

# ***HOM coupler design\****

***Hans-Walter Glock, Carsten Potratz***

***Universität Rostock - Institut für Allgemeine Elektrotechnik***

***22.9.2011***

***Meeting: HOM issues in the ESS SC linac, ESS Lund***

***\* work supported by German Federal BMBF, contracts 05H09HR5 and 05K10HRC***

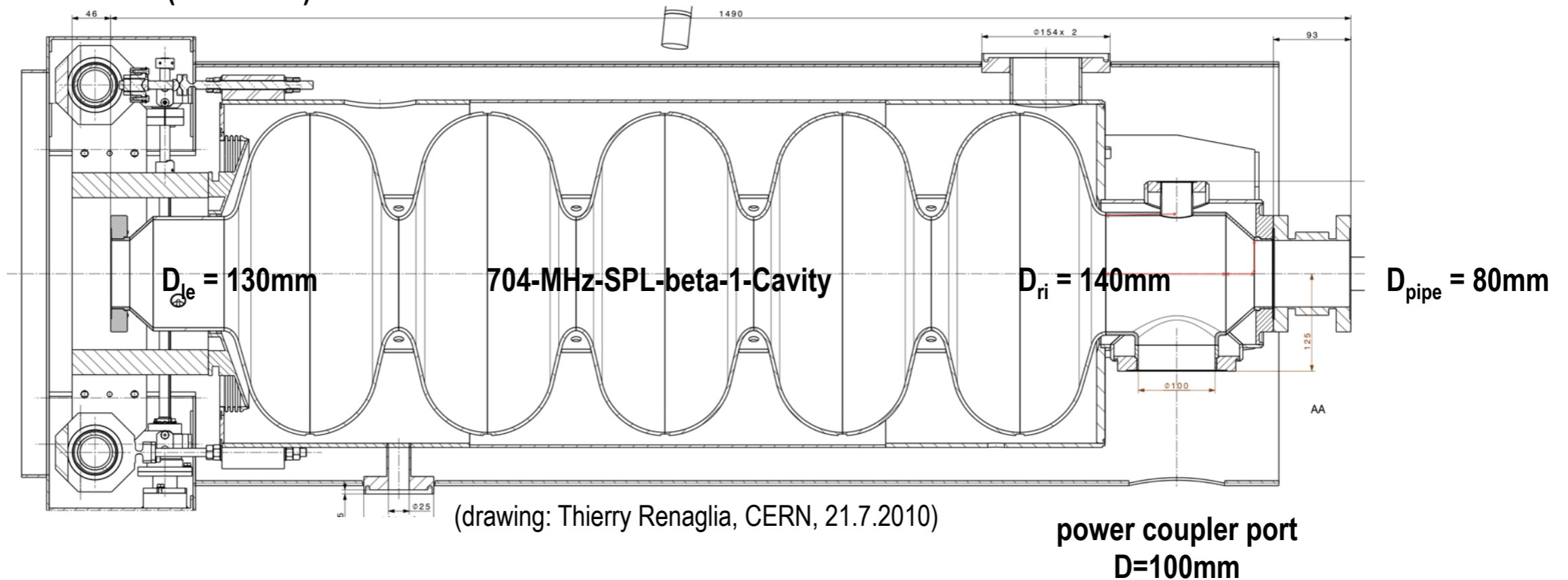
## Overview

- "Once upon a time" - brief history of Rostock's coupler design ...
- ... including some more general experiences
- Latest SPL/(ESS) version
- Sensitivity to main parameters, current density, ideas of mechanical resonances
- Some coupler olympics: Rama's TESLA-style in comparison with hook-type
- explicitly excluded: multipacting issues (hook type has an issue!) => Steve's/Rob's talk

## "History" of developments I (primarily focussed on SPL $\beta=1$ )

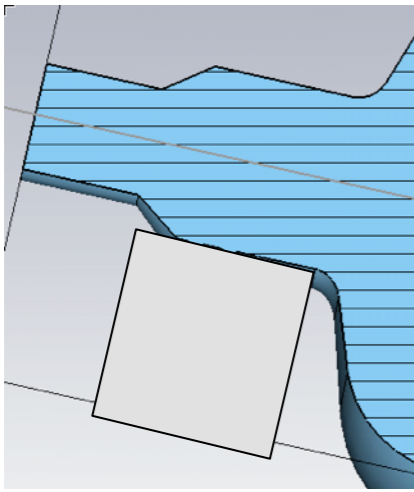
"left" side HOM coupler port  
rotated by  $60^\circ$  towards you  
(not drawn)

"right" side HOM coupler port  
opposed to power c.,  $D=36\text{mm}$

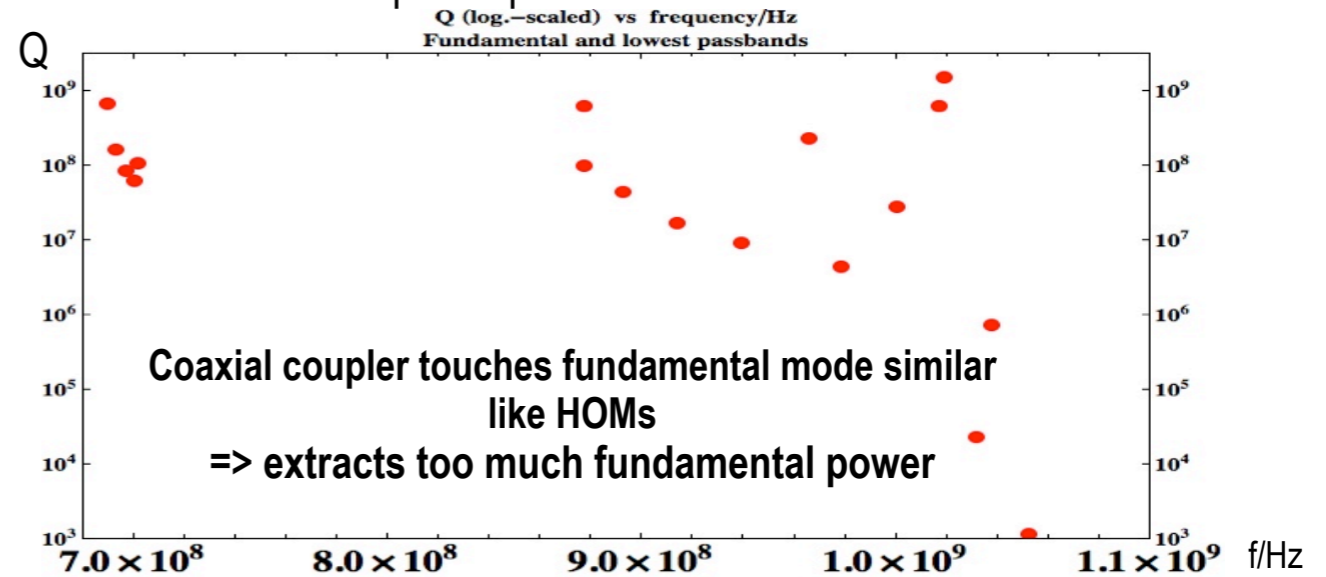
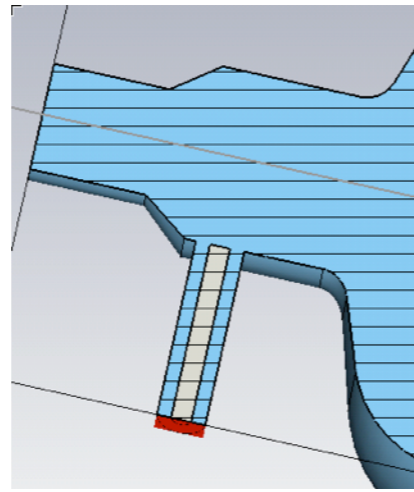


# "History" of developments I (primarily focussed on SPL $\beta=1$ )

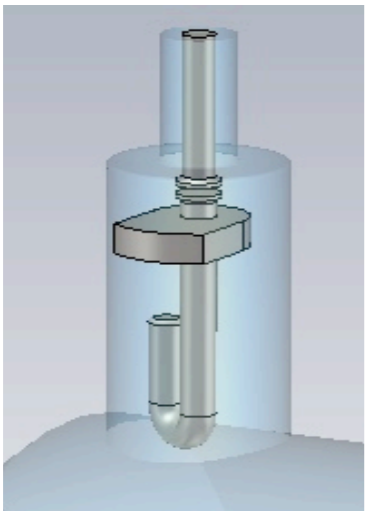
① when  $Q_{\text{HOM}} > 10^7$  seemed to be acceptable



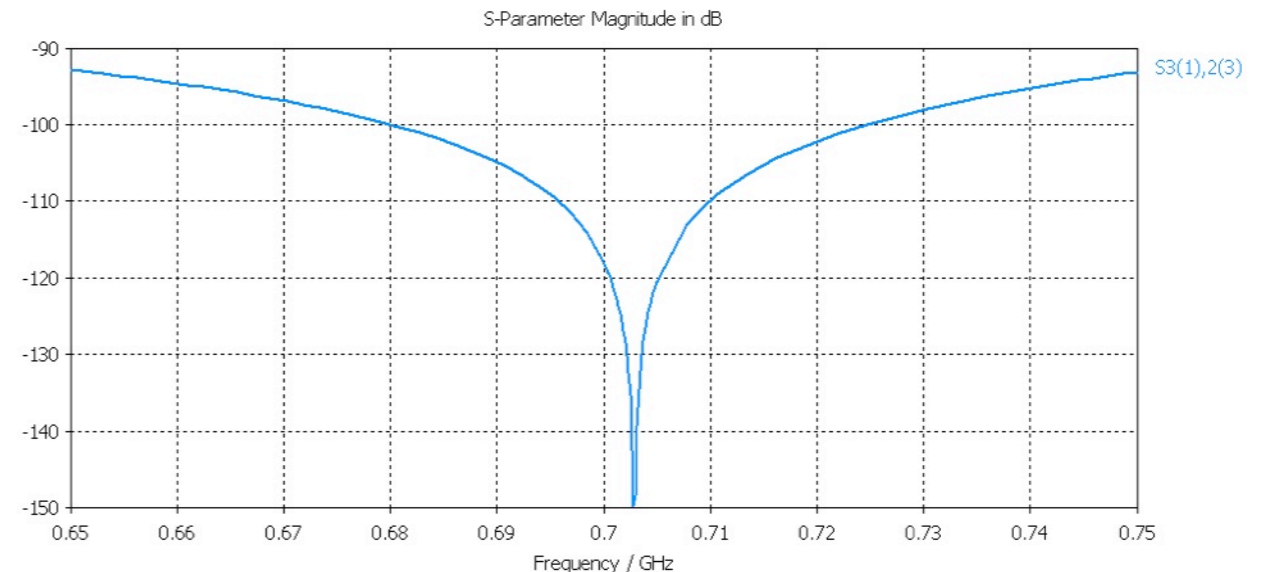
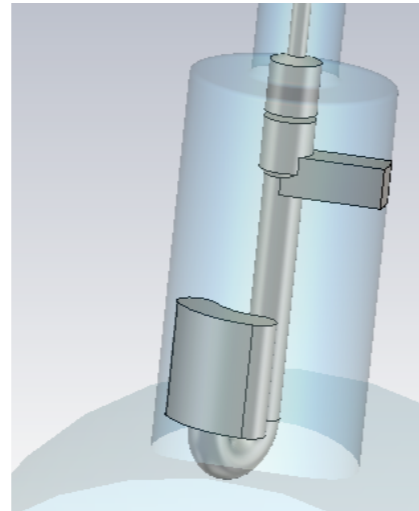
② when couplers seem unavoidable but should as simple as possible



③ when we therefore introduced a fundamental mode filter

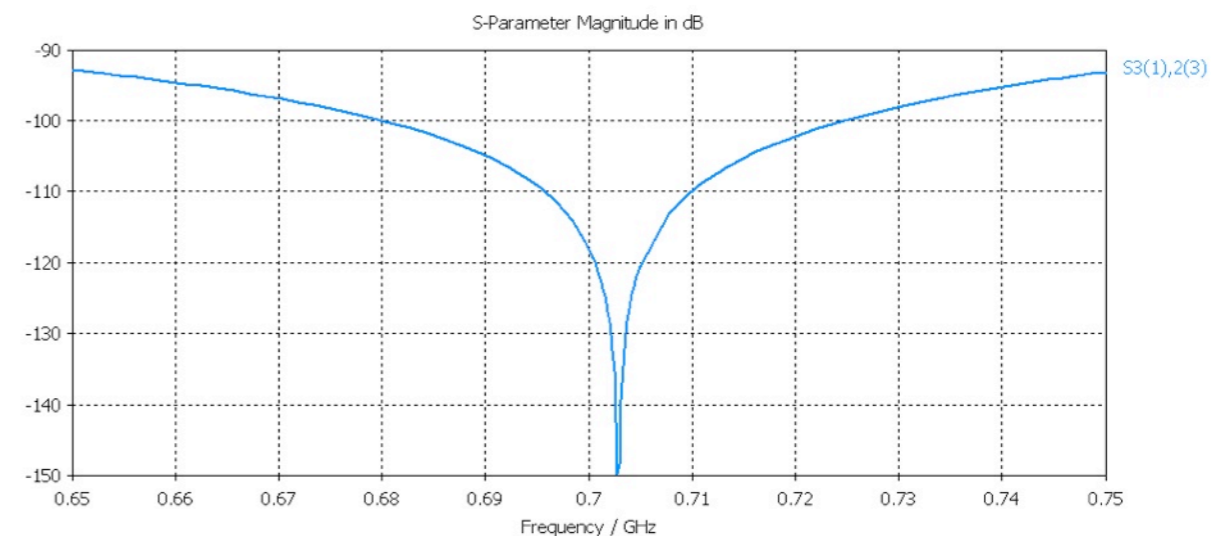
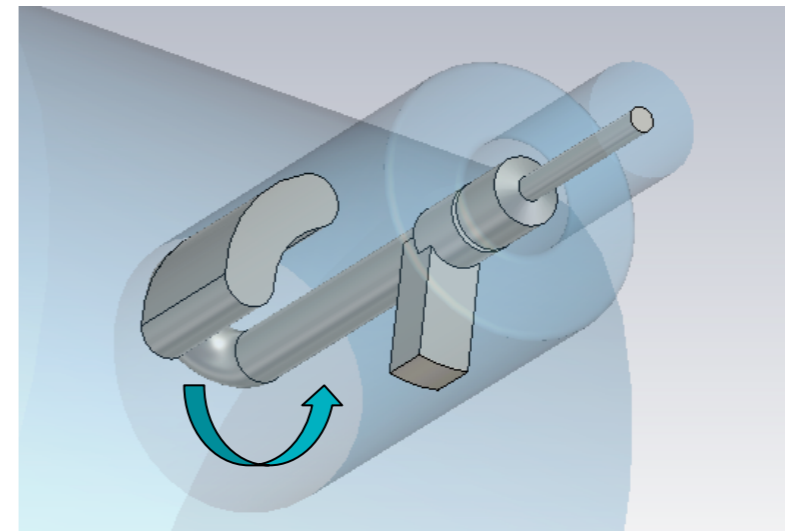


④ where we had to add a less elegant capacity enhancement for adjusting notch frequency



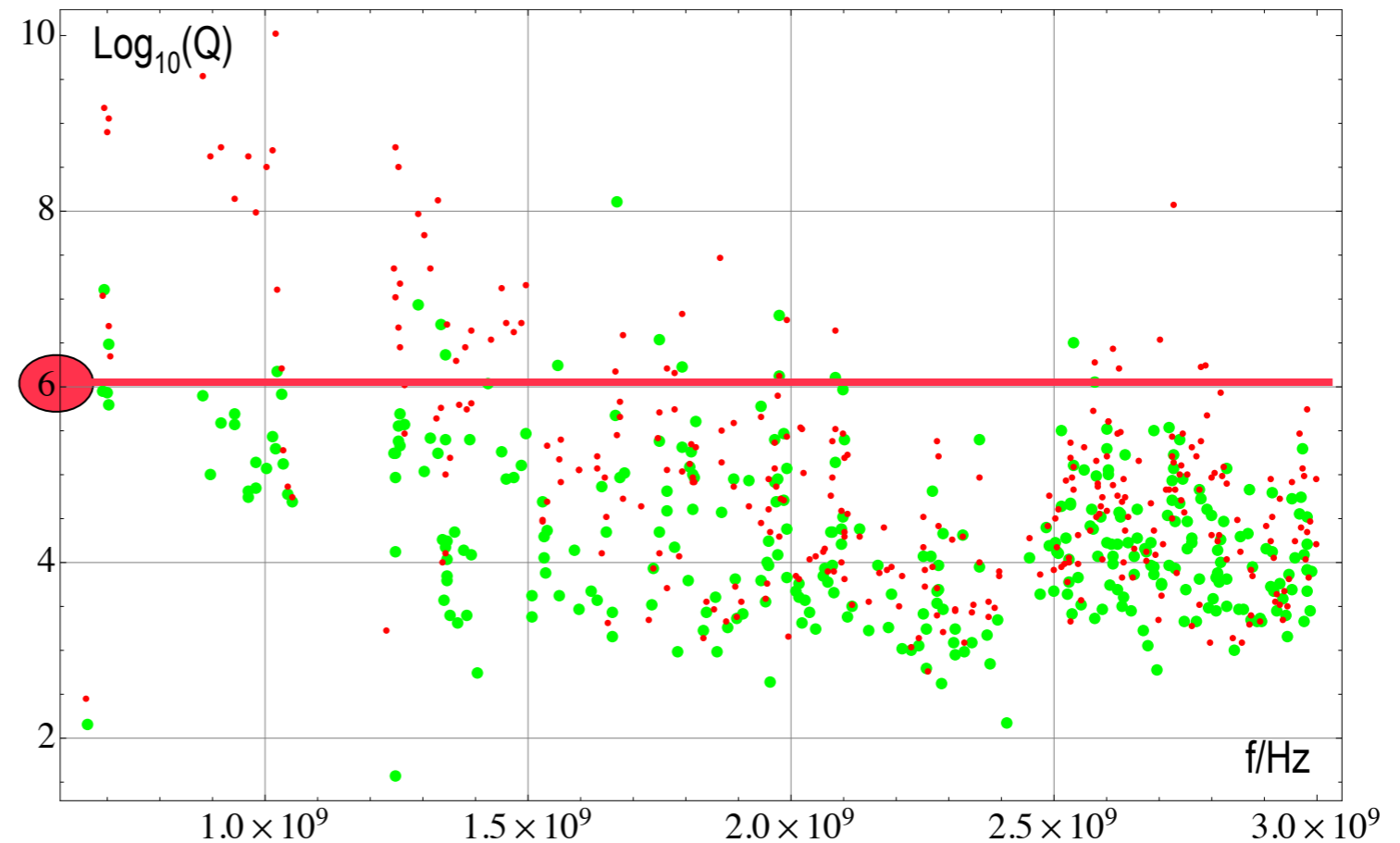
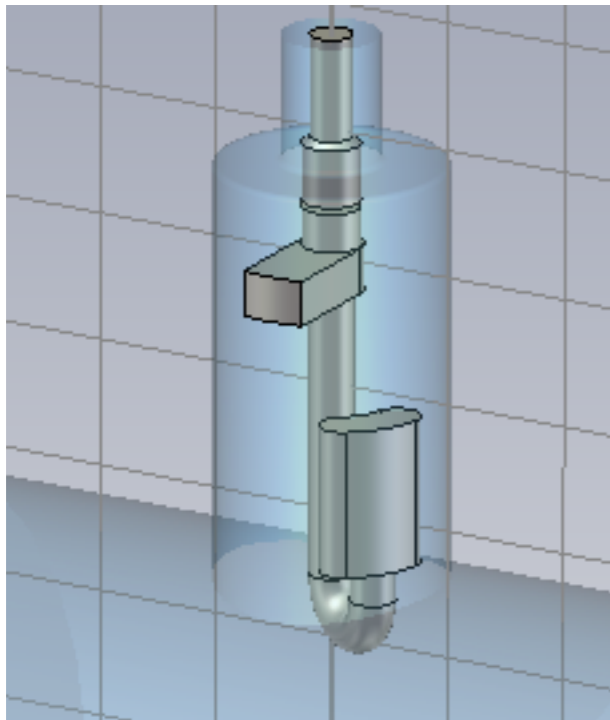
## History II: Lessons learned about fundamental mode notch filter

- 1.) Tuning rather sensitive both against capacity surface AND rotation angle ( $\sim 5$  MHz/Degree  $\Leftrightarrow 30$  dB/Degree)
- 2.)  $\Rightarrow$  notch filter understood as combination of resonance AND "directional coupler"-effect: certain E-H-correlation causes cancelation
- 3.) This demands for external re-tuning capability after mounting (e.g. rotation)



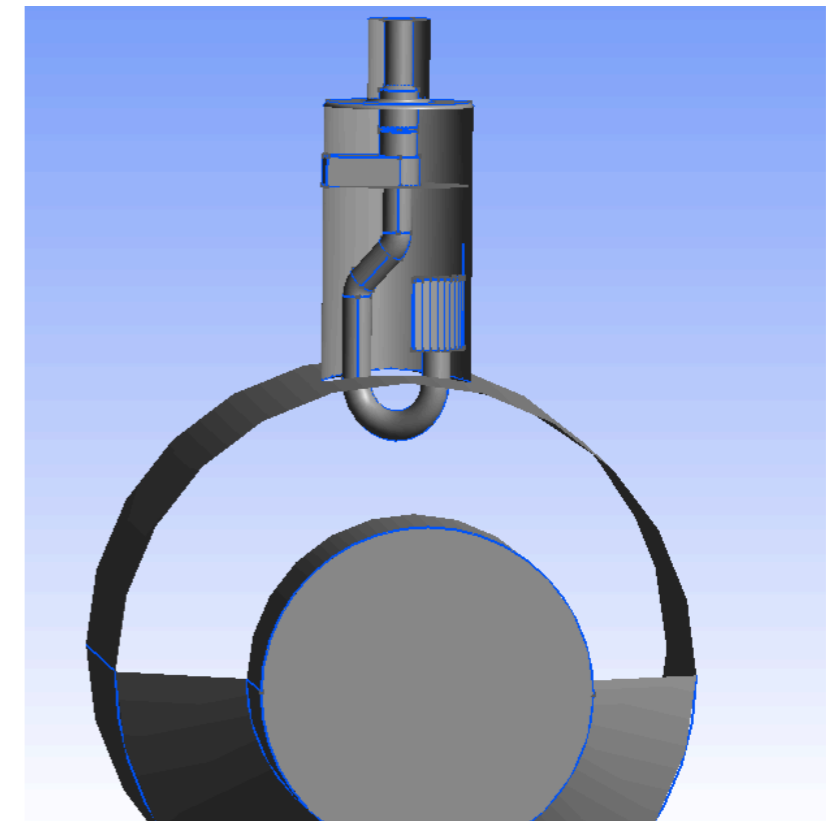
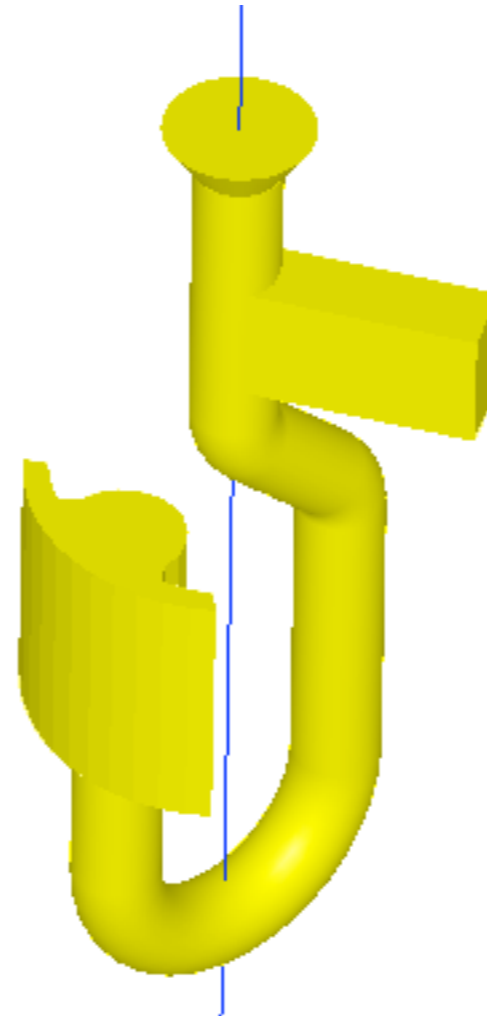
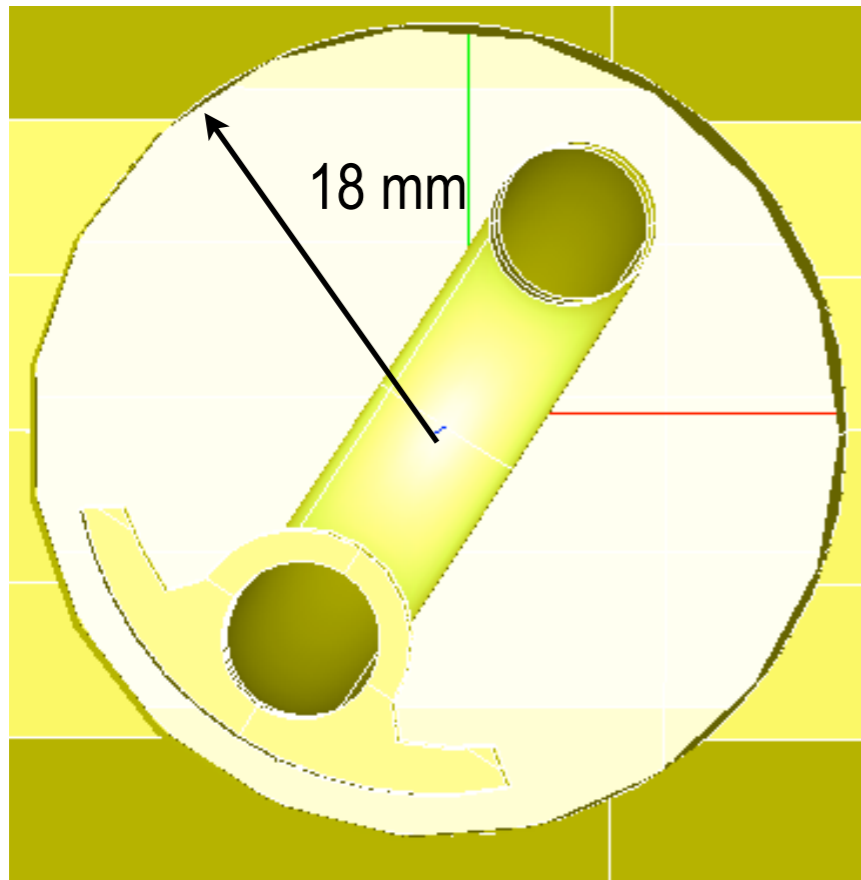
# "History" of developments III

still version ④:



- red: HOM coupler only, green with matched power coupler
- main power coupler helps significantly below 2 GHz; still  $Q_s > 10^6$  found
- => extend depth and width of loop to increase coupling

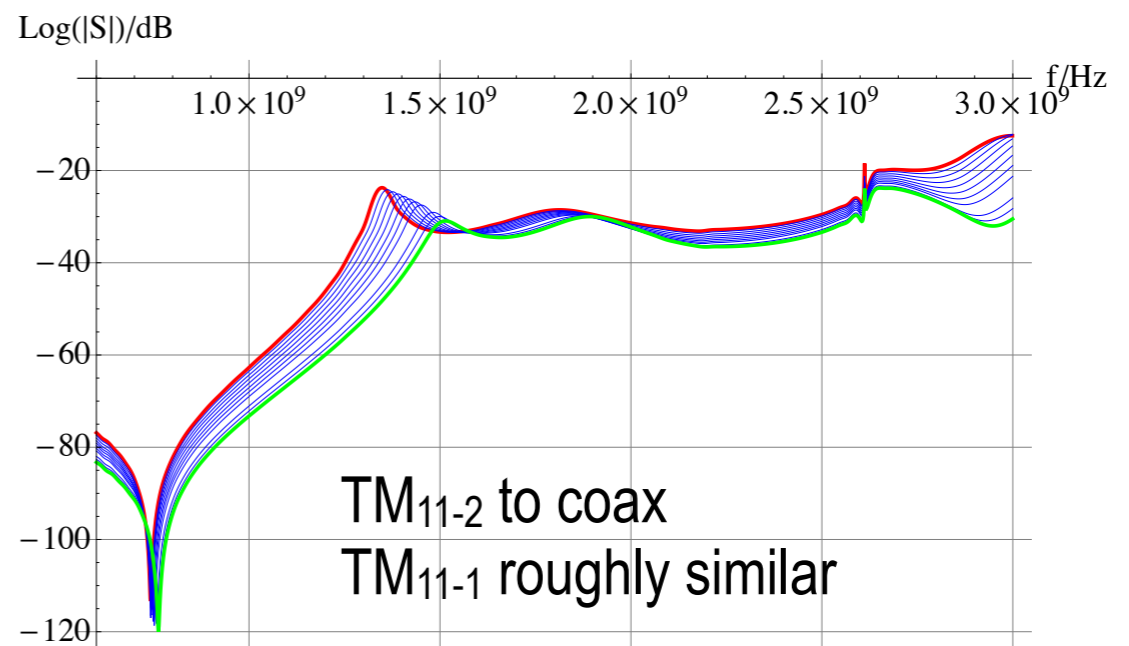
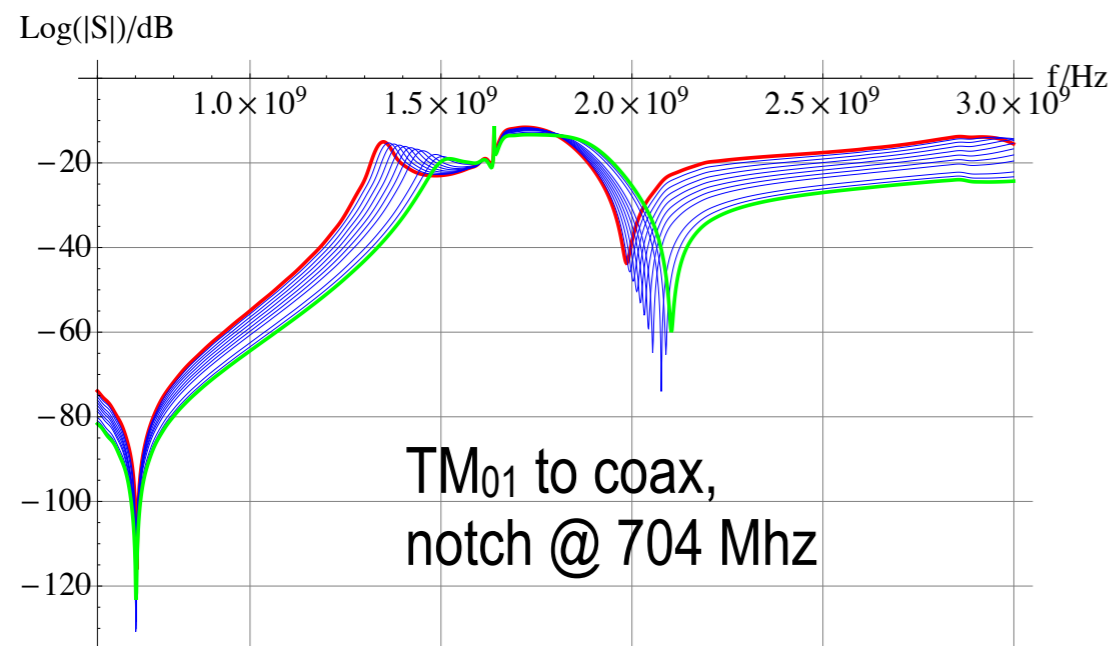
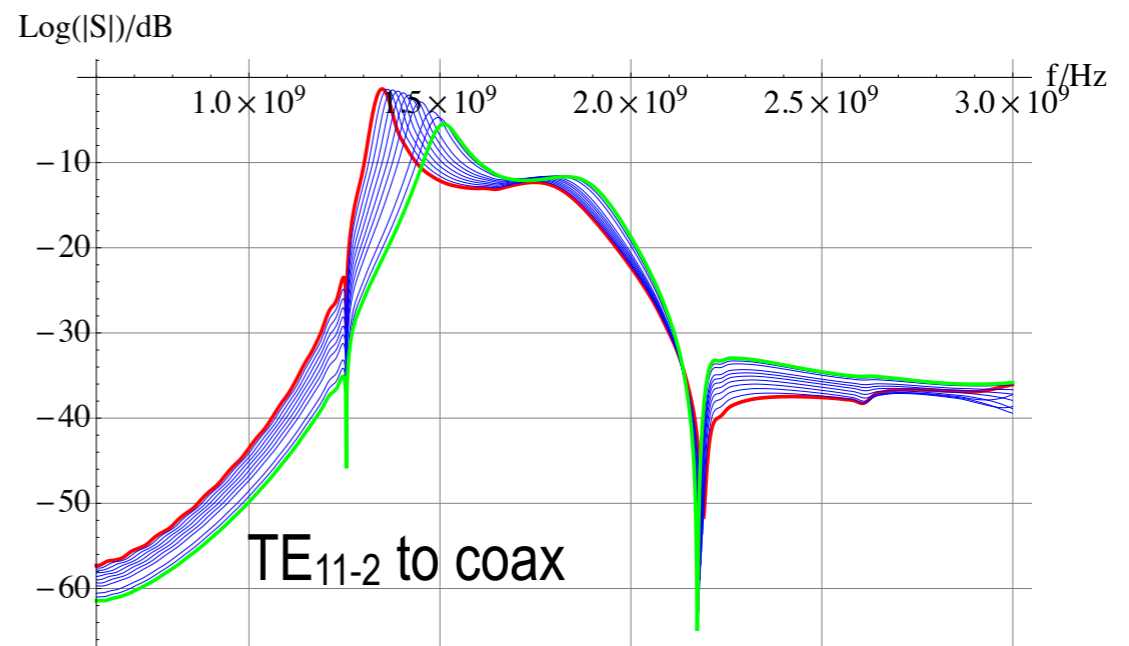
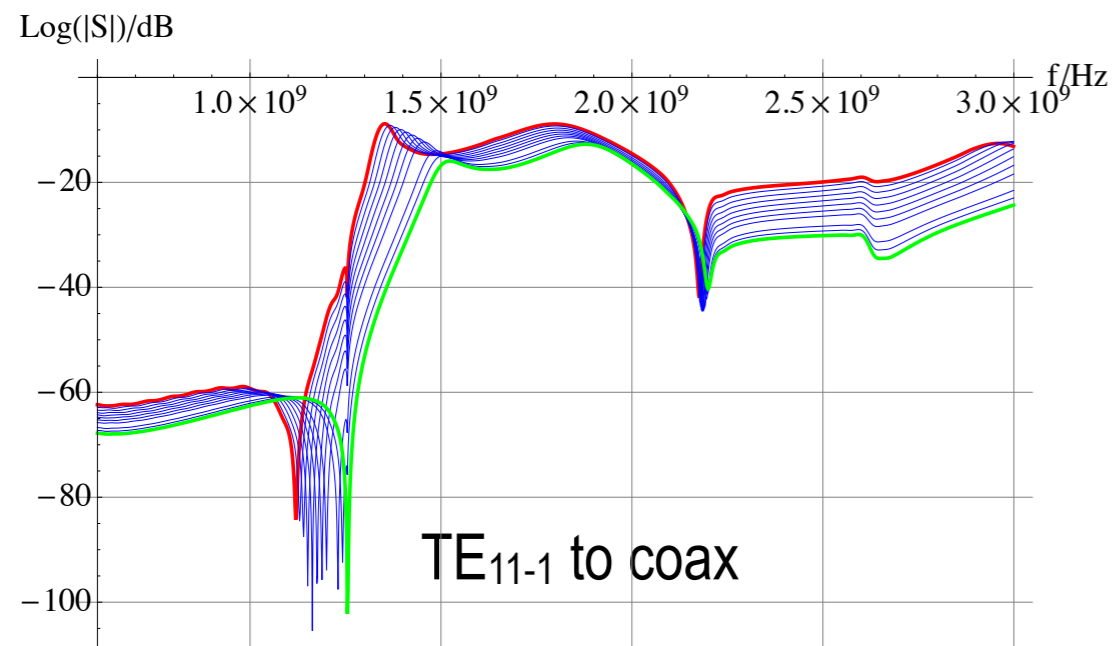
## History IV: Version ⑤



- several things changed: loop shape and lower bow radius, depth, capacity profile ...
- ... and computing method\* (SALOME + NUDGE-based CUDA parallelized time domain iterator, C. Potratz)\*

\*: C. Potratz et. al.: Time Domain Field and Scattering Parameter Computation in Waveguide Structures by GPU-Accelerated Discontinuous-Galerkin Method, accepted for publishing in IEEE-MTT

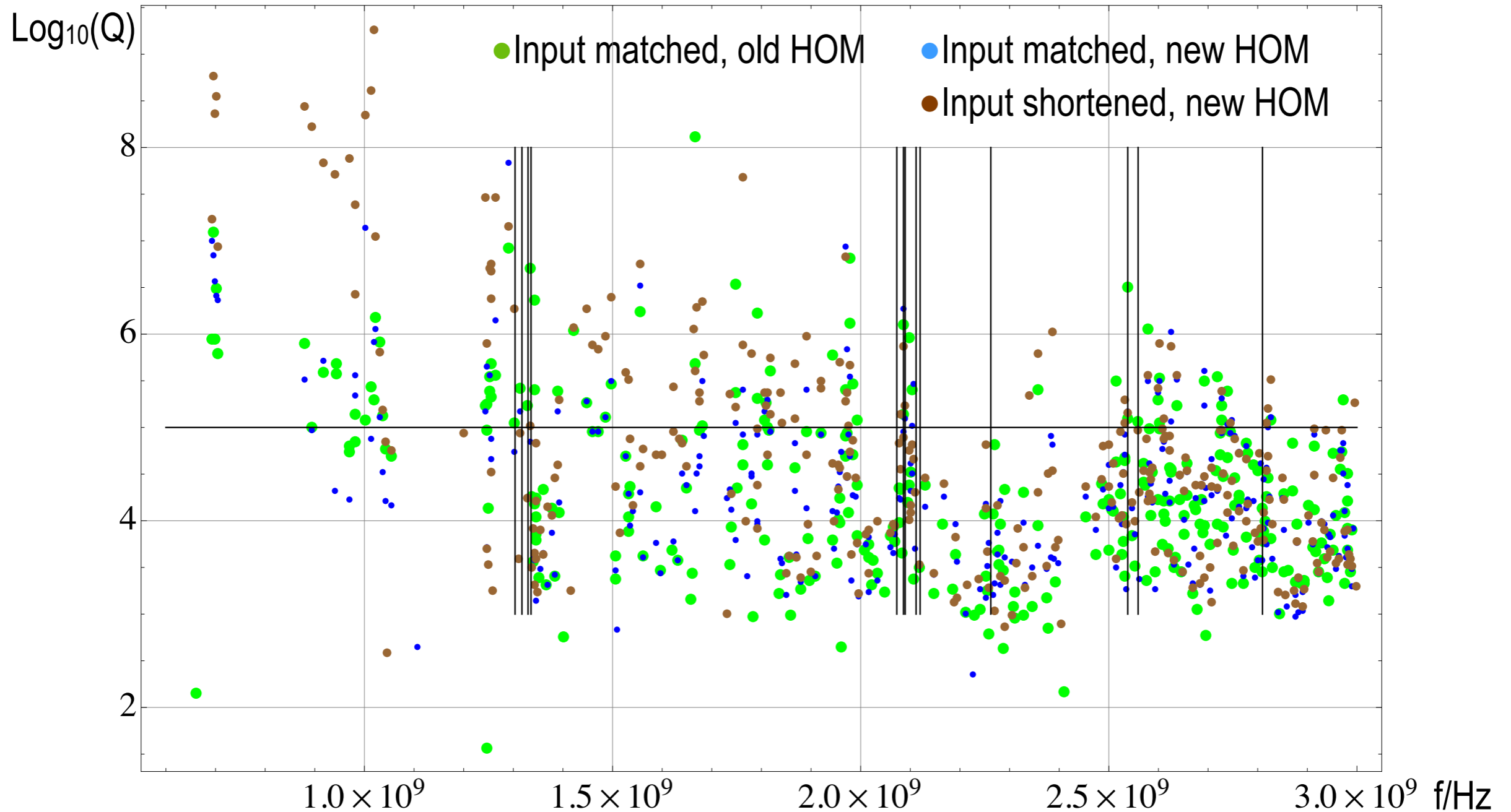
# Coupling dependency on penetration depth (Version ⑤)



**10 mm additional depth ~ very roughly +10dB coupling**

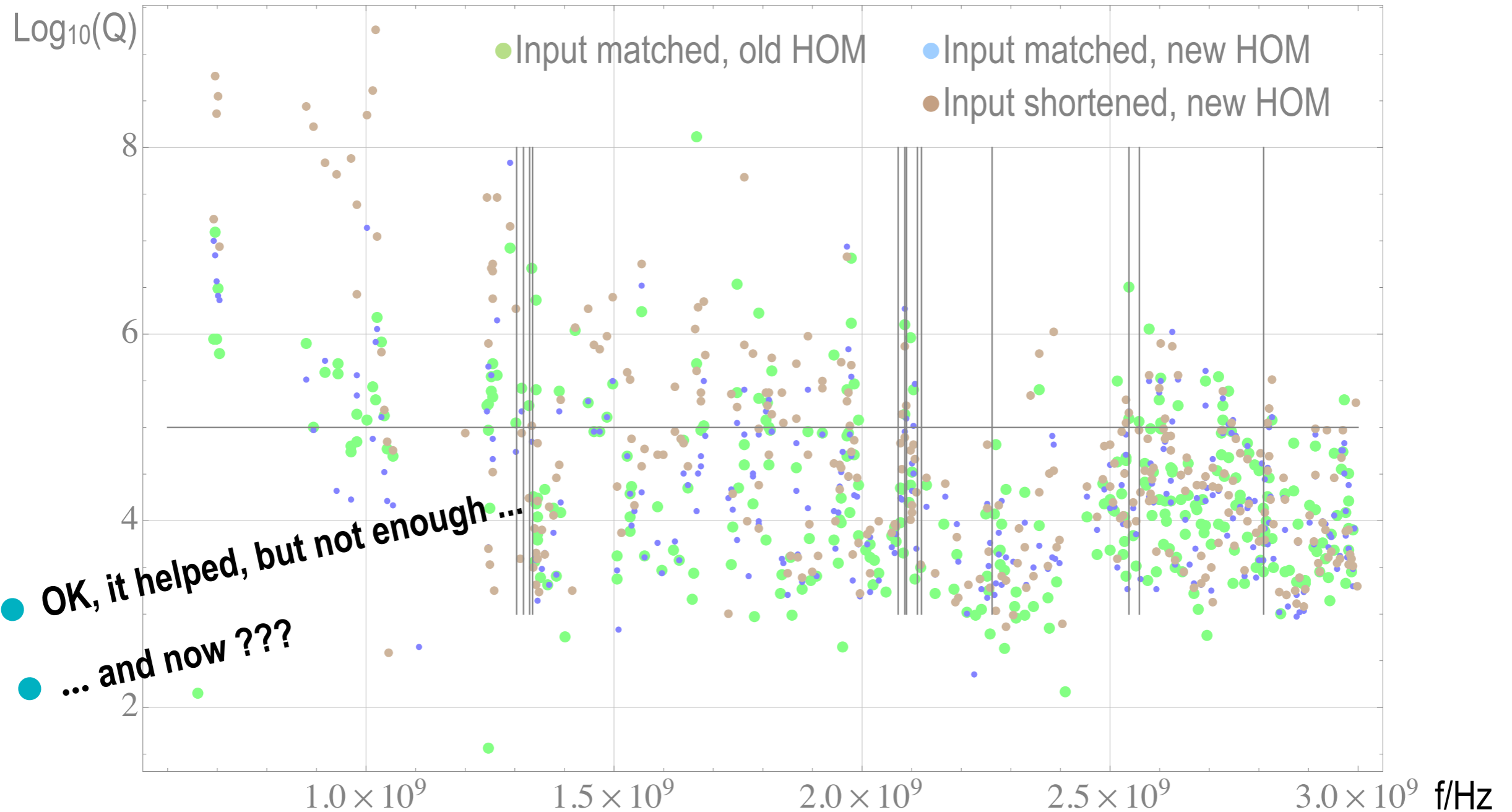


## Q-value distribution versions ④ and ⑤



vertical lines: prominent monopoles according J. Plouin: SPL cavity design by CEA-Saclay, 3rd SPL Coll.-Meeting, 11-13 Nov 2009, CERN

## Q-value distribution versions ④ and ⑤



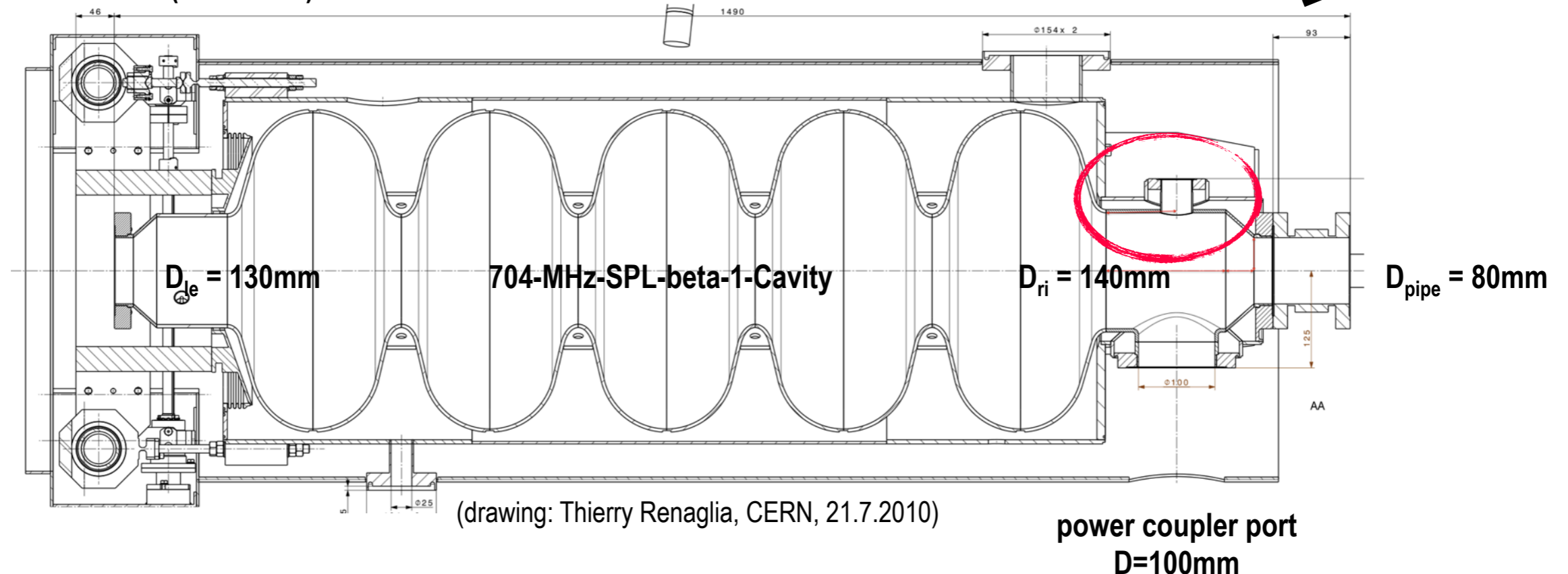
vertical lines: prominent monopoles according J. Plouin: SPL cavity design by CEA-Saclay, 3rd SPL Coll.-Meeting, 11-13 Nov 2009, CERN

## We were lucky:

- CERN decided (~June 2011) to increase HOM port diameter from 36 mm to 45 mm
- which also allowed an increased hook pipe diameter

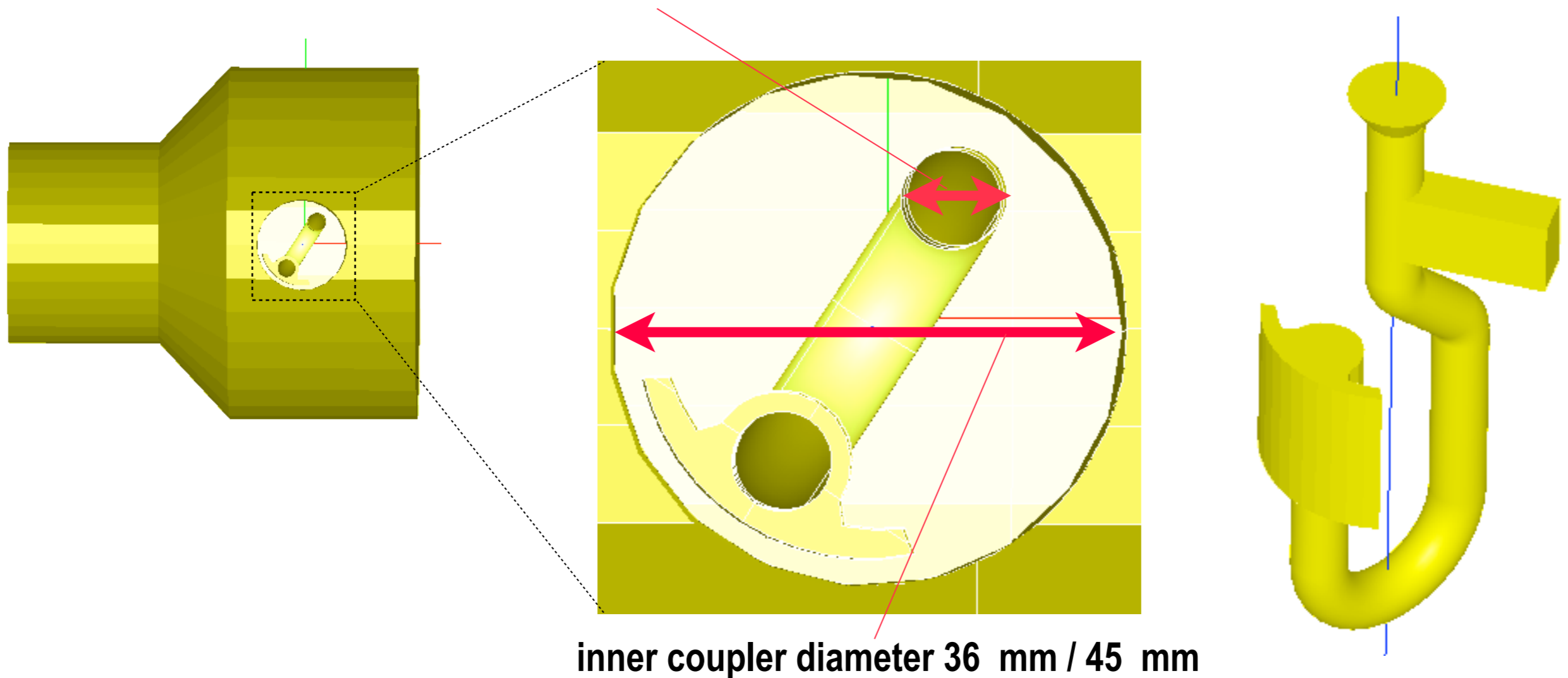
"left" side HOM coupler port  
rotated by 60° towards you  
(not drawn)

"right" side HOM coupler port  
opposed to power c., ~~D=36mm~~ 45mm

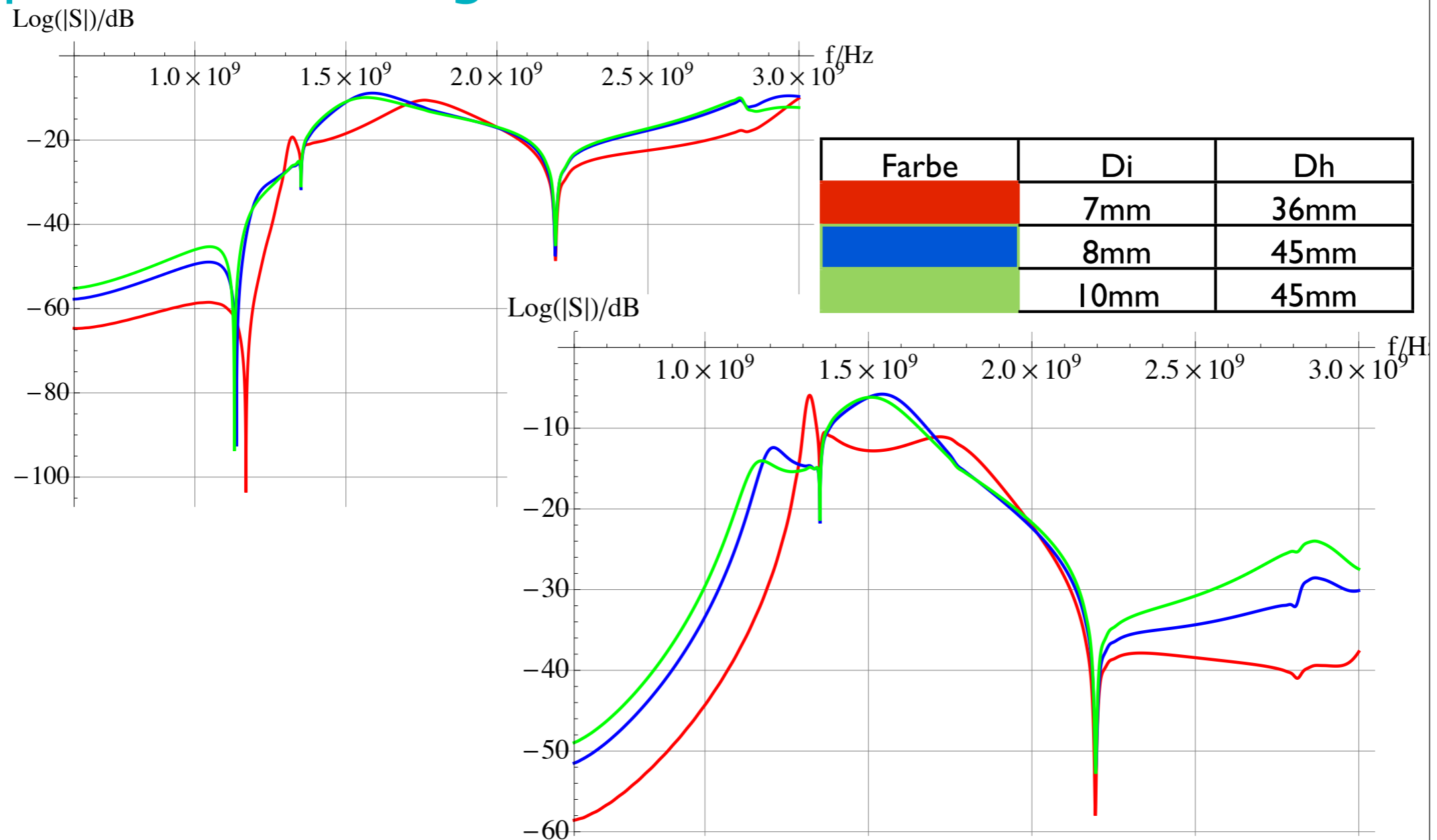


## Compare design alternatives

hook pipe diameter 7 mm @ 36 mm // 8 mm @ 45 mm // 10 mm @ 45 mm

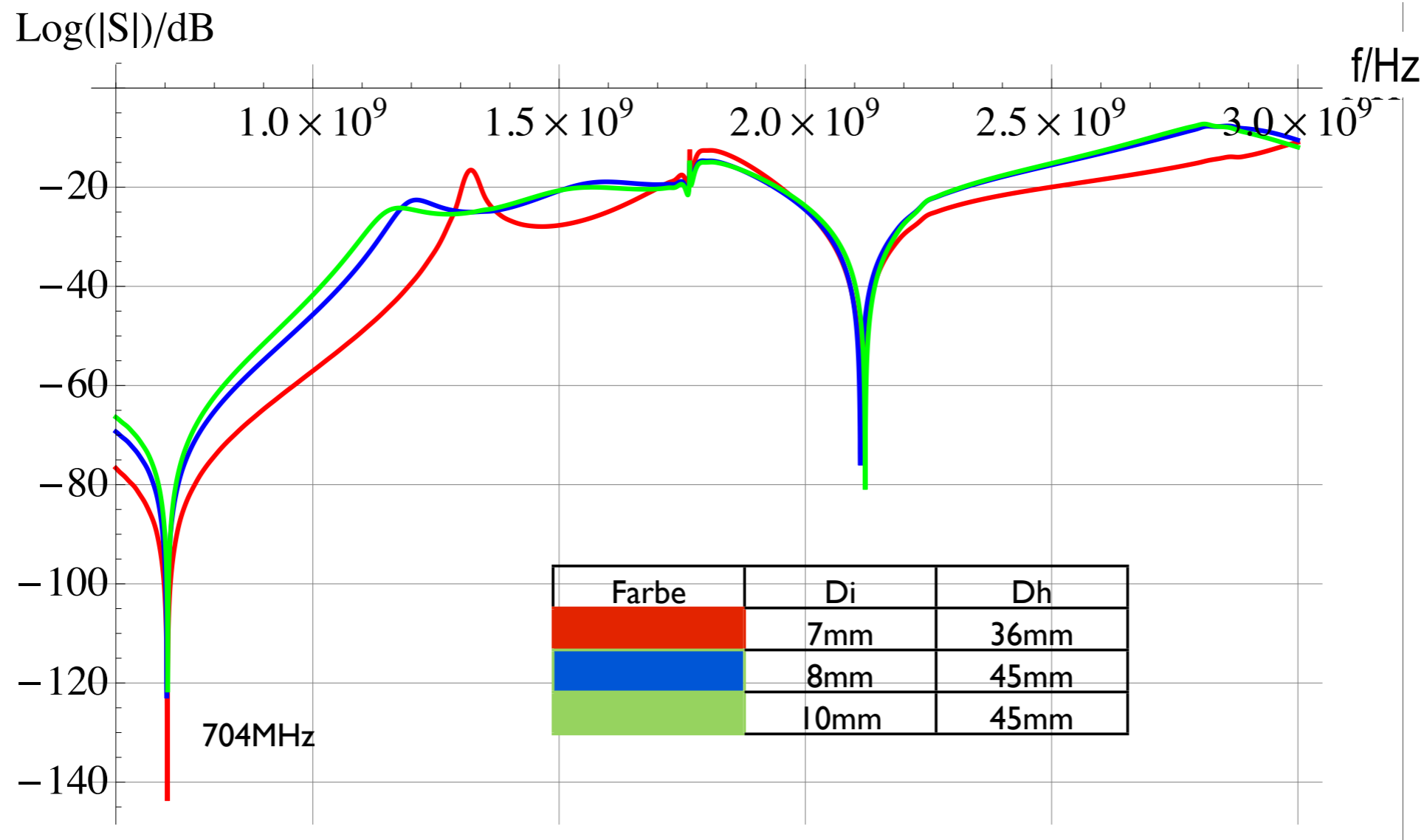


## Comparison of design alternatives TE<sub>11</sub>-coax



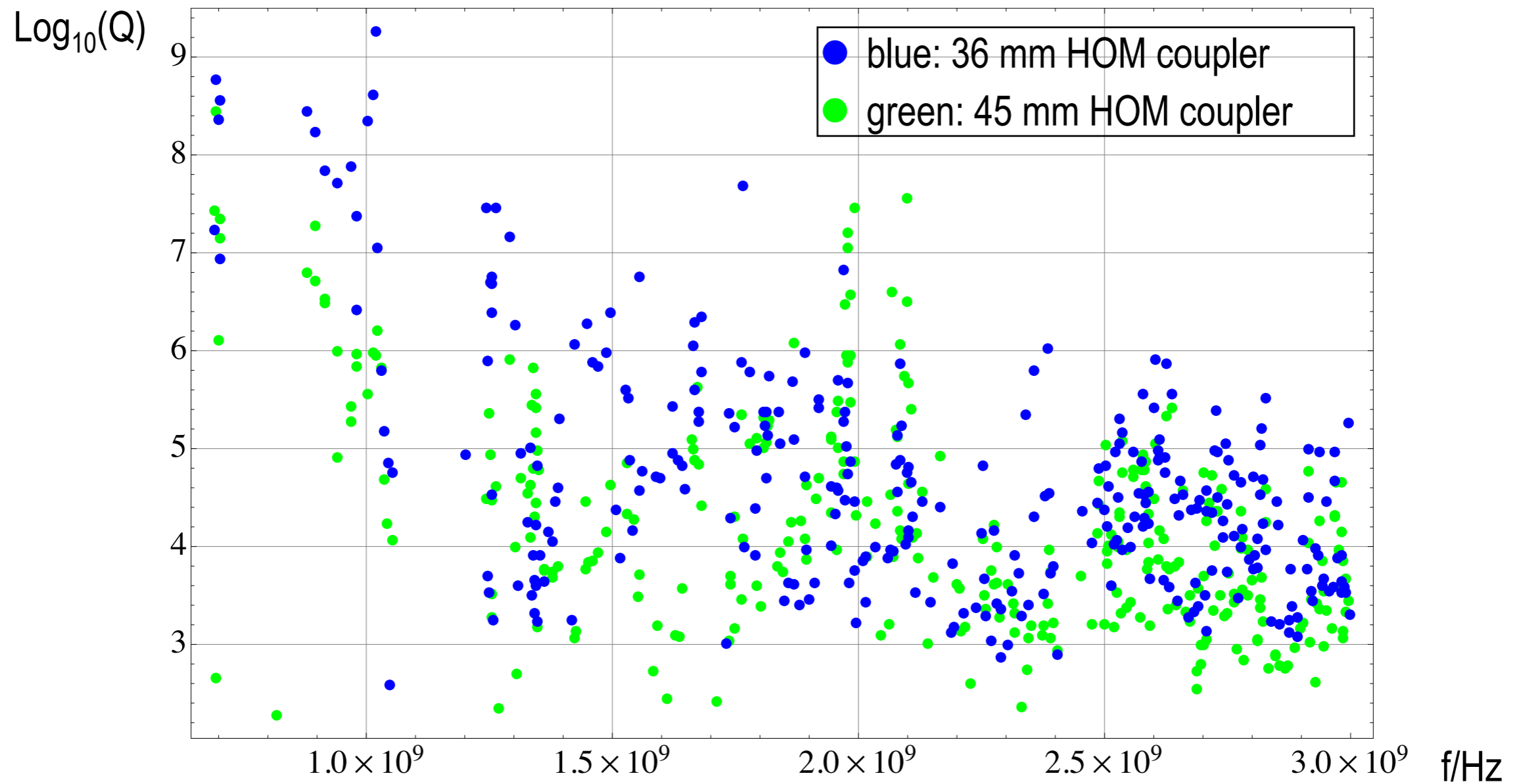
- 36 mm => 45 mm significantly improves coupling in the low frequency range
- 8 mm => 10 mm inner diameter has little effect

## Comparison of design alternatives TM<sub>01</sub>-coax



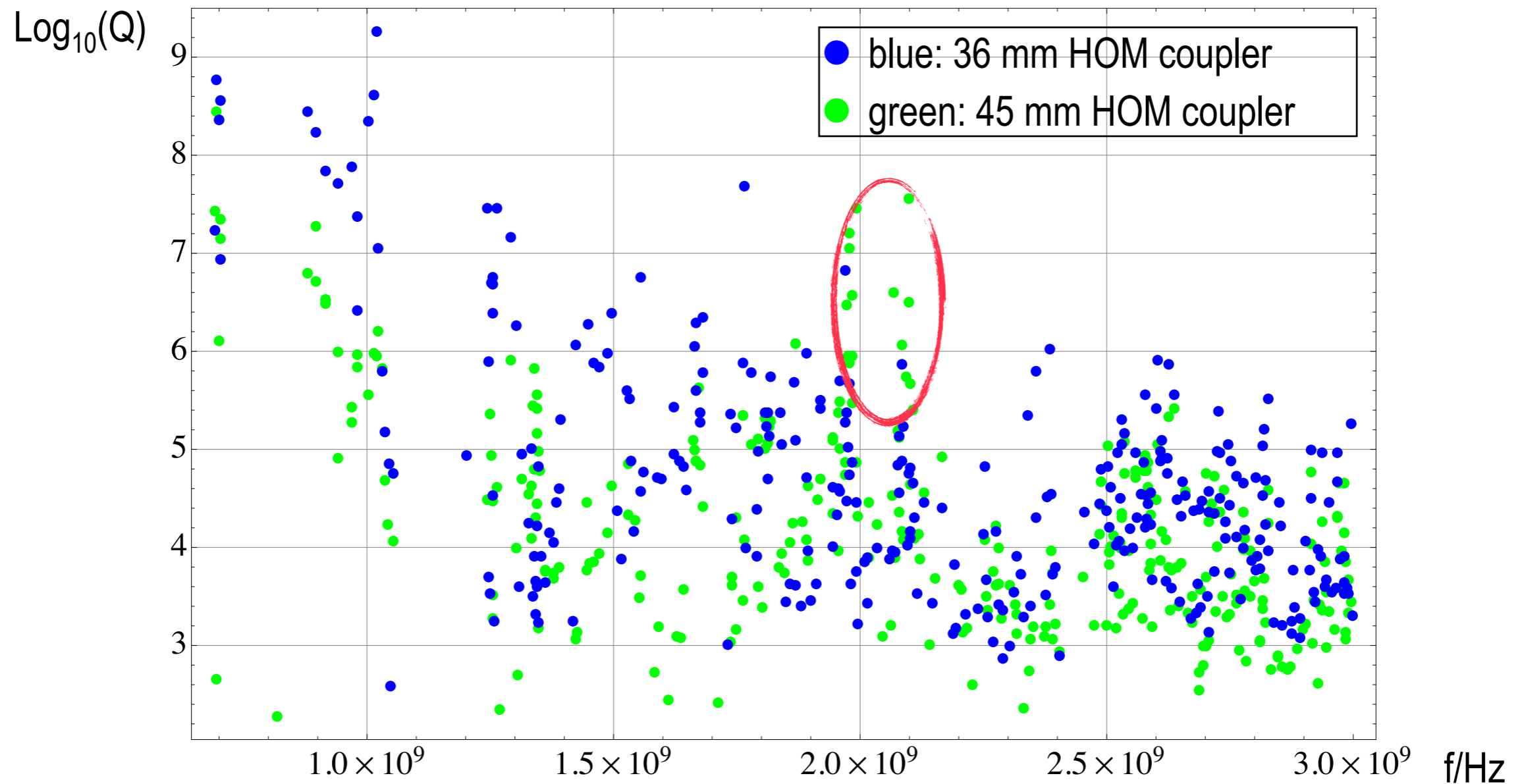
- 36 mm => 45 mm significantly improves coupling in the low frequency range
- 8 mm => 10 mm inner diameter has little effect

## $Q_{\text{ext}}$ : two 45/8-HOM-couplers, power coupler shortened



- significant improvement 36 mm  $\Rightarrow$  45 mm, around 1 GHz  $Q \sim$  one order of magn.

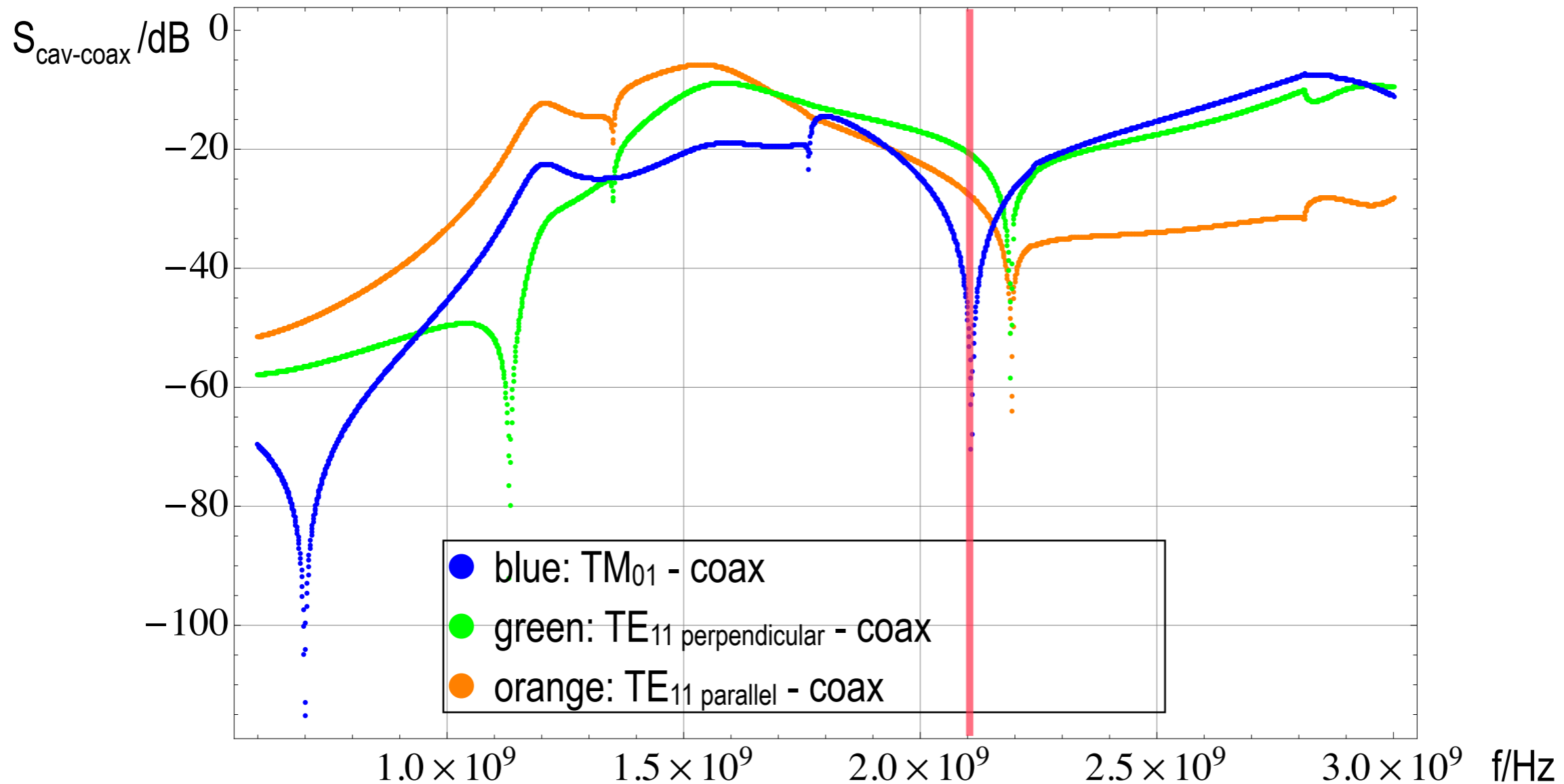
## $Q_{\text{ext}}$ : two 45/8-HOM-couplers, power coupler shortened



● but a few exceptions (not to be explained by failure of pole-fitting)



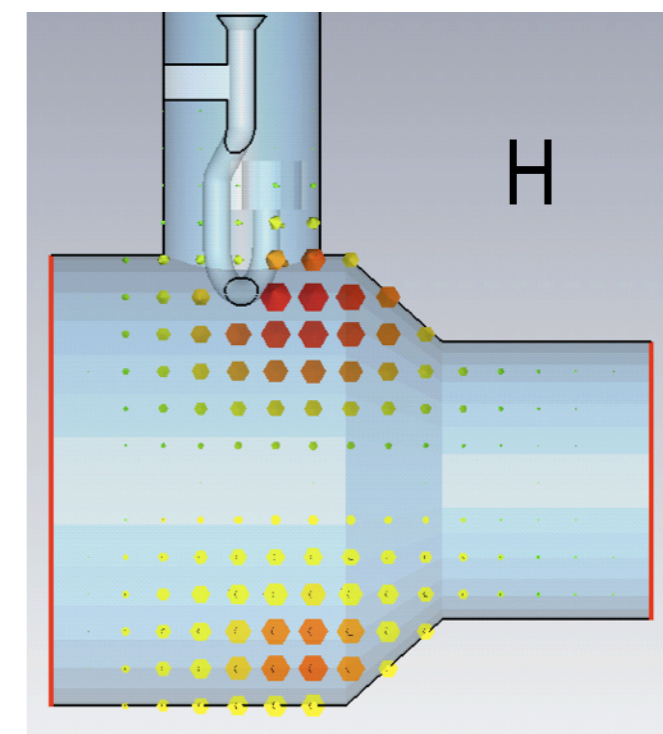
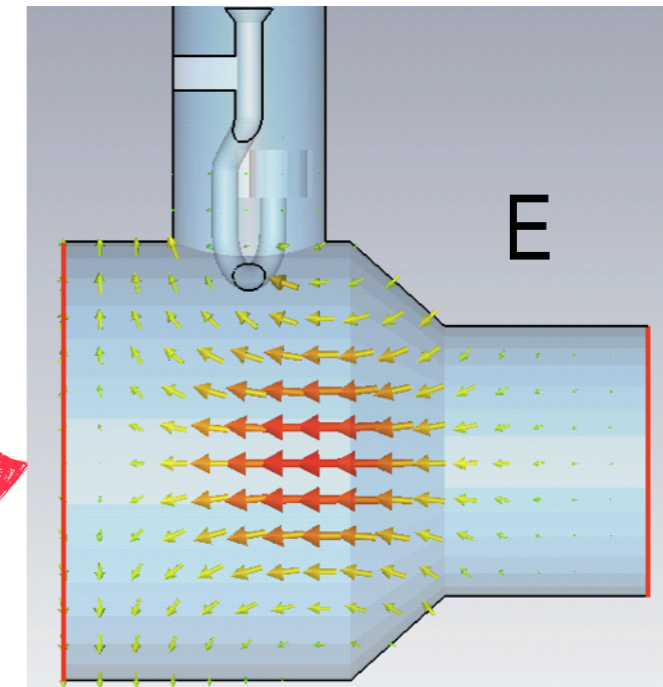
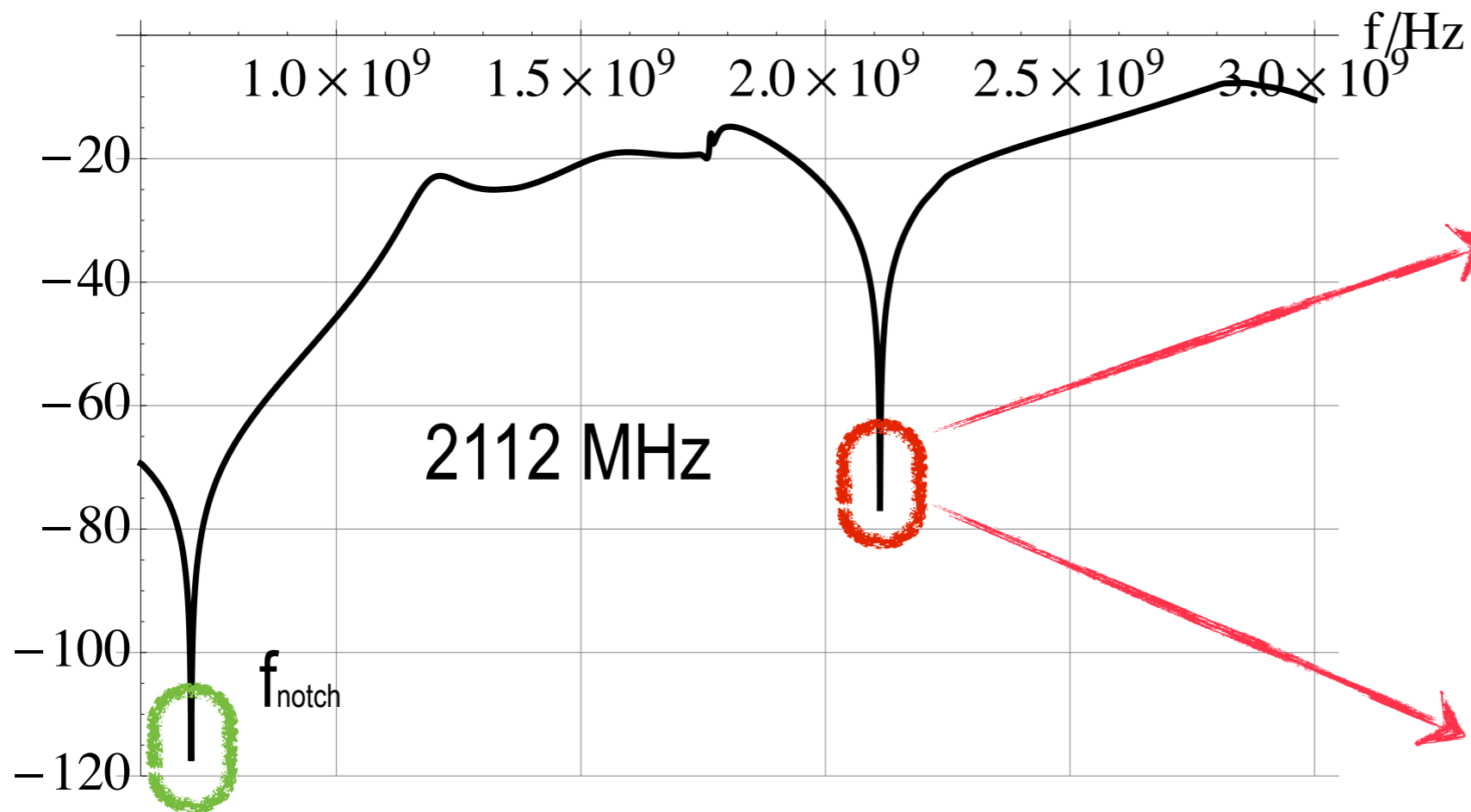
## Most important transmissions HOM-only-(45/8)



- **BAD LUCK:** 2nd TM<sub>01</sub> notch @ **2.112 GHz = 3.00 x 704 MHz**
- does NOT apply to HOM-Input-coupler

## Additional resonances in the coupler section $TM_{01}$

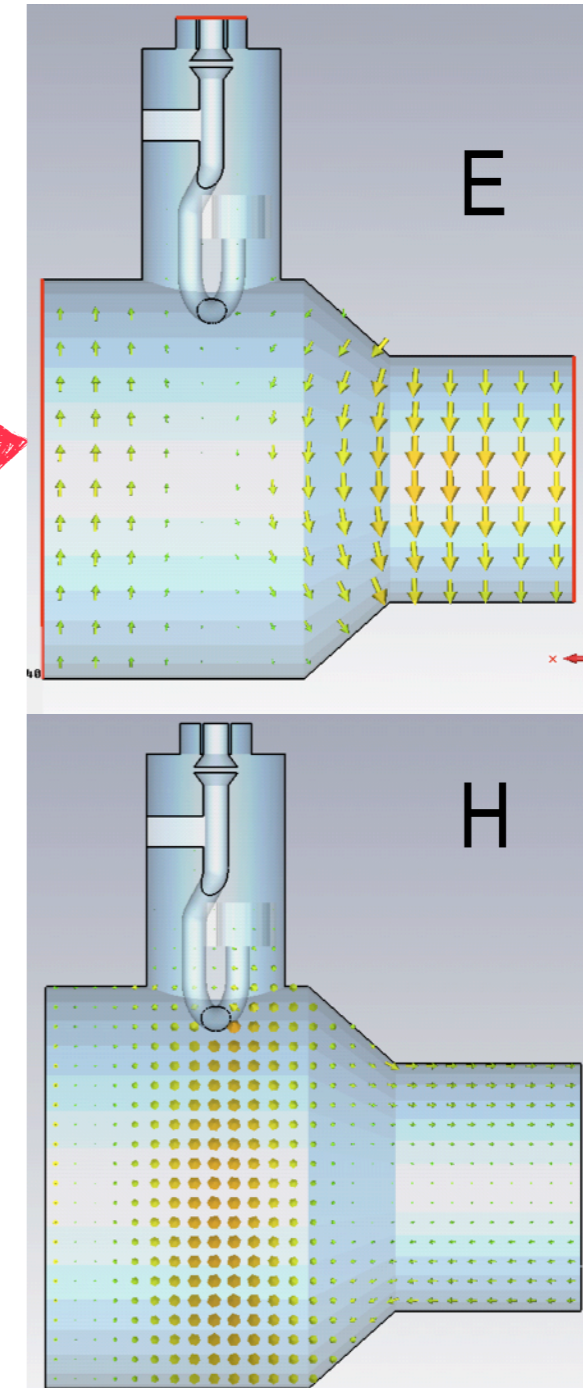
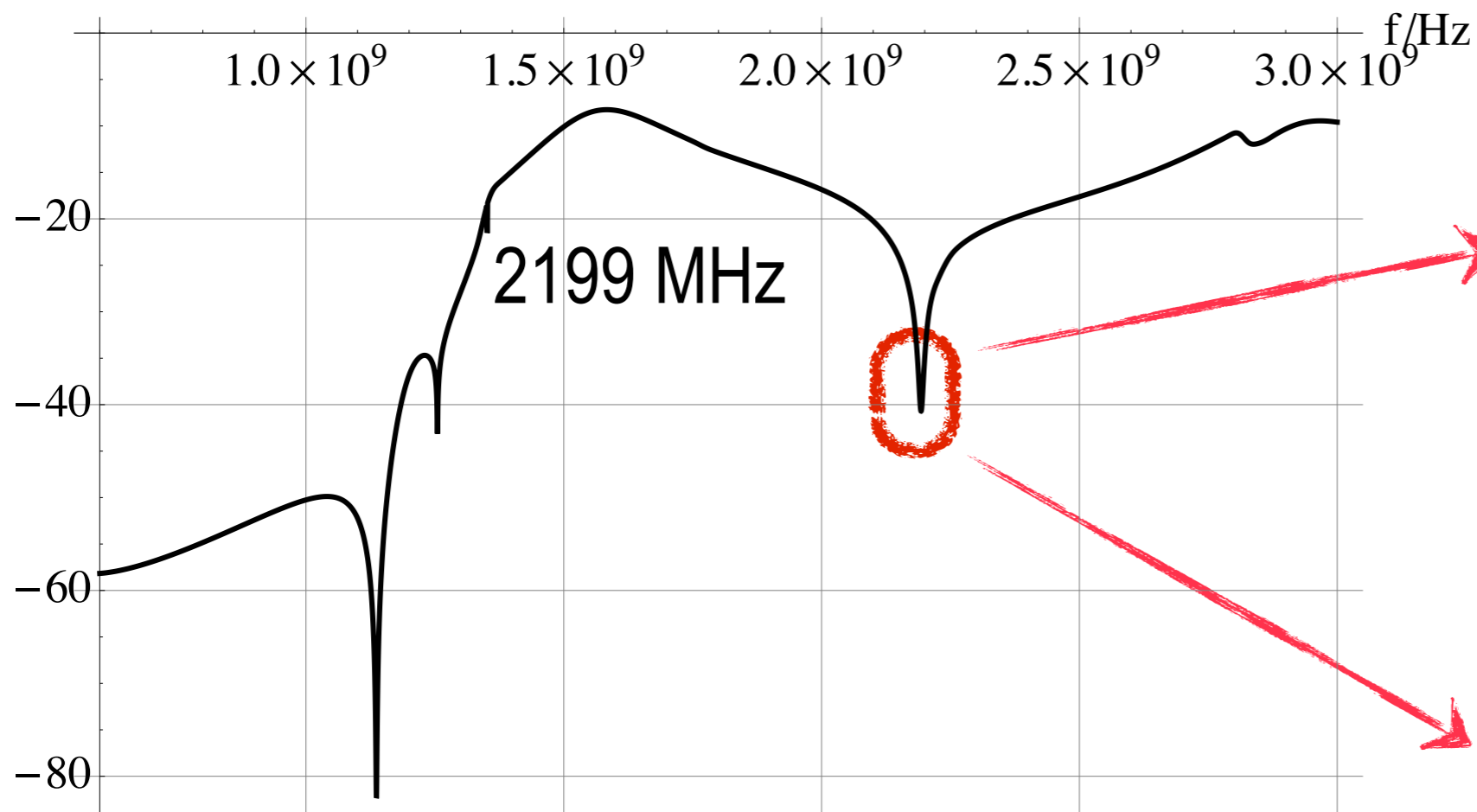
$\text{Log}(|S_{3(1)2(3)}|)/\text{dB}$



Parasitical notch frequency determined by distance between taper and HOM coupler.

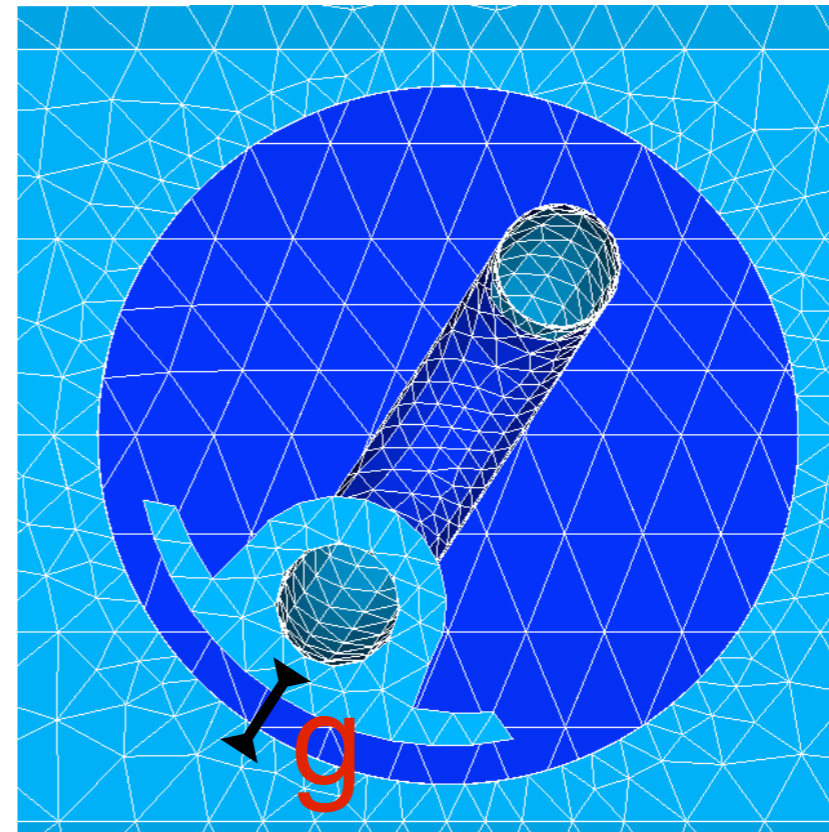
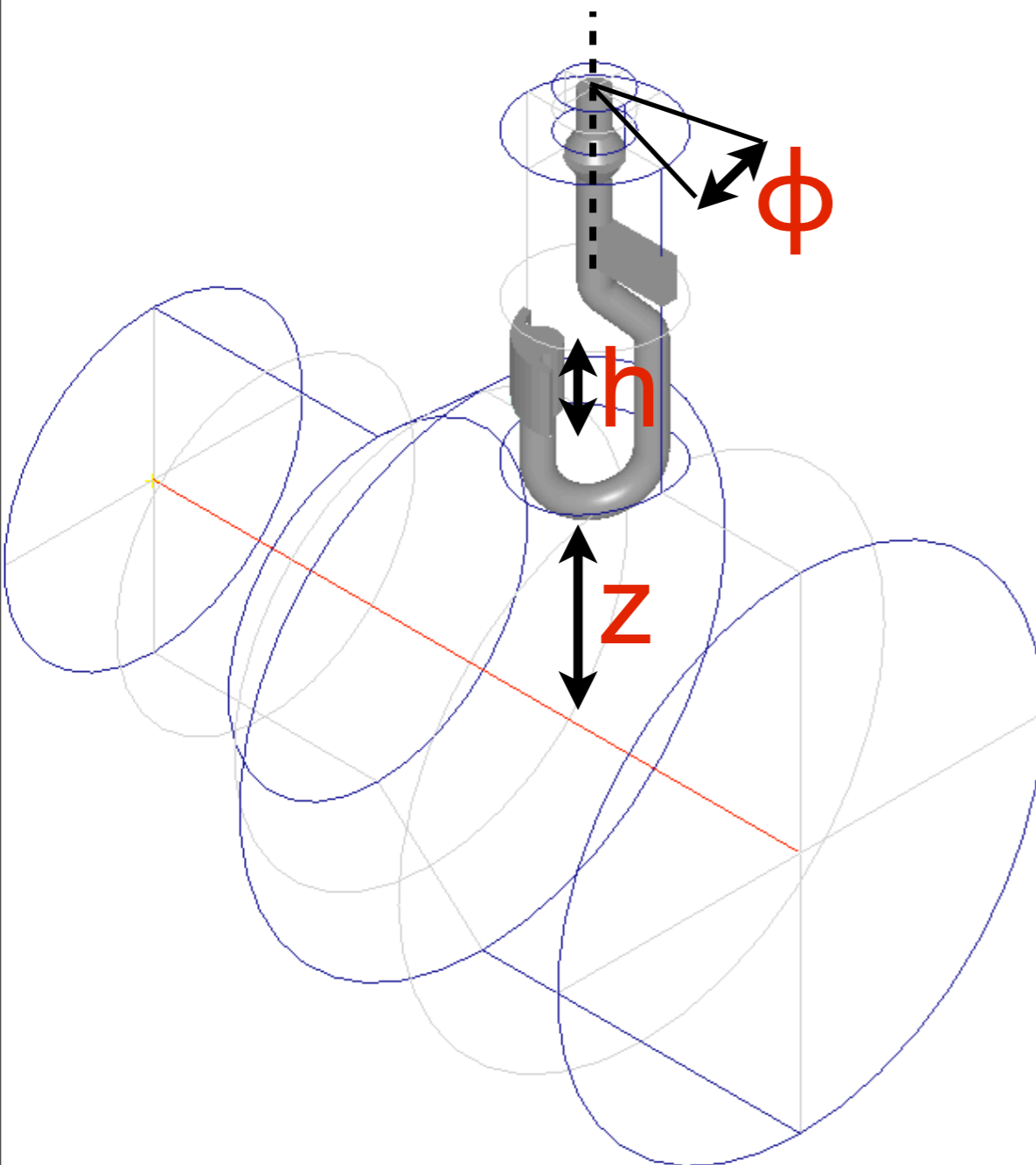
## Additional resonances in the coupler section $TE_{11}$ (not at a dangerous frequency)

$\text{Log}(|S_{3(1)2(1)}|)/\text{dB}$



Parasitical notch frequency determined by distance between taper and HOM coupler.

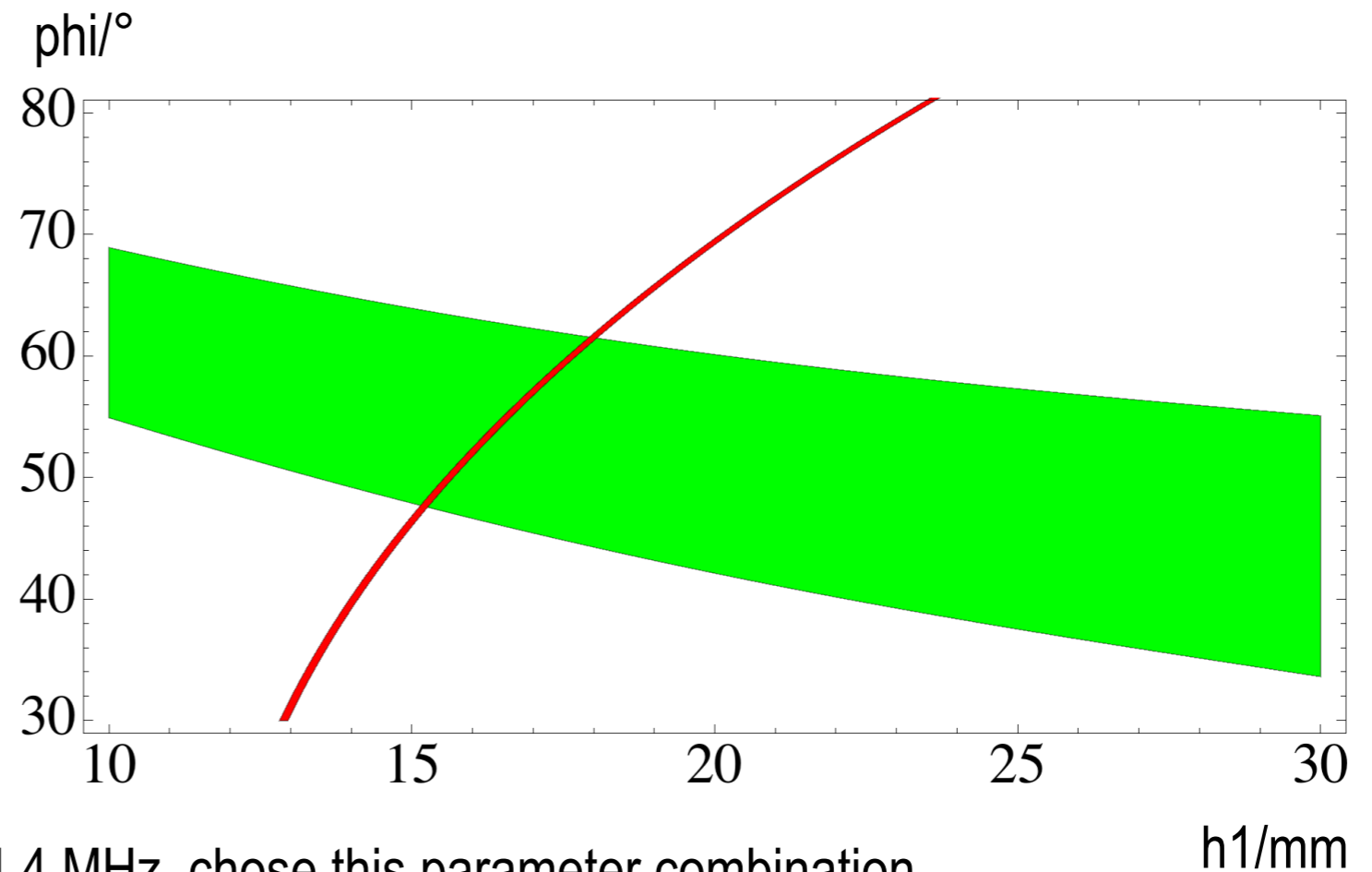
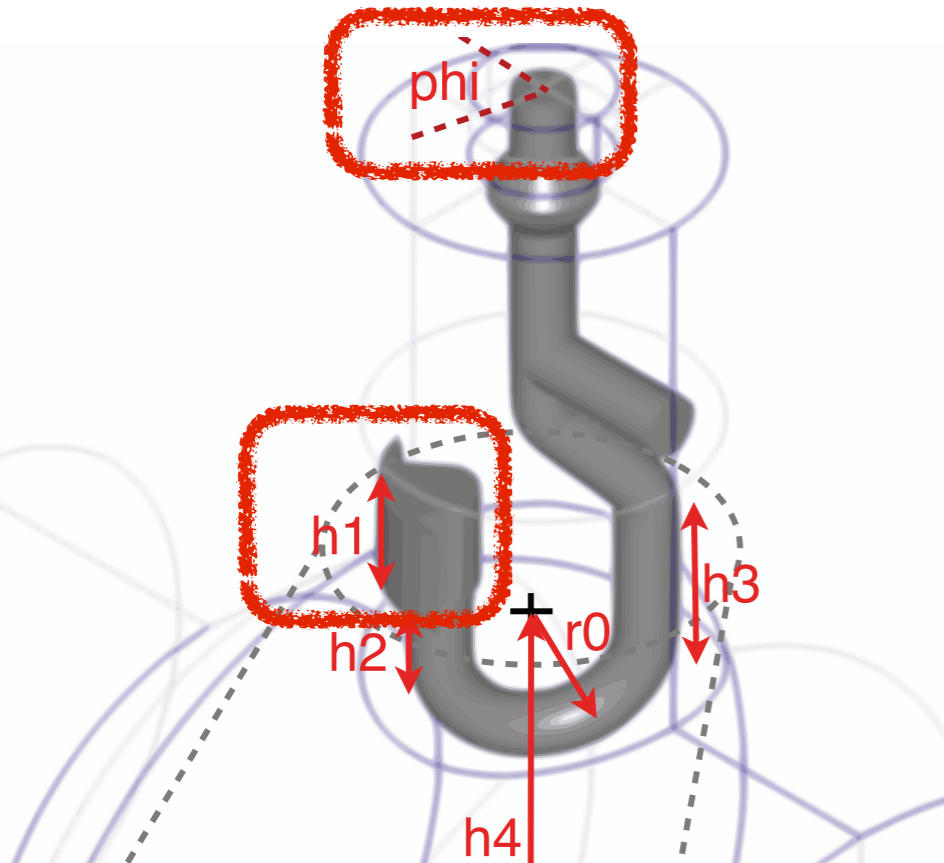
## Sensitivity of notch-filter frequency



$\partial f_n / \partial z$	9.6 MHz/mm
$\partial f_n / \partial h$	-1.75 MHz/mm
$\partial f_n / \partial g$	101.2 MHz/mm *
$\partial f_n / \partial \phi$	2.8 MHz/°

$$*f_n \propto \sqrt{g}$$

## Use angular sensitivity to avoid 2.112 GHz notch?

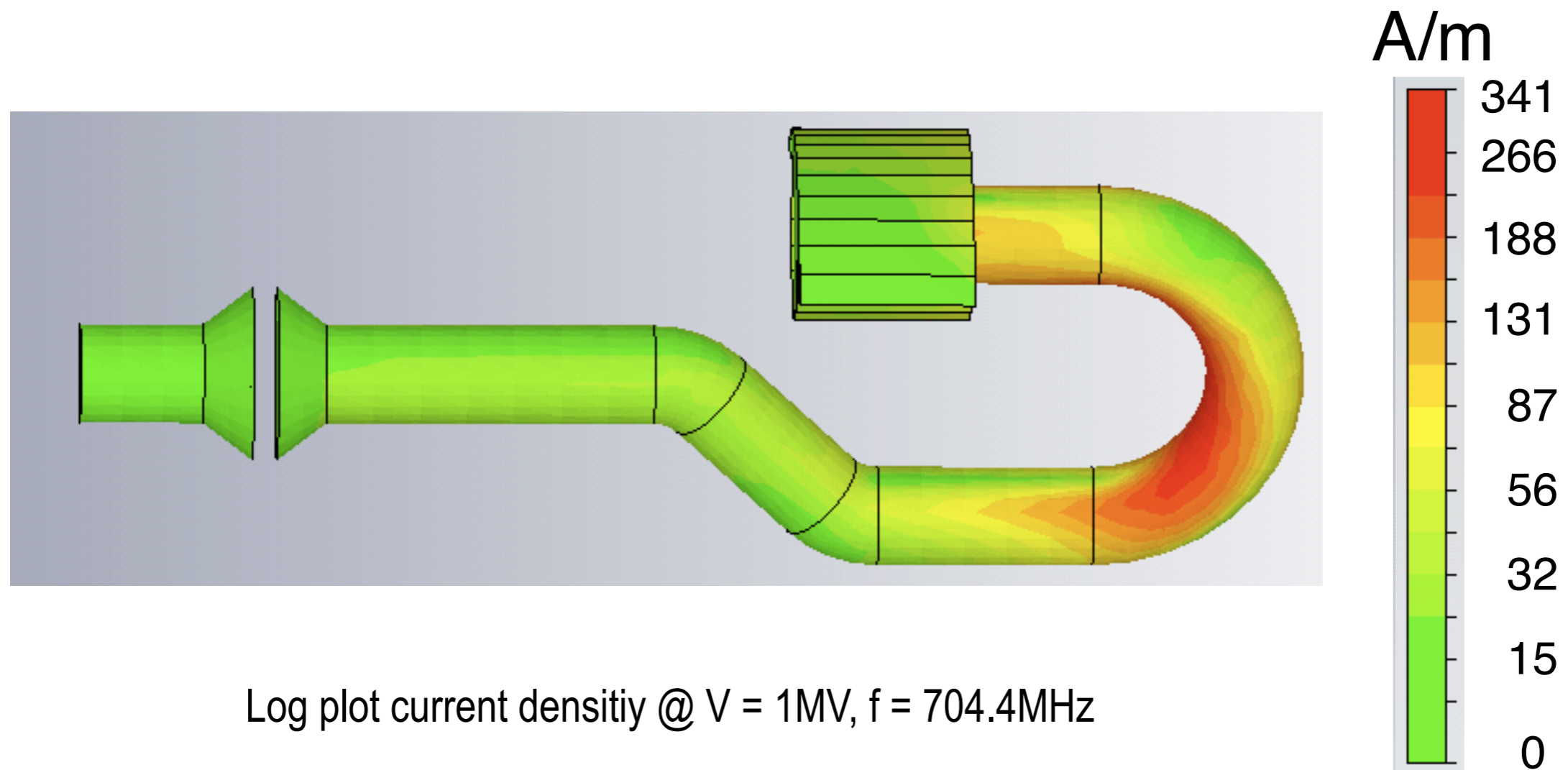


**red:** TM01-notch fixed @ 704.4 MHz, chose this parameter combination

**green:** BAD LUCK-notch @ 2112±20 MHz; avoid this combination

or put the port closer to the cavity (which also increases (all) couplings)

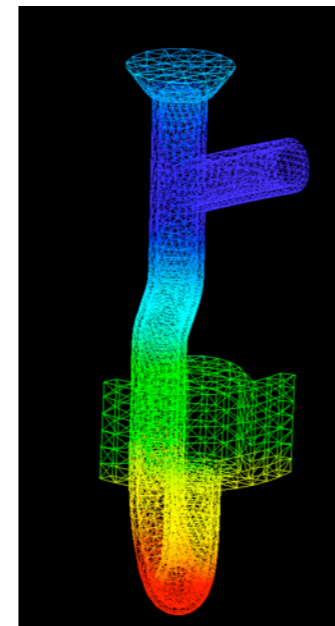
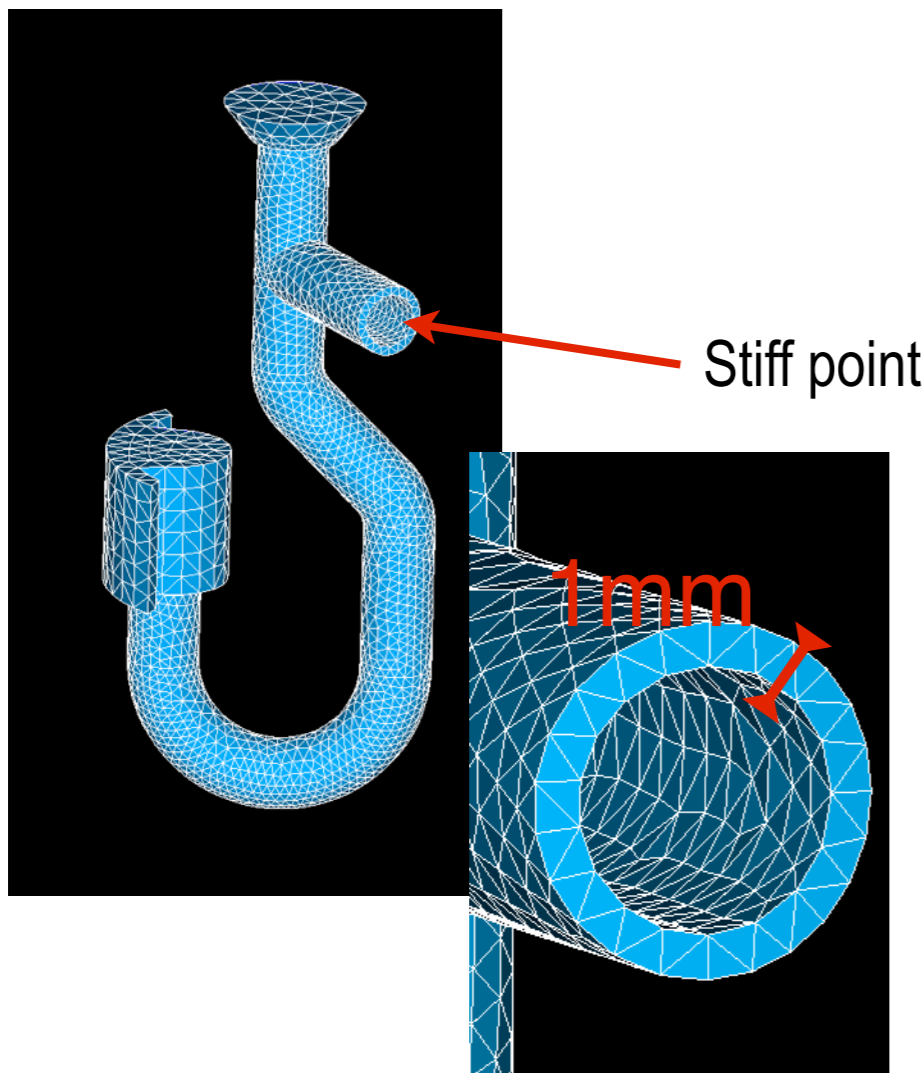
## Surface current density at $V_{acc} = 1\text{MV}$ , $f = 704.4\text{ MHz}$



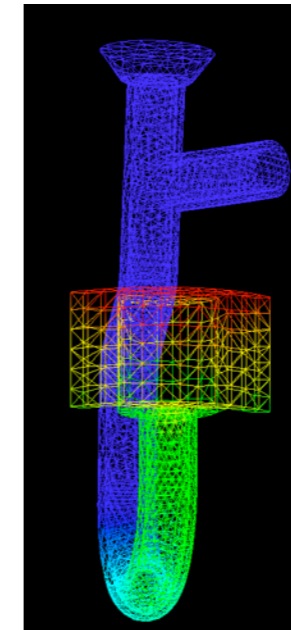
Log plot current density @  $V = 1\text{MV}$ ,  $f = 704.4\text{MHz}$

\*insider communication: current densities ~50% of (36/7)-design => power ~25%

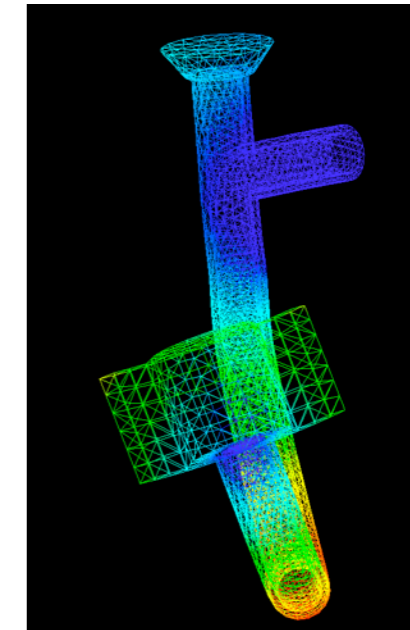
## Mechanical modes



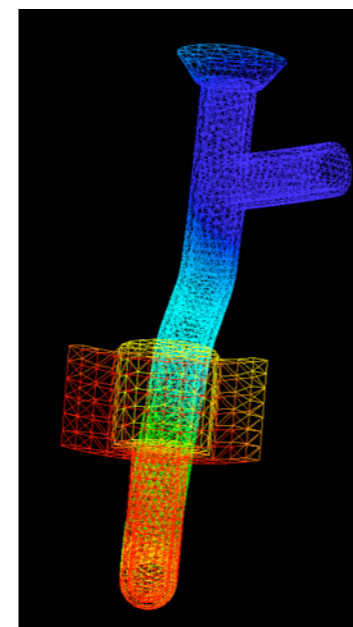
93 Hz



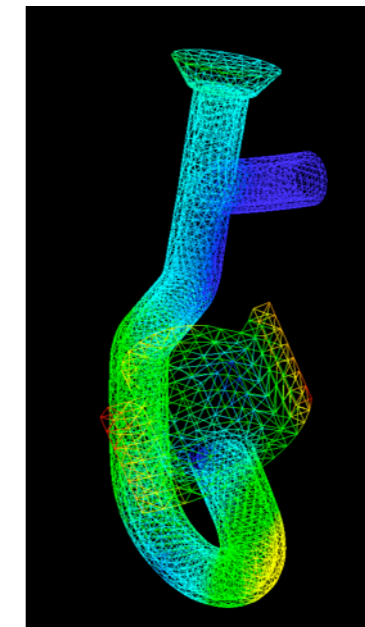
106 Hz



150 Hz



194 Hz



439 Hz

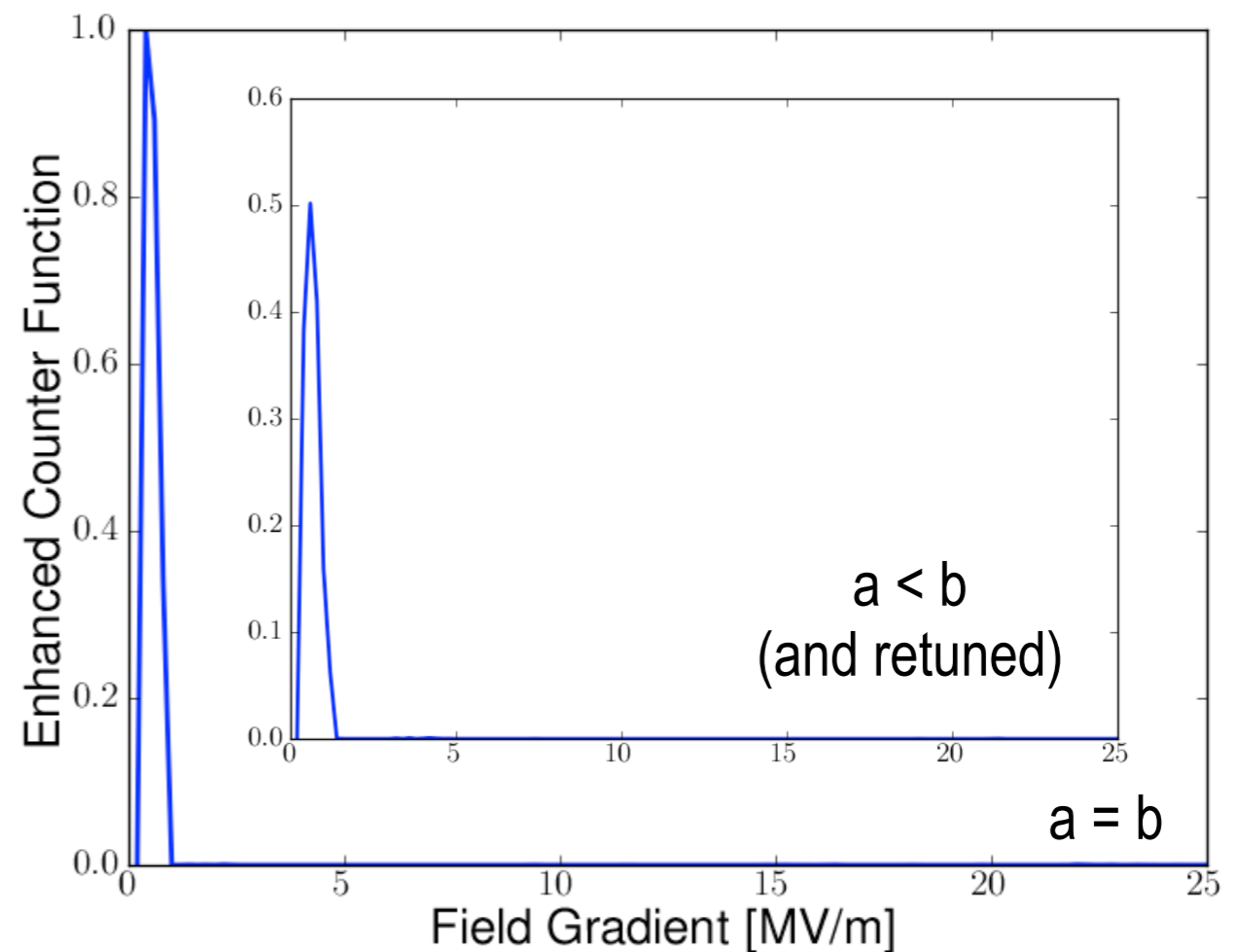
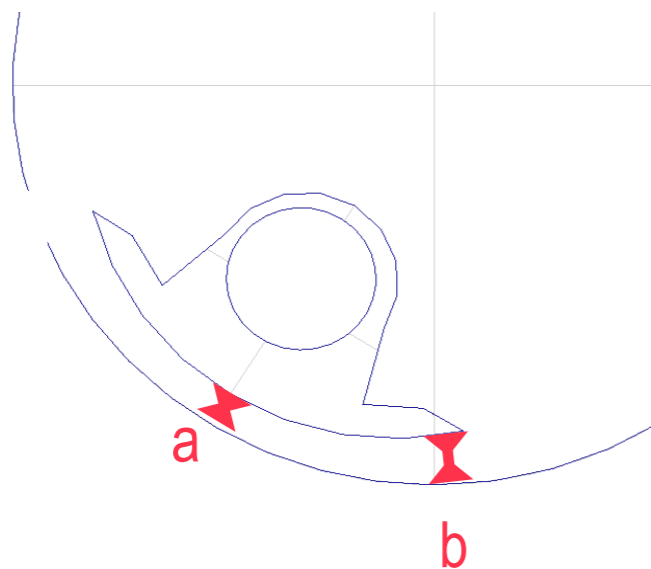
- simulated with Code Aster
- reach low frequency range, excitation spectrum?

## Hook-style coupler summary

- Design based on LEP-, LHC-, Saclay- ideas ...
- ... and "mountability" was one primary concern ( $\Rightarrow$  no fixed connection downside to flange)
- Step by step coupling was improved to  $Q < 10^6$ , ( $10^5$  ?) everywhere (except  $\sim 900$  MHz) ...
- ... exploiting the 45 mm port diameter.
- fundamental mode notch based on combined directional coupler-/filter effect and attaches TM<sub>0</sub> only  $\Rightarrow$  sensitive to orientation (problem or property?  $\Leftrightarrow$  rotational tuning possible, but needed)
- cooling concept under discussion @CERN (hope for conduction, do not exclude active cooling)
- BAD LUCK-3\*704.4 MHz-notch can be avoided

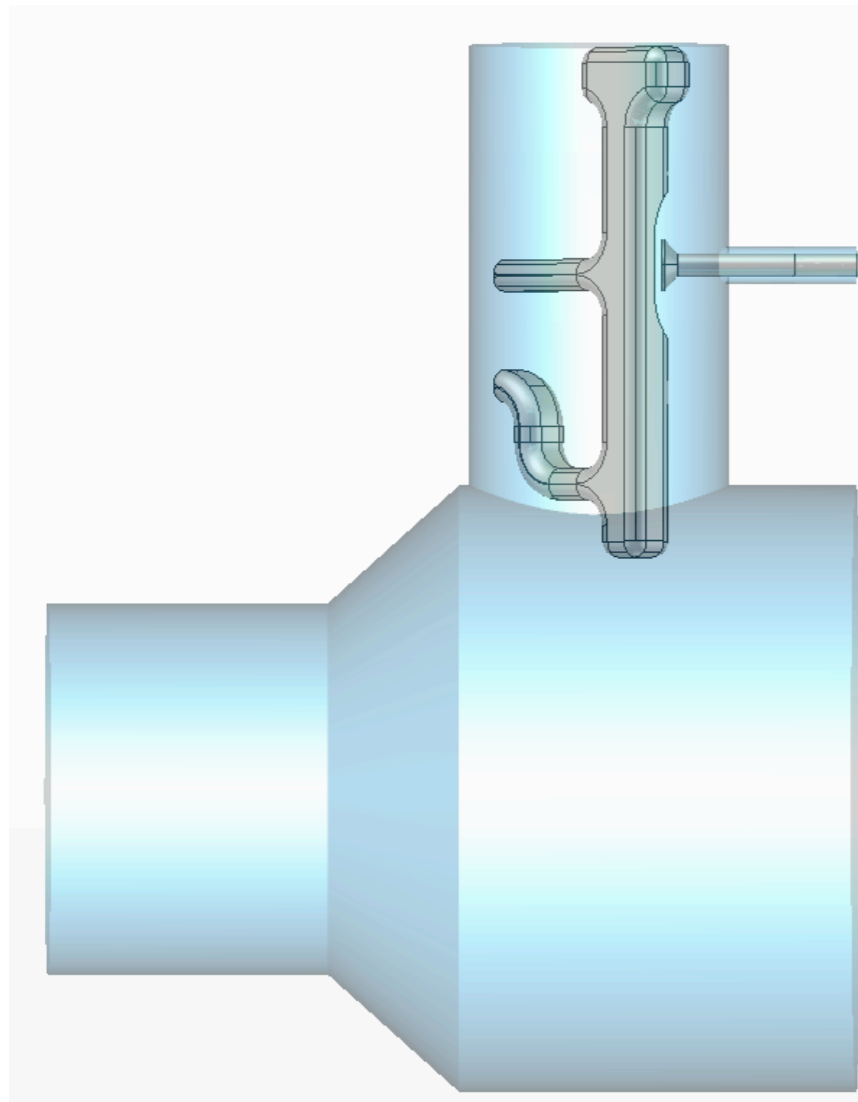


## Improved multipacting characteristics (following Guillaume's suggestion)

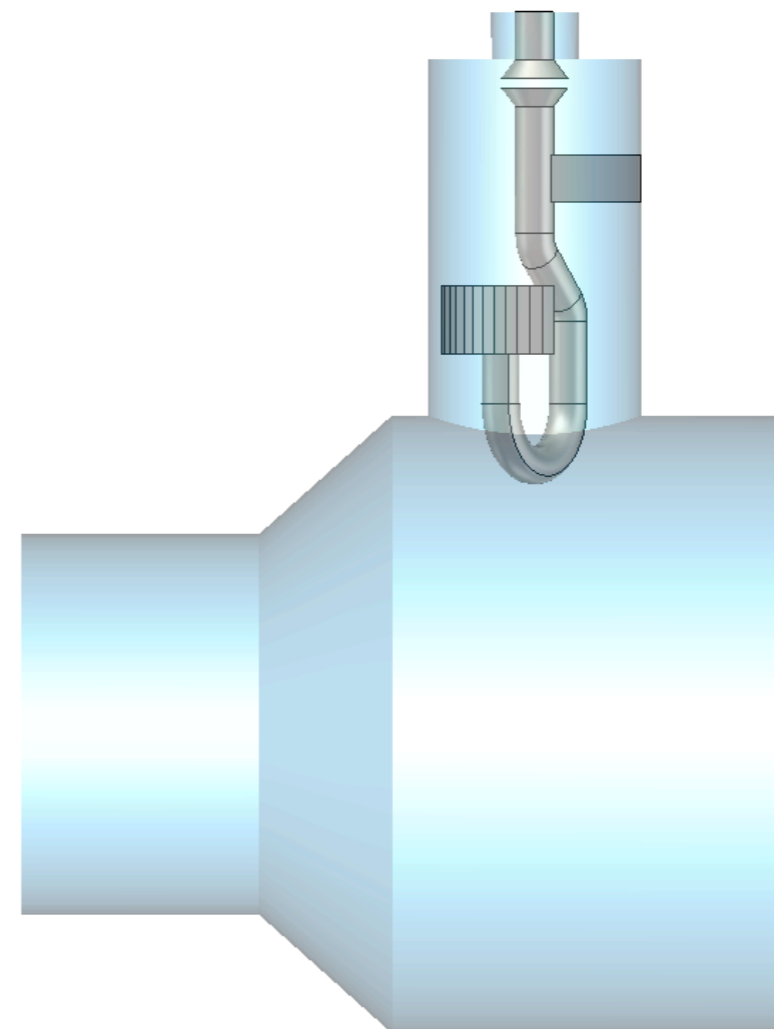


- Original UROS design has a severe (???) MP barrier around 1MV/m
- Slightly modified capacitive element (after retuning) reduces the ECF by a factor of 2

## Coupler olympics



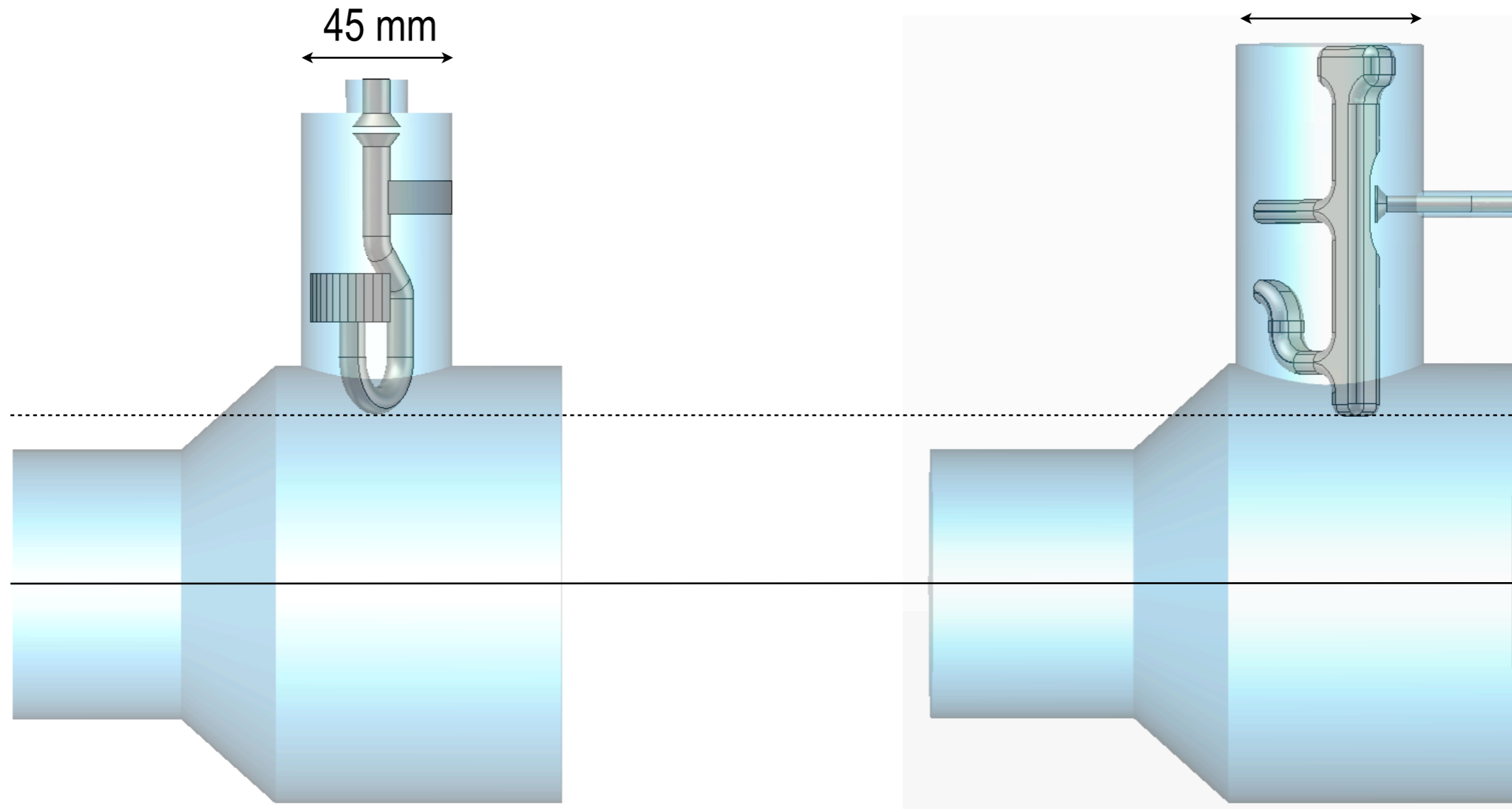
Scaled TESLA\*



Hook style

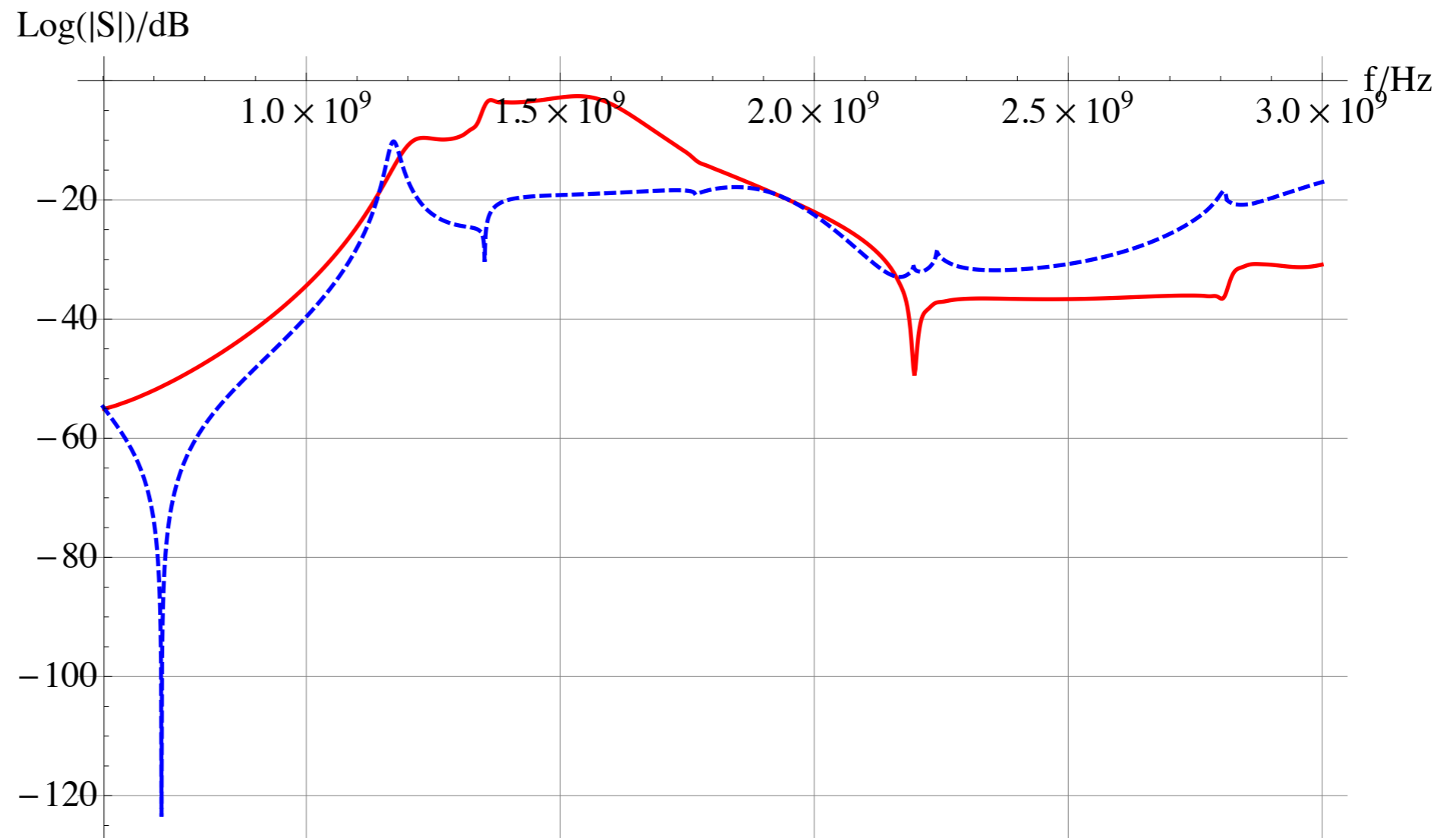
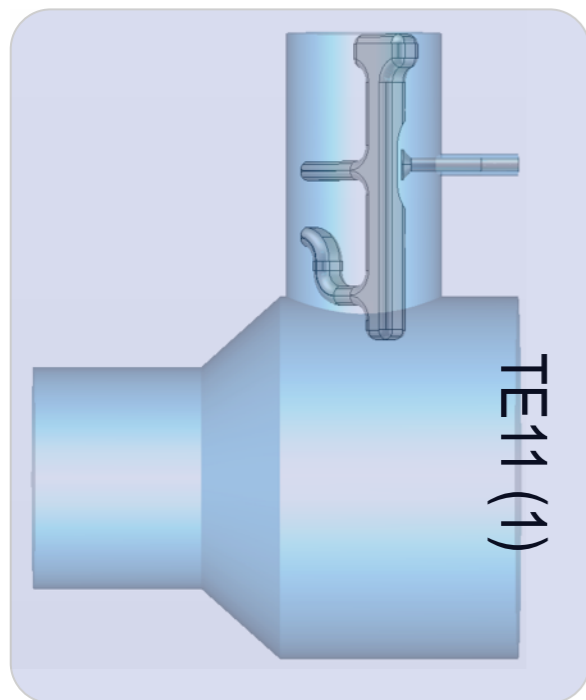
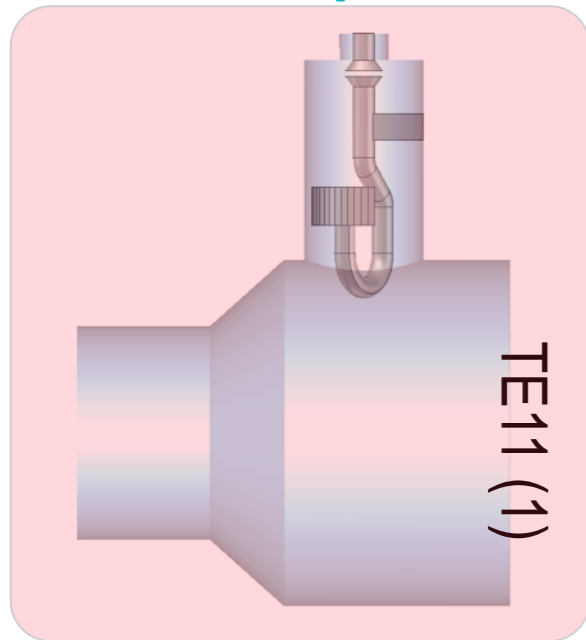
\*Coupler geometry courtesy Rama R. Calaga: <http://rcalaga.web.cern.ch/rcalaga/704MHz/HOM.html>

## Slightly different "weight" classes

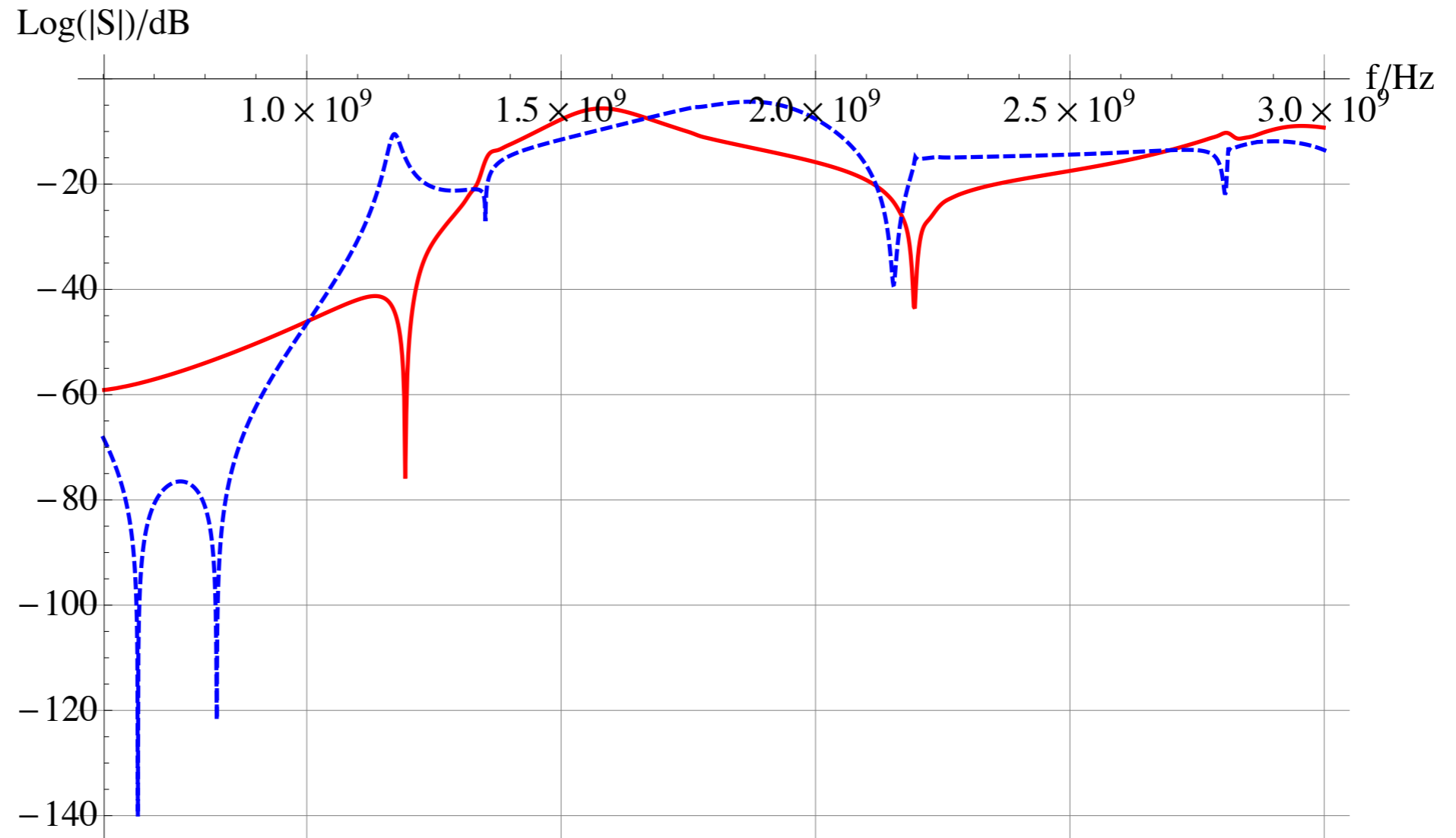
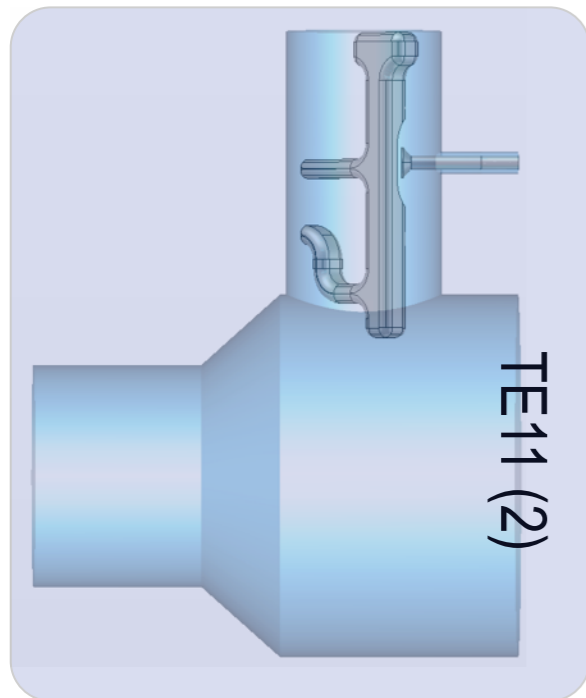
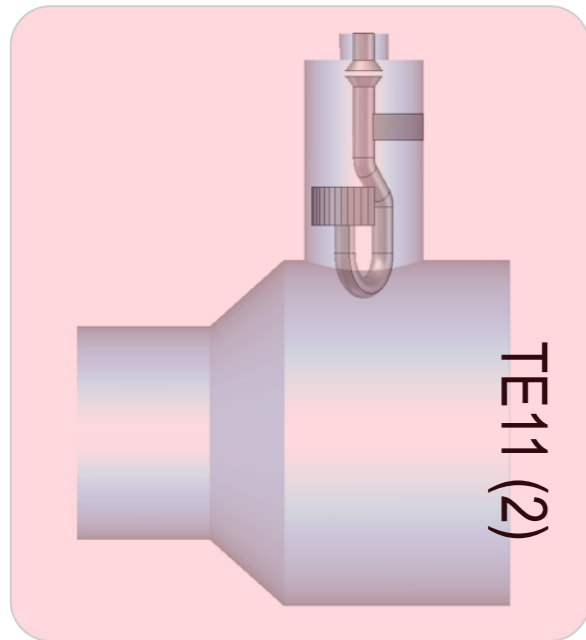


- Same distance to beam axis
- Different HOM-coupler diameter

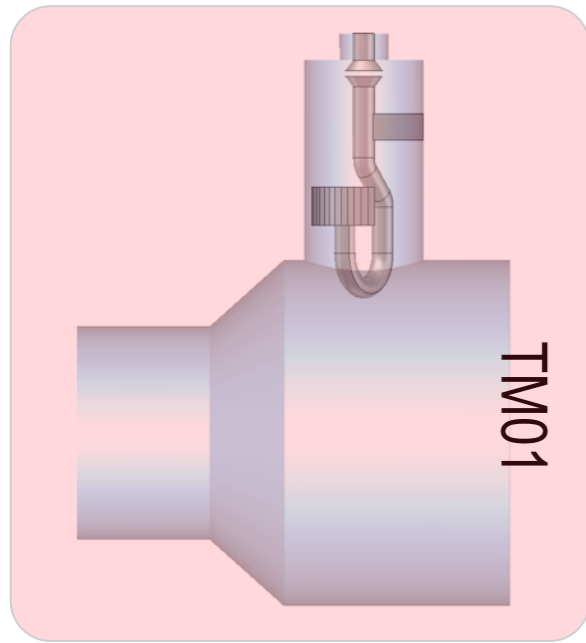
## Comparison of transmission characteristics $TE_{11}(1)$



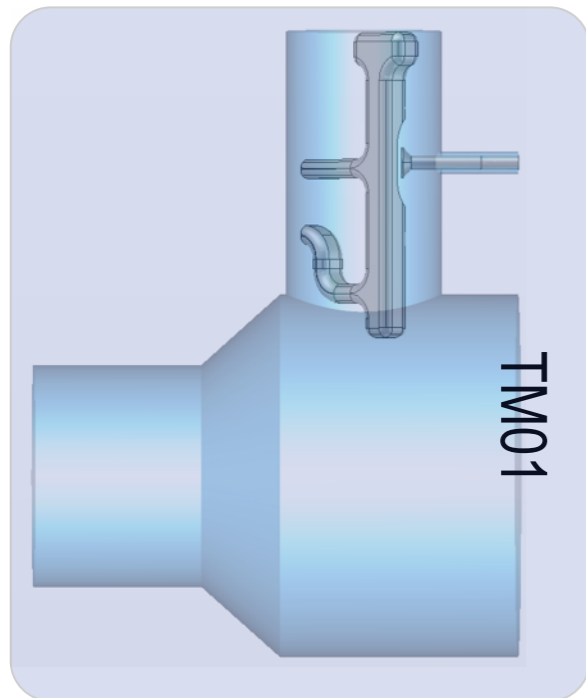
## Comparison of transmission characteristics $TE_{11}(2)$



## Comparison of transmission characteristics $TM_{01}$

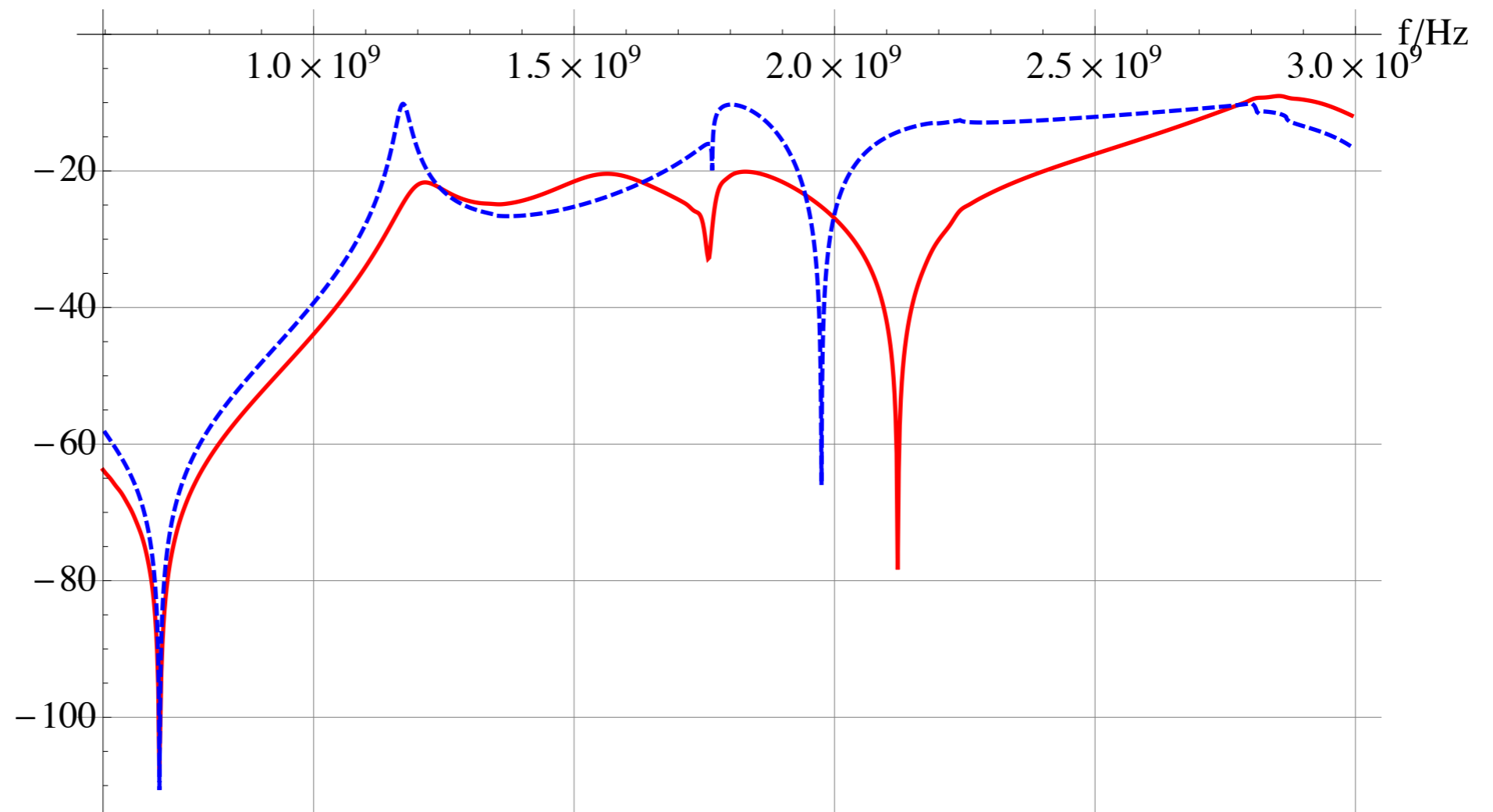


$TM_{01}$

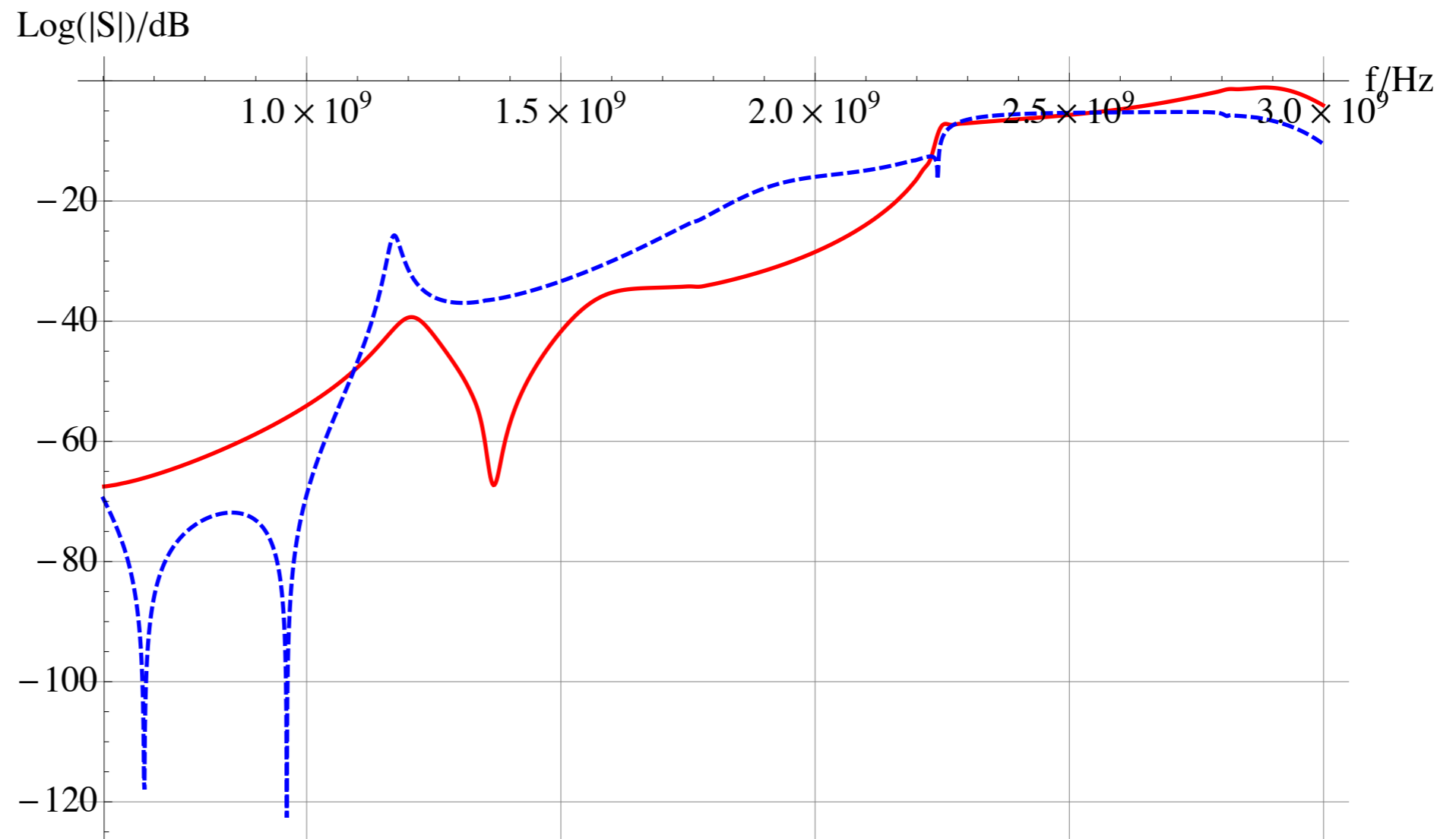
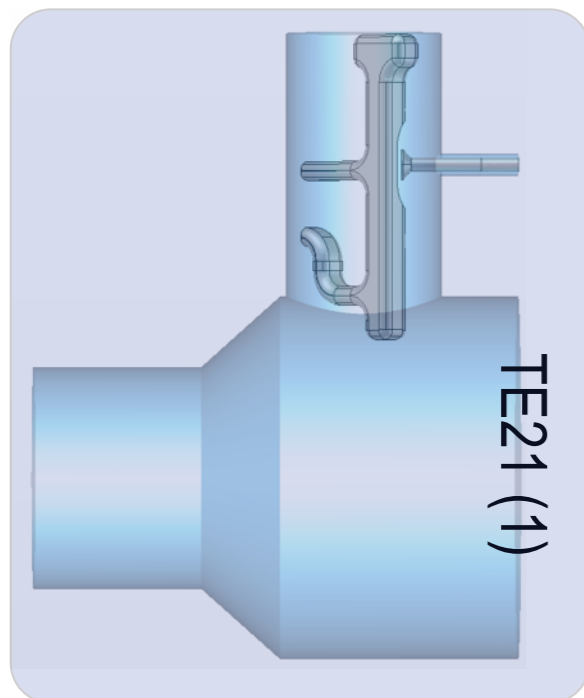
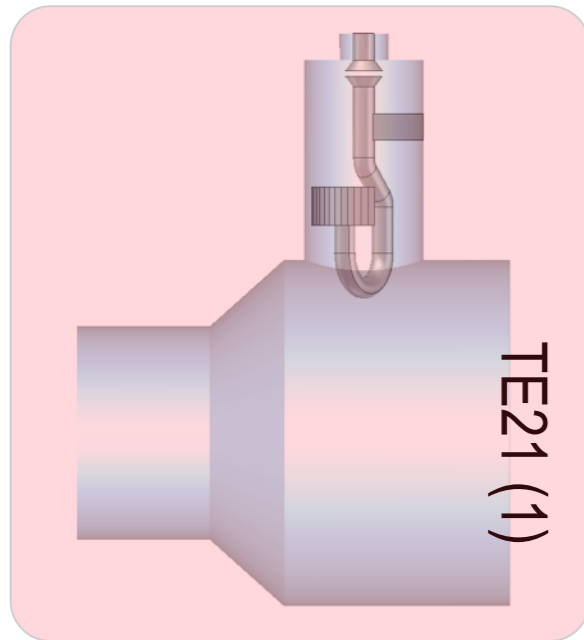


$TM_{01}$

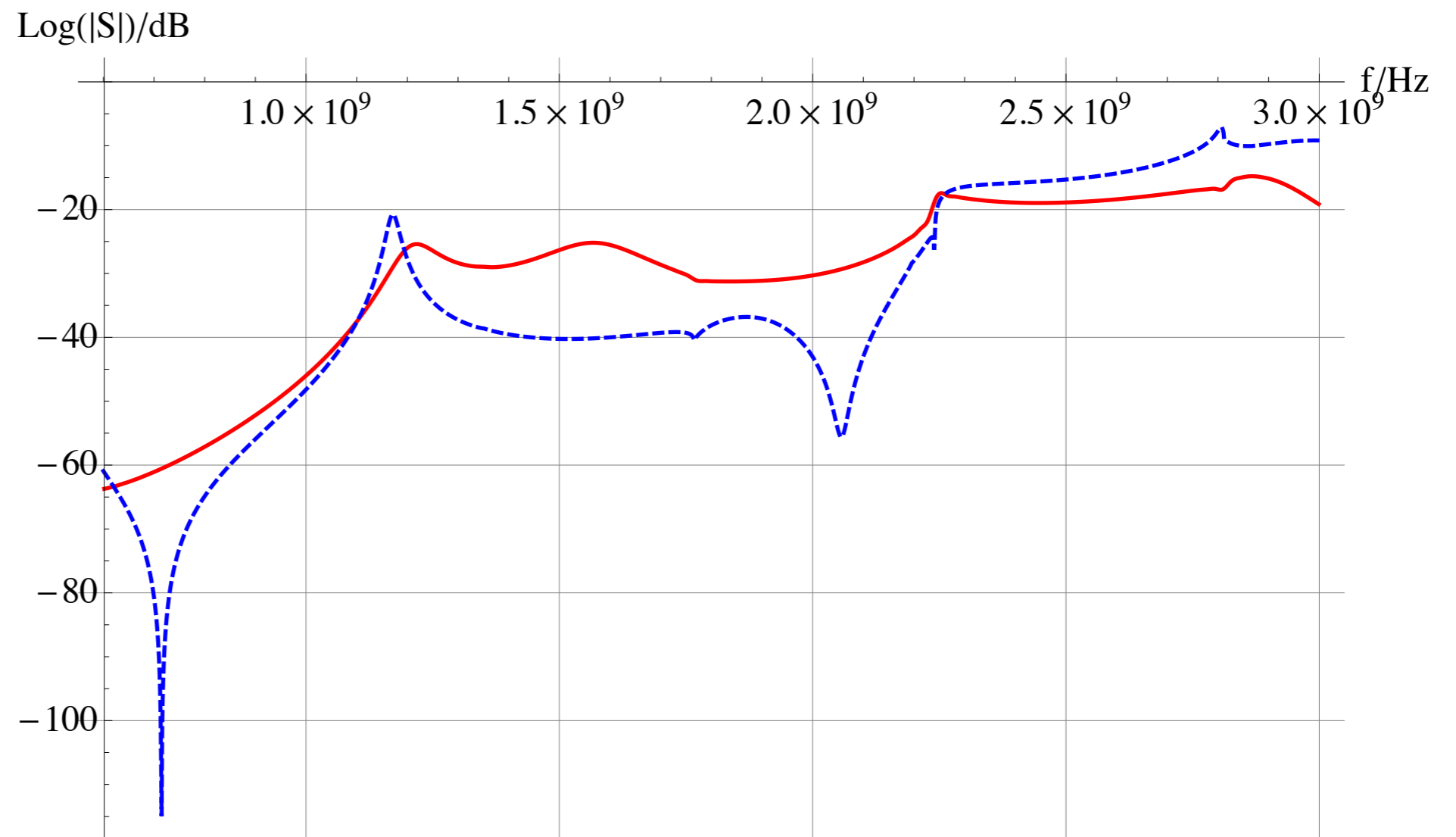
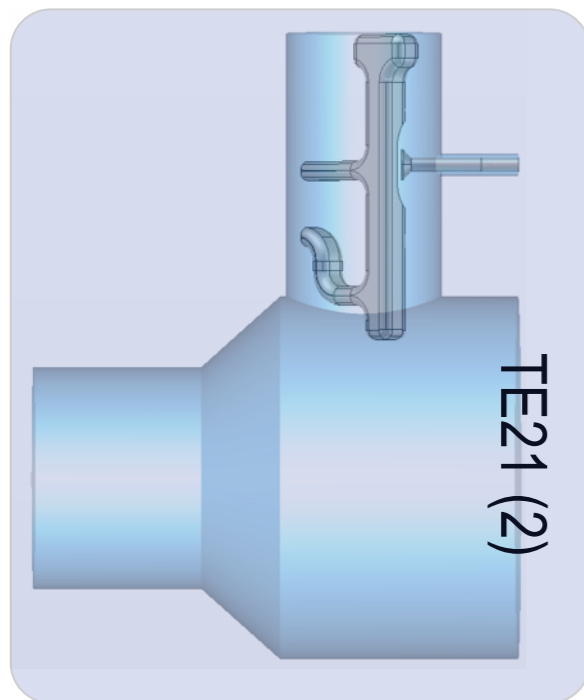
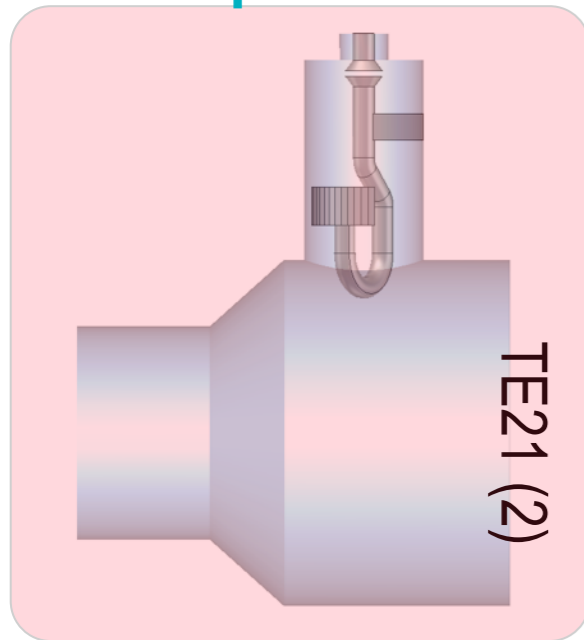
$\text{Log}(|S|)/\text{dB}$



## Comparison of transmission characteristics TE<sub>21</sub>(1)

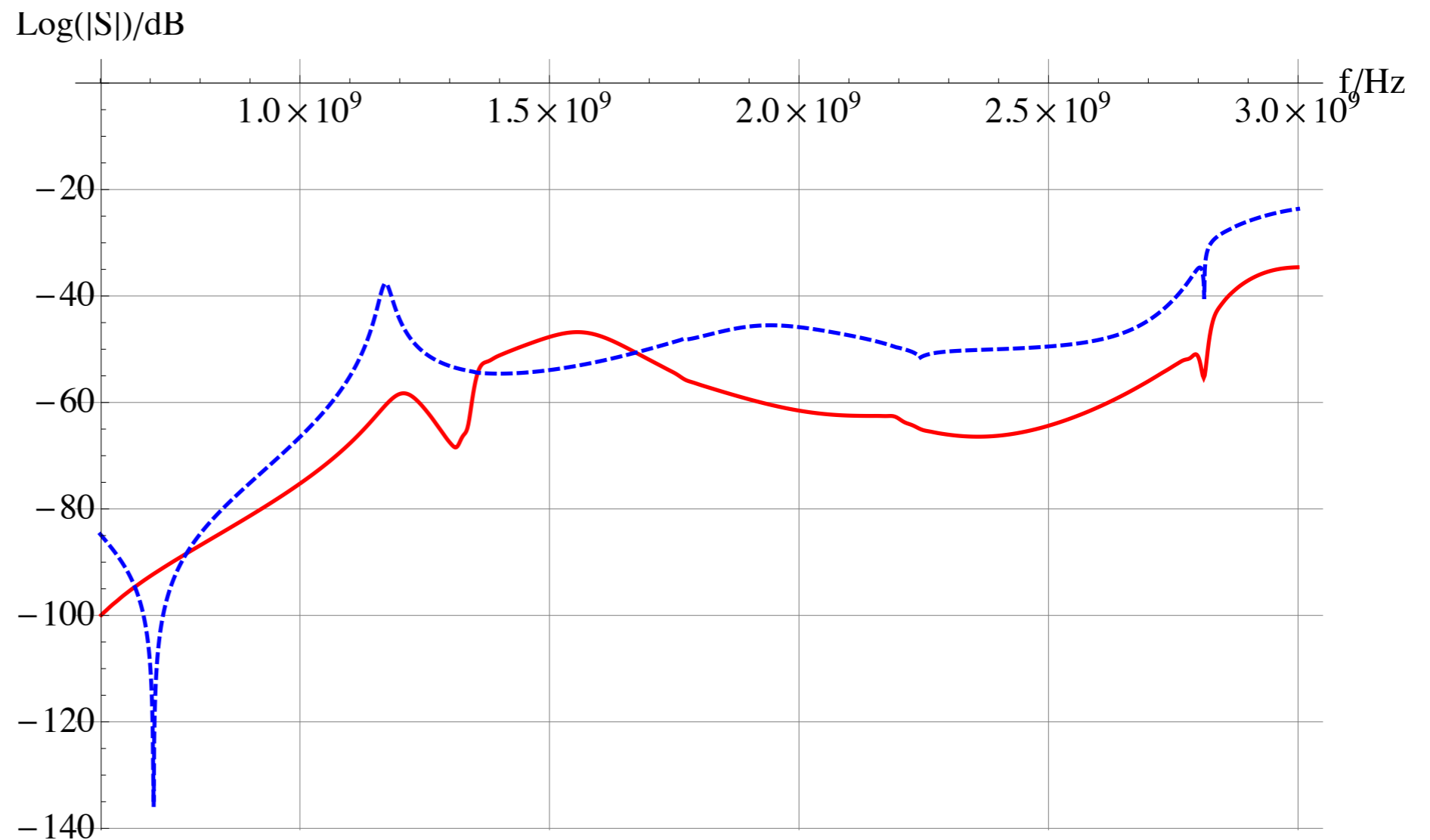
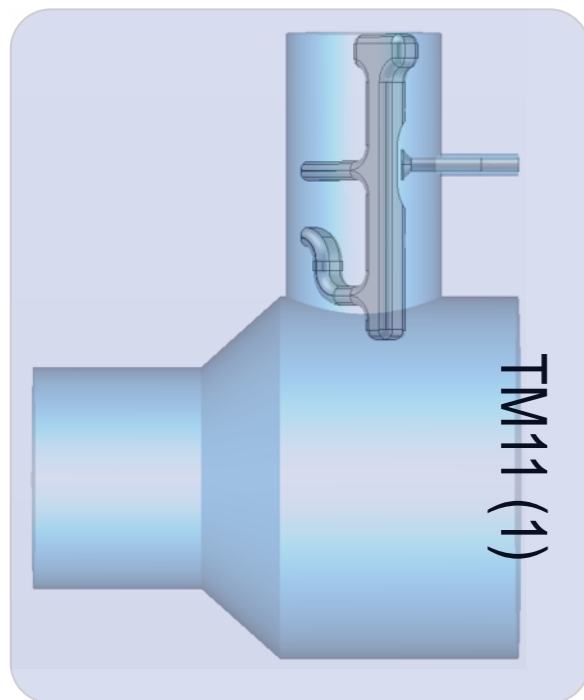
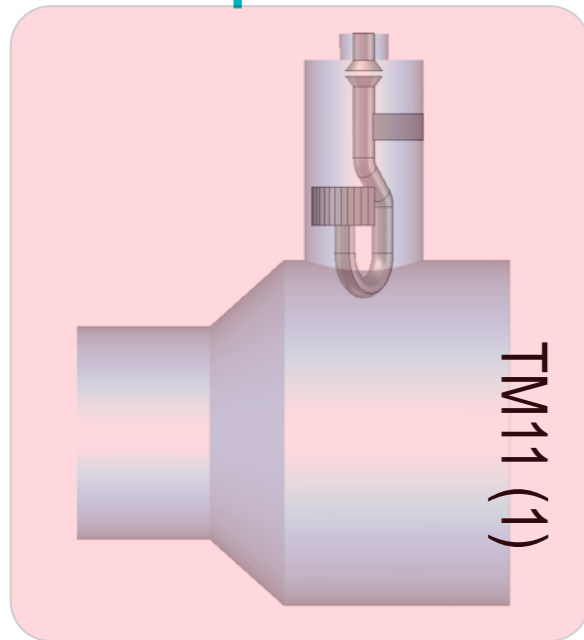


## Comparison of transmission characteristics TE<sub>21</sub>(2)

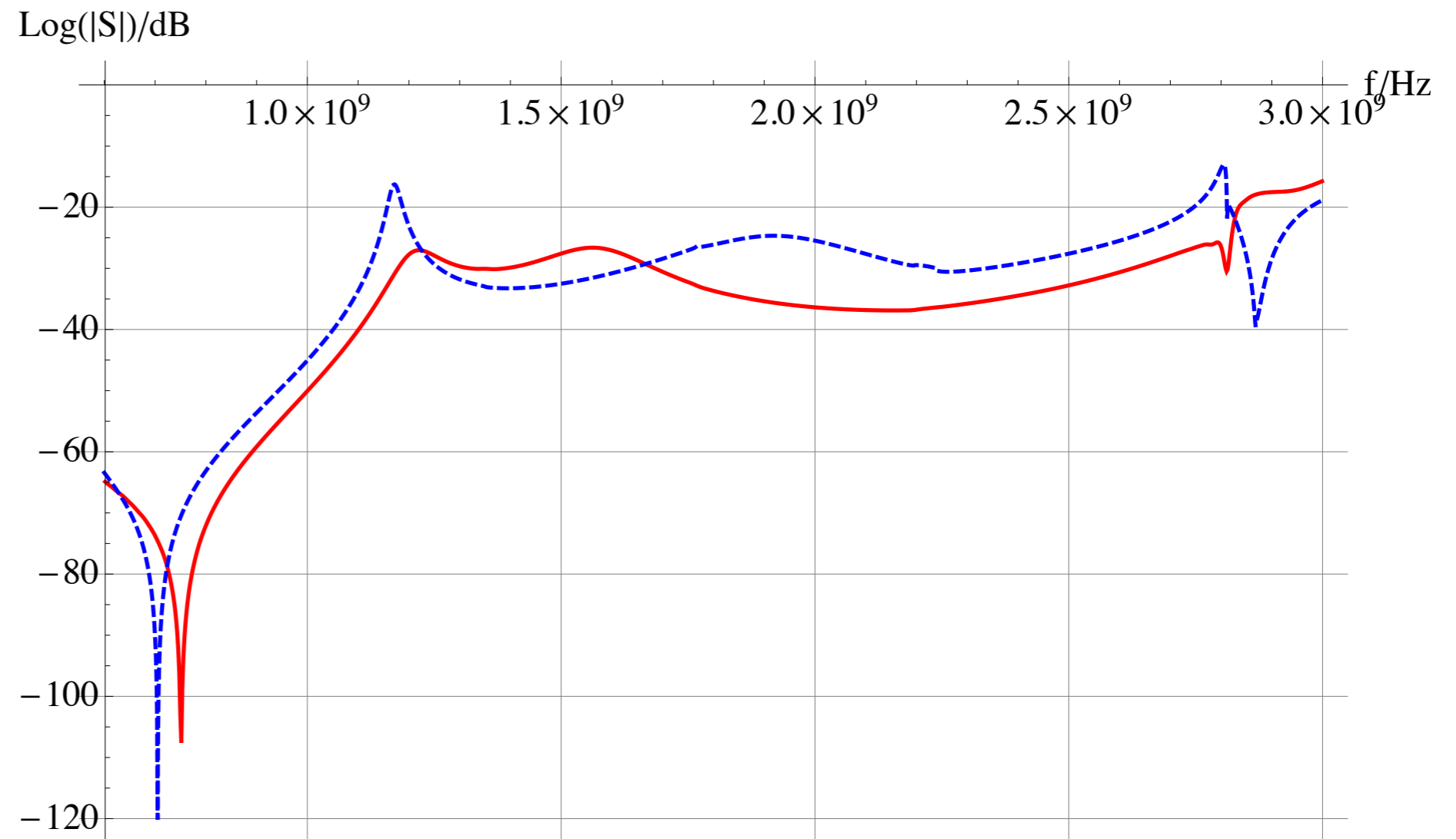
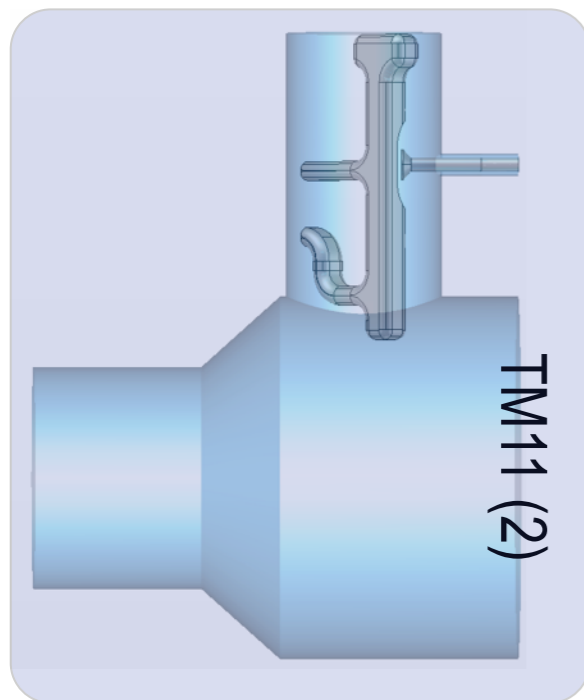
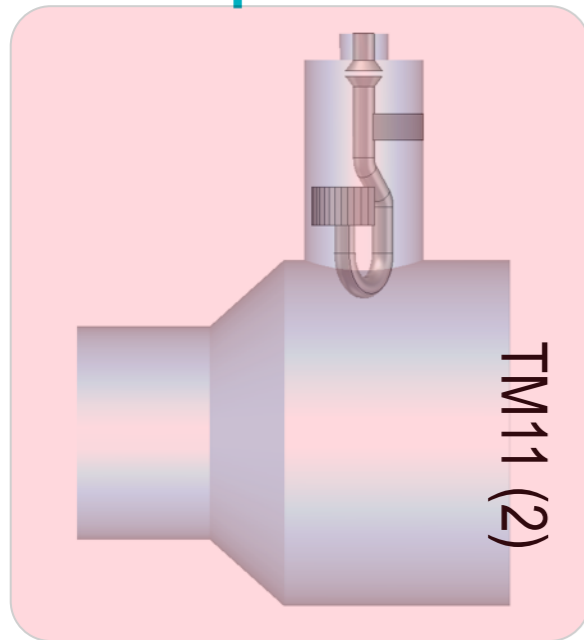




## Comparison of transmission characteristics $TM_{11}(1)$



## Comparison of transmission characteristics $TM_{11}(2)$



## Comparison of transmission characteristics and more

- similar transmission performance, some advantage for TESLA-style, esp. @  $TM_{11}(1)$  ...
- ... but using bigger port (56 mm instead of 45 mm), gaining ~ 5 mm reduced distance for evanescent beam pipe modes
- Hook-style appropriate for de-/mounting
- TESLA-style appropriate for cooling loop (two stems)
- Notch in TESLA-style based on internal resonance => independent from beam pipe mode, high field intensities inside, re-tuning with end-plate capacity
- Notch in Hook-style based on combination of internal resonance and directional coupling, re-tuning by rotation
- TESLA-style stiffer
- MP/FE => Steve, Rob et.al.

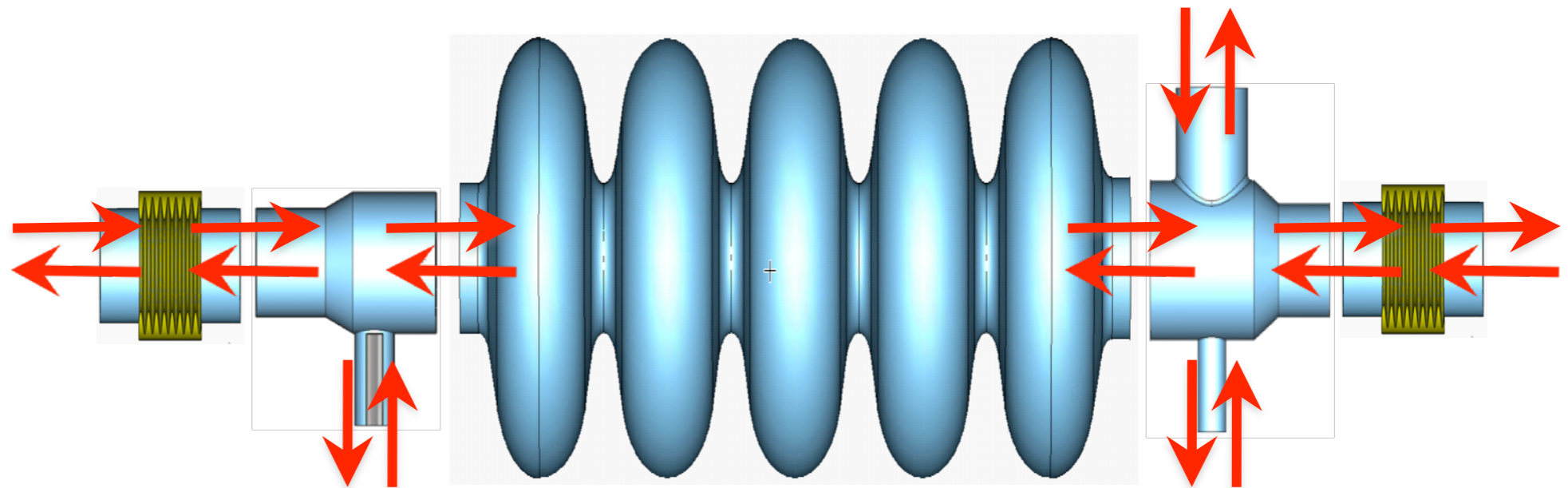


Thank you for your attention



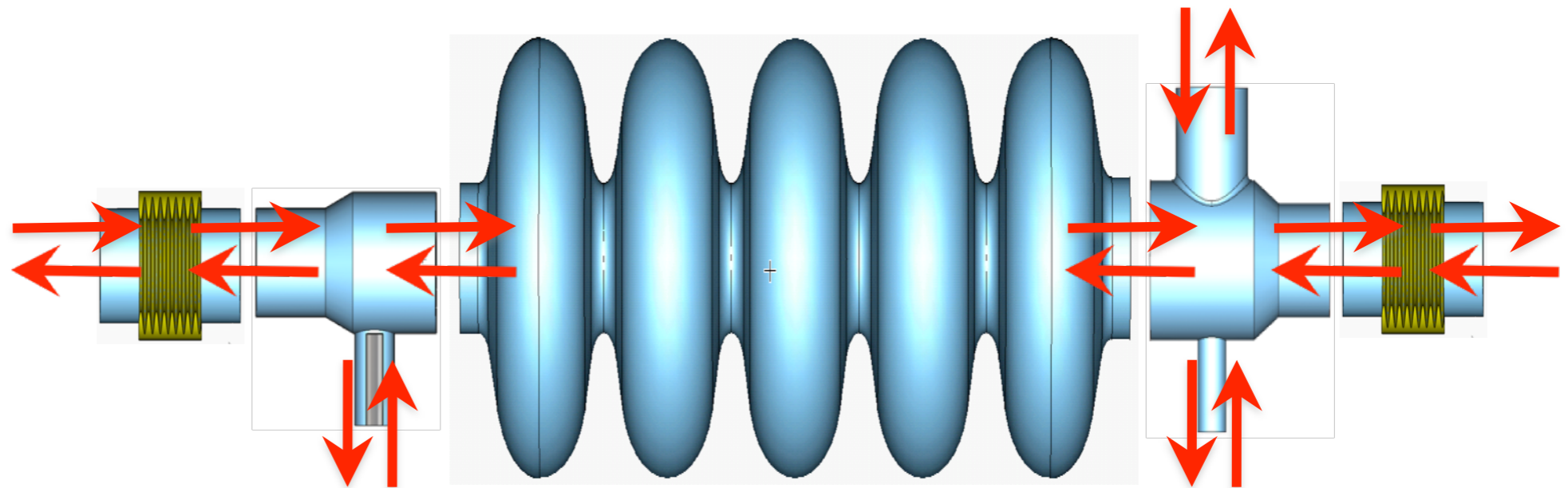
# Spare Slides

## Concatenation procedure based on scattering properties: Coupled S-Parameter Computation = CSC



- Split structure in sections
- Compute scattering (S-) parameters of all sections individually with appropriate solvers
- Compute overall S-parameters as function of  $f$  with special algorithm\*, applicable to any structure topology and mode number
- \*: **e.g.:** H.-W. Glock, K. Rothemund, U. van Rienen: "CSC - A System for Coupled S-Parameter Calculations", TESLA-Report 2001-25 or K. Rothemund, H.-W. Glock, U. van Rienen: "Eigenmode Calculation of Complex RF-Structures using S-Parameters", IEEE Transactions on Magnetics, Vol. 36, (2000): 1501-1503 and references therein

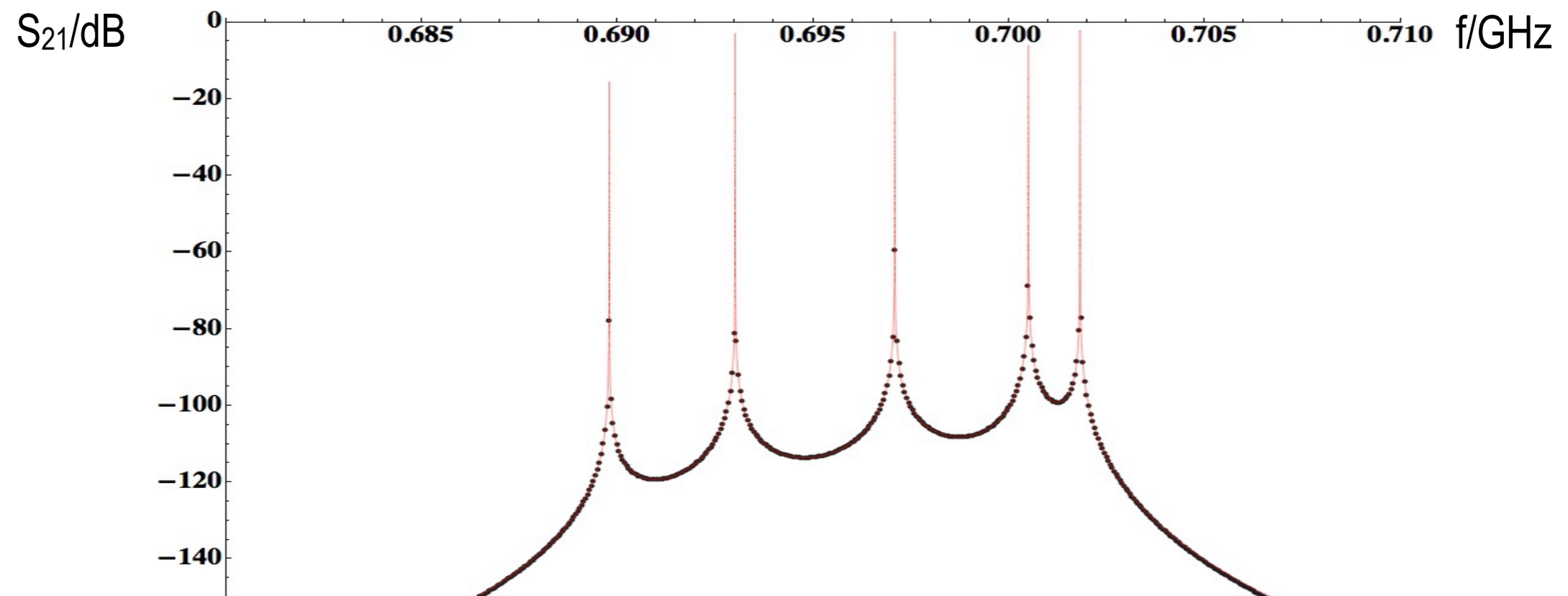
## Concatenation procedure based on scattering properties: Coupled S-Parameter Computation = CSC



- Split structure in sections
- Compute scattering (S-) parameters of all sections individually with appropriate solvers
- Compute overall S-parameters as function of  $f$  with special algorithm\*, applicable to any structure topology and mode number
- Derive loaded Q-values from S-parameter spectra

## Using Pole-fitting algorithm\* to determine loaded Q's

$$S_{21}(f) = \sum_k \frac{a_k}{2\pi i f - p_k} \quad Q_k = -\frac{\text{Im}\{p_k\}}{2\text{Re}\{p_k\}}$$

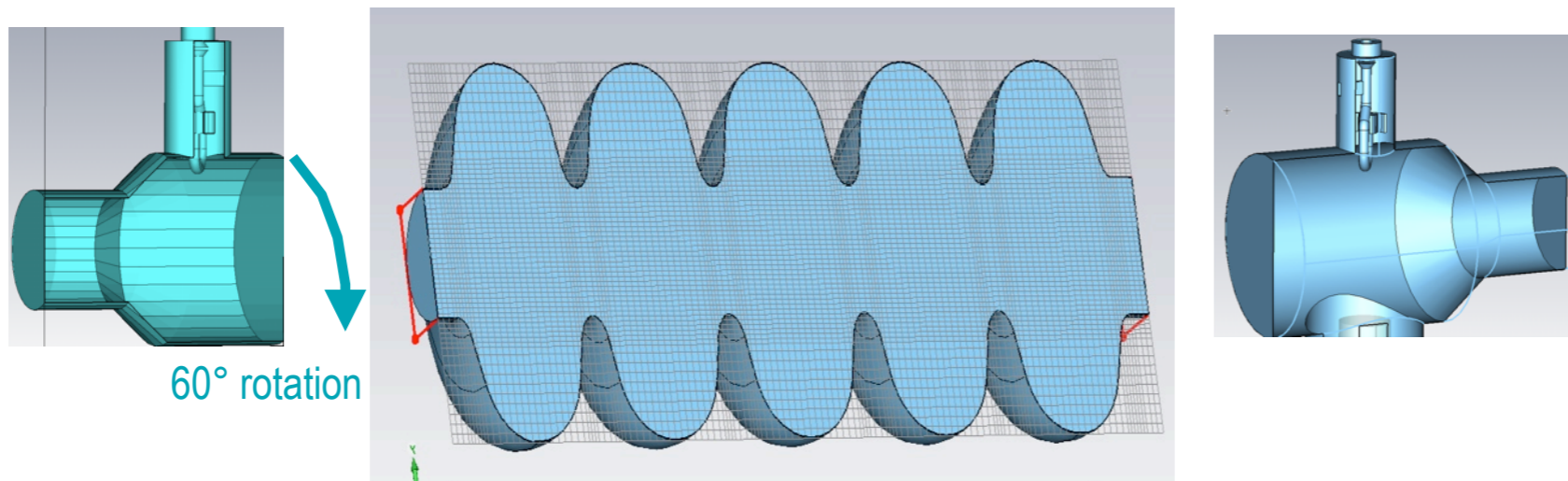


fundamental mode passband - dots: cstStudio© computation - line: fit result

\*: Hecht, Rothmund, Glock, van Rienen: "Computation of RF properties of long and complex structures", Proc. EPAC 2002



## $Q_{\text{ext}}$ computations based on:



- CSC\*-coupling of 8 modes in  $D=130\text{mm} / 140\text{mm}$ -coupler-cavity-connections ( $TE_{11}, TM_{01}, TE_{21}, TM_{11}, TE_{01}$ ), frequency range 0.6 ... 3 GHz
- right coupler with power coupler (100mm coax, 50 Ohm, penetration depth freely chosen, coax termination short / match)
- $D=80\text{mm}$  beam pipes left open (but there most below cut-off:  
 $TE_{11} - 2.196\text{ GHz}$ ,  $TM_{01} - 2.869\text{ GHz}$ ,  $TE_{21} - 3.643\text{ GHz}$ ,  $TM_{11}/TE_{01} - 4.571\text{ GHz}$ )

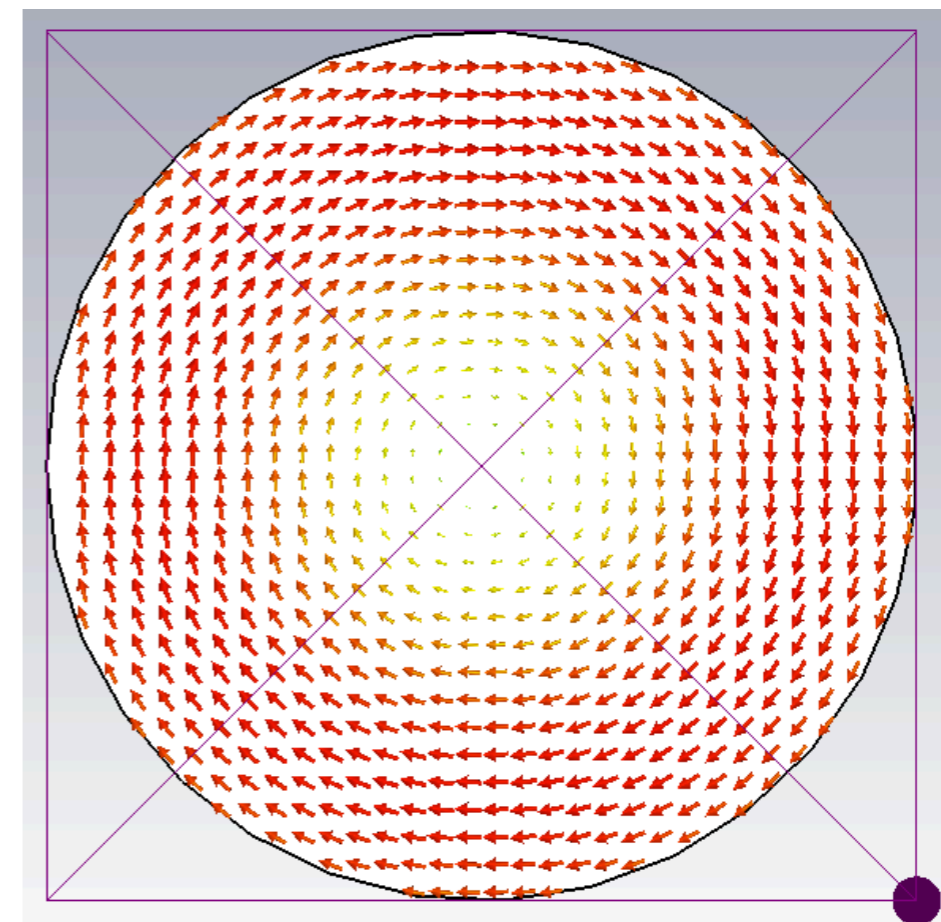
\*: compare appendix

## Why looking for a single S-parameter during tuning?:

Fundamental mode is of  $TM_0$ -type (no dependence on angle around axis).

Thus it directly couples only to  $TM_{01}$ ,  $TM_{02}$ , ... waveguide modes, all below cut-off @ 704 MHz.

Neglecting all higher types than  $TM_{01}$  causes very small error but significantly reduces numerical effort.

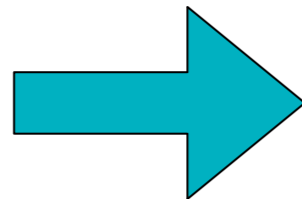


TM01-Mode H-field

## Well-established Time-Domain/Fourier Procedure - Now\*\* in Discontinuous Galerkin FEM\*-Formulation

Maxwell

$$\begin{aligned}\partial_t \mathbf{E} &= \frac{1}{\varepsilon} \nabla \times \mathbf{H} \\ \partial_t \mathbf{H} &= -\frac{1}{\mu} \nabla \times \mathbf{E}\end{aligned}$$



DG-FEM discretized Maxwell

$$\begin{aligned}\partial_t \underline{\mathbf{E}} &= \frac{1}{\varepsilon} \underline{\mathbf{M}}_k^{-1} \underline{\mathbf{S}}_k \underline{\mathbf{H}} + \frac{1}{\varepsilon} \underline{\mathbf{M}}_k^{-1} \underline{\mathbf{L}}_k \cdot (\mathbf{n} \times (\mathbf{F}_E - \mathbf{F}_E^*)) \\ \partial_t \underline{\mathbf{H}} &= -\frac{1}{\mu} \underline{\mathbf{M}}_k^{-1} \underline{\mathbf{S}}_k \underline{\mathbf{E}} - \frac{1}{\mu} \underline{\mathbf{M}}_k^{-1} \underline{\mathbf{L}}_k \cdot (\mathbf{n} \times (\mathbf{F}_H - \mathbf{F}_H^*))\end{aligned}$$

local on each element

coupling with adjacent elements

- Strong locality => well suited for massive parallel computations
- High order approximation of field quantities => minimizes numerical dispersions effects
- Fully explicit, even on unstructured grids => Geometric flexibility and computational high efficiency

\*J. Hesthaven and T. Warburton, "Nodal high-order methods on unstructured grids: I. time-domain solution of maxwell's equations," Journal of Computational Physics, no. 181, pp. 186–221, 2002.

\*\*C. Potratz, H.-W. Glock, U. van Rienen: Time Domain Field and Scattering Parameter Computation in Waveguide Structures by GPU-Accelerated Discontinuous-Galerkin Method, submitted to IEEE-MTT

## The Cluster (current configuration)



NVidia GPGPU (General Purpose Graphic Processing Unit, source: [www.nvidia.com](http://www.nvidia.com))

- Currently six nodes with 4 Cores and 2 GPUs
- Combined peak performance 18 TFlop/s (GPU only)  $\approx$  9k€
- Used for 3D-CEM simulation (our DG-FEM code running the GPUs) for parallel optimization
- Mechanical/Thermal simulations using Code Aster\* (CPU only)
- Pre/Postprocessing (triangulation, scattering parameter computation, etc.)

\*[www.code-aster.com](http://www.code-aster.com)

# Influence of the fixture onto the HF properties

Log(|s<sub>23</sub>|)/dB

