





Electronics for Wakefield Monitor

Alexandra Andersson CERN





Wake Field Monitor == Structure BPM

- In addition to the wanted longitudinal accelerating mode, accelerating structures tend to have unwanted transverse dipole modes which are kicked up by off-centre beams.
- These modes need to be damped to prevent their interference with later bunches.
- Some part of these modes can also be coupled out and measured to give information on beam position relative to the electrical centre of the structure.







WFM in CLIC

- To preserve luminosity in CLIC accelerating structures need to be aligned to 5 μm.
- This will be achieved by measuring structure misalignment via wake field monitors, followed by structure alignment via micromovers.



Electronics requirement in CLIC

- Frequency: 18 GHz (possible 26 GHz)
- Accuracy: 5 μm
- Resolution: 5 μm
- Range: ±2 mm (?)
- Factor 400 in Voltage \rightarrow 52 dB dynamic range
- Cheap! There are 144k accelerating structures to equip...





Implications of CLIC requirements

- Simple rather than sophisticated
- 1 readout per pulse, average information only
- Only magnitude, not direction information
 - At the moment we are not sure to have a global, phase stable reference for CLIC
 - We cannot use a Local Oscillator and reference to the beam RF at 12 GHz because of incompatibility with the 18 GHz Wake Field mode.
 - Would need a monopole mode pickup at 2, 6, or 18 GHz.

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TD24 structure (CERN design)

- 24 cells + 2 matching cells + high power couplers separeted from the cells
- No rf absorbers in the damped waveguides
- Diamond machining and diffusion bonding of the disks
- o Tuning studs in every cell
- WFM = 90 deg waveguide bend + pick-ups + rf absorber
 - o Implemented on the middle cell
 - o Screwed on the structure with special feature to ensure good electrical contact
- Cooling circuit
 - Brazed on the structure

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Absolute maximum offset: ~ 3 mm (Limited by Iris radius) Produces output of ~ 2.4 kW for TM mode, 240 W for TE mode

10000







WFM Electronics for qualifying structure prototypes to be installed in CTF3

- Would like to have both phase and amplitude information for TM mode (18 GHz) and TE mode (24 GHz) signals.
- Phase stable reference distribution exists at 3 GHz.
- Can use one single board for both frequencies if LO for first down-mixing stage is 21 GHz.
- Generated IF at 3 GHz is detected with IQ demodulators for phase and amplitude information, and with logarithmic power detectors for large dynamic range







Electronics Block diagram









PCB prototype

- RF laminate: RO4350B
 0,25mm thickness
- Prototype designed with some connections via RF connectors (SMP) to be jumpered with small semi-rigid cables.







PCB prototype measurements

- Measurements only at 18 GHz for now. (24 GHz source was unavailable)
- Input reflections on SMP connectors are terrible. Between 4 and 6 dB...
- Is it the connectors themselves, or a problematic connection to the board?
- Nevertheless, output response is not completely horrible.
- Still, looking to improve input matching in future versions.







Log detector linearity



Quite linear from around 3 dBm to -40 dBm. Useful range a bit wider



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IQ detector linearity



About the same range as the LOG detectors



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Frequency response on the low offset side is terrible.

Phase balance between I and Q is very good. The constant slope phase curve suggest a pathlength error on the order of 50 ps



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ER 12 GHz reference for phase at 18 GHz?

$$V_{LO} = \sin(\omega_{6G}t + \varphi_0)$$

$$V_{LO2} = \sin(2\omega_{6G}t + 2\varphi_0) \qquad V_{Beam} = \cos(2\omega_{6G}t)$$
$$V_{LO3} = \sin(3\omega_{6G}t + 3\varphi_0) \qquad V_{WFM} = \sin(3\omega_{6G}t + 3\varphi_{WFM})$$

$$V_{LO2} \times V_{Beam} = \frac{1}{2} \left(\sin(2\varphi_0) + \sin(4\omega_{6G}t + 2\varphi_0) \right)$$

$$V_{LO3} \times V_{WFM} = \frac{1}{2} \left(\sin(3\varphi_0 - \varphi_{WFM}) + \sin(6\omega_{6G}t + 3\varphi_0 + \varphi_{WFM}) \right)$$

$$\varphi_0 = \varphi + \pi$$

$$2\varphi_0 = 2\varphi + 2\pi = 2\varphi$$

$$3\varphi_0 = 3\varphi + 3\pi = 3\varphi + \pi$$

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