

POCPA3 – DESY, May 21-23, 2012

Elettra digital control PS

Bipolar digital control 30 A – 20 V PS

Denis Molaro

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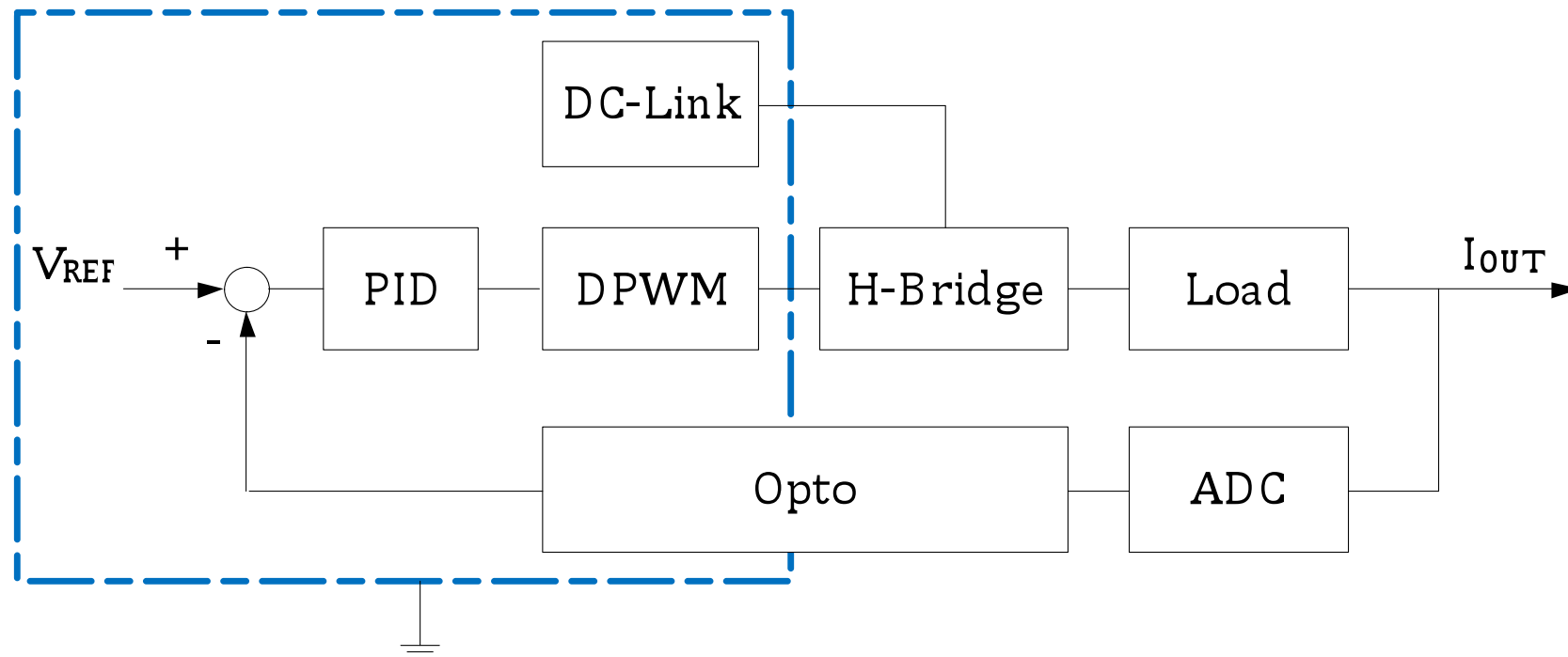
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Fermi@Elettra's PS requirements

- Four-quadrant: ± 5 A, ± 10 V and ± 20 A, ± 20 V
- Max Ripple: 30 ppm/FS
- Long term stability (8h): ± 50 ppm/FS
- Standard Ethernet interface
- Output current resolution: 16 bit
- Current set point update rate: 50 Hz (for feedback)
- Load range: 0.5 – 2 Ω and 0.5 – 400 mH
- Number of channels: more than 300

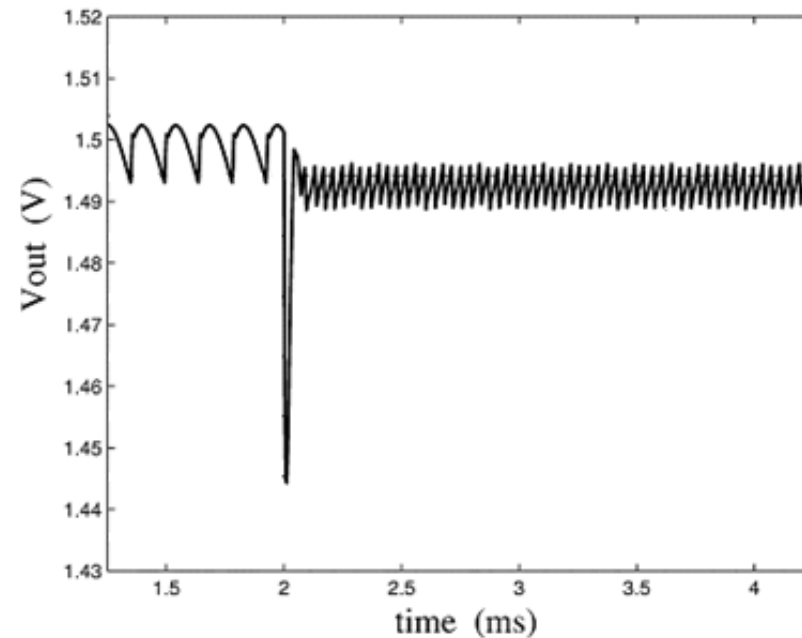
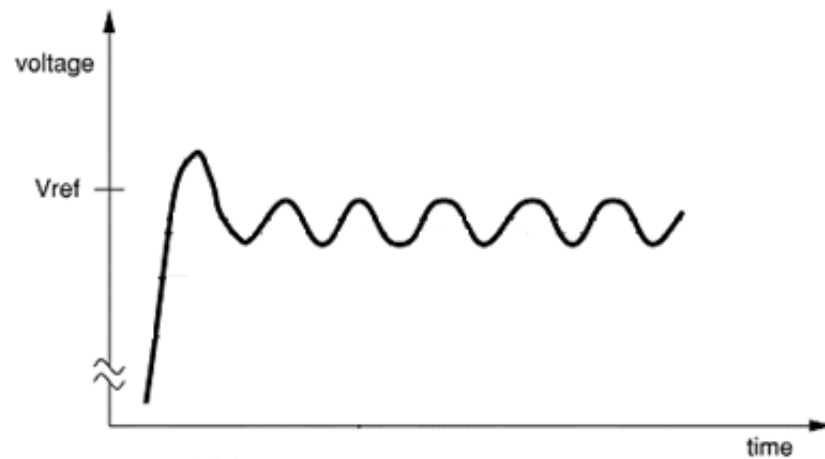
Why digital?

- Easy of PID modification (wide load range for the same current range)
- Step response optimization for all PS-magnet combination.
- Galvanic isolation between current sensor and control electronics



Resolution & Limit Cycles

Digital controls are affected by limit cycles



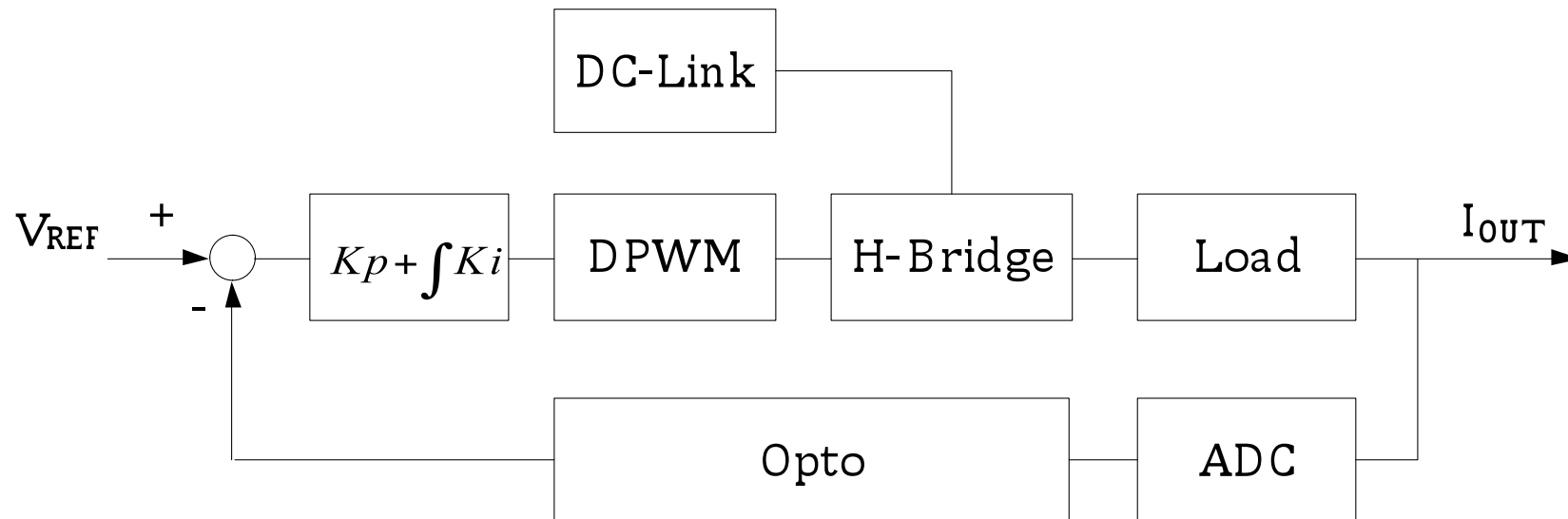
Resolution & Limit Cycles

Three conditions are required to avoid Limit Cycles^[1]

- Output min step < reading min step
(i.e. output resolution > reading resolution)
- $0 < \text{Integral Gain } (K_i) < 1$
- Loop stability (with ADC gain)

} **Hardware**

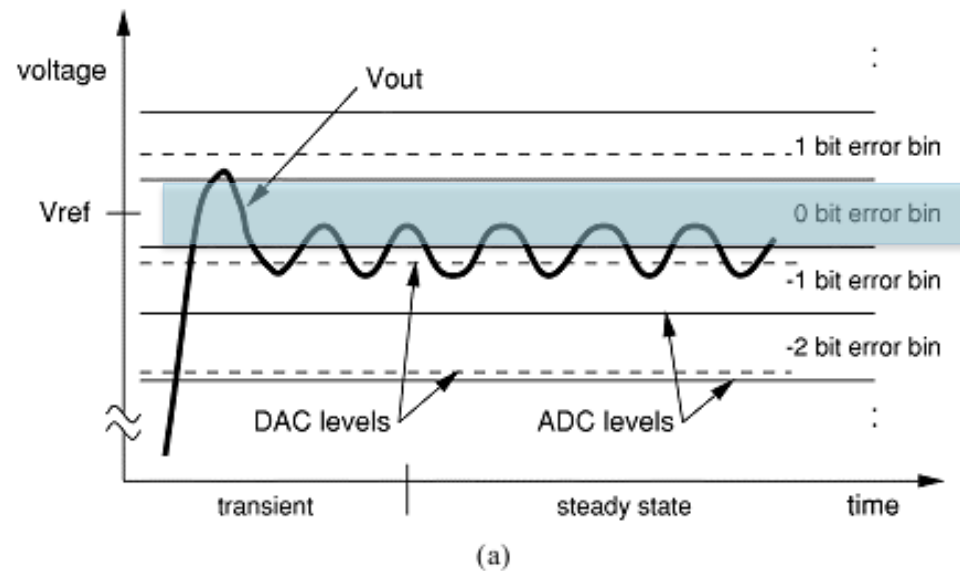
} PID parameters



[1] A.V. Peterchev, S.R. Sanders: Quantization resolution and limit cycling in digitally controlled PWM converters IEEE Transaction on Power Electronics, Volume 18, Issue 1, Part 2, Jan. 2003, pp. 301-308

Resolution & Limit Cycles

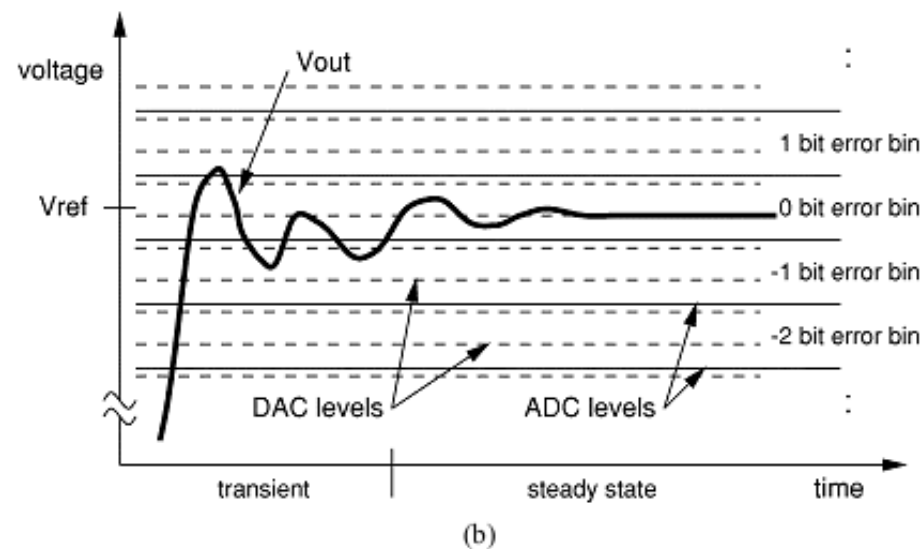
When output min step > reading min step →



Limit cycles

Resolution & Limit Cycles

When output min step < reading min step →



Stady state stable

Resolution & Limit Cycles

Output min step (H-Bridge):

$$I_{out_MIN-STEP} = \frac{2 \cdot V_{DC-Link}}{R_{LOAD} \cdot (2^{DPWM-RES} - 1)}$$

Reading min step:

$$I_{read_MIN-STEP} = \frac{I_{out_RANGE}}{2^{ADC-RES}}$$

Example ($\pm 5A$ PS):

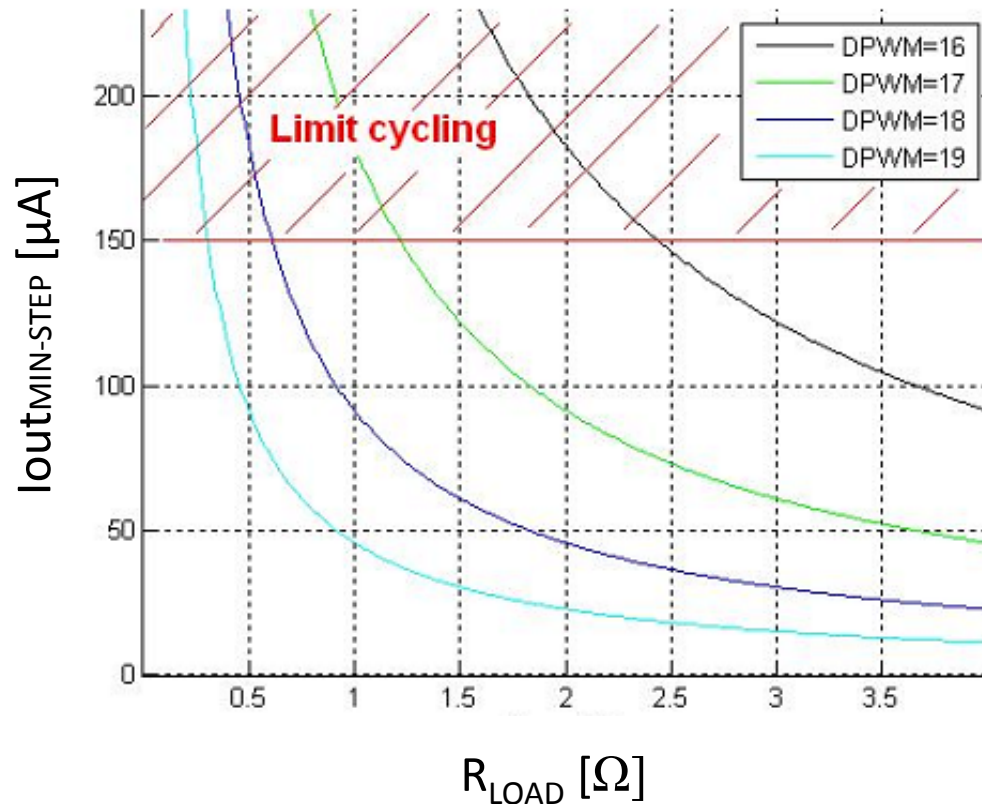
$$V_{DC-Link} = 12 \text{ V}$$

$$I_{out_RANGE} = 10 \text{ A}$$

$$ADC-RES = 16 \text{ bit}$$

Resolution & Limit Cycles

Limit cycle condition for $V_{DC-Link} = 12\text{ V}$, $I_{out_RANGE} = 10\text{ A}$ and $ADC-RES = 16\text{ bit}$



$$I_{ADC-RES} = \frac{10}{2^{16}} = 153\ \mu A$$

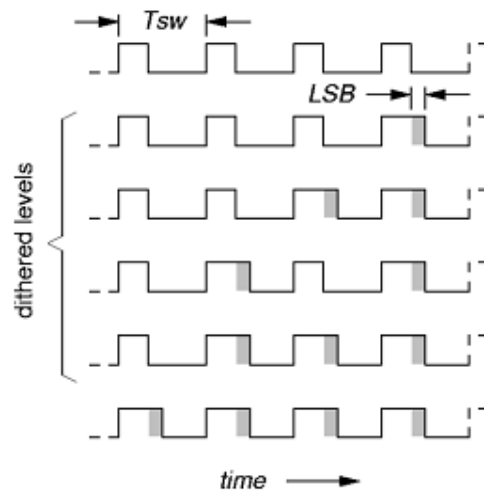
Due to wide range of resistive load a 19 bit of DPWM resolution is required

DPWM Resolution

- Delay-line DPWM has been used (TMS320F2808)

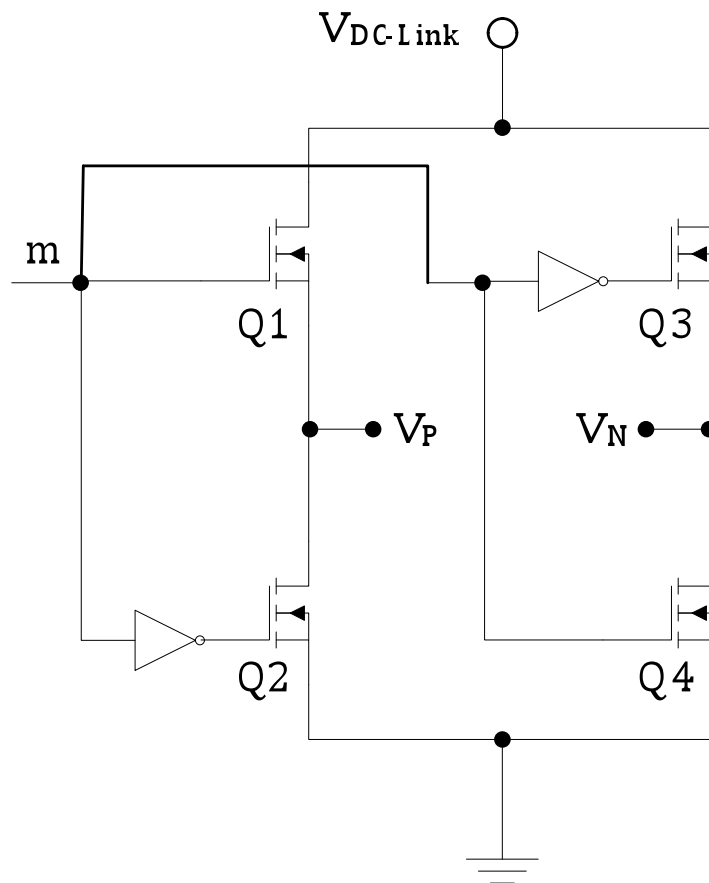
PWM Freq (kHz)	Regular Resolution (PWM)		High Resolution (HRPWM)	
	Bits	%	Bits	%
20	12.3	0.0	18.1	0.000
50	11.0	0.0	16.8	0.001
100	10.0	0.1	15.8	0.002

- Using 4 PWM cycles of dithering
17.8 bit resolution has been achieved



H-Bridge Toggle Technique

- A further bit of resolution has been added by a Toggle Technique (Patent Pending) between the two legs of the H-Bridge.



$$m \in \{0, 1\}$$

$$I_{out} = \frac{V_P - V_N}{R_{LOAD}} = \frac{(2m - 1) \cdot V_{DC-Link}}{R_{LOAD}}$$

$$I_{out_{MIN-STEP}} = \frac{2 \cdot m_{LSB} \cdot V_{DC-Link}}{R_{LOAD}}$$

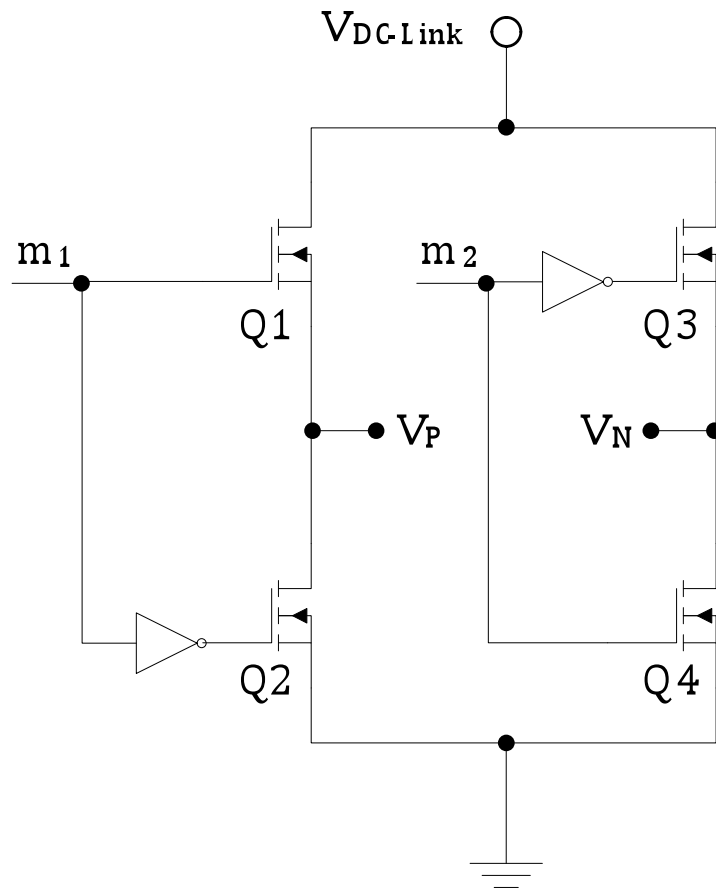
For instance with

$$V_{DC-Link} = 12, R_{LOAD} = 1 \text{ and } m_{LSB} = 1/2^{17.8}$$

$$I_{out_{MIN-STEP}} = 105 \mu A$$

H-Bridge Toggle Technique

- A further bit of resolution has been added by a Toggle Technique (Patent Pending) between the two legs of the H-Bridge.



For: $m_1 \neq m_2$

$$I_{out} = V_P - V_N = \frac{(m_1 + m_2 - 1) \cdot V_{DC-Link}}{R_{LOAD}}$$

$$I_{out_{MIN-STEP}} = \frac{m_{LSB} \cdot V_{DC-Link}}{R_{LOAD}}$$

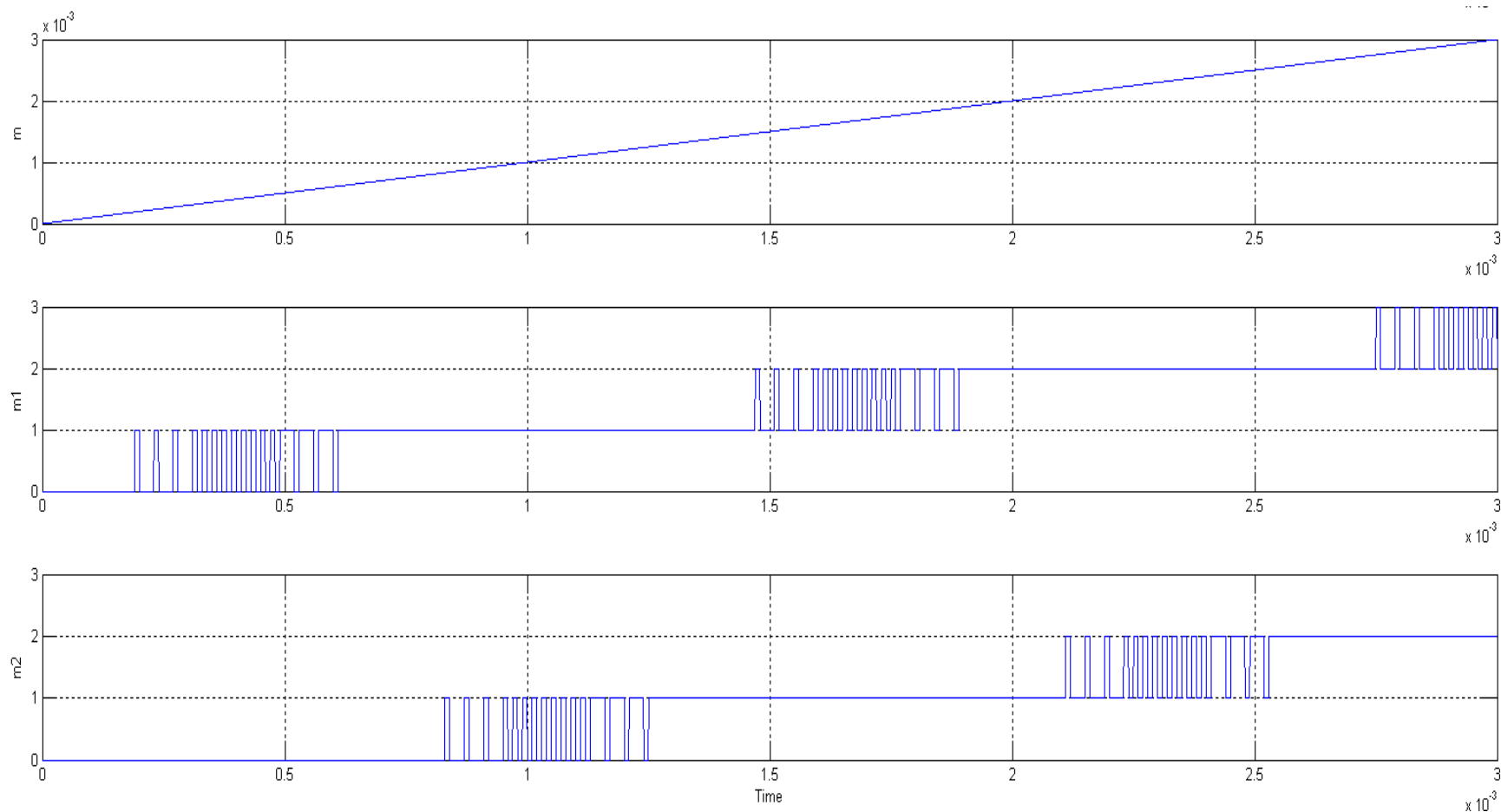
Keeping m_1 constant, varying m_2 (or vice versa)

$$V_{DC-Link}=12, R_{LOAD}=1 \text{ and } m_{LSB}=1/2^{17.8}$$

$$I_{out_{MIN-STEP}} = 52.5 \mu A$$

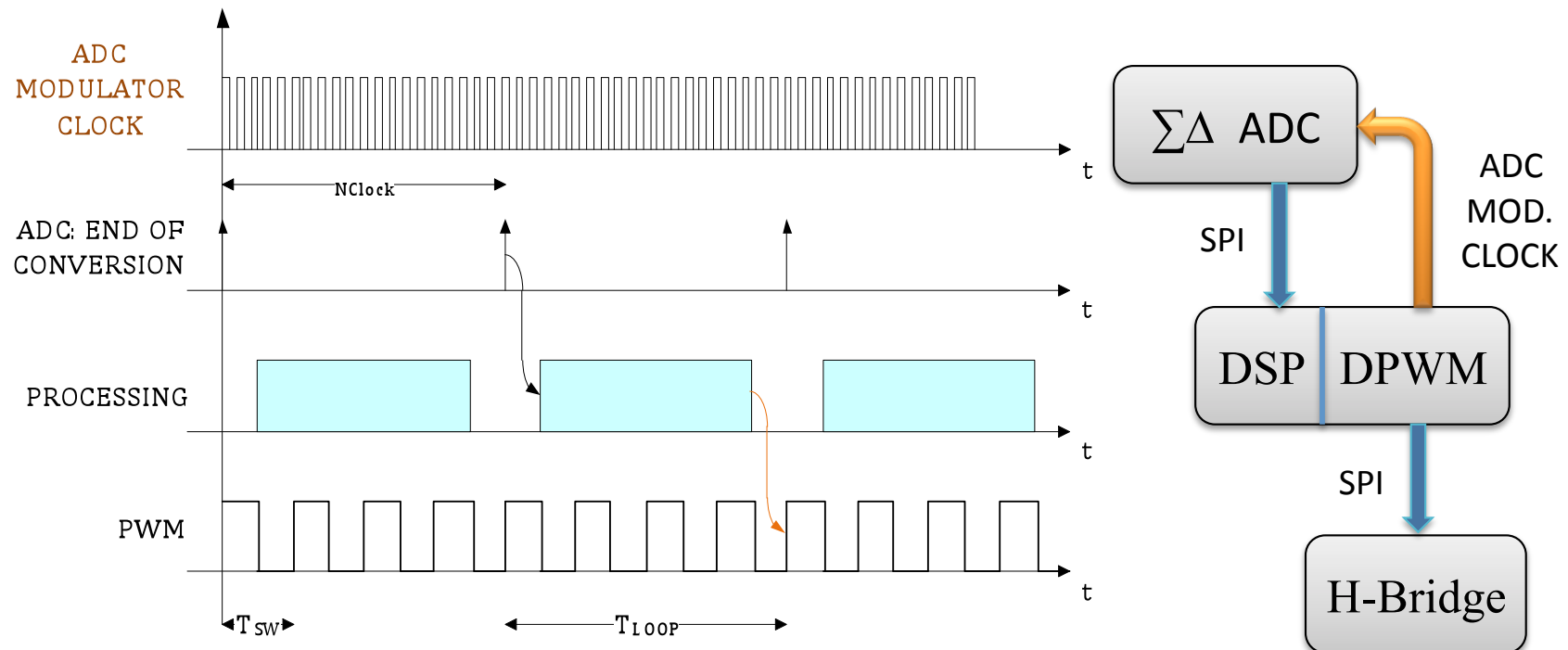
H-Bridge Toggle Technique

- Example of Toggle Technique with 2 bit of dithering.

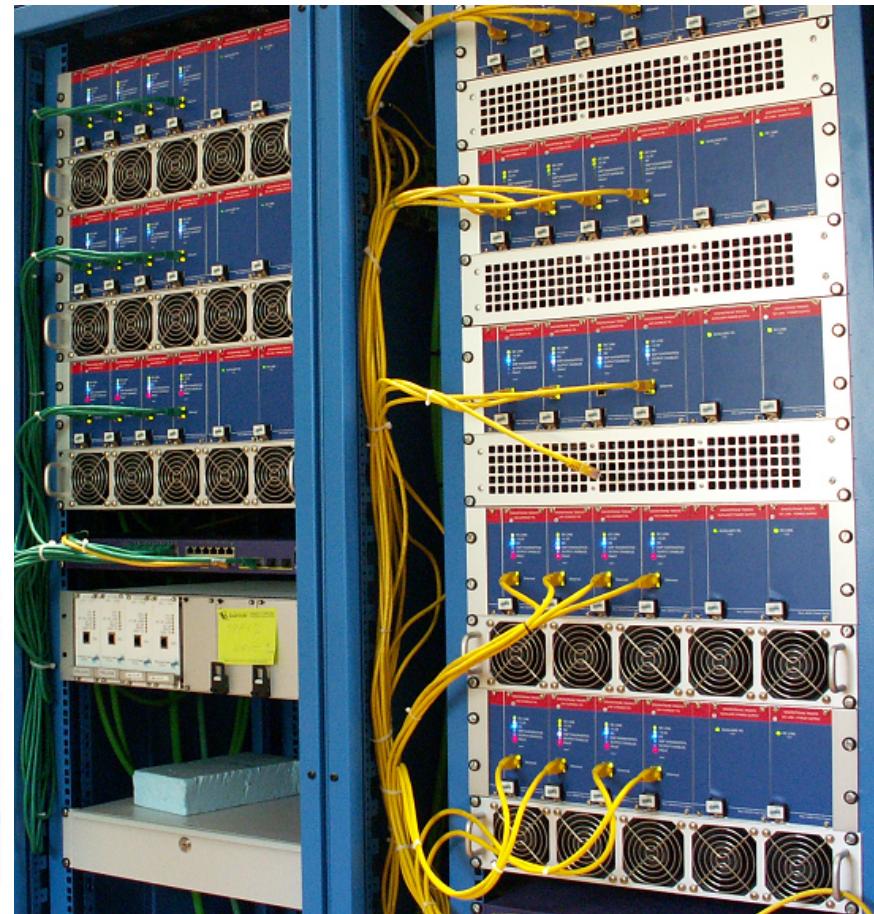
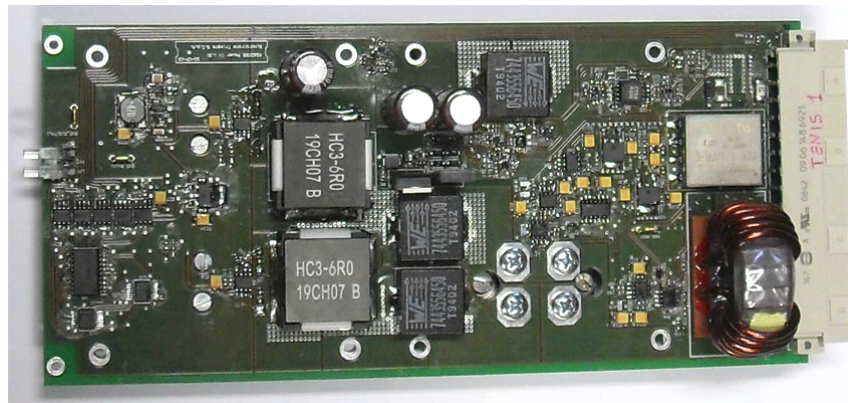


ADC Synchronization

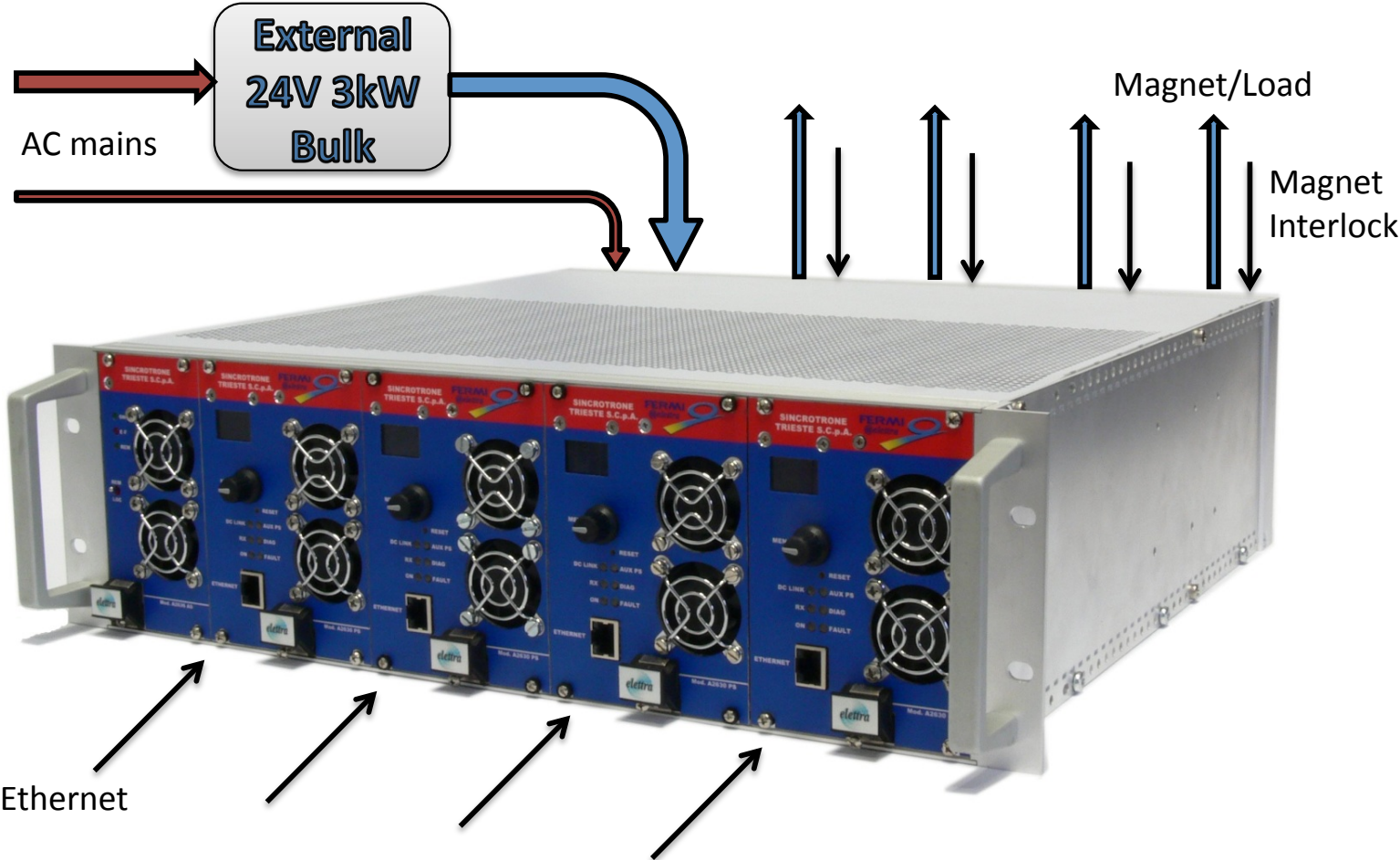
- $\Sigma\Delta$ converter has been used (ADS1252: 24 bit, SPI, SOIC 8 package)
- Problem: ADC hasn't a start of conversion



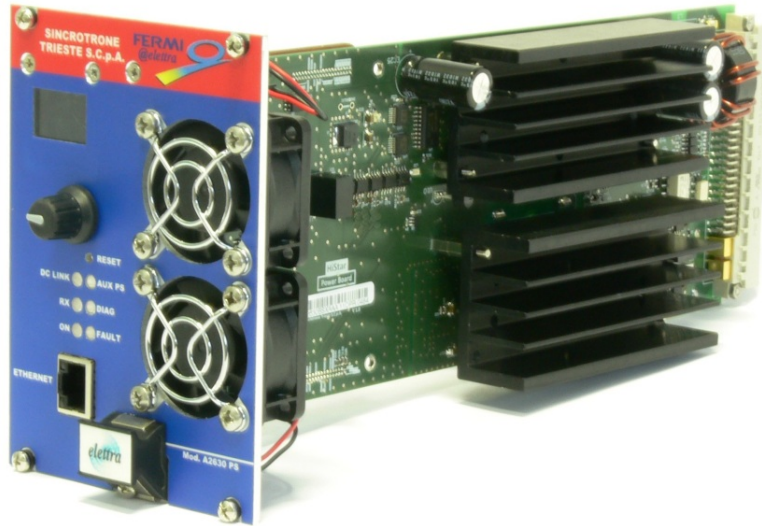
Fermi@Elettra 5A and 20A PS



INFN Frascati - SPARC 30A PS



INFN Frascati - SPARC 30A PS



Output Current	± 30 A
Output Voltage	± 20 V
Output Current Resolution	16 bit
Output Current Ripple	30 ppm / FS
Output Current Stability	50 ppm / FS
DC/DC efficiency (full load)	95%
Switching frequency	104 kHz
Close Loop Bandwidth	1,5 kHz
Accuracy	0.05 %
Max set point update rate	300 Hz

INFN Frascati - SPARC 30A PS



Output Current	± 30 A
Output Voltage	± 20 V
Output Current Resolution	16 bit
Output Current Ripple	30 ppm / FS
Output Current Stability	50 ppm / FS
DC/DC efficiency (full load)	95%
Switching frequency	104 kHz
Close Loop Bandwidth	1,5 kHz
Accuracy	0.05 %
Max set point update rate	300 Hz

Local Control / Monitor	Graphic Color Display, 6 LEDs and Encoder
Extra-Features	Hot-Swap Point-by-Point Current Waveform Loading User-definable interlock thresholds, active levels and timings Remote Updates
Auxiliary ADC	DC Link Voltage Ground Leakage Current MOSFETs Temperature Shunt Temperature
Hardware protection	Input Fuses Earth Fuse Over-Voltage
Internal Interlocks	DC Link Under-Voltage MOSFETs Over-Temperature Shunt Over-Temperature Over-Current Over-Voltage Earth Fault Current Regulation Fault Excessive Current Ripple
External Interlocks	8 Inputs: "dry" contacts 3 Outputs: relay-type
Cooling	Self-Regulated Fans

Conclusion & Consideration

- Digital control is very useful for standardize different range of PS
- Simplifies the additions of extra features (communication, waveforms...)

BUT

- Requires digital hardware and software experts for development and modifications
- Introduces electronic components not easy to repair-replace (FPGA)

SO

- Could we roll back to the old and simple POWER SUPPLY and not waveform generator with OLED display, please?



That's all Folks!

Thank You