

Crab Cavity RF Noise: Effect, issue, theory and mitigations. A status

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Why CC RF noise matters so much ?

- As CC acts in the transverse plane, its RF noise will increase the beam transverse emittance
- This results in loss of luminosity as it is inversely proportional to the normalized transverse emittance
- We have been given a budget for the loss of integrated luminosity (over a fill) caused by CC RF noise: 1 % ([1],[3] tables 6-9, egrowth < 0.05 μm/h for 2.5 μm emittance)
- This corresponds to a maximum of 2%/hour emittance growth due to CC at lowest β* (15 cm)
- This is very small -> study of CC RF noise was encouraged from beginning of HL-LHC.



Transverse Emittance Growth. Theory

$$\begin{aligned} \frac{d\epsilon_n}{dt} &= N_{cavities}\gamma\beta_{cc} \left(\frac{eV_o f_{rev}}{2E_b}\right)^2 \left\{ e^{-\sigma_{\phi}^2} \left[I_o \left[\sigma_{\phi}^2 \right] + 2\sum_{l=1}^{\infty} I_{2l} \left[\sigma_{\phi}^2 \right] \right] \right\} \sum_{k=-\infty}^{\infty} \int_{-\infty}^{\infty} S_{\Delta\phi} \left[(k \pm \nu_b) f_{rev} \right] \rho \left(\nu_b \right) d\nu_l \\ &= N_{cavities}\gamma\beta_{cc} \left(\frac{eV_o f_{rev}}{2E_b}\right)^2 C_{\Delta\phi}(\sigma_{\phi}) \frac{2\sigma_{\Delta\phi}^2}{f_{rev}} \\ &= \frac{1}{\beta^*} \left[N_{cavities}\gamma \left(\frac{ec\theta_{cc} f_{rev}}{4\omega_{RF}}\right)^2 \right] C_{\Delta\phi}(\sigma_{\phi}) \frac{2\sigma_{\Delta\phi}^2}{f_{rev}} \\ \frac{d\epsilon_n}{dt} &= N_{cavities}\gamma\beta_{cc} \left(\frac{eV_o f_{rev}}{2E_b}\right)^2 \left\{ e^{-\sigma_{\phi}^2} \sum_{l=0}^{\infty} I_{2l+1} \left[\sigma_{\phi}^2 \right] \right\} \sum_{k=-\infty}^{\infty} \int_{-\infty}^{\infty} S_{\Delta A} \left[(k \pm \nu_b \pm \nu_s) f_{rev} \right] \rho \left(\nu_b \right) d\nu_b \\ &= \frac{1}{\beta^*} \left[N_{cavities}\gamma \left(\frac{ec\theta_{cc} f_{rev}}{4\omega_{RF}}\right)^2 \right] C_{\Delta A}(\sigma_{\phi}) \frac{4\sigma_{\Delta A}^2}{f_{rev}} \end{aligned}$$

- Transverse emittance growth due to RF noise as derived in [2], Eq. (20) (22).
- Operational parameters: Little or no control. This term is effectively inversely proportional to $1/\beta^*$ for constant full crabbing angle θ_{cc} . According to [3], we use 380 µrad full crabbing angle from the beginning of physics.
- Bunch length dependence: Effectively constant over operational range.
- RF noise: Depends on RF and LLRF technology (to be determined).



Maximum allowable RF noise

- With the analytical model we can compute the allowable RF noise to remain below 2%/h:
 - End of physics conditions: V_{cc} =3.3 MV, β_{cc} =3800, 4 cavities/plane, 1 ns 4 σ bunch length), ϵ_{n} = 2.5 μ m
 - The ADT damper will provide some reduction of phase noise (no effect on amplitude noise). With 50 turns damping ([3], tables 5,6,8,9), analytical formulas ([2], Fig.9) give a reduction factor of 0.32
 - Phase noise σ_{Δφ}=14 µrad will result in emittance growth 0.92 %/h with damper on
 - Amplitude noise $\sigma_{\Delta A} = 14 * 10^{-6}$ will result in emittance growth 1.08%/h.



Mitigation 1. Low noise LLRF

- The LLRF includes a proportional RF feedback that must reduce the cavity impedance at the fundamental by >100 linear. This results in a 136 kHz regulation BW [4] (Sec. 1)
- Comparison of the ACS phase noise and the CC target
- We aim at -143 dBc/Hz SSB phase noise and amplitude noise in the 3 kHz-136 kHz band [4]
- That is 10 dB better than ACS
- This will result in
 - **7.6%/h** e-growth due to phase noise
 - 9%/h due to amplitude noise
- Factor 10 excess!



- Fixed-frequency clocks and LO, we can use narrow phase-lock loops to improve the demodulator LO and thereby reduce the RF phase noise at the first two betatron sidebands
- Using IOTs instead of klystrons
- Reducing the RF demodulator noise by at least 10 dB



Mitigation 2. CC feedback

- Dedicated feedback system to counteract crab cavity noise could be developed [6],[8] to provide the extra factor 10
 - Such a system could work in conjunction with the ADT
 - Its performance will be limited by the pickup measurement noise (pickup specs later in this presentation.
 - Theory and simulations have shown very promising performance [6][8].



Feedback system using CC as kicker



 This system is very promising in simulations



- But the performance is limited by the measurement noise level.
 Emittance growth rate curves with varying magnitudes of measurement error in the presence of both phase and amplitude noise
- Need for a low-noise bunch displacement (mode 0) and tilt (mode 1) measurement chain.









- For single bunch we get 440 nm for mode 0 and 4.5 μrad for mode
 1
- As the CC noise spectrum extends to 136 kHz only, while measurement noise is white (25 ns spacing -> 20 MHz BW), an optimal filter will reduce measurement noise by 12 linear-> in batch mode
 - **5.3** μ**m** for mode 0
 - 55 μrad for mode 1
- See [6] for analytical derivations and more simulations.



LLRF processing (tentative)

- Except for the novel use of a CC as kicker, it is a *classic* transverse feedback with mode 0 (displacement) and 1 (tilt)
- We plan to follow processing shown in [7] Eq. (16) to extract mode 0 and 1 signals, at least for SPS test bench
 - Delta/Sigma signals from WB PU
 - Filtering with 400 MHz BPF
 - Analog mixer with 375 MHz LO
 - ADC clocked at 100 MHz
 - I/Q demodulation
 - Optimal filter to increase SNR
 - Then we compute Delta/Sigma. The signal has both dipole (real-valued I = mode 0) and tilt (imaginary Q = mode 1) info. See [7]

$$X_N = \frac{I_{\Delta}I_{\Sigma} + Q_{\Delta}Q_{\Sigma}}{I_{\Sigma}^2 + Q_{\Sigma}^2} + j\frac{Q_{\Delta}I_{\Sigma} - I_{\Delta}Q_{\Sigma}}{I_{\Sigma}^2 + Q_{\Sigma}^2}$$

- We then apply phase shift (around betatron tune) to have 90 degrees, including latency and PU-CC phase advance, plus BPF for SNR
- We modulate CC set-point in phase (phase fdbk) and amplitude (amplitude fdbk)

To be tested in SPS in 2024.

CC feedback PU

- We have a PU next to each cryostat (cavity pair)
- We have 2 available PU candidates
 - Button
 - Stripline
- We consider operation (demodulation) at 400 or 800 MHz-> 120 mm stripline (green) and button are good options
- The frequency responses of the two Pus are very similar
- The stripline gives ~20
 dB more signal. Can we make use of it ?



Courtesy of M. Krupa

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Signals from button PU before demodulation



- Single bunch, 1.05 ns, 2.3e11 ppb
- For the required resolution (5.3 um and 55 urad) the mode 0 and 1 signals have similar peak amplitude. Good
- But they are **4000-5000** below common mode
- Assuming 20 dB rejection from delta hybrid (can we get more?) we would still have common mode 400-500 times larger than mode 0 or 1 measurements





Signals from button PU after 400 MHz BPF



- Again, for the required resolution (5.3 μm and 55 μrad) the mode 0 and 1 signals have similar 400 MHz component. Good
- But they are still 4000-5000 below common mode
- Note that the mode 0 and mode 1 signals, after 400 MHz BPF, are indeed in quadrature.





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Plans. Near future (2023-mid 2024)

- 1. Design an RF feedback that fulfills the noise requirement: -143 dBc/Hz max on all betatron lines from first (3 kHz) to fdbk BW (136 kHz)
- 2. Test the CC fdbk in SPS
 - 1. Cabling existing (button) PU, add hybrid and 400 MHz BPF
 - 2. Check/deploy demodulation for the Sigma Delta PU signal in CavLoop module
 - 3. Implement the Delta/Sigma operation, extract mode 0 and 1 signals
 - 4. Design the optimal filter
 - 5. Implement BPF with proper phase shift on the mode0 and 1 signals
 - 6. Feedback on CC voltage set point

NB: In the SPS CC the RF noise is much higher than HL-LHC goal, with measured SSB around -125 dBc/Hz at first betatron band. Plus we can inject RF noise -> we can design CC FDBK tests with much larger PU measurement noise level

- 3. Select PU and PU front end for the HL-LHC CC
 - 1. Button or coupler? Urgent
 - 2. Study the front-end. Can we live with common mode? Do we still have enough resolution for the 5.3 μ m, 55 μ rad precision?

Thank you for your attention. Questions? Comments?

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