Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\rm OO}$

SGWB reconstruction for a non-equilateral and unequal-noise LISA constellation

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Data analysis challenges for SGWB workshop

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Outline

Introduction

- Measuring SGWBs with LISA
- Time Delay Interferometry (TDI) and TDI variables
- Data generation and pre-processing
- 2 Towards realistic data
 - Responses, spectra and strain noise
 - Noise and signal correlations
 - Impact on parameter estimation
- A new idea for SGWB data analysis
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Introduction Towards realistic data

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Measuring SGWBs with LISA

SGWBs detection

The data \tilde{d} (in frequency space) can be expressed as

$$\tilde{d} = \tilde{s} + \tilde{n}$$

For an isotropic SGWB $\longrightarrow \langle h_{\lambda}(\vec{k}) h_{\lambda'}^{*}(\vec{k'}) \rangle = P_{h}^{\lambda}(k)(2\pi)^{3}\delta_{\lambda\lambda'}\delta(\vec{k}-\vec{k'})$

Assuming $\langle \tilde{s}\tilde{n}\rangle=0$ and Gaussian signal and noise

$$\left|\left\langle \tilde{\mathbf{d}}^{2}\right\rangle = \left\langle \tilde{\mathbf{s}}^{2}\right\rangle + \left\langle \tilde{\mathbf{n}}^{2}\right\rangle = \mathcal{R} \, \mathbf{P}_{h}^{\lambda} + \mathbf{N} \equiv \mathcal{R} \left[\mathbf{P}_{h}^{\lambda} + \mathbf{S}_{n}\right]$$

where we have introduced

- The response function of the instrument \mathcal{R}
- The signal power spectrum P_h^{λ} (in 1/Hz)
- The noise power spectrum N (in 1/Hz)
- The (square of the) Strain sensitivity S_n (in 1/Hz)

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In order to compare with cosmological predictions it's customary to introduce

$$\Omega_{\rm GW} \equiv \frac{1}{3H_0^2 M_\rho^2} \frac{\partial \rho_{\rm GW}}{\partial \ln f} = \frac{4\pi^2}{3H_0^2} f^3 \sum_{\lambda} P_h^{\lambda} \qquad \text{and} \qquad \Omega_n(f) = \frac{4\pi^2}{3H_0^2} f^3 S_n(f) \;,$$

where $H_0 \simeq 3.24 \times 10^{-18} h_0 \, {\rm Hz}$ is the Hubble constant today.

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Measuring SGWBs with LISA

Some common assumptions...

LISA will have three (correlated) data streams dubbed XYZ basis ↓ Diagonalize to get the (uncorrelated) AET basis



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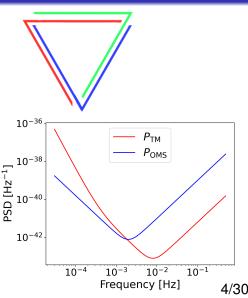
Time Delay Interferometry (TDI) to reduce **huge** laser noise

Low frequencies are dominated by Test Mass (TM) noise large frequencies by Optical Metrology System (OMS) noise:

$$P_{TM}(f, A) = A^2 \times 10^{-30} \times F_{TM}(f) ,$$

$$P_{OMS}(f, P) = P^2 \times 10^{-24} \times F_{OMS}(f)$$

where $F_{TM}(f)$, $F_{OMS}(f)$ are some functions of frequency.



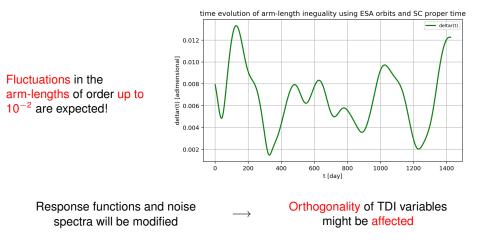
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Measuring SGWBs with LISA

In reality LISA won't have equal arms...



An accurate description is necessary to avoid biases in the analysis

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Measuring SGWBs with LISA

... nor the same noise levels in all links!

Let us have a closer look at the problem of noise characterization (still stick with TM and OMS noise only with known templates)

Each spacecraft contains two test masses and two lasers

12 (6 Acc +6 OMS) independent noise components are expected!

$$\begin{pmatrix} A & 0 \\ 0 & P \end{pmatrix} \longrightarrow \begin{pmatrix} A_{2} & A_{2} & D_{2}A_{3} & D_{2}A_{3} & D_{3}A_{3} & D_{4}A_{3} & D_{$$

Several complications are added in the problem:

- Noise components propagate differently in different TDI variables
- Higher dimensionality of the parameter space
- Correlations between the noise parameters

Again, an accurate analysis of the scenario is required

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Time Delay Interferometry (TDI) and TDI variables

Single link signal model

The observable is the difference in the travel time of a photon from *j* to *i*:

$$\Delta t_{ij}(t) \simeq \int_0^{L_{ij}} rac{\hat{l}_{ij}^a \, \hat{l}_{ij}^b}{2} h_{ab}(t(s), ec{x}(s)) \, \mathrm{d}s \,, \quad ext{or alternatively}, \quad \eta_{ij}^{\mathrm{GW}}(t) \equiv rac{\mathrm{d}}{\mathrm{d}t} \Delta t_{ij}(t) \;.$$

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To compute this quantity we first expand the signal in plane waves:

$$h_{ab}(\vec{x},t) = \int_{-\infty}^{\infty} \mathrm{d}f \int \mathrm{d}\Omega_{\hat{k}} \, \mathrm{e}^{2\pi i f(t-\hat{k}\cdot\vec{x})} \sum_{A} \tilde{h}_{A}(f,\hat{k}) \boldsymbol{e}_{ab}^{A}(\hat{k}) \,,$$

and assuming homogeneity, isotropy and no-chirality for the SGWB:

$$\langle \tilde{h}_{A}(f,\hat{k}) \, \tilde{h}_{B}^{*}(f',\hat{k}') \rangle = \delta(f-f') \delta(\hat{k}-\hat{k}') \delta_{AB} \frac{P_{h}^{AB}(f)}{16\pi} , \qquad \langle \tilde{h}_{A}(f,\hat{k}) \, \tilde{h}_{B}(f',\hat{k}') \rangle = 0 .$$

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The cross-spectral density for the single link measurement thus reads

$$S_{ij,mn}^{\eta,\mathrm{GW}}(f) \equiv \sum_{A} \mathcal{R}_{ij,mn}^{A} \mathcal{P}_{h}^{AA}(f) = \frac{f^2}{f_{ij}f_{mn}} e^{-2\pi i f(\mathcal{L}_{ij}-\mathcal{L}_{mn})} \sum_{A} \mathcal{P}_{h}^{AA}(f) \Upsilon_{ij,mn}^{A}(f) ,$$

where $\mathcal{R}_{ij,mn}^{A}$ is the single link response and we introduced:

$$\Upsilon^{A}_{ij,mn}(f) = \int \frac{\mathrm{d}\Omega_{\hat{k}}}{4\pi} \,\mathrm{e}^{-2\pi i f \hat{k} \cdot (\vec{x}_{i} - \vec{x}_{m})} \,\xi^{A}_{ij}(f,\hat{k}) \,\xi^{A}_{mn}(f,\hat{k})^{*} \,.$$
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Single link noise model

As we did for the signal, we want to express the single link noise η_{ij}^N .

In general this might be quite complicated.

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Assuming TDI cancels primary noises and only TM and OMS are present:

 $\eta_{ij}^{\mathsf{N}} = \mathcal{D}_{ij} \mathbf{n}_{ji}^{\mathsf{TM}}(t) + \mathbf{n}_{ij}^{\mathsf{TM}}(t) + \mathbf{n}_{ij}^{\mathsf{OMS}}(t) ,$

where \mathcal{D}_{ij} is the delay operator.

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where \mathcal{D}_{ij} is the delay operator.

Assuming stationarity, zero-mean and uncorrelated:

$$\langle \tilde{n}_{ij}^{\mathsf{TM}}(f) \, \tilde{n}_{ij}^{\mathsf{TM}^*}(f')
angle = \frac{1}{2} \, S_{ij}^{\mathsf{TM}}(f) \, \delta(f - f') \; ,$$

 $\langle \tilde{n}_{ij}^{\mathsf{OMS}}(f) \, \tilde{n}_{ij}^{\mathsf{OMS}^*}(f')
angle = \frac{1}{2} \, S_{ij}^{\mathsf{OMS}}(f) \, \delta(f - f') \; ,$

the only non-zero cross spectral densities are:

$$egin{aligned} S^{\eta,\mathrm{N}}_{ij,ij}(f) &= S^{\mathsf{TM}}_{ij}(f) + S^{\mathsf{TM}}_{ji}(f) + S^{\mathsf{OMS}}_{ij}(f) \;, \ S^{\eta,\mathrm{N}}_{ij,ji}(f) &= e^{2\pi i t \mathcal{L}_{ji}} S^{\mathsf{TM}}_{ij}(f) + e^{-2\pi i t \mathcal{L}_{ij}} S^{\mathsf{TM}}_{ji}(f) \;, \end{aligned}$$

where $S_{ji}^{\text{TM}}(f) = A_{ij}^2 \times F_{TM}(f)$ and $S_{ji}^{\text{OMS}}(f) = P_{ij}^2 \times F_{OMS}(f)$.

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Time Delay Interferometry (TDI) and TDI variables

Time Delay Interferometry

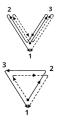
Compare measurements at different times (delay operator)

Noise reduction!

Let us start with two possibilities:

- Michelson variables, dubbed XYZ, defined as (YZ are permutations): $X \equiv (1 - D_{13}D_{31})(\eta_{12} + D_{12}\eta_{21}) + (D_{12}D_{21} - 1)(\eta_{13} + D_{13}\eta_{31})$
- Sagnac variables, dubbed $\alpha\beta\gamma$, defined as ($\beta\gamma$ permutations): $\alpha \equiv \eta_{12} + D_{12}\eta_{23} + D_{12}D_{23}\eta_{31} - (\eta_{13} + D_{13}\eta_{32} + D_{13}D_{32}\eta_{21})$

 D_{ij} delay operator, η_{ij} measurement in *i* coming from *j*



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Time Delay Interferometry

Compare measurements at different times (delay operator)

Noise reduction!

 $\longrightarrow \begin{pmatrix} -1/\sqrt{2} & 0 & 1/\sqrt{2} \\ 1/\sqrt{6} & -2/\sqrt{6} & 1/\sqrt{6} \\ 1/\sqrt{3} & 1/\sqrt{3} & 1/\sqrt{3} \end{pmatrix}$

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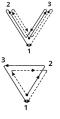
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 D_{ij} delay operator, η_{ij} measurement in *i* coming from *j*

Both signal and noise are correlated in these variables!

For equal arms, diagonalization via: (diagonal variables dubbed AET and \mathcal{AET})



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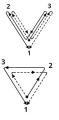
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Both signal and noise are correlated in these variables!

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Let us also introduce the fully symmetric ζ variable: $\zeta = D_{12}(\eta_{31} - \eta_{32}) + D_{23}(\eta_{12} - \eta_{13}) + D_{31}(\eta_{23} - \eta_{21}).$



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Time Delay Interferometry (TDI) and TDI variables

Single link to TDI in a more formal way

It is convenient to define TDI variables using vector notation:

$$ilde{\mathcal{V}}(f) = \sum_{ij \in \mathcal{I}} \, oldsymbol{c}_{ij}^{\mathcal{V}}(f) \; ilde{\eta}_{ij}(f) \; ,$$

i.e., starting from the (FTs of the) 6 single link measurement $\tilde{\eta}_{ij}(f)$, we use a 3 × 6 matrix $(c_{ij}^{V}(f))$, to project onto any TDI basis.

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i.e., starting from the (FTs of the) 6 single link measurement $\tilde{\eta}_{ij}(f)$, we use a 3 × 6 matrix ($c_{ij}^{V}(f)$), to project onto any TDI basis.

In this formalism, the cross spectral densities

$$\langle \tilde{U}(f) \ \tilde{V}^*(f') \rangle = \frac{1}{2} \ S^{UV}(f) \ \delta(f-f') \ ,$$

can be expressed as:

$$egin{aligned} &\langle ilde{U}(f) ilde{V}^{*}(f')
angle &= \sum_{ij,mn \in \mathcal{I}} oldsymbol{c}_{ij}^{U}(f) oldsymbol{c}_{mn}^{V*}(f') \langle ilde{\eta}_{ij}(f) ilde{\eta}_{mn}^{*}(f')
angle \;, \ &= rac{1}{2} \underbrace{\sum_{ij,mn \in \mathcal{I}} oldsymbol{c}_{ij}^{U}(f) oldsymbol{c}_{mn}^{V*}(f)}_{oldsymbol{c}_{ij,mn}^{U}(f)} oldsymbol{S}_{ij,lm}^{\eta}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,mn \in \mathcal{I}} oldsymbol{c}_{ij,mn}^{U}(f)}_{S^{UV}(f)} oldsymbol{S}_{ij,lm}^{\eta}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,mn \in \mathcal{I}} oldsymbol{S}_{ij,mn}^{UV}(f)}_{S^{UV}(f)} oldsymbol{S}_{ij,lm}^{\eta}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,m \in \mathcal{I}} oldsymbol{S}_{ij,lm}^{UV}(f)}_{S^{UV}(f)} oldsymbol{S}_{ij,lm}^{\eta}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,m \in \mathcal{I}} oldsymbol{S}_{ij,lm}^{UV}(f)}_{S^{UV}(f)} oldsymbol{S}_{ij,lm}^{\eta}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,m \in \mathcal{I}} oldsymbol{S}_{ij,lm}^{UV}(f)}_{S^{UV}(f)} oldsymbol{S}_{ij,lm}^{\eta}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,m \in \mathcal{I}} oldsymbol{S}_{ij,lm}^{UV}(f) \,\,\delta(f-f') \ &\underbrace{\sum_{ij,m \in \mathcal{I}} oldsymbol{S}_{ij,lm}^{UV}(f) \,\,\delta(f-f')}_{S^{UV}(f)} \ &\underbrace{\sum_{ij,m \in \mathcal{I}} oldsymbol{S}_{ij,lm}^{UV}(f) \,\,\delta(f-f') \,\,\delta(f') \,\,$$

Study the properties of $C_{ij,mn}^{UV}(f) \rightarrow$ make general statements on TDI basis. 10/30

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Data generation and pre-processing

Data generation (in frequency domain)

Assuming signal and noise to be Gaussian, stationary, and isotropic

independent realizations for each data segment and frequency are drawn:

$$egin{aligned} & ilde{s}_{c}(f_{i}) = rac{\mathcal{N}(0,\sqrt{\Omega_{\mathrm{GW}}(f_{i})}) + i\,\mathcal{N}(0,\sqrt{\Omega_{\mathrm{GW}}(f_{i})})}{\sqrt{2}} \ & ilde{n}_{c}(f_{i}) = rac{\mathcal{N}(0,\sqrt{\Omega_{n}(f_{i})}) + i\,\mathcal{N}(0,\sqrt{\Omega_{n}(f_{i})})}{\sqrt{2}} \end{aligned}$$

so that the spectra (Ω_{GW} and Ω_n) quantify the variance of fluctuations

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Given that:

- LISA will be operating for 4yrs (and assume also 75% efficiency)
- We choose data segments of roughly 12 days

in practice we have:

- Roughly 95 independent measurements at each frequency.
- A frequency resolution of around 10⁻⁶Hz

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Data generation and pre-processing

Data modeling and likelihood

Which likelihood should we use for our dataset?

Whittle likelihood should work (see any of the existing reviews):

$-2 \ln \mathcal{L}(\tilde{d}|\theta) \propto \sum_{f} \tilde{d}_i C_{ij}^{-1}(f|\theta) \tilde{d}_j^* + \ln \left[\det C_{ij}(f|\theta)\right],$

where C_{ij} is the covariance matrix for $\langle \tilde{d}_i \tilde{d}_j^* \rangle$ given by:

$$C_{ij}(f,\theta) = S_{ij}\left(f,\,ec{ heta}_{s}
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ight) \;.$$

For fast estimates of the errors on parameter reconstruction use Fisher:

$$F_{ij} \equiv \left. - \frac{\partial^2 \ln \mathcal{L}(\tilde{d}|\theta)}{\partial \theta_i \partial \theta_j} \right|_{\theta_0} = \sum_{f} = \operatorname{Tr} \left[C^{-1} \frac{\partial C}{\partial \theta_i} C^{-1} \frac{\partial C}{\partial \theta_j} \right] ,$$

However, this cannot tell us anything about possible biases!

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Let us start by defining $D_c(f_i)$ (our new data), as:

 $D_c(f_i) \equiv \langle \tilde{d}_c^2(f_i) \rangle = \langle (\tilde{s}_c(f_i) + \tilde{n}_c(f_i))^2 \rangle = \langle \tilde{s}_c^2(f_i) \rangle + \langle \tilde{n}_c^2(f_i) \rangle \;.$

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 We average over the (95) data segments: This leaves us with some D(f_i) (the averaged data) and an estimate of the error σ(f_i) (the standard deviation or the data).

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We can reduce the complexity of the problem by performing two operations:

- We average over the (95) data segments: This leaves us with some D(f_i) (the averaged data) and an estimate of the error σ(f_i) (the standard deviation or the data).
- We coarse grain (i.e. bin) the data (in frequency):
 i.e. from the initial linear 10⁻⁶Hz spacing (~ 5 × 10⁵ points)
 → we go to some final (and less dense) set of frequencies *f_i*. This leaves us with the final data set *f_i*, *D_i* and errors *σ_i*.

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angle + \langle \tilde{n}_c^2(f_i)
angle \;.$

We can reduce the complexity of the problem by performing two operations:

- We average over the (95) data segments: This leaves us with some D(f_i) (the averaged data) and an estimate of the error σ(f_i) (the standard deviation or the data).
- We coarse grain (i.e. bin) the data (in frequency):
 i.e. from the initial linear 10⁻⁶Hz spacing (~ 5 × 10⁵ points)
 → we go to some final (and less dense) set of frequencies f_i
 This leaves us with the final data set f_i, D_i and errors σ_i.

To gain a factor $\sim \mathcal{O}(100 \times 100)$ in computation time!

Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\rm OO}$

Data generation and pre-processing

An accurate likelihood for the compressed data

A Gaussian likelihood would give a systematic low bias!

(astro-ph/9808264, astro-ph/0205387, astro-ph/0302218, 0801.0554)

Consider the Gaussian likelihood:

$$\ln \mathcal{L}_{G}\left(\vec{\theta},\vec{n}\right) \propto -\frac{N_{chunks}}{2} \sum_{i,j} \sum_{k} w_{ij}^{(k)} \left(\frac{D_{ij}^{(k)} - h^{2}\Omega_{GW}\left(f_{ij}^{(k)},\vec{\theta}\right) - h^{2}\Omega_{n,ij}\left(f_{ij}^{(k)},\vec{n}\right)}{h^{2}\Omega_{GW}\left(f_{ij}^{(k)},\vec{\theta}\right) + h^{2}\Omega_{n,ij}\left(f_{ij}^{(k)},\vec{n}\right)}\right)^{2}$$

and the Lognormal likelihood:

$$\ln \mathcal{L}_{LN}\left(\vec{\theta},\vec{n}\right) \propto -\frac{N_{chunks}}{2} \sum_{i,j} \sum_{k} w_{ij}^{(k)} \ln^2 \left(\frac{\hbar^2 \Omega_{GW}\left(f_{ij}^{(k)}, \vec{\theta}\right) + \hbar^2 \Omega_{n,ij}\left(f_{ij}^{(k)}, \vec{n}\right)}{D_{ij}^{(k)}}\right)$$

Then we define our likelihood as (astro-ph/0302218, 2009.11845)

$$\ln \mathcal{L} = \frac{1}{3} \ln \mathcal{L}_G + \frac{2}{3} \ln \mathcal{L}_{LN}$$

which removes the skewness contributions and thus is more accurate.

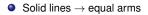
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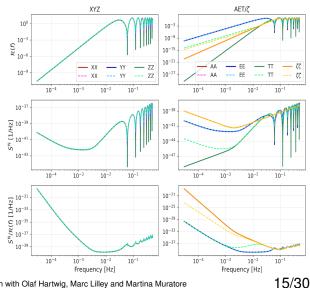
Conclusions and future perspectives

Responses, spectra and strain noise

Signal and noise Michelson



- Dashed lines \rightarrow unequal arms
- ٠ Most TDI variables are mildly affected
- Slight displacement of the zeros
- ٠ T Michelson looses signal orthogonality!
- ٠ ζ is clearly more robust to unequal arm corrections!



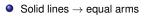
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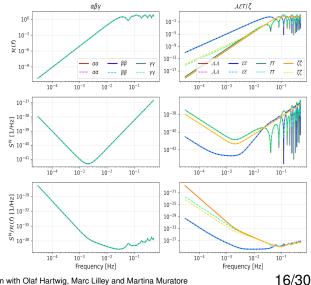
Conclusions and future perspectives

Responses, spectra and strain noise

Signal and noise Sagnac



- Dashed lines \rightarrow unequal arms
- ٠ Sensitivity is similar to Michelson variables (but less zeros)
- \mathcal{T} is way more robust than T! ۲
- ۲ \mathcal{T} and ζ 's signal orthogonality are comparable!



Towards realistic data •**00**00000

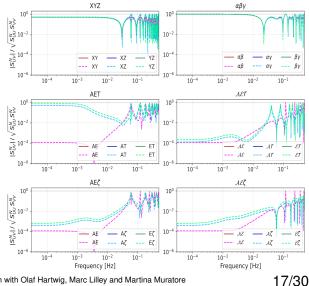
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Conclusions and future perspectives

Noise and signal correlations

Noise correlations

- ٠ Solid lines \rightarrow equal arms
- ۲ Dashed lines \rightarrow unequal arms
- Unequal arms break ۲ orthogonality
- ٥ Correlations involving T are most affected
- ٠ \mathcal{AET} is again more robust
- ۰ Replacing T with ζ might help
- ٠ No real benefit in trading \mathcal{T} for ζ



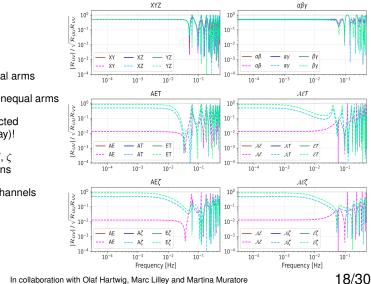
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Conclusions and future perspectives

Noise and signal correlations

Signal correlations



- Solid lines \rightarrow equal arms
- Dashed lines \rightarrow unequal arms ٠
- All bases are affected ۲ (\sim in the same way)!
- ٢ Leakage into T, T, ζ induces correlations
- ۲ Signal-sensitive channels stay orthogonal

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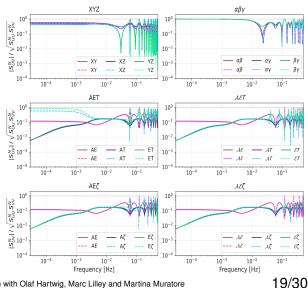
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Conclusions and future perspectives

Noise and signal correlations

Noise correlations v2 (unequal noise levels)

- Solid lines \rightarrow equal arms
- Dashed lines \rightarrow unequal arms
- ٥ Equal arms also have non-zero correlations!
- Correlations in all TDI bases
- 0 Again, Michelson performs worse than other bases





Let's test the impact of including these effects on parameter reconstruction.

Let us consider the following scenario:

- Constant but unequal armlenghts
- Noise described by a template (controlled by some parameters)
- Constant (check both equal and unequal) noise parameters
- Signal described by a power law $h^2 \Omega_{GW} = 10^{alpha} (f/f_*)^{n_T}$
- Injection with quite large SNR ($\alpha = -11.5$, $n_{\rm T} = 0 \rightarrow {\rm SNR} \sim 270$)
- Fisher matrix to get uncertainties and assess the impact of CSDs
- MCMC to validate Fisher results and look for biases
- Compressed data only uses the diagonal (still can check biases)

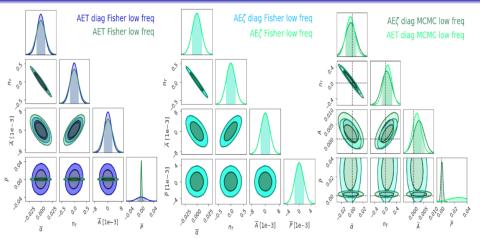
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Conclusions and future perspectives $_{\rm OO}$

Impact on parameter estimation

Low frequency (equal noise) analysis



• MCMC (diagonal only): Fisher approximation works quite well

- Fisher: Neglecting correlations is dangerous for AET, but not much for AEζ!
- MCMC/Fisher: AET performs much worse than AEζ

In collaboration with Olaf Hartwig, Marc Lilley and Martina Muratore

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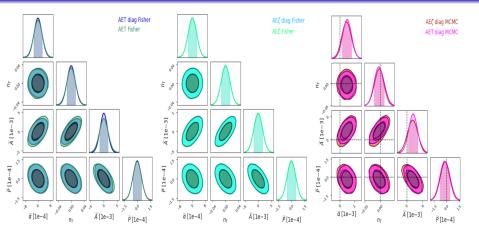
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Conclusions and future perspectives $_{\rm OO}$

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Impact on parameter estimation

Full frequency (equal noise) analysis



- MCMC (diagonal only): Fisher approximation works quite well
- Fisher: Neglecting correlations only marginally affects the reconstruction
- MCMC/Fisher: AET slightly underestimates errors (wrt AEζ)

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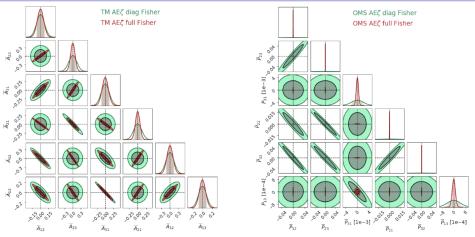
Towards realistic data

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Conclusions and future perspectives $_{\rm OO}$

Impact on parameter estimation

Full frequency (unequal noise) analysis



• MCMC (diagonal only): (not shown here but) Fisher approximation works quite well

- Fisher: Neglecting correlations might be dangerous for both bases (here only $AE\zeta$)
- MCMC/Fisher: more impact on worsley constrained (subdominant) parameters! In collaboration with Olaf Hartwig, Marc Lilley and Martina Muratore 23/30

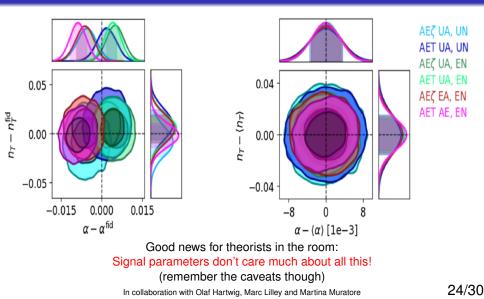
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Conclusions and future perspectives $_{\rm OO}$

Impact on parameter estimation

Signal parameters



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Conclusions and future perspectives $_{\rm OO}$

ML for SGWB data analysis

Traditional methods (MCMC, nested sampling, whatever) are quite efficient and guaranteed to converge (in some cases)

but

scale poorly with number of parameters and require explicit likelihoods

Towards realistic data

A new idea for SGWB data analysis • 000 Conclusions and future perspectives $_{\rm OO}$

ML for SGWB data analysis

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Can alternative approaches perform better in some cases?

Towards realistic data

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Conclusions and future perspectives $_{\rm OO}$

ML for SGWB data analysis

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Can alternative approaches perform better in some cases?

Normally, with Bayesian inference, we try to study the posterior probability:

$$p(\theta|d) = rac{p(d|\theta)\pi(\theta)}{p(d)} \equiv r(d,\theta)\pi(\theta) \; ,$$

where we have introduced:

$$r(d, heta) \equiv rac{p(d| heta)}{p(d)} = rac{p(heta|d)}{\pi(heta)} = rac{p(heta, d)}{p(d)\pi(heta)} \; ,$$

i.e. $r(d, \theta)$ is the ratio between joint probability and marginal probability.

Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\rm OO}$

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i.e. $r(d, \theta)$ is the ratio between joint probability and marginal probability. Given a pair (θ, d) , $r(d, \theta)$ can be used to assess whether θ can generate d!

This can be cast in a minimization problem that can be solved with ML the approach is typically referred to as Neural Ratio Estimation.

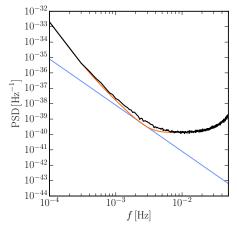
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Conclusions and future perspectives

Recover previous results I ...

Assume we inject a power law signal: Can we recover it with the same level of accuracy?



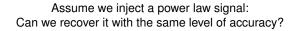
Towards realistic data

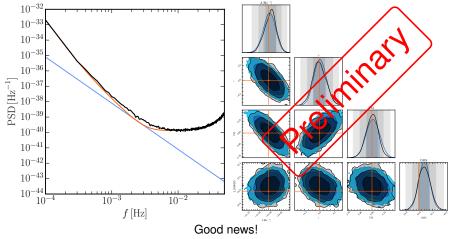
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Conclusions and future perspectives $_{\rm OO}$

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Recover previous results I ...





In collaboration with James Alvey, Uddipta Bhardwaj, Christoph Weniger and Valerie Domcke

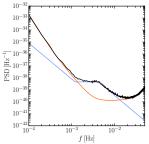
Towards realistic data

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Conclusions and future perspectives

Recover previous results II...



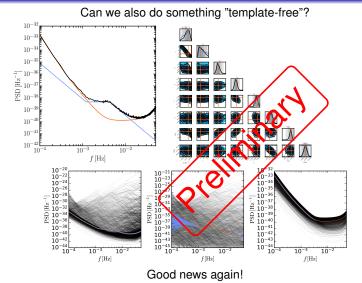


Towards realistic data

A new idea for SGWB data analysis $\circ \circ \circ \circ$

Conclusions and future perspectives $_{\rm OO}$

Recover previous results II...



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Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\rm OO}$

... plus something completely new!

What if there's something else beyond SGWB and noise?

For example, assume some sources slightly below the threshold for detection are randomly injected.

Would this still work??

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A new idea for SGWB data analysis 0000

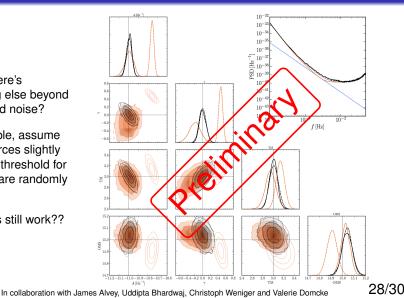
Conclusions and future perspectives

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Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\bullet \circ}$

Conclusions and future perspectives

Conclusions

- Unequal arms jeopardize the standard orthogonalization procedure
- Some TDI variables are more robust to unequal arm corrections
- Similarly, unequal noise levels also introduce correlations
- Signal reconstruction looks mildly affected (with caveats...)
- ML techniques for SGWB look quite promising...

Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\bullet \circ}$

Conclusions and future perspectives

Conclusions

- Unequal arms jeopardize the standard orthogonalization procedure
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- Similarly, unequal noise levels also introduce correlations
- Signal reconstruction looks mildly affected (with caveats...)
- ML techniques for SGWB look quite promising...

Future perspectives

- Application to concrete cases (inflation, phase transitions, ...)
- Keep improving on detector modeling
- More realistic noise model and data generation procedure
- Drop stationarity (both for signal and noise)
- Include anisotropies
- New techniques?

Towards realistic data

A new idea for SGWB data analysis

Conclusions and future perspectives $_{\odot \bullet}$



The End

Thank you

