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Time-Delay-Based Analog Front-End for Monitored Drift Tubes in 65 nm CMOS with < 200 ns Baseline Recovery Time and Coherent Time-over-Threshold

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Monitored Drift Tube (MDT) chambers for muons detection at ATLAS Experiment adopt analog front-end read-out electronics for precise-tracking/momentum measurements of detected particles [1]. State-of-the-art has historically used bipolar shaping electronics (at approximately 800 ns baseline recovery time [2]) that scarcely fits with High-Luminosity (HL) requirements where small-MDT (sMDT [3]) are used at 180 ns input signal time-width. This brief presents 4-channels Application Specified Integrated Circuit in 65 nm CMOS for sMDT signal amplification and shaping that performs < 200 ns baseline recovery time and avoids distortion of Time-over-Threshold (ToT) characteristic due to multiple close input charge pulses composing sMDT signal.

Summary (500 words)

Introduction. ATLAS (s)MDT charge-to-voltage conversion is performed by Charge-Sensitive-Preamplifier (CSP) feeding analog signal processing stages differentiated by Unipolar or Bipolar shaping circuits. Due to complex requirements of unipolar shaping circuits bipolar shaping schemes are preferred.

However, in classical readout electronics [2], Bipolar shaping circuits suffer from long Baseline Recovery Time (BLR), higher than 800 ns, unfit for >1 MHit/sec required by HL-sMDT experiments [3]. Another drawback with current readout electronics is that the secondary pulses due to main track superimpose the main pulse, hence the ToT is not correlated to the main pulse. These problems are even more critical in sMDT chamber which has a drift-time of 180 ns. To profit from high data rate sMDT performance, BLR time must be less than the drift-time and ideally with no pile-up effects.

Method. The proposed ASIC is based on an innovative time-delay-based technique that significantly accelerates the BLR time performance. The qualitative block-scheme and time-domain evolution of the proposed approach are shown in Fig. 2. The incoming sMDT multiple charge signal is processed by <10 ns rising time CSP for charge-to-voltage conversion. Key performance parameters of CSP are fast peaking-time and noise performance which are optimized by using high input device transconductance (i.e. $= 25$ mA/V). CSP output feeds a Low-Pass-Filter (LPF) whose differential input signal is composed by voltage difference between the CSP output and a delayed replica. In this way, along the rising time of the CSP output voltage, the LPF output signal maintains the fast shaping while along the decay phase accelerates the BLR time, removing some electrical power from CSP main pulse output signal.

Delay Stage is composed by simple phase-shift R-C network feeding a voltage buffer for driving LPF input impedance. This way, the LPF performs both single-to-differential and unipolar-to-bipolar conversion of the CSP output voltage signal. Accuracy of the phase shift is essential and for this reason the capacitor of the Delay Stage is designed by binary weighted 5-bit programmable capacitor array to achieve a programmable time delay. Finally, the resulting < 180 ns BLR time bipolar signal is fed into comparator which compares it with a programmable threshold value to generate ToT signal. The complete electrical scheme of the integrated analog front-end is shown in Fig. 3.

Results. Single channel design is validated by both ideal δ -Dirac input pulse (Fig. 4) and sMDT Garfield signal (Fig. 5-6) (swept in 5-100 fC range) by Post-Layout Extracted View (PEX) simulations.

Fig. 4 displays a peaking time of 13 ns with sensitivity (Output voltage/Input charge) of 8 mv/fC. BLR time is reduced, by 78% as compared to classical models, to 165 ns that is even less than drift time of SMDT chamber. Fig. 4-7 shows time domain evolution of the output signals of CSP, Delay Stage, LFP and discriminator vs input signal. Fig 8 shows ToT range as input charge pulse varies from 5-100 fC. Secondary pulses crossing threshold value are suppressed by asynchronous finite state machine logic triggered by falling edge of ToT.

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