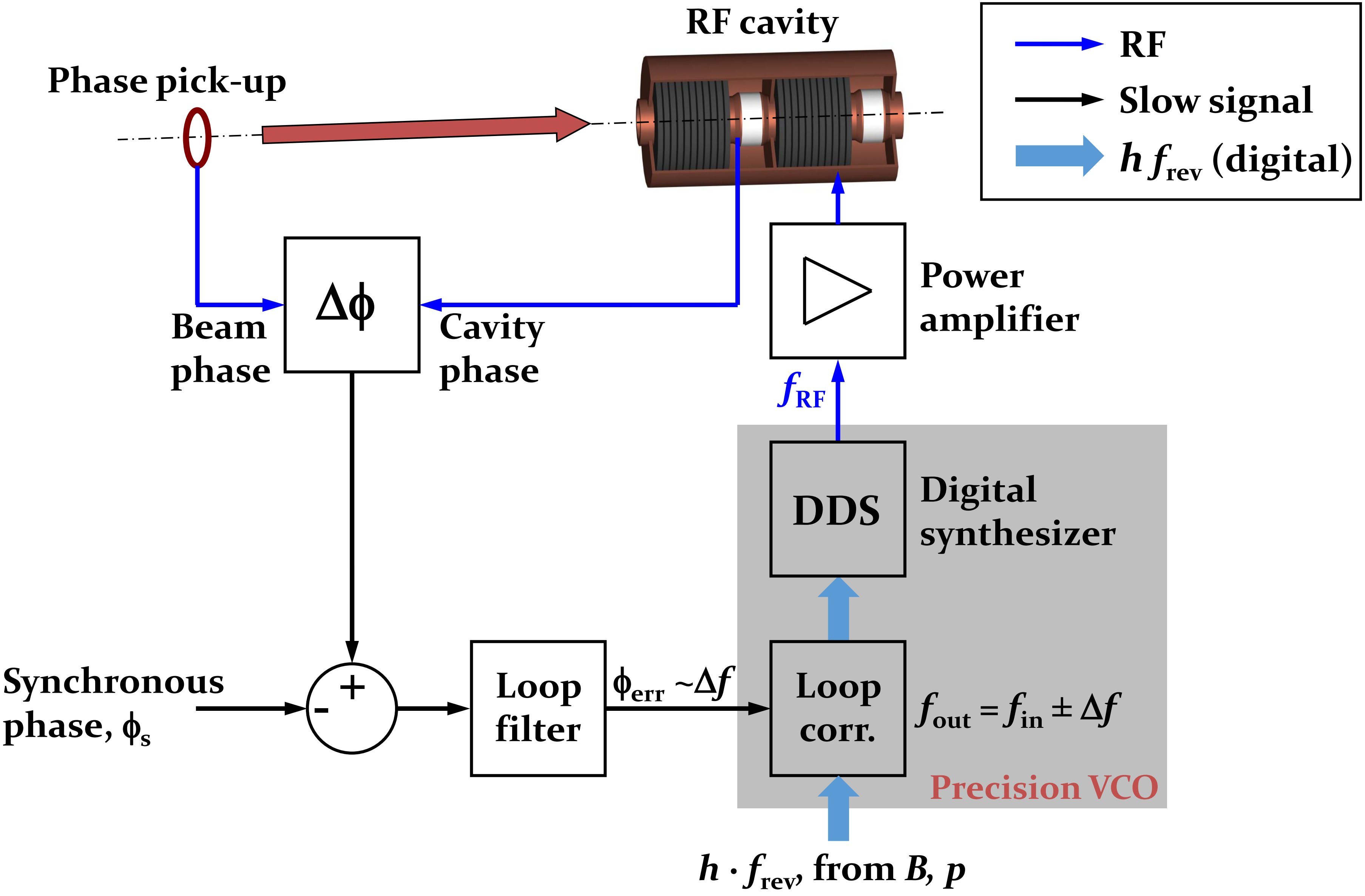
$$\boxed{\frac{\Delta R}{R} = \frac{\gamma^2}{\gamma_{\mathrm{tr}}^2 - \gamma^2}\frac{\Delta f}{f}}$$

→ **Need beam-based frequency correction**



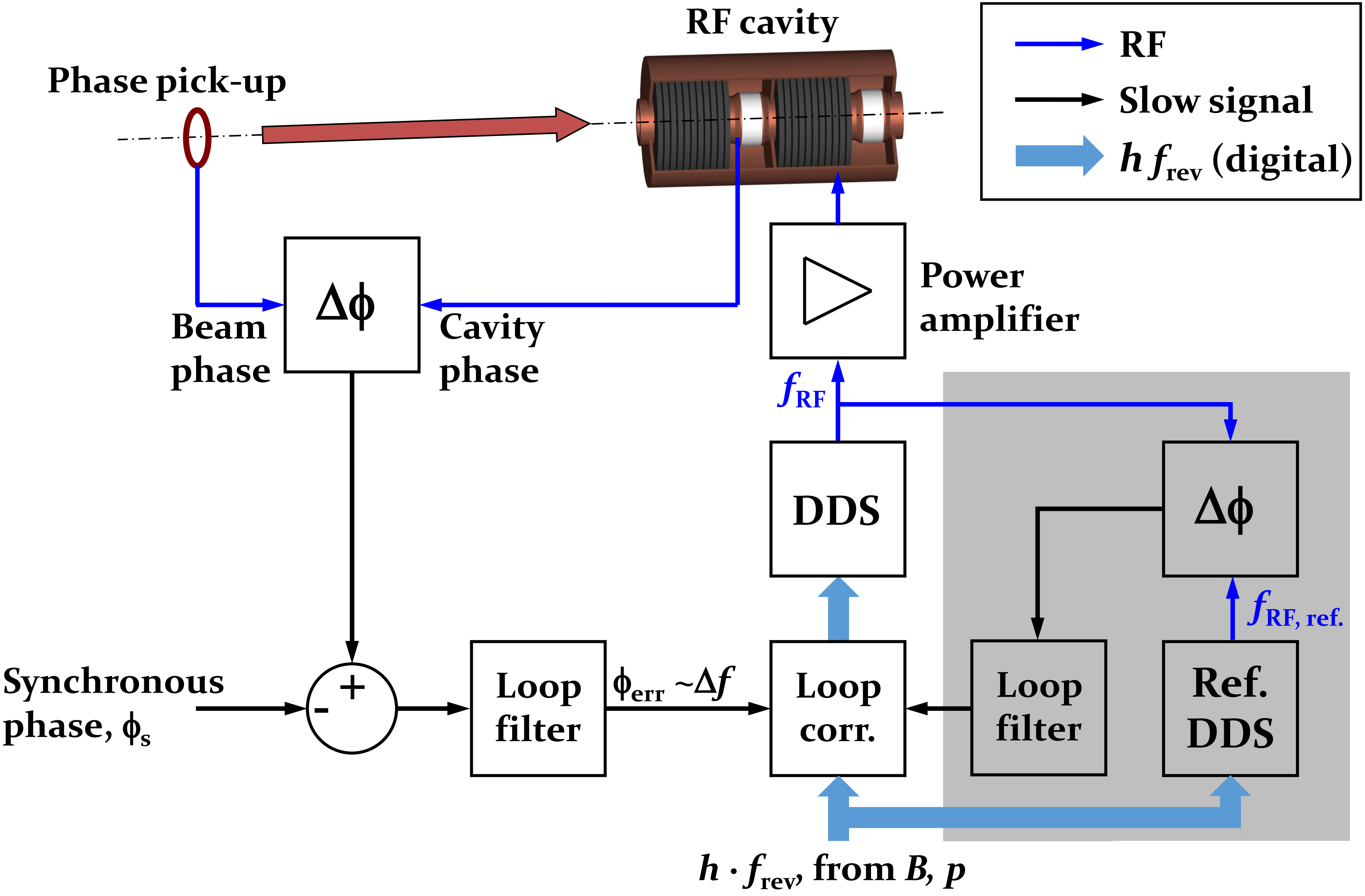
Synchro(nization) loop

Beam phase loop



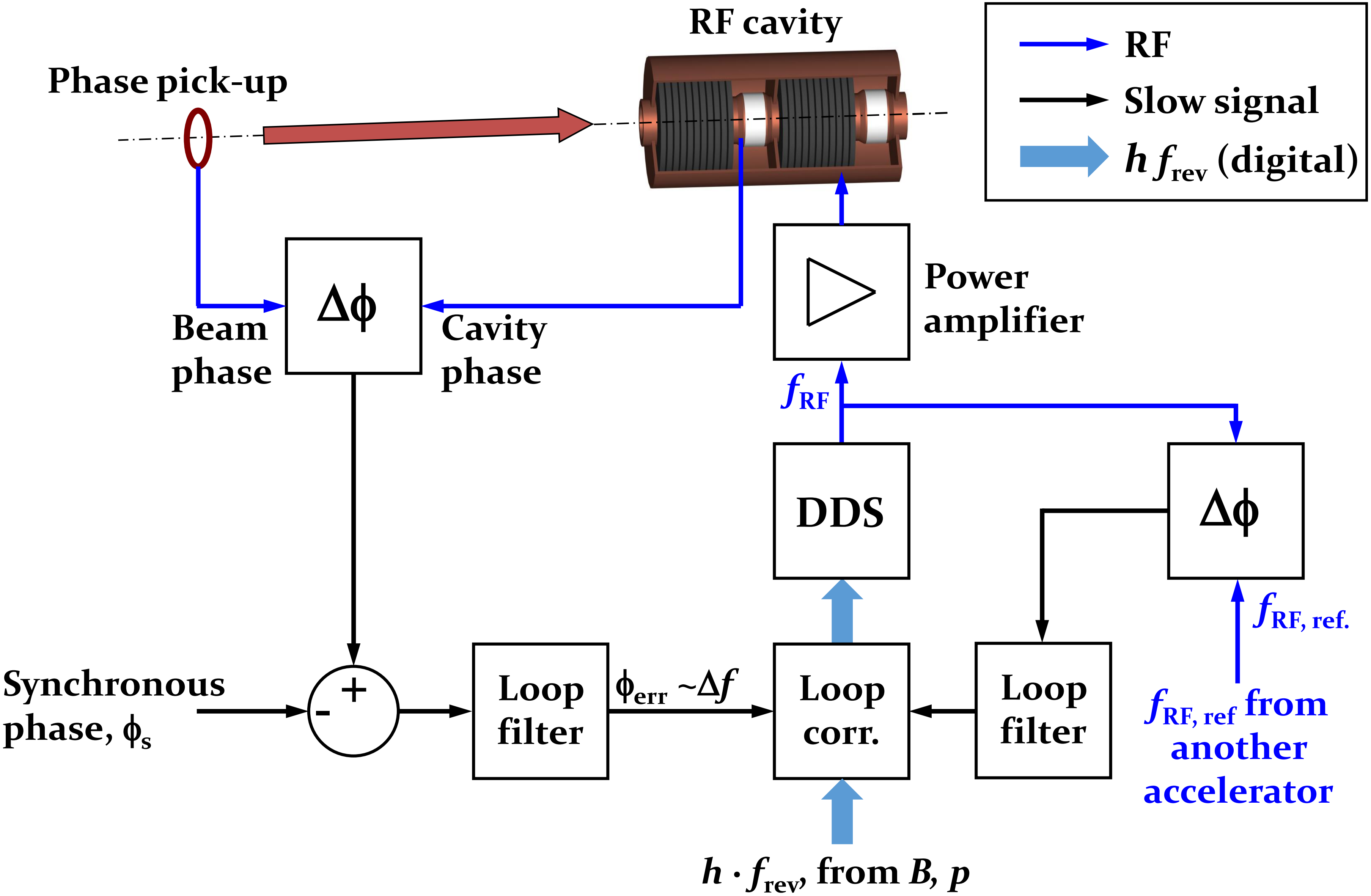
→ Fast control of RF frequency to cavities, but no slow corrections

Synchronization loop, internal reference



→ Avoids noise from radial detection when not crossing transition

Synchronization loop, external reference

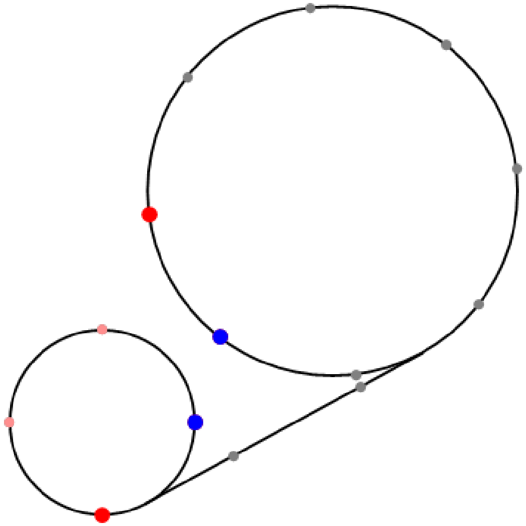


→ Synchronize between accelerators for transfer

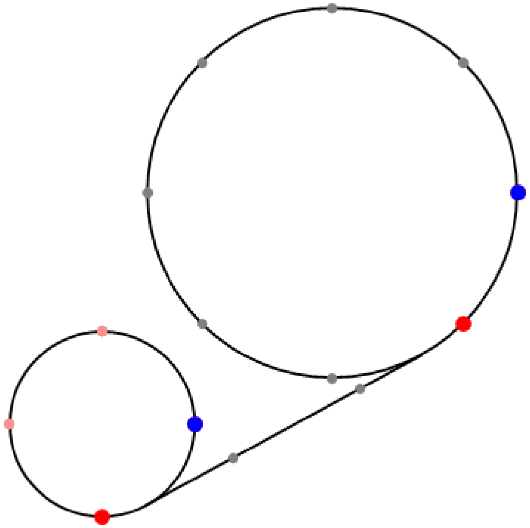
Before synchronization

- **Simple test case of circumference ratio 2: $C_2 = 2C_1$**

Target accelerator is master at transfer



Target accelerator is master at transfer

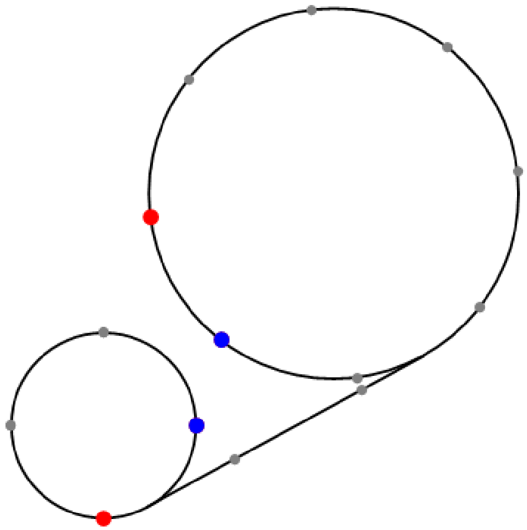


→ Synchronize both accelerator to force: $f_{\mathrm{rev},1} = 2f_{\mathrm{rev},2}$

After synchronization

- Simple test case of circumference ratio 2: $C_2 = 2C_1$

Source or target accelerator
is master at transfer



→ Revolution frequencies coupled: $f_{\mathrm{rev},1} = 2f_{\mathrm{rev},2}$

→ Ready to extract during every turn of the target accelerator

Summary

- RF system parameters
- Parameters of RF cavities
- Power amplifier
- Local feedbacks

  → Direct and 1-turn delay feedback
- Building blocks of low-level RF systems

  → Phase comparison, RF sources and receivers
- Basic global feedback loops

  → Beam phase, radial and synchronization loops

→ Make the beam feel comfortable!

A big Thank You

to all colleagues providing support, material and feedback

Maria-Elena Angoletta, Philippe Baudrenghien, Thomas Bohl, Giorgia Favia, Jörn Jacob, Erk Jensen, John Molendijk, Eric Montesinos, Gerry McMonagle, Mauro Paoluzzi, Damien Perrelet, Bernhard Schriefer, Lukas Stingelin, Fumihiko Tamura, Frank Tecker, Daniel Valuch and many more...

Thank you very much for your attention!

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Normalized Hamiltonian representation

- **For a single harmonic RF system**

$$H(\phi,\dot{\phi}) = \frac{1}{2}\dot{\phi}^2 + \frac{\omega_s^2}{\cos\phi_S}\left[\cos\phi_S - \cos\phi + (\phi-\phi_S)\sin\phi_S\right]$$

with $\phi = \phi_S + \Delta\phi$ **it becomes**

$$H(\Delta\phi,\dot{\phi}) = \frac{1}{2}\dot{\phi}^2 + \frac{\omega_s^2}{\cos\phi_S}\left[\cos\phi_S - \cos(\phi_S+\Delta\phi) - \Delta\phi\sin\phi_S\right]$$

using

$$\cos(\phi_S+\Delta\phi) = \cos\phi_S\cos\Delta\phi - \sin\phi_S\sin\Delta\phi$$

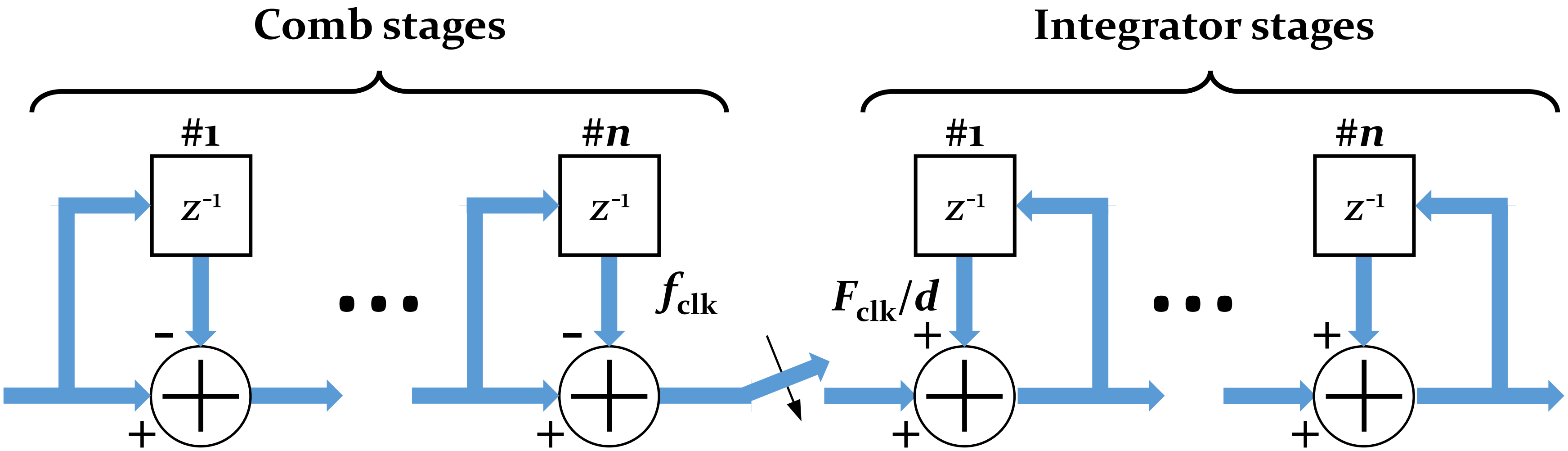
$$\simeq \cos\phi_S\left(1-\frac{1}{2}\Delta\phi^2\right) - \sin\phi_S\Delta\phi$$

this simplifies to

$$H(\Delta\phi,\dot{\phi}) \simeq \frac{1}{2}\dot{\phi}^2 + \frac{1}{2}\omega_s^2\Delta\phi^2$$

Cascaded integrator-comb filter (CIC)

- **Efficient implementation of low pass filter**
- **Standard form with sampling rate decimation: $f_{\mathrm{clk}} \rightarrow f_{\mathrm{clk}}/d$**



$$H(z) = \left(\frac{1-z^{-d}}{1-z^{-1}}\right)^{n}$$

***n*: filter order**

***d*: decimation ratio**

$$z = e^{2\pi i \cdot f / f_{\mathrm{clk}}}$$

→ Easy to implement in programmable logic: no multipliers

→ Only adders and shift registers

Cascaded integrator-comb filter (CIC)

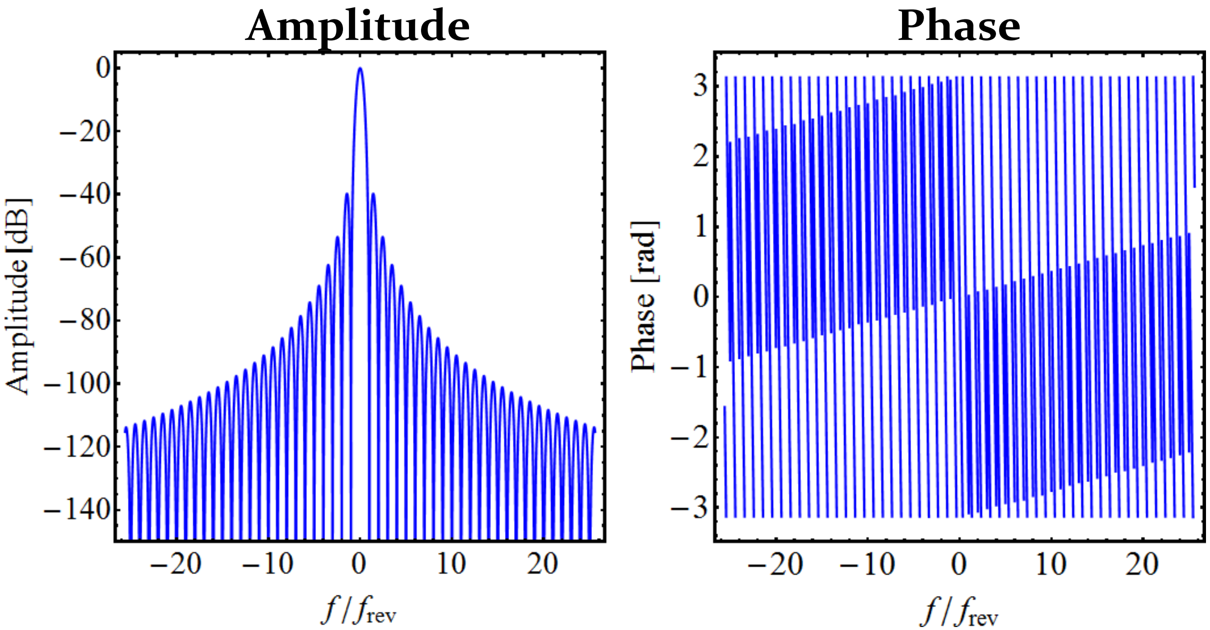
Why particularly interesting for circular accelerators?

- **Choose clock frequency, $f_{\text{clk}} = 2^m f_{\text{rev}}$ and decimation $d = 2^m$**

→ **Notches at all multiples of f_{rev} except zero**

→ **Linear phase $\phi(f)$ → filter behaves like a constant delay**

Example:
$f_{\text{clk}} = 128 f_{\text{rev}}$,
$d = 128$,
$n = 3$



👍 **Ideal low-pass filter in digital receivers**

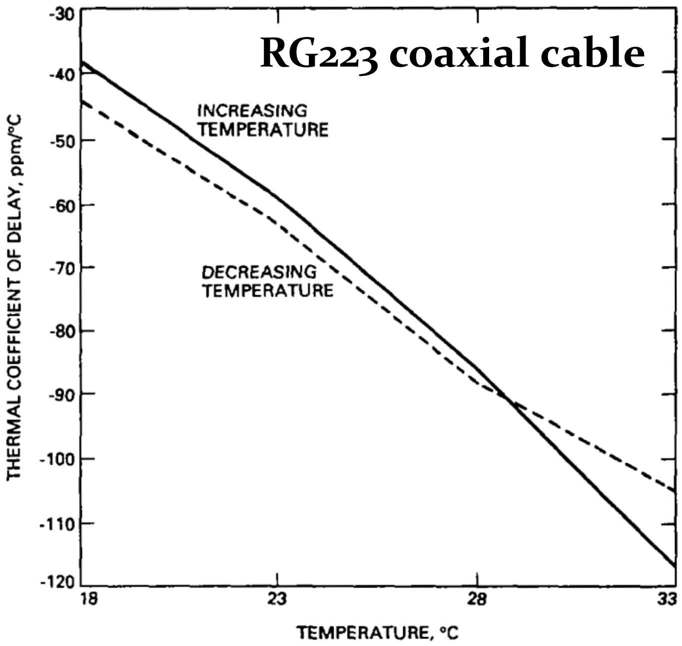
→ **Filter selected multiple of f_{rev} while suppressing all others**

Transmission of reference signals

- **Thermal drift of long coaxial cables or optical fibres**

- **Thermal coefficient of delay:**

$$\mathrm{TCD} = \frac{\Delta\tau}{\tau}\cdot\frac{1}{\Delta T} = \frac{\Delta\phi}{\phi}\cdot\frac{1}{\Delta T}$$



Lutes/Diener, NASA TDA Progress Report 42-99

- **Example: 2 km long RG223 cable with ~10 µs delay**

→ **ΔT of only 1° C (room temperature) changes delay by ~0.5 ns**

→ **1.8° at 10 MHz (CERN PS), but 73° at 400 MHz (LHC)**

- **Optical fibres are typically 10…100 times more stable**
- **What to do if this is still not sufficient?**

Simple synchronization process

1. **Move beam to off-momentum (*B* const.):** $\frac{df}{f} = \frac{\gamma_{\mathrm{tr}}^2 - \gamma^2}{\gamma^2\gamma_{\mathrm{tr}}^2}\frac{dp}{p}$
   - → Well defined frequency difference between accelerators
2. **Measure azimuth error, when beam at correct azimuth**
   - → Close synchronization loop
   - → Moves beam to ref. momentum