

Wide Bandgap – Wide Bandwidth?

Thoughts about multi-GHz readouts for thin WBG sensors

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HEPHY

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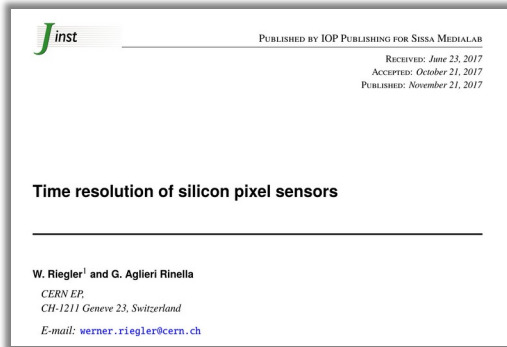
Contents

- Theoretical background
- Design considerations for high-BW readouts
- Prototype results from HEPHY

Two Questions

- 1) For WBG sensors with higher drift velocities than silicon, what is the optimal readout for optimal timing?
- 2) If we want to apply TCT to thin ($< 100 \mu\text{m}$) WBG sensors, how can we achieve the required readout bandwidths (irrespective of noise), in order to extract features such μ_e, μ_h ?

Time Resolution



Analytical treatment
by [Riegler et. al.](#)

$$\sigma_t^2 = \sigma_{\text{Jitter}}^2 + \sigma_{\text{Ionization}}^2 + \sigma_{\text{Distortion}}^2 + \sigma_{\text{TDC}}^2$$

Can be
minimized

$(1/f_{\text{sample}}) / \sqrt{12}$
Negligible for fast
oscilloscopes or can use
interpolation

Very sensitive to readout electronics!

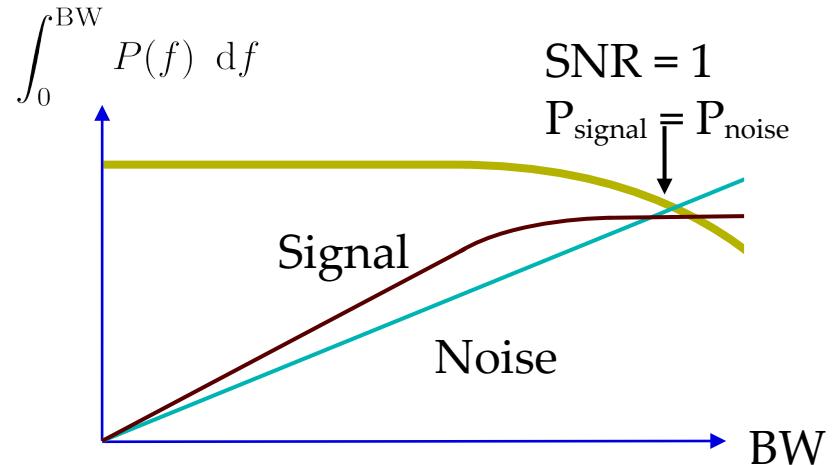
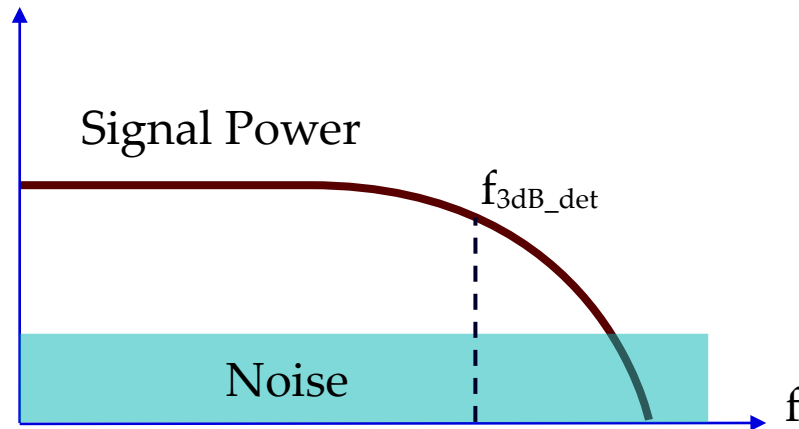
$$\sigma_{\text{Jitter}} = \frac{N}{dV/dt} \approx \frac{t_r}{S/N}$$

Decreases monotonically for thinner
detectors and high drift velocities

SNR vs Bandwidth

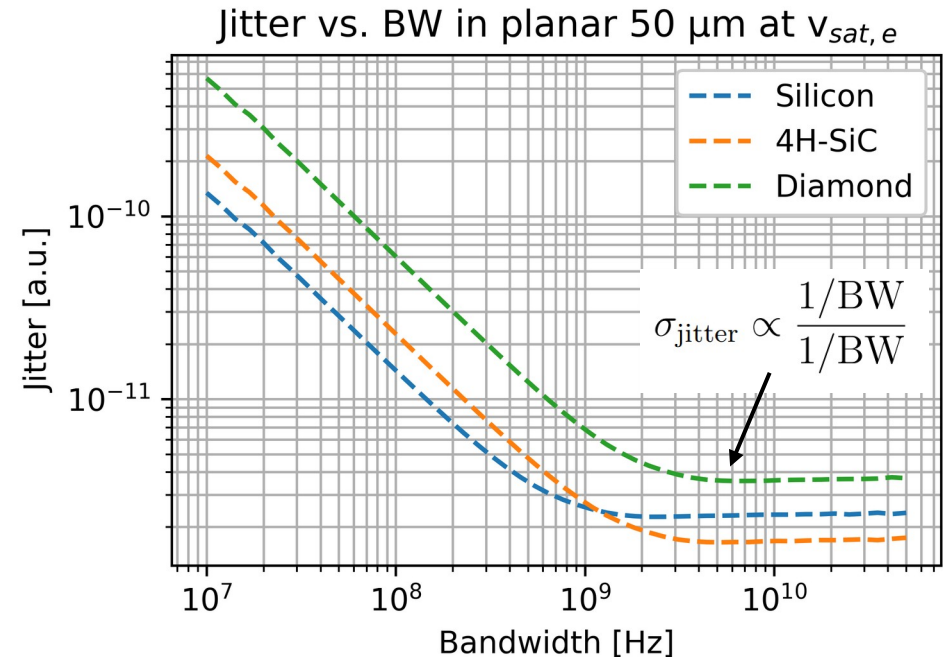
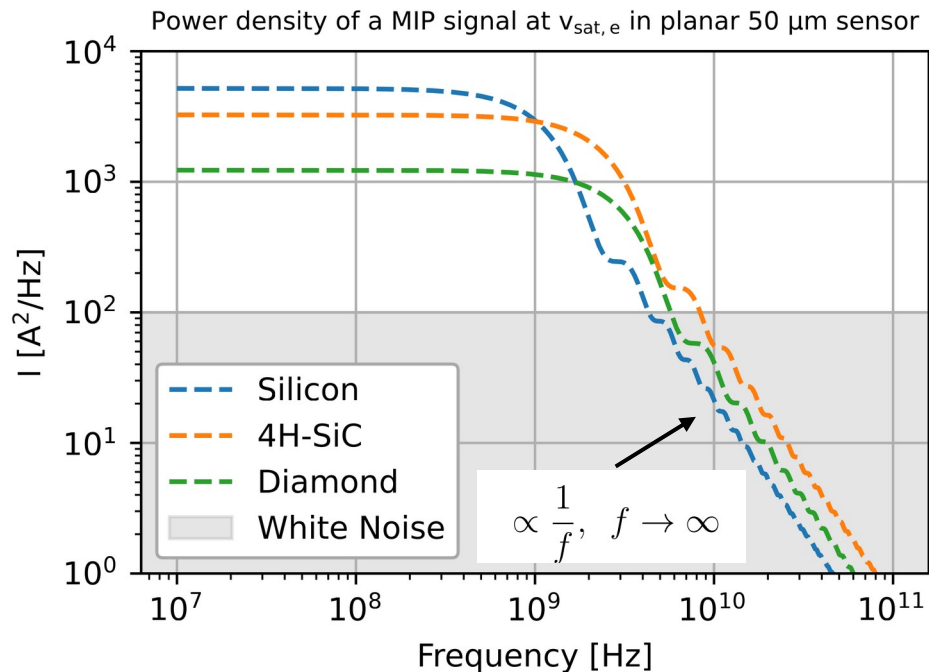
- For white noise, the SNR is constant with frequency, up to the detector signal frequency cutoff
- After the limit of detector bandwidth, the SNR drops below 1

$P [A^2/Hz]$



Jitter vs Bandwidth

- Toy model for planar sensors, $v_e = v_h = v_{sat}$, see [2nd DRD3 Week](#)
- In the limit of high frequencies, jitter is constant (rise time grows as fast as SNR is decreasing)

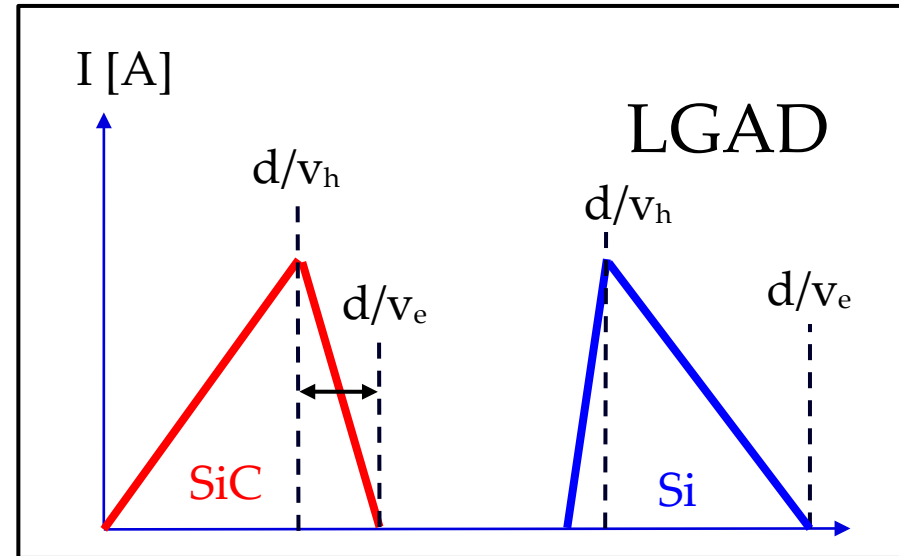
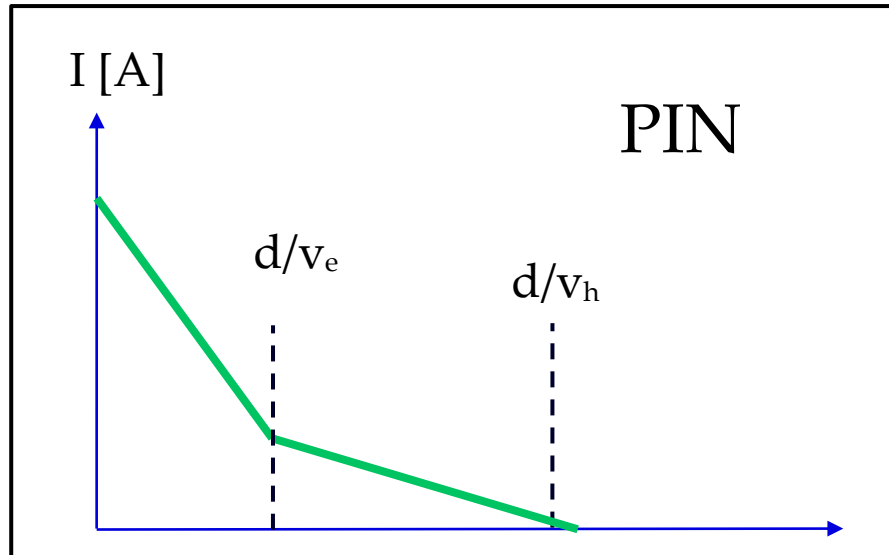


WBG PIN and LGADs

- In PIN devices : maximum current at $t = 0$
- For LGADs, maximum current after charge carrier drift to gain layer
- For 4H-SiC : n-type, hole multiplication
- P-LGAD : p-type, electron multiplication

$$f_{\text{signal}} \propto v_{e,h} / d$$

Diamond , SiC : $v_e = 2 \times v_{e(\text{Si})}$



WBG Readouts

Timing readout (2-3 GHz):

- Minimal noise
- Rather small, low-power components to allow for multi-channel readout
- Waveforms can be shaped or (optimally) filtered, waveform shape is irrelevant

TCT readout (> 5 GHz)

- Noise can be much higher (averaging laser signals, large charge depositions, i.e. alpha source)
- Single or a few channels (can use large, power-hungry components)
- Want constant gain and impedances up to multiple GHz to be able to reconstruct signal shape

Future Readouts

Timing readout (2-3 GHz):

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TCT readout (> 5 GHz)

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High BW Readout Recipe

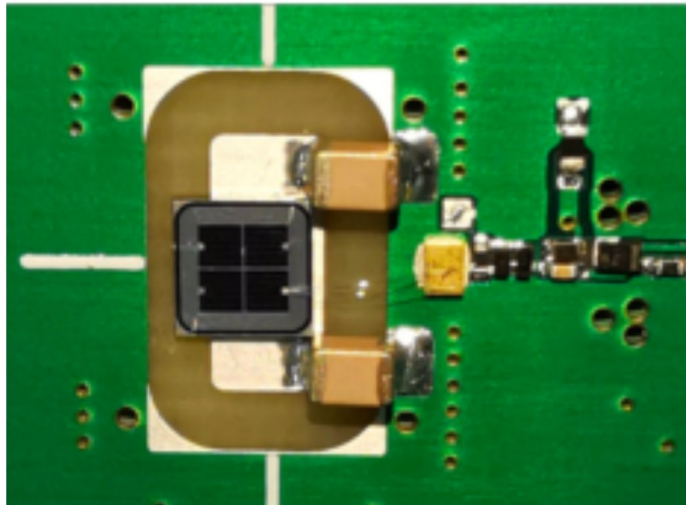
- Using discrete components not really feasible anymore, especially for feedback circuits (parasitics of SMD components)
- Move to MMIC (micro-wave monolithic integrated circuit) amplifiers / ASICs ?
- Keep impedance matched to 50Ω (or have input impedance $< 50 \Omega$?)
- Minimize detector capacitance and parasitic capacitances
- Take care of AC return path (Vias, etc.)
- Use adequate PCB substrates (Rogers)
- Bondwires as short as possible (and multiple of them $1/L_{\text{tot}} = 1/L_1 + 1/L_2 + \dots$)

Challenge for WBG : Do all of this for high voltages ($\geq 1\text{kV}$)

(HV to full deplete sensors and to arrive a v_{sat})

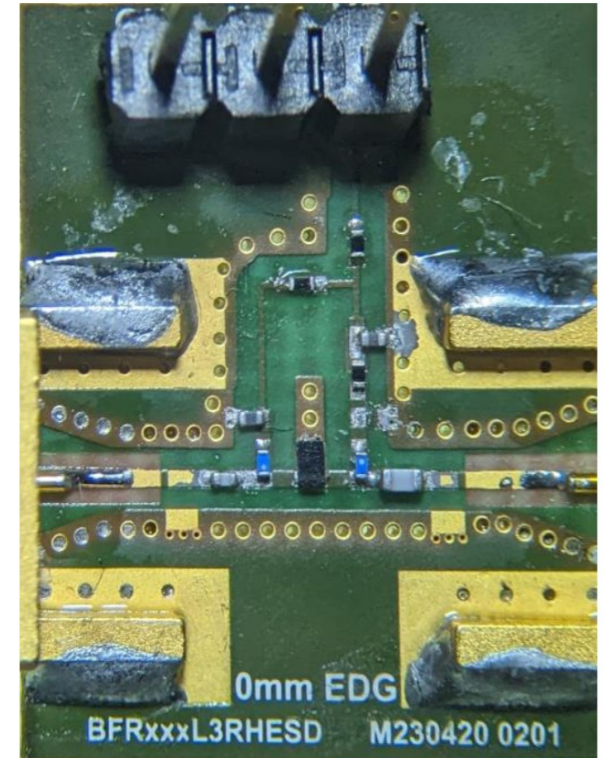
Readout Boards in the Community

- BFR840 based circuit introduced by SCIPP
- Widely used (Chubut board, 2-stage design at UZH)
- Other solutions : Galli amplifiers (FNAL 16ch), Cividec
- All optimized for < 2 GHz



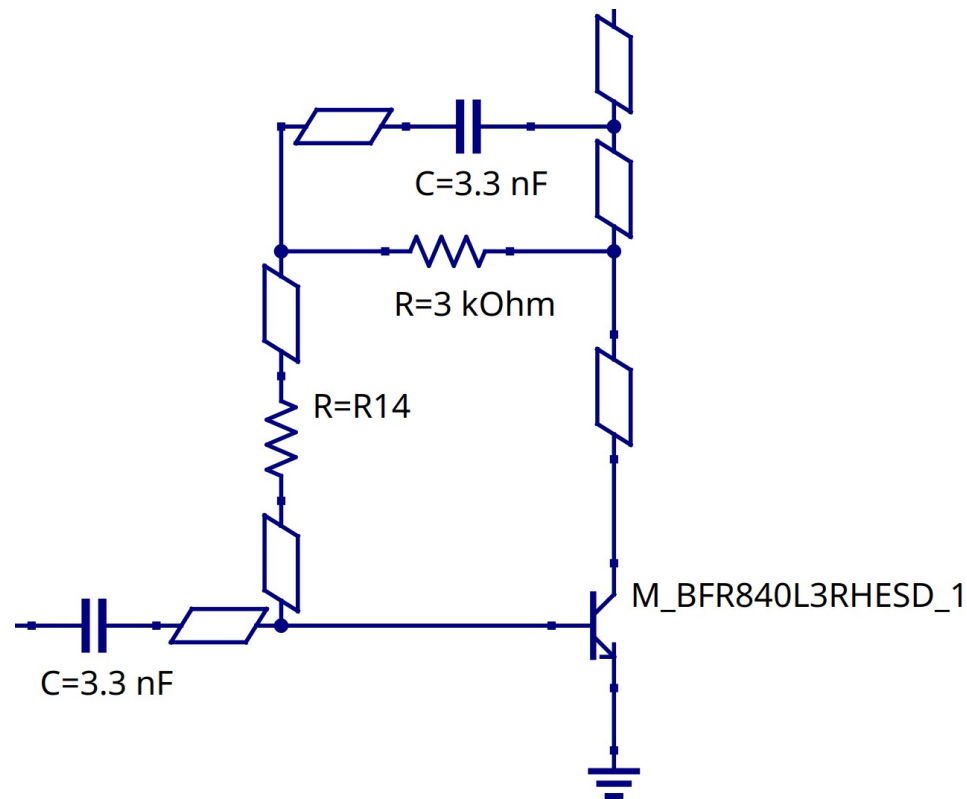
UCSC LGAD
board

BFR840L3RHESD Infineon Reference Design



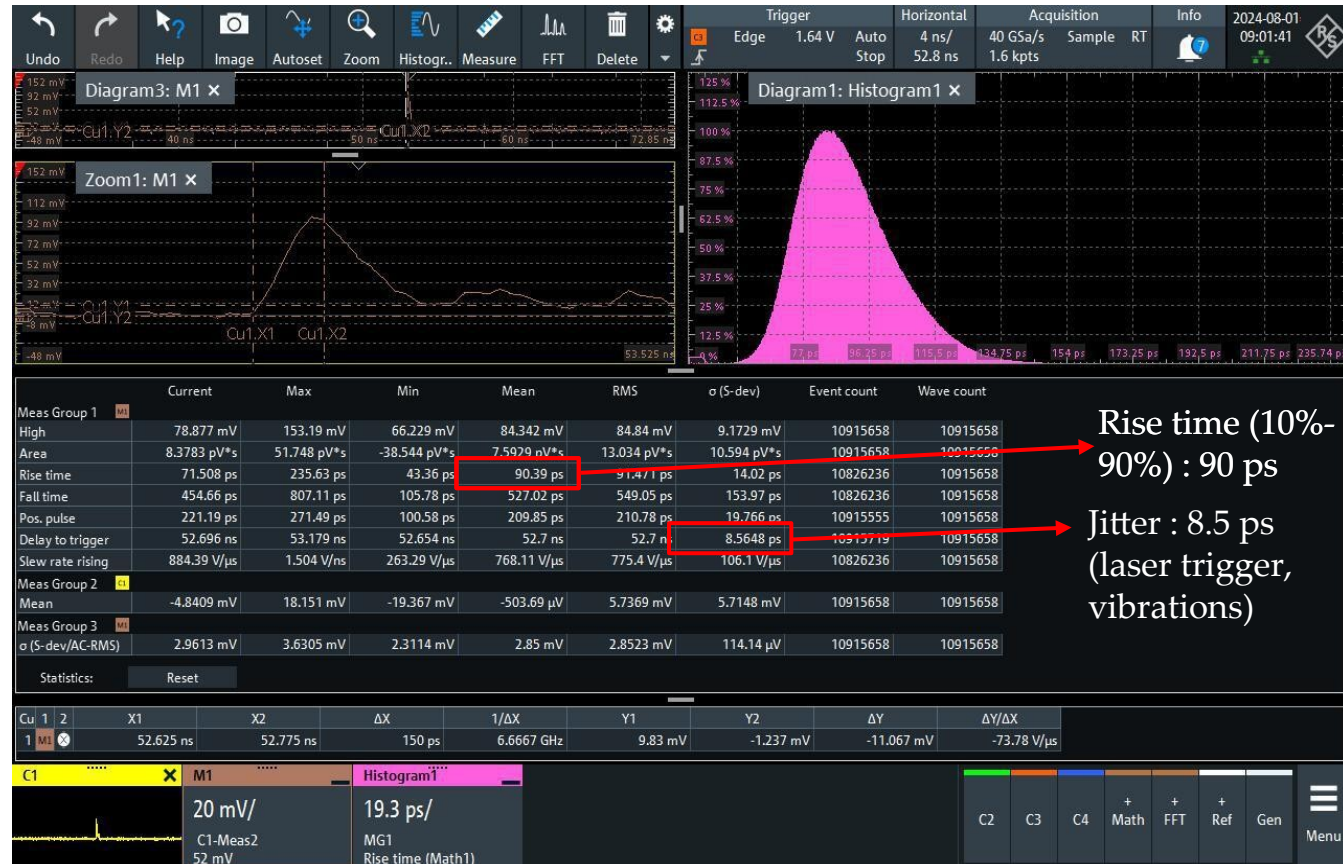
BFR840 Based Readouts

- How far can we go with the BFR840?
- Transimpedance (i.e. gain) and BW determined by feedback resistor R_{14}
- Simply reduce R_{14} and get more BW (with worse SNR)?



UCSC Board Tests

- UCSC board with feedback $R_{14} = 82 \Omega$
- 4H-SiC detector with $C_{det} = 0.5 \text{ pF}$, UV-TCT with $\sim 3 \text{ ps}$ jitter
- Achieved 90 ps risetime!
- Eqv. BW = 3.9 GHz
- However, oscillations in signal, esp. for $R_{14} < 82 \Omega$
- Would need a more HF adequate design and HF resistors



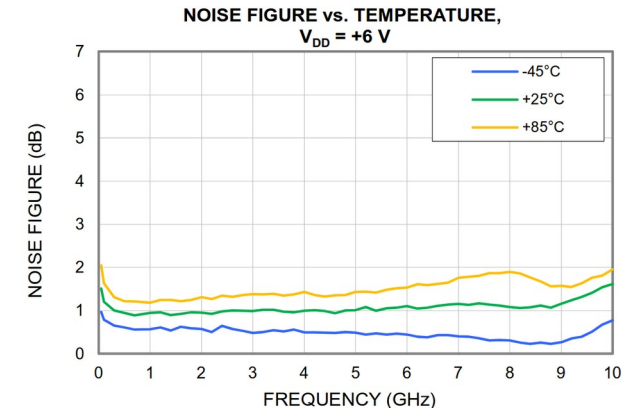
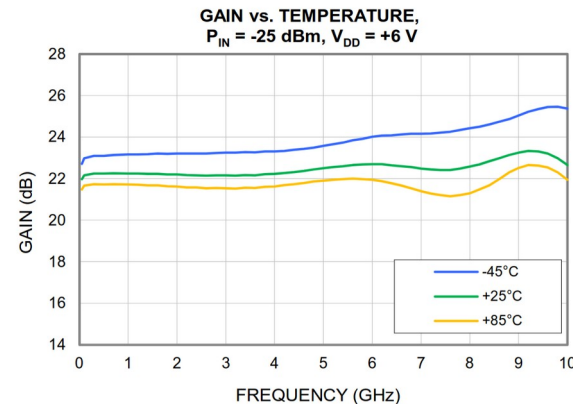
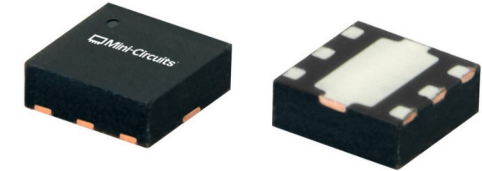
Rise time (10%-90%) : 90 ps

Jitter : 8.5 ps (laser trigger, vibrations)

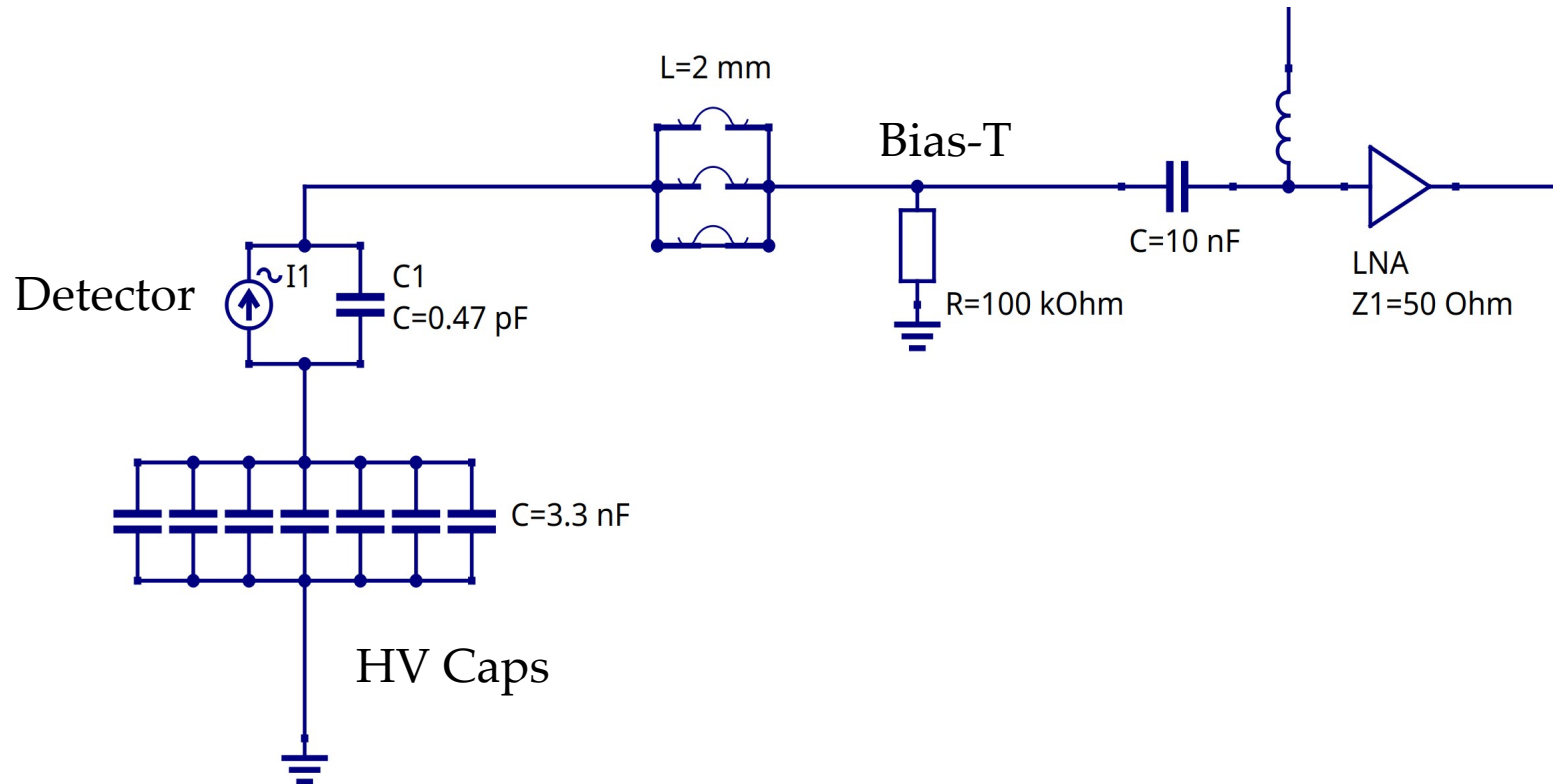
$$1/(2 * \pi * 0.5 \text{ pF} * 82 \text{ Ohm}) = 3.88 \text{ GHz}$$

MMIC Prototype

- If the transistor feedback circuit is challenging at high frequencies, why not use an IC / MMIC?
- Treat as a 50Ω black box
- Wire-bond as directly as possible (bare dies available!)

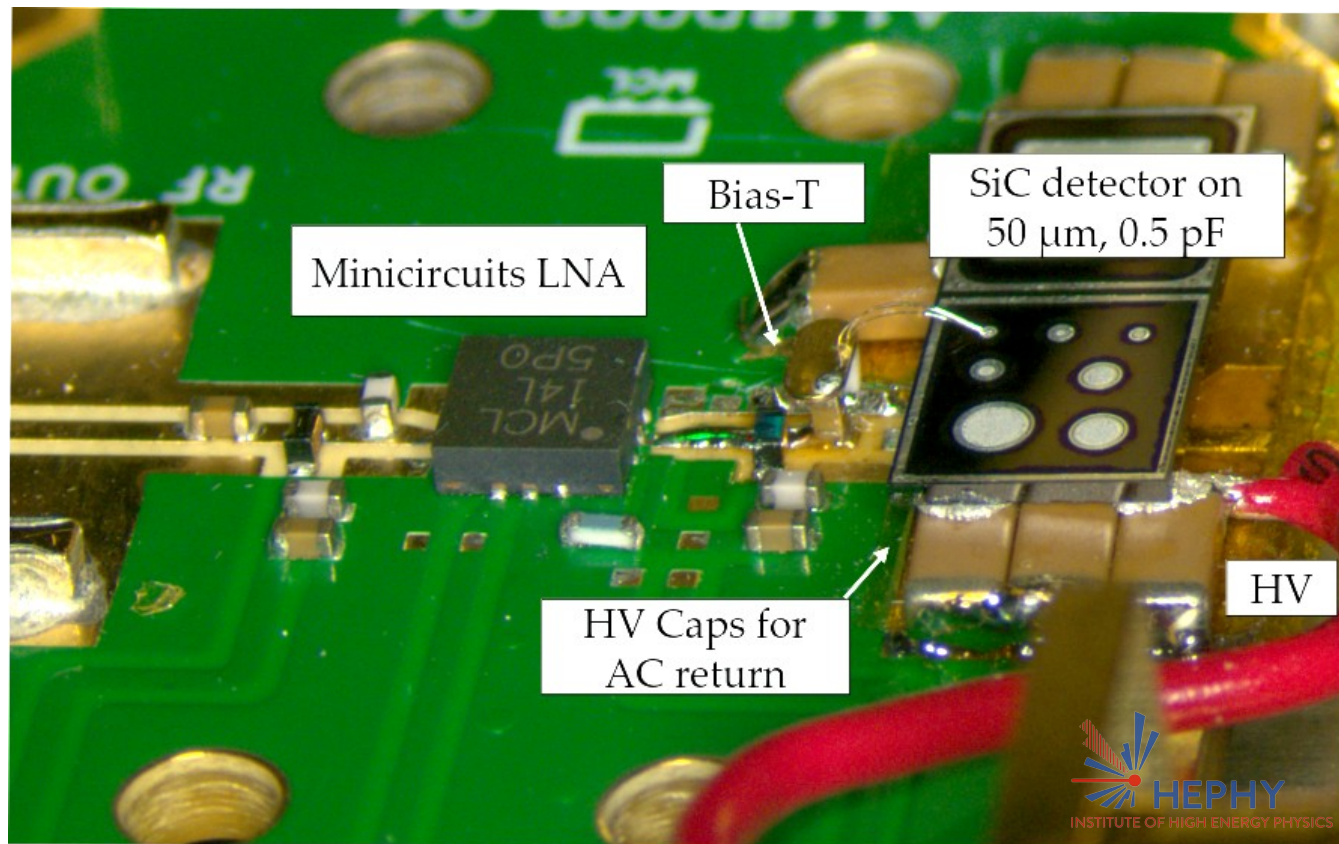


MMIC Prototype Circuit



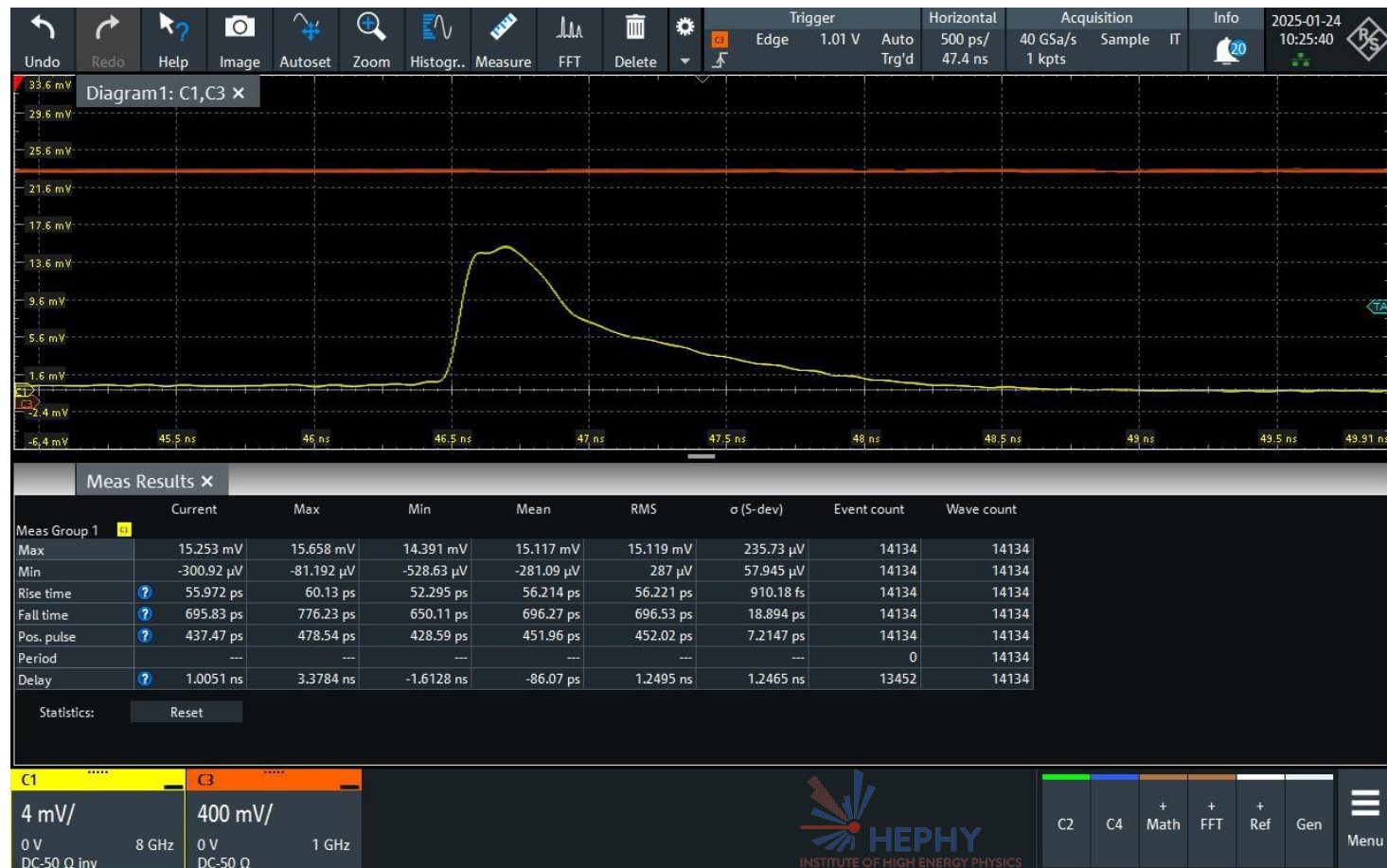
MMIC-based prototype

- SiC sensor mounted on LNA board (RO4350B)
- Minicircuits LNA
50 Ω , stable gain up to 10 GHz, NF = 1.1 dB
- RC-bandwidth :
6.3 GHz
- Bondwires : 3x 2mm



MMIC Prototype Results

- Very fresh results (few hours old)
- UV-TCT laser (370 nm)
- Risetime : 55 ps (~ 6.4 GHz BW)
- Partially limited by laser pulse (45 ps FWHM)
- (Almost) no reflections!



Next Steps

- Try to extract v_e/v_h from voltage scans / alpha measurements
- Design new PCB to optimize bond wire length
- Test with different low-capacitance sensors (small area planar, pixels, 3D sensors)