

Magnetic Measurements with the PS B-train

Recent results and planned developments

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Introduction

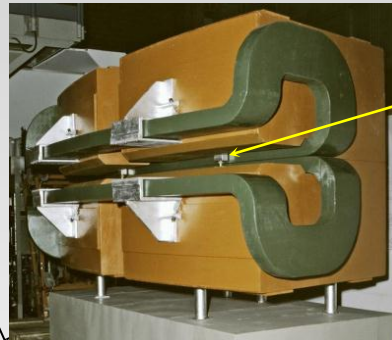


real-time dipole field measurement for accelerator operation

sometimes $B(t) \neq f(I, dI/dt, I(\tau \leq t))$

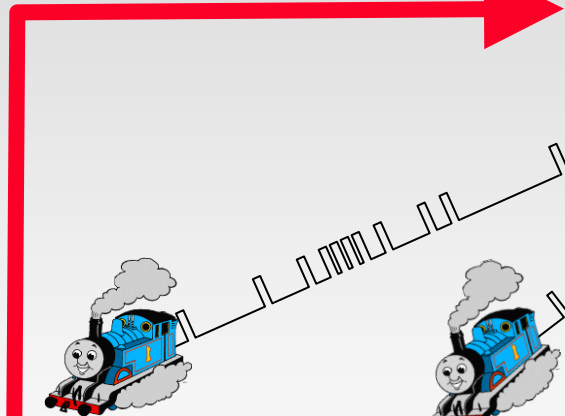
- Hysteresis (“Pulse to Pulse Modulation”)
- Temperature
- Eddy currents
- Ageing of iron yoke
- $I(t)$ measurement accuracy
- Timing jitter

reference dipole
(in the ring or external + series excitation)

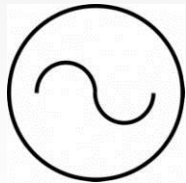
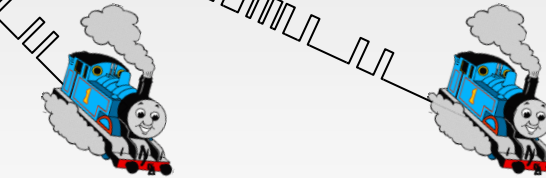


local measurement $\Rightarrow \oint B dl$
whole ring

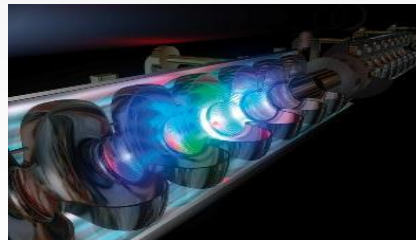
$I(t)$



$B(t)$



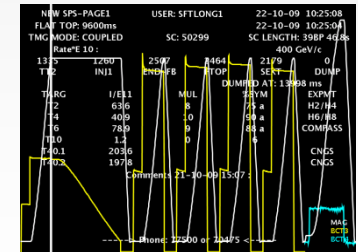
power supply control (B or dB/dt)



RF control (mandatory)

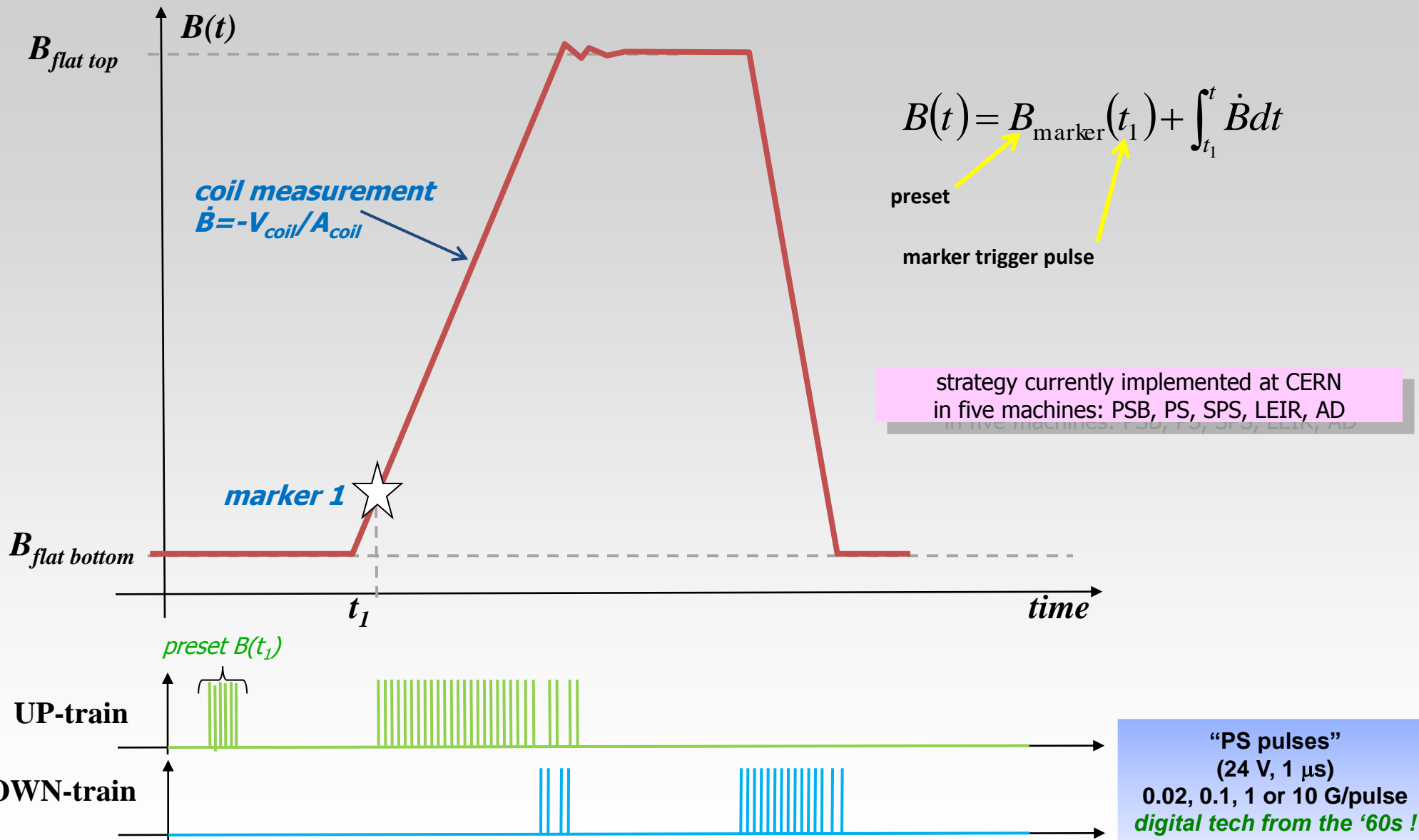


beam diagnostics
(beam current DCCT)



qualitative feedback to operation

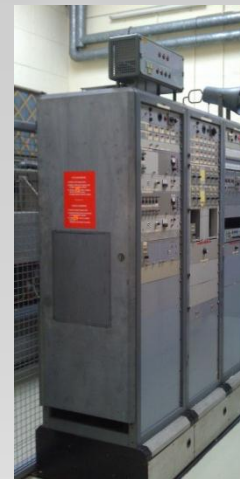
B-train working principle – dynamic marker



3 systems in the lifetime of PS:

- **Mark I (1960-1980)**

- included in the original magnet design
- only one side of U101 (block 4) equipped with peaking strips @ 50 G + flux coils
- 3 channels @ 0.1, 1 and 10 G;



Mark I



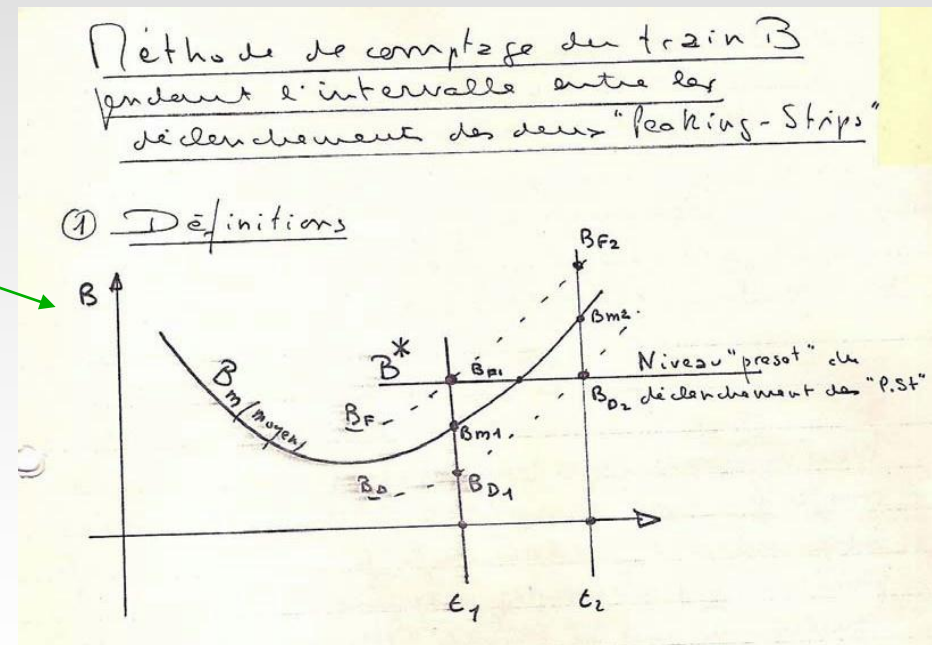
Mark II & III

- **Mark II (1980-2000)**

- motivation: adapt to multi-batch filling + new F8L
- different F/D fields → sensors in both halves
- new peaking strips designed for up to 400 G + new flux coils
- extra logic to interpolate B(t) between triggers

- **Mark III (2000 -)**

- motivation: improved precision and reliability
- peaking strips @ 50 G (D only) + flux coils (D + F)
- peaking strip F removed from B-train, after a few tests, in order to be compatible with the newly installed PSB/AD systems

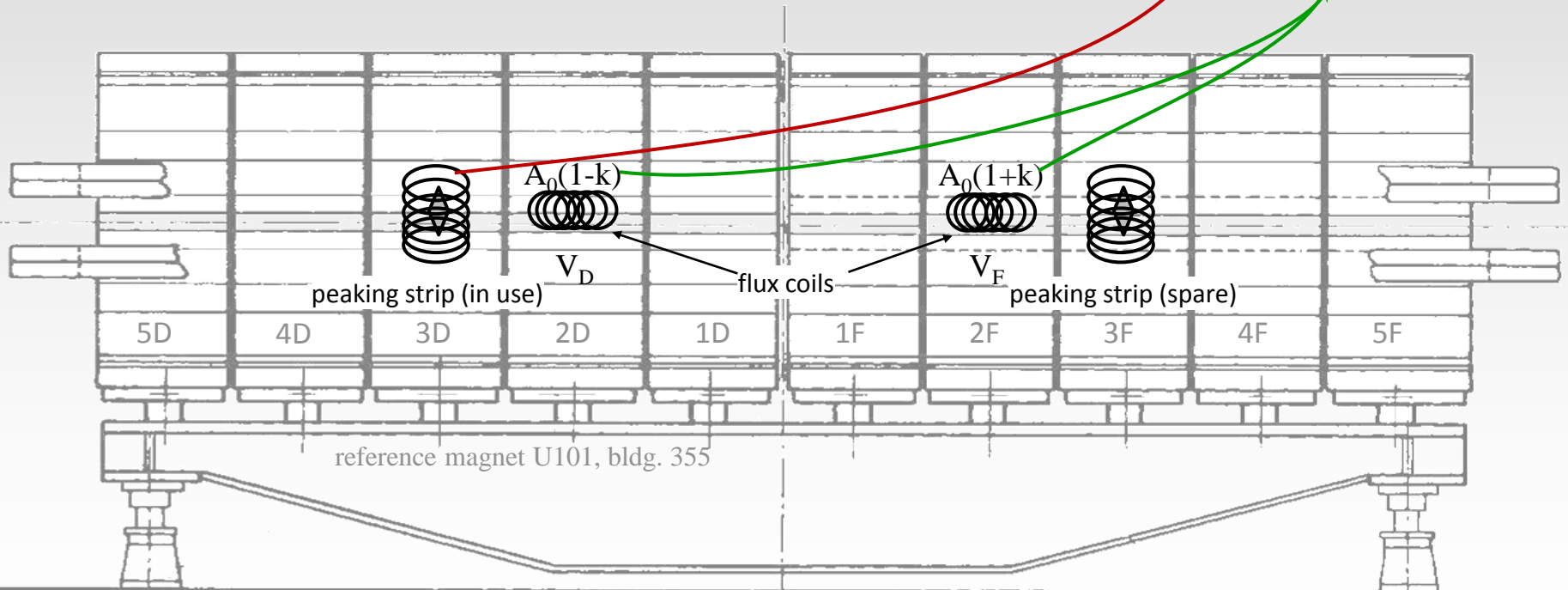


R. Gouiran, 1979

CERN PS B-train: current configuration

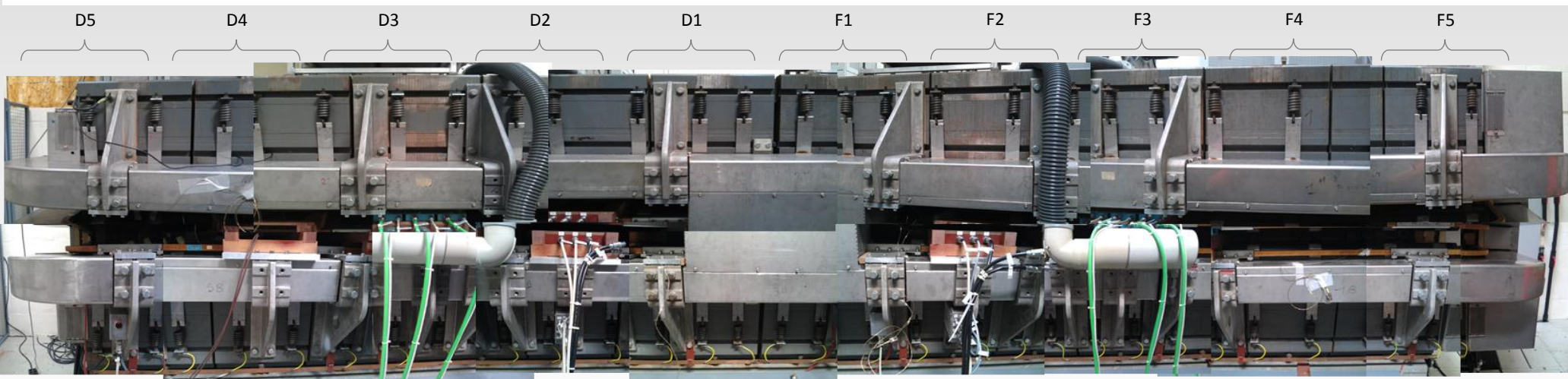
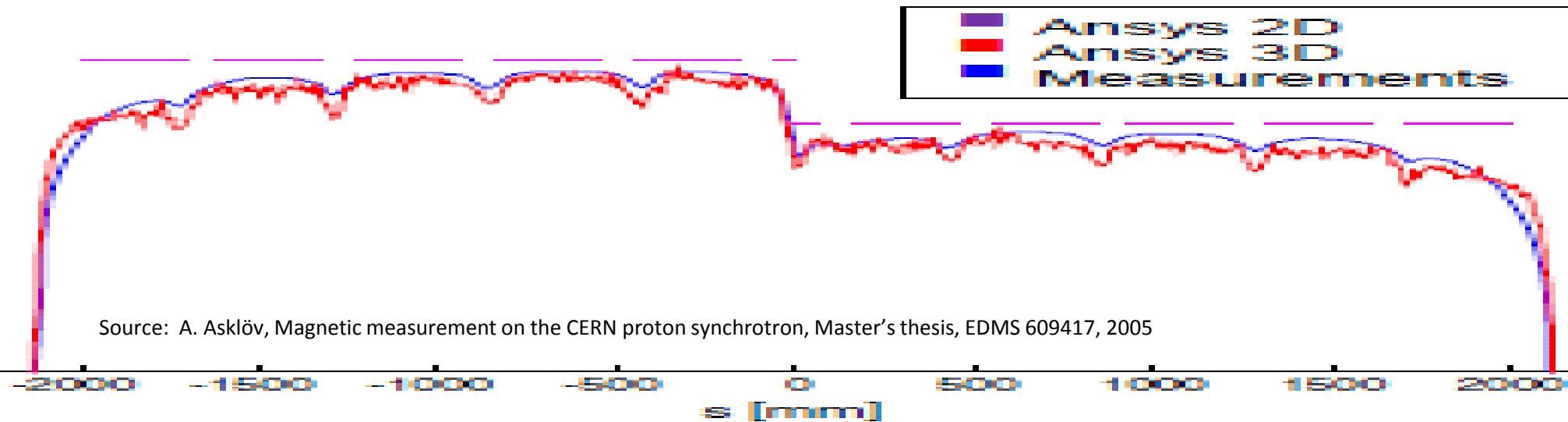
- B-train = **weighted average** of the local field in F/D halves
- **In-built assumptions:** $B_D(t_0)=B_F(t_0)=4.98 \text{ mT}$, $k=0.091^\dagger$
- Only one field marker in use (the second one is foreseen only for diagnostics)
- When these assumptions are not satisfied \rightarrow **B-train error** \rightarrow **beam may become unstable**
- Integrator is reset at the end of every magnet cycle (synch to control system timing)
- Integrator internal auto-calibration every 15 cycles (*not synch to timing*)

$$B_{avg} = \frac{1}{2} ((1-k)B_D + (1+k)B_F) = 4.98 \text{ mT} + \frac{1}{2A_0} \int_{t_0}^t (V_D + V_F) dt$$



[†] k takes into account an oscillating dipole component generated by the PFW in 3-current mode. See: R. Guirán, Enroulements polaires de l'aimant du CPS, Note Interne PS/SM 76-1

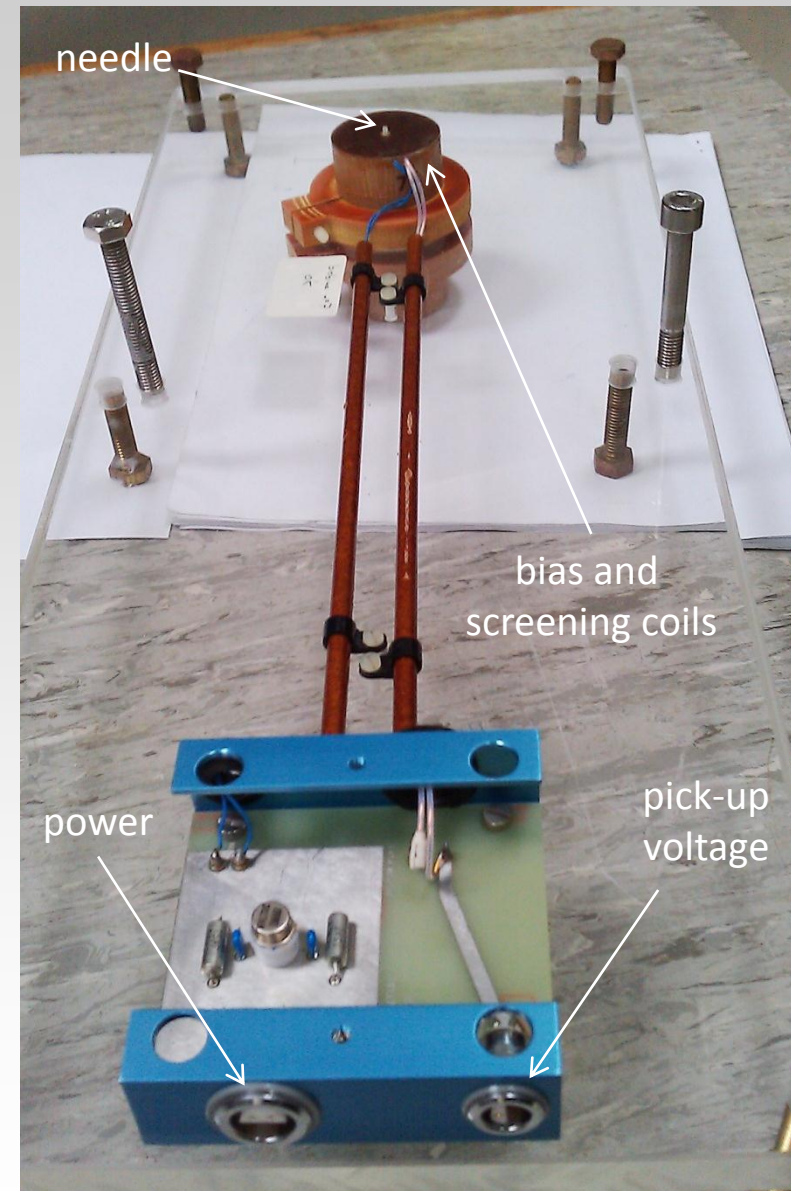
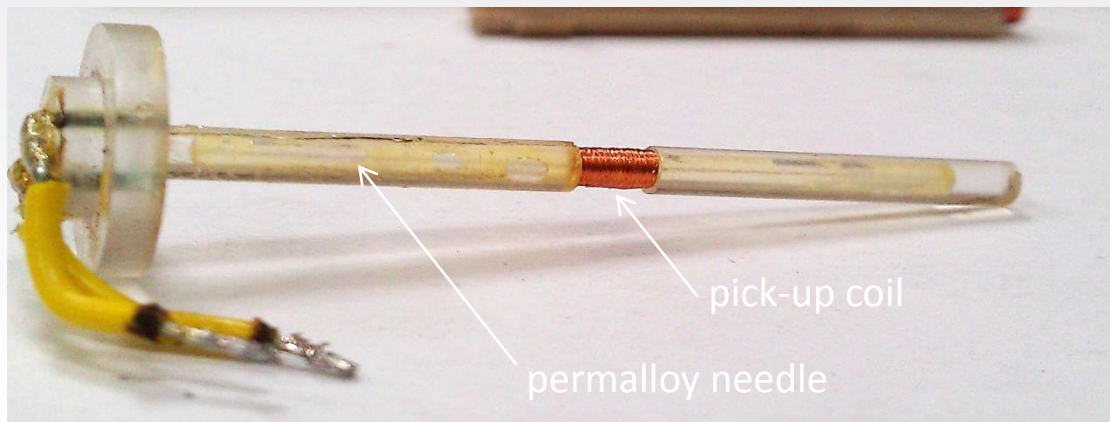
Reference magnet U101

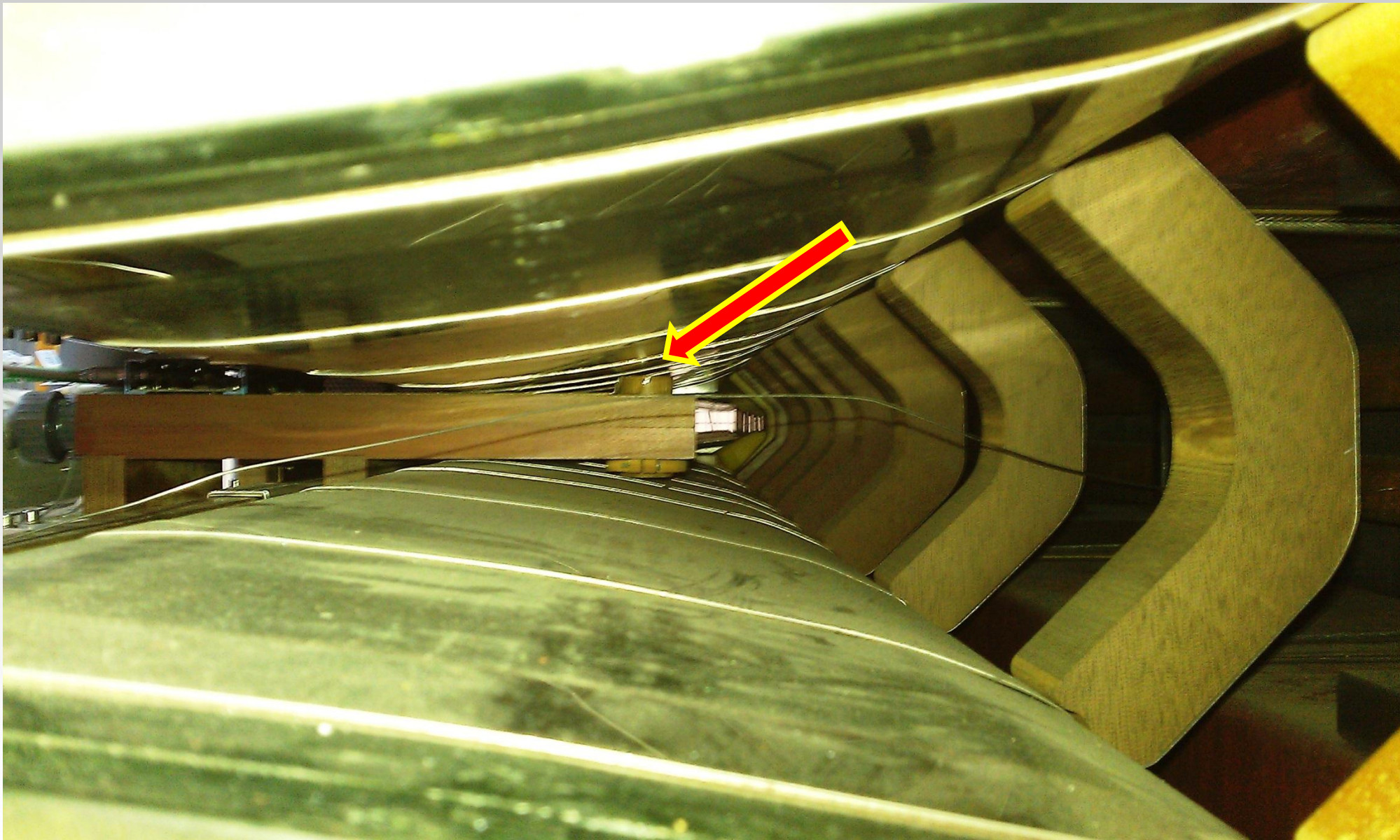


- Longitudinal field distribution varies with I_{F8L} , I_{PFW} saturation
- Calibration vs. beam depends on sensor position
- Sensors may not work properly in end blocks (inhomogeneous/3D field)

Existing field marker: peaking strips

- Developed at CERN in 1956 specifically for combined-function PS magnets.
- Based on a **pre-stressed bi-stable magnetic needle**: magnetization flips over at a preset level and generates a pulse detected by a pick-up coil
- Two coils powered in series for **bias** and **screening field**, pulsed to avoid overheating
- Main constraints: **bias coil heating at high field** (> 5 mT); does not work **at too high or too low dB/dt**
- Experience shows that this sensor **is very stable** (drift < 50 μ T in 20 years),
- Very few spares available, all different from each other (making new ones is difficult)







- Based on National Instruments hardware: PXIe chassis, RT controller, 2× 6366 DAQ cards (8×2 MHz parallel 16-bit channels + 32×10MHz digital I/O channels)
- Acquisition of all sensors and B-train signals
- Introduced in 2010 to track the magnetic state in F/D halves
- Used to check the consistency of sensor and B-train signals, timing and currents
- Testbed for the design of new sensors and acquisition algorithms

NB: high sample rate in the 200 kHz range is necessary for:

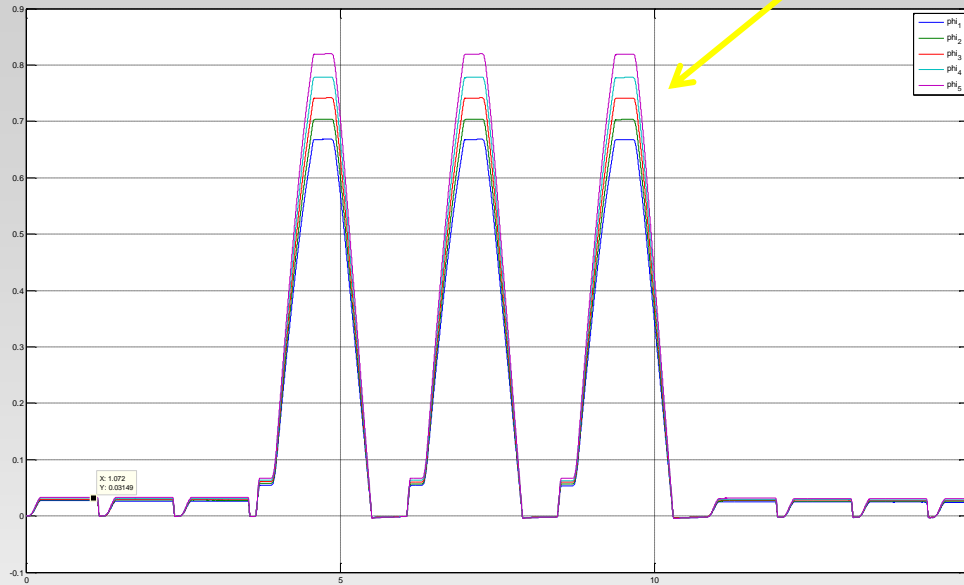
- accurate integration of flux coil output
- correlation of current/field to few 0.01 G



- New 5-coil, 102 mm fluxmeter for the dynamic measurement of normal harmonics $B_1 \rightarrow B_5$
- First test carried out 16/17 feb
- Support can be quickly repositioned in all free F or D blocks (optimal position to be defined)
- Second symmetrically placed unit foreseen

5× measured integrated fluxes

wanted harmonics
(normal only)



$$\{\Phi_j(t)\} = [k_{jn}]\{B_n(t)\}$$

$$k_{jn} = \frac{A_j}{n w_j r_{ref}^{n-1}} \left[\left(r_j + \frac{w_j}{2} \right)^n - \left(r_j - \frac{w_j}{2} \right)^n \right]$$

weight matrix

(NB: problem intrinsically ill-conditioned due to its polynomial-fitting nature)

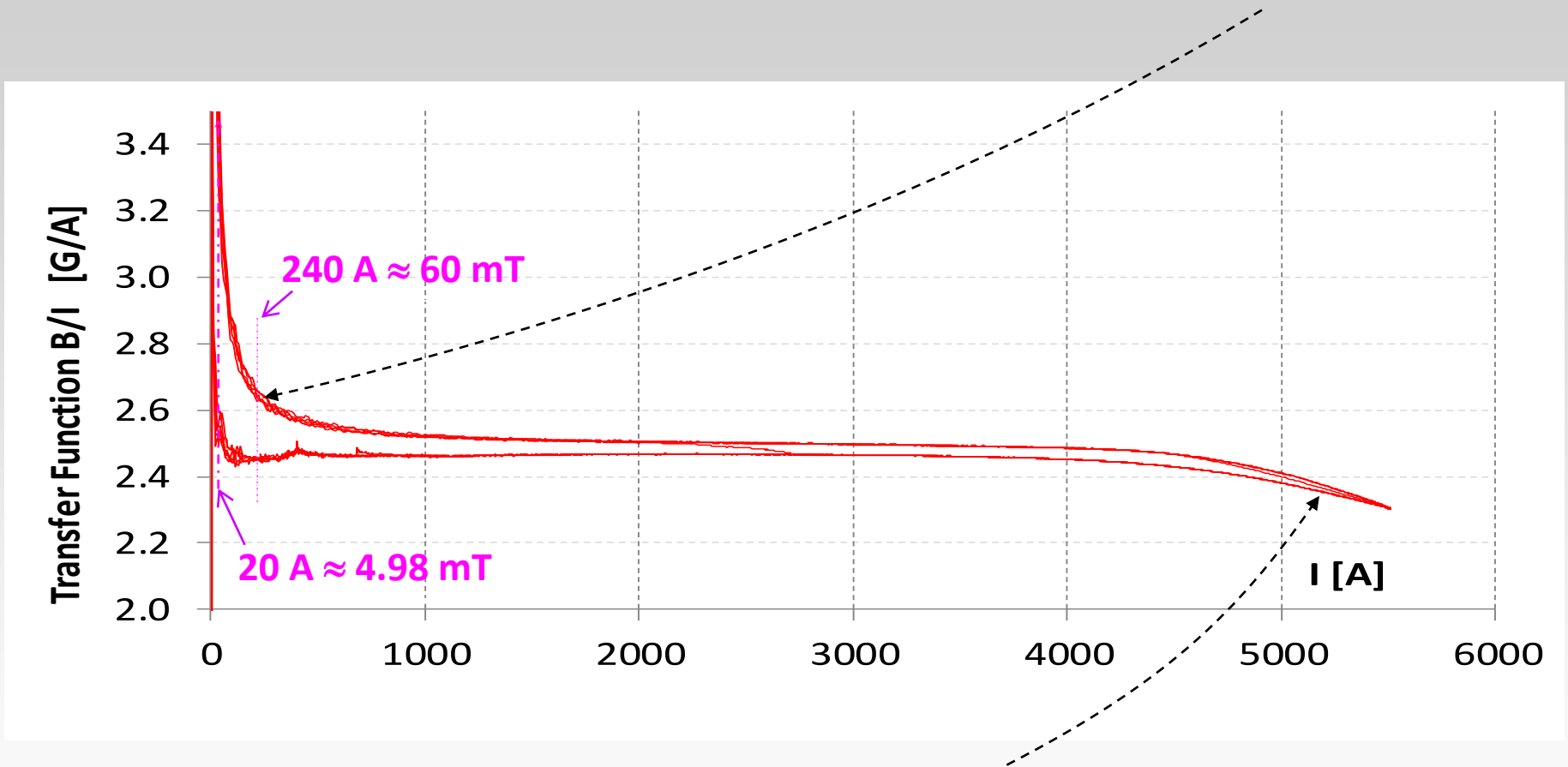
Open issues:

- integrator drift is very sensitive to calibration errors → only B1 and B2 OK for the moment (more accurate calibration in preparation)
- truncation error due to neglected $n \geq 6$ harmonics to be evaluated
- major improvements planned:
 - analog bucking of main components
 - circular array of tangential coils (well conditioned calibration + skew harmonics + possibility of rotation to measure accurately remanent harmonic components)

Test results highlights



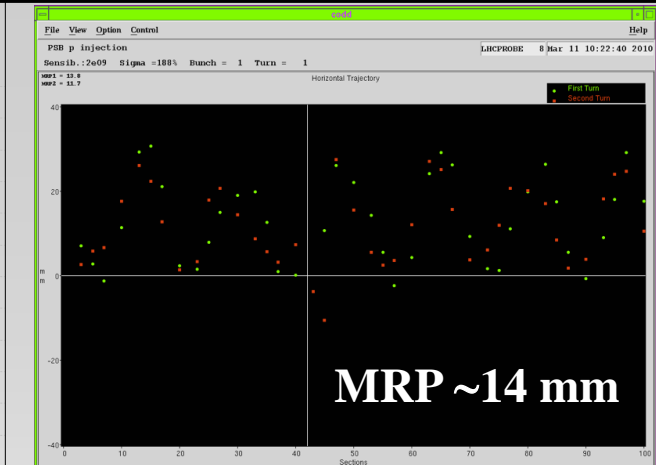
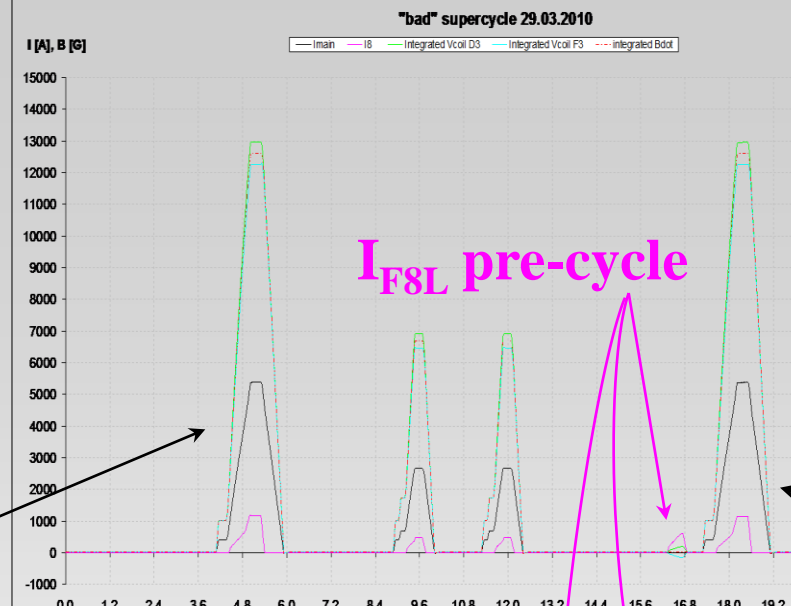
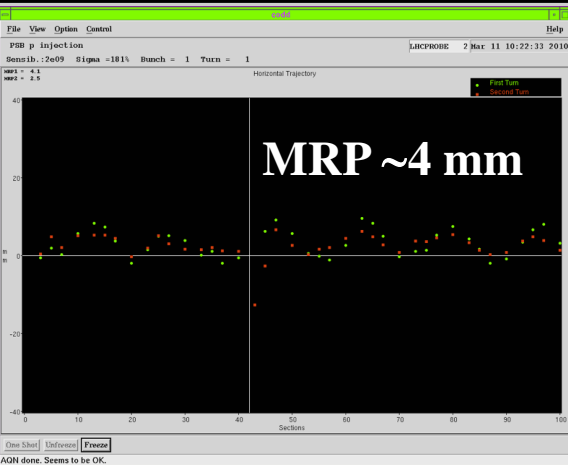
large fluctuations due to history-dependent residual field



going into saturation erases the previous magnetic history

Pre-cycling at high field + higher minimum current → better field stability

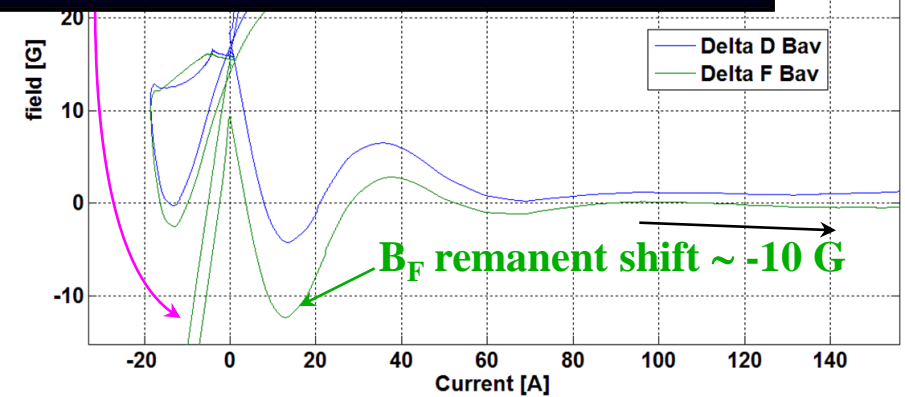
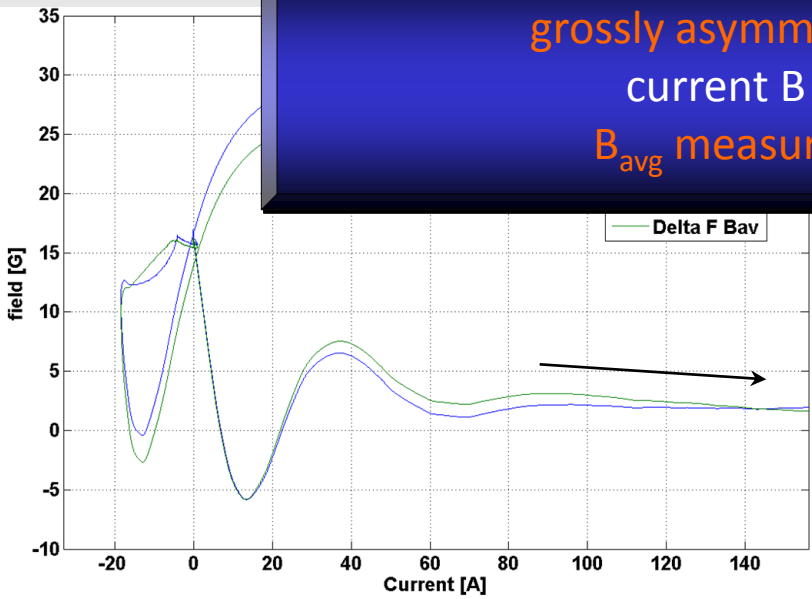
PS beam stability test – April 2010



"good" cycle

"bad" cycle

grossly asymmetric powering imbalances B_F and B_D
 current B train system is blind to initial B_F
 B_{avg} measurement error \rightarrow beam MRP error



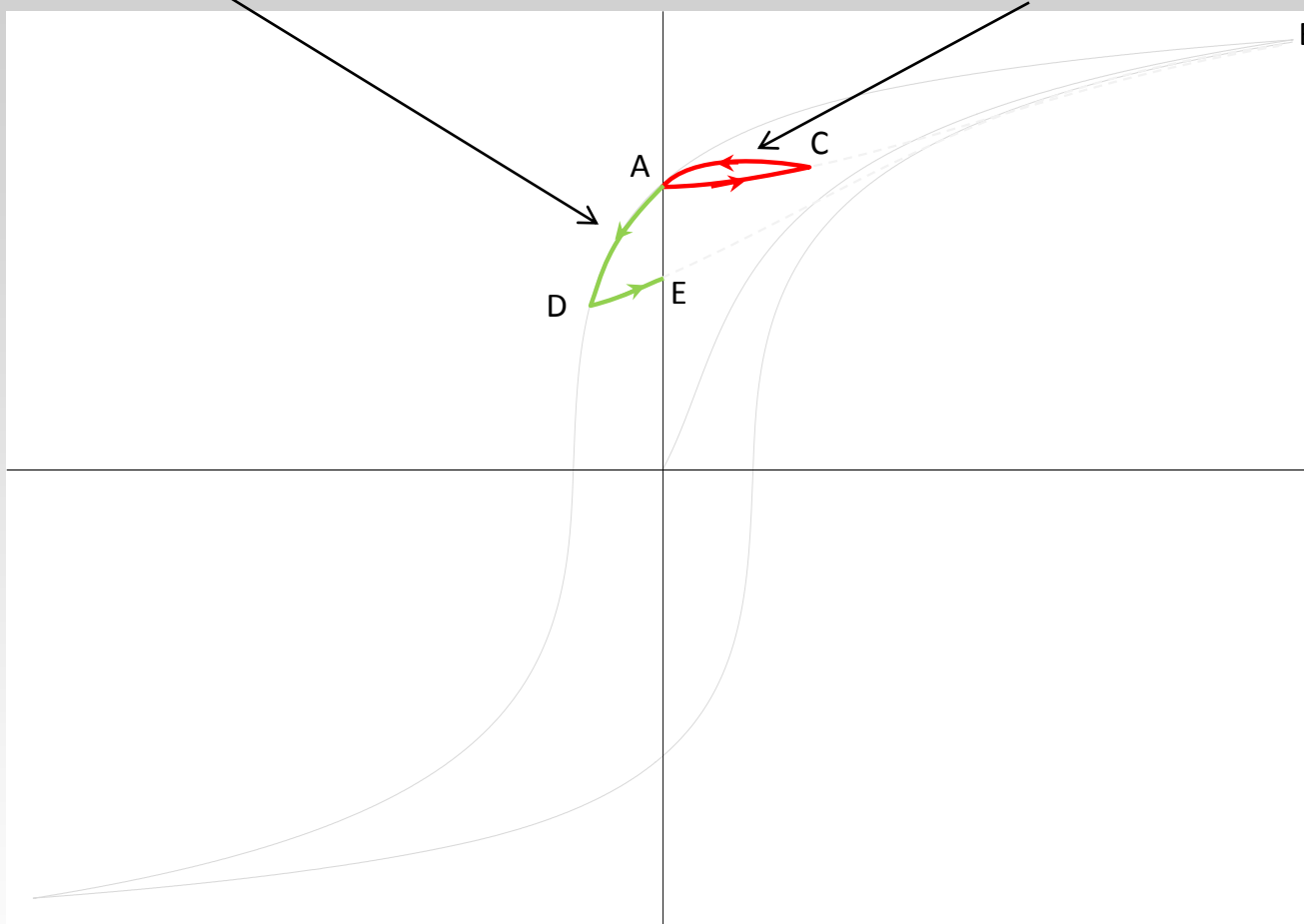
Hysteresis curves $B_F(I)$ and $B_D(I)$ – Difference w.r.t. straight line fit



Field imbalance

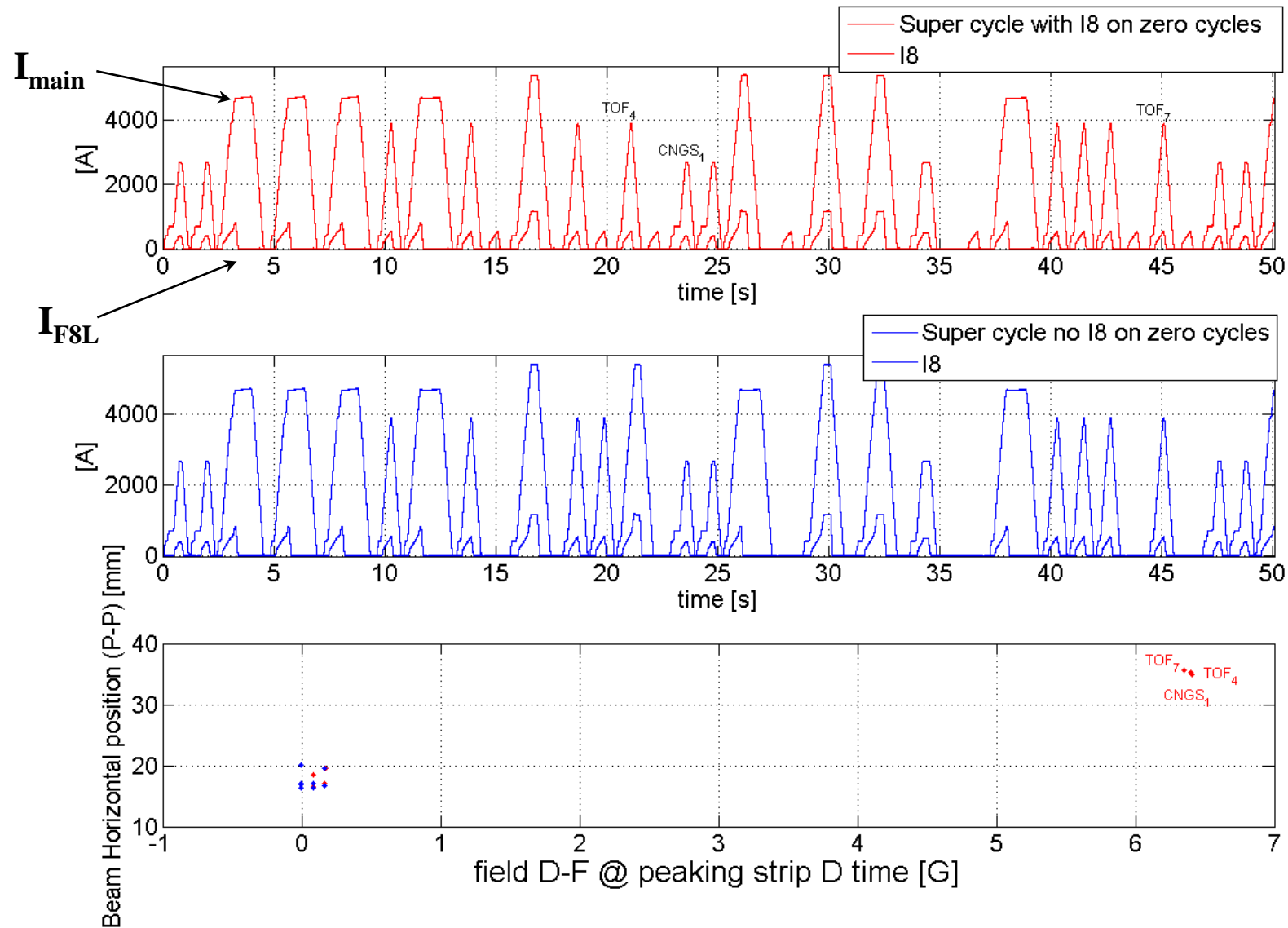
negative quadrupole increment in the defocusing half
the field changes along the downwards limit cycle

positive quadrupole increment in the focusing half
the field changes towards the upwards limit cycle



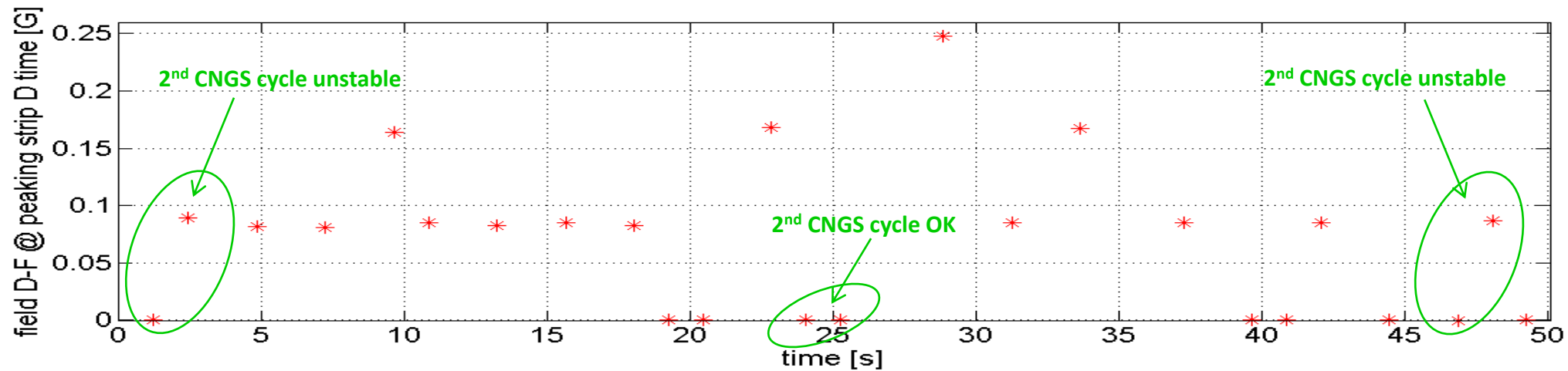
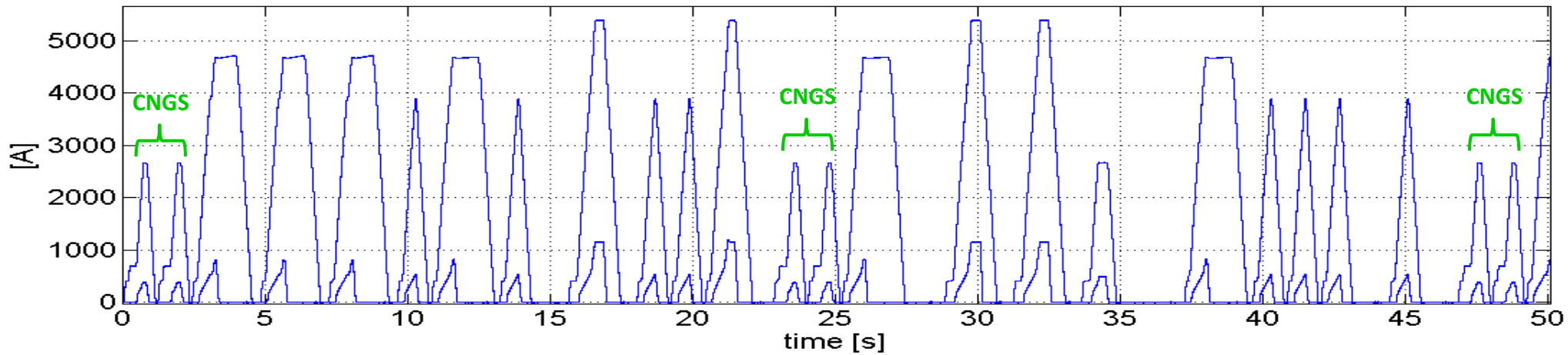
- dipole and quadrupole component are always proportional → tune trim circuit operation (I_{F8L}) at low field causes field imbalance in the two magnet halves
- Independent field marking in the two halves is necessary for correct $B(t)$ integration

PS beam stability test @ injection – June 2011



systematic correlation between initial field imbalance (\rightarrow B-train error) and large RP error

Super cycle no 18 on zero cycles

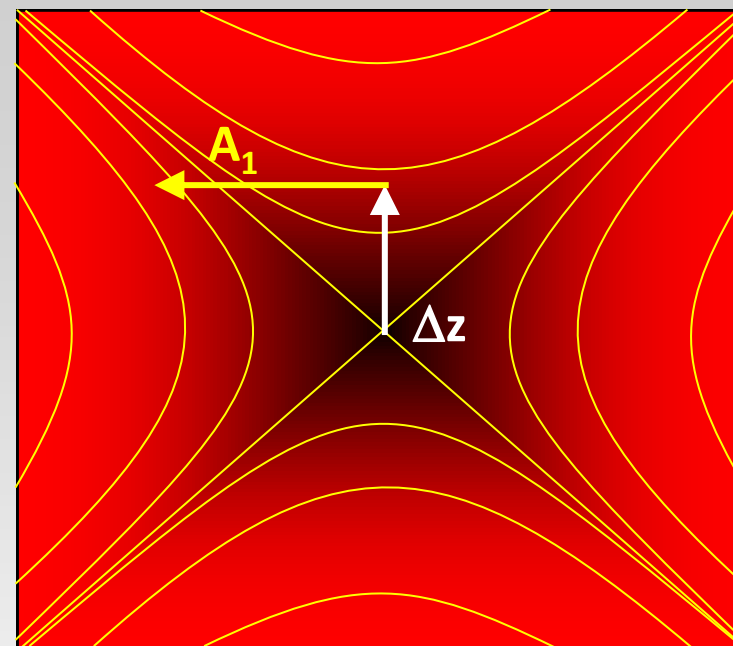
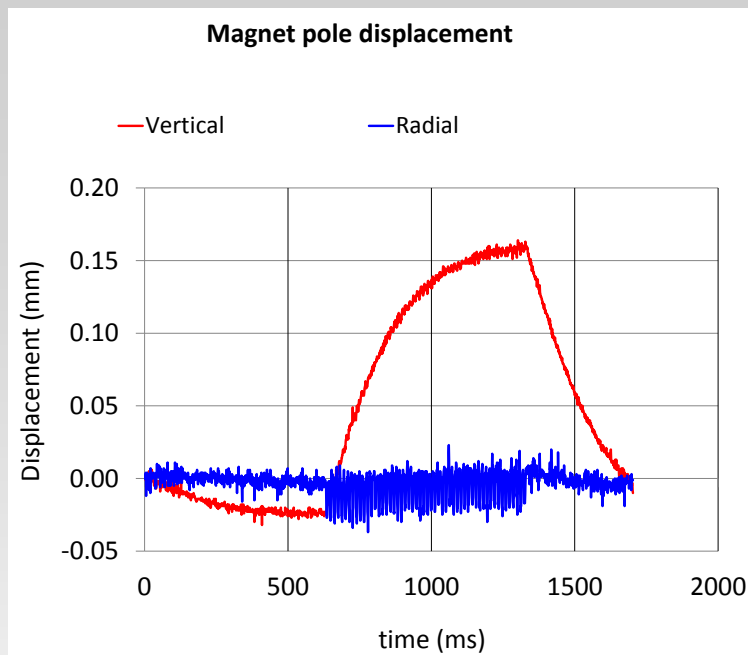


specific powering cycle sequences lead to reproducible MRP errors
 now we can observe correlation with reproducible magnetic errors

Courtesy A. Beaumont



Laser tracker measurements done on U17 to test stability of NMR marker support in 2008



- Vertical displacement of bottom pole = 0.16 mm @ 5400 A (gap closes due to magnetic forces)
- All B-train sensors fixed to bottom pole
- Top/bottom asymmetry → possible vertical displacement of sensors w.r.t. to ideal axis
- **Feed-down → skew spurious harmonics** e.g. 0.1 mm → 5G
(systematic effect as a function of main current)

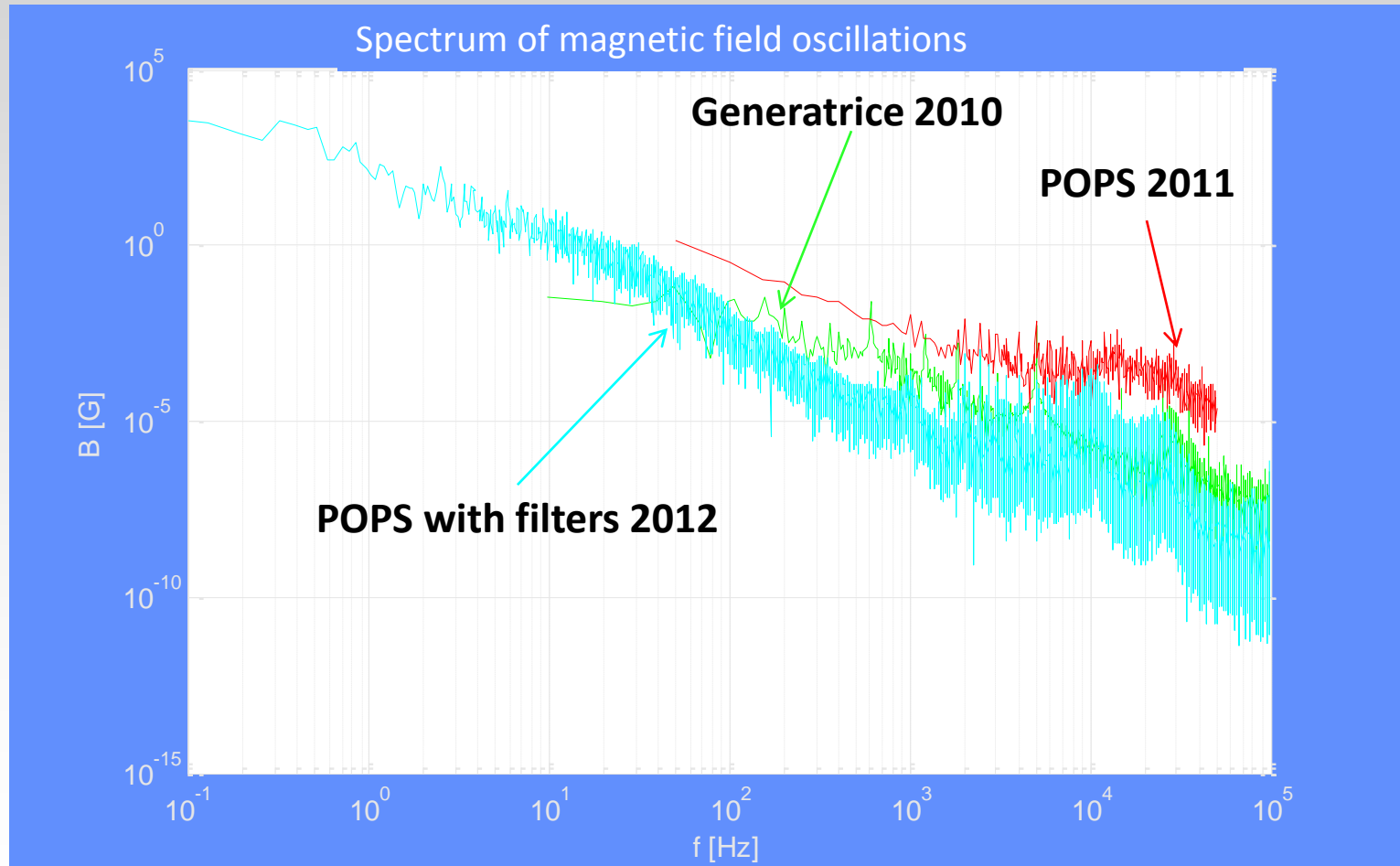
Abnormally high noise on Bdot reported by Ops in Feb 2011 (concern for RF).

- B(t) noise at $f > 1$ kHz found to be up to 40 dB higher in 2011 than 2010
- Observations consistent with actual (small) B fluctuations (see EDMS 1168725)
- Noise levels dropped back well below 2010 levels in 2012

Field spectrum inferred from flux coil measurements in U101:

(spurious resonances $> 10^4$ Hz possible)

$$\frac{B}{V_{coil}} = \frac{1}{i2\pi f A_{coil}}$$

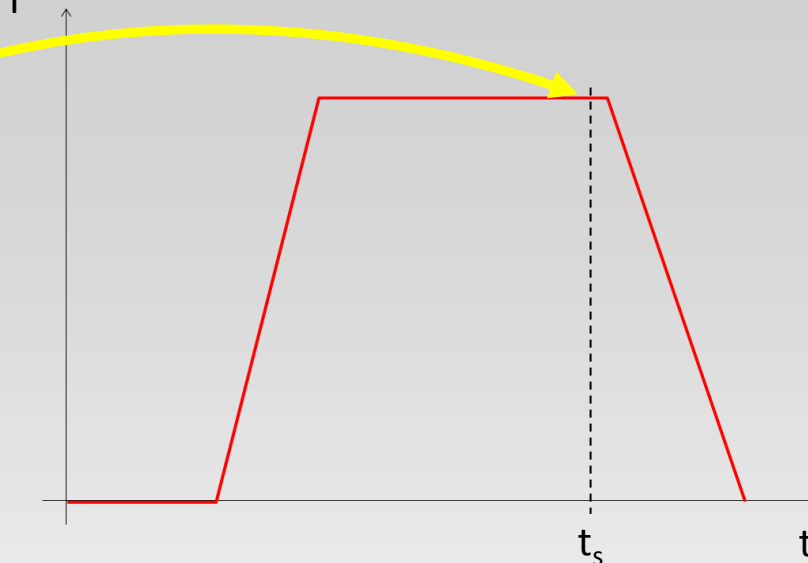
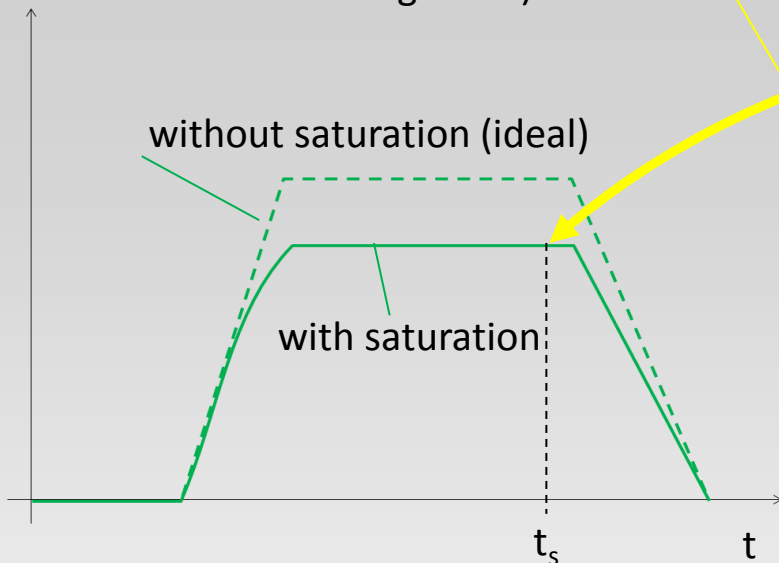


Dynamic effects (eddy currents)

Eddy current measurement method

Scale so that the curves coincide at the end of the flat-top

B (obtained from flux coil integration)



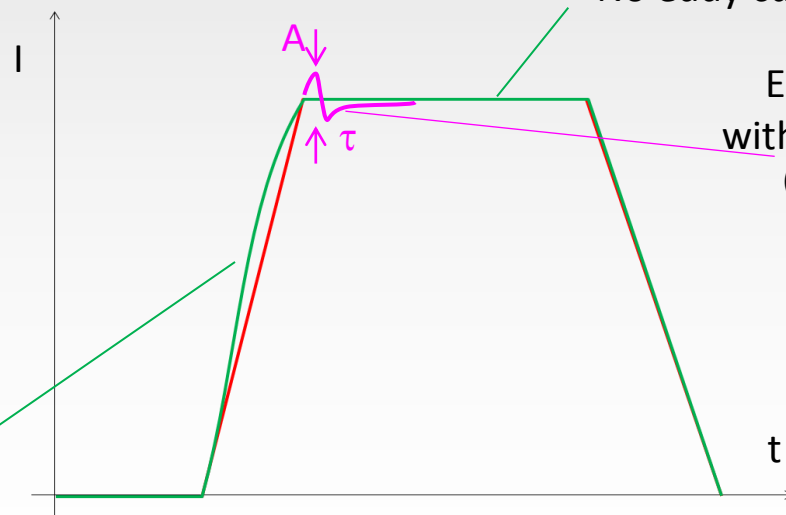
equivalent current:

$$I^*(t) = \frac{I(t_s)}{B(t_s)} B(t)$$

relative field error

$$\varepsilon(t) = \frac{I^*(t)}{I(t)} - 1$$

apparent field advance
due to the scaling procedure

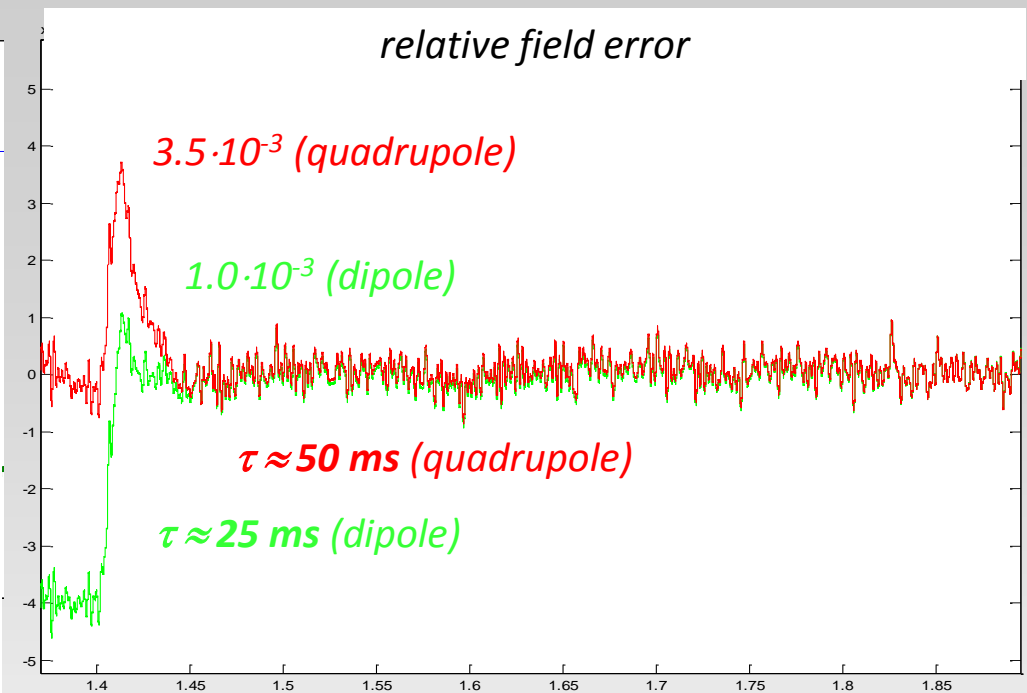
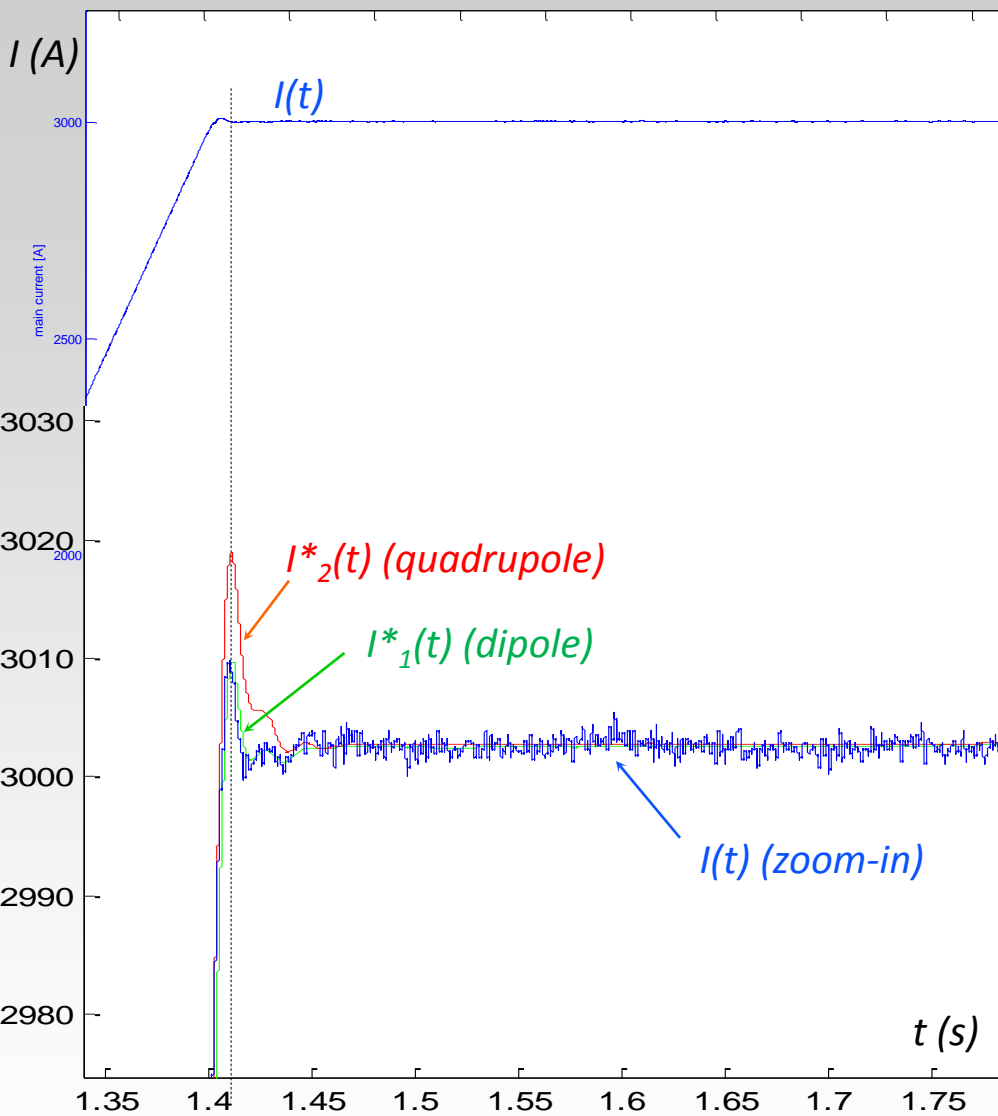


No eddy currents $\rightarrow I^* \equiv I$

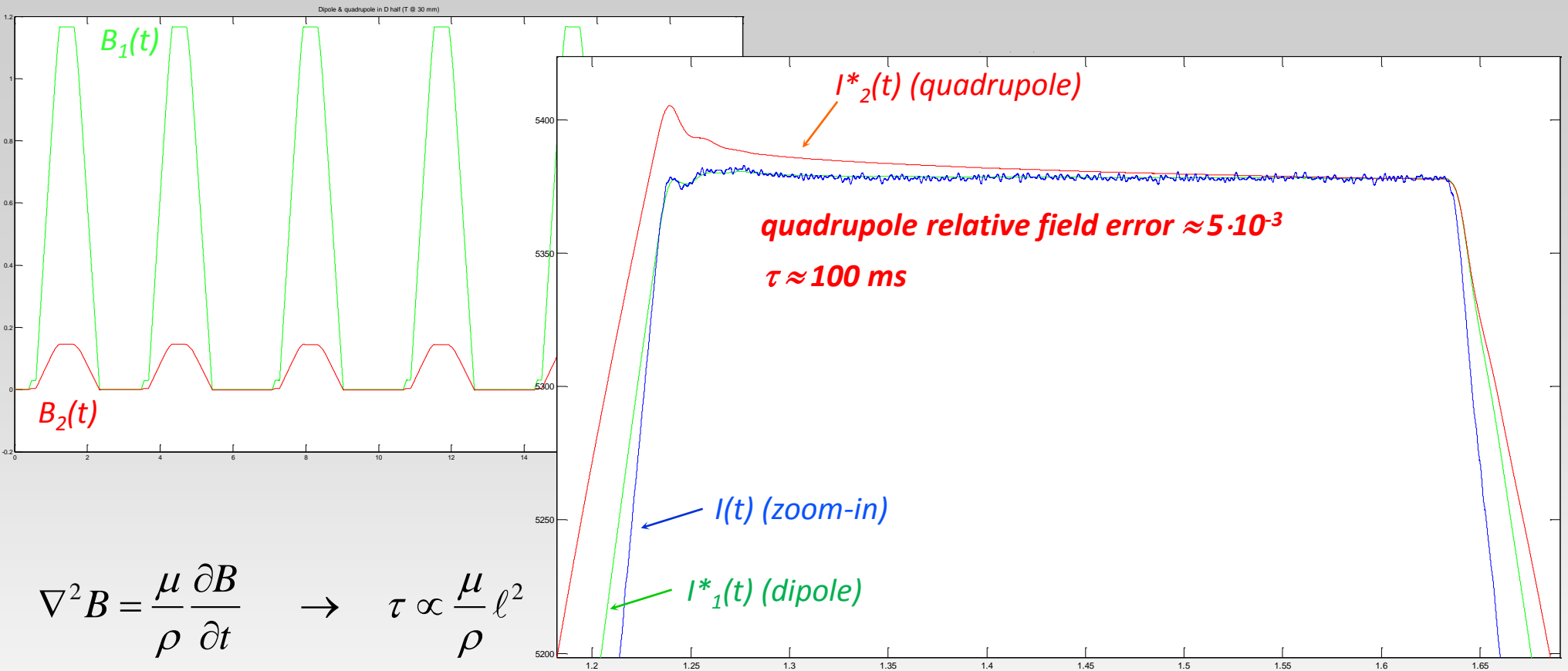
Eddy currents show up as a transient
with exponential decay at end of ramp-up
(NB: direction depends on current overshoot!)

Ideal model:

- Transient amplitude $A \propto dB/dt$
- Decay time $\tau = \text{constant}$



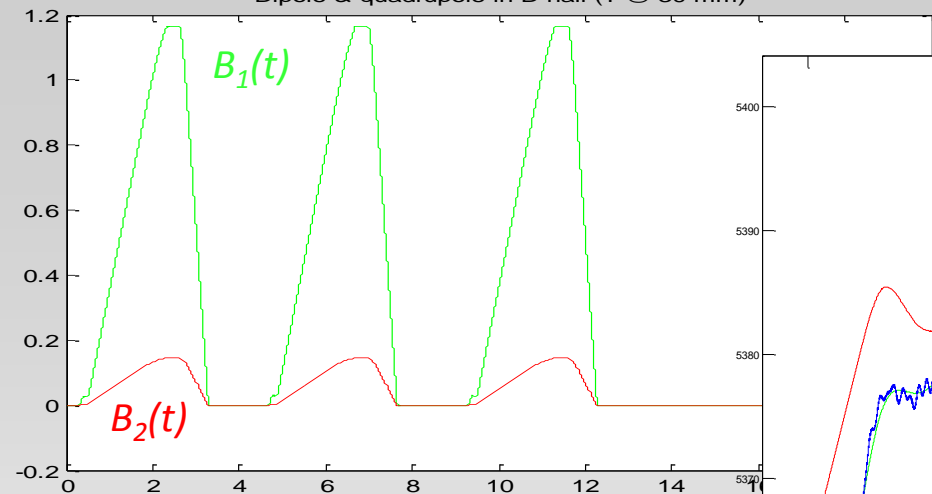
- At low field (linear range) eddy current effects are relatively small and fast
- Quadrupole effects significantly larger than on dipole



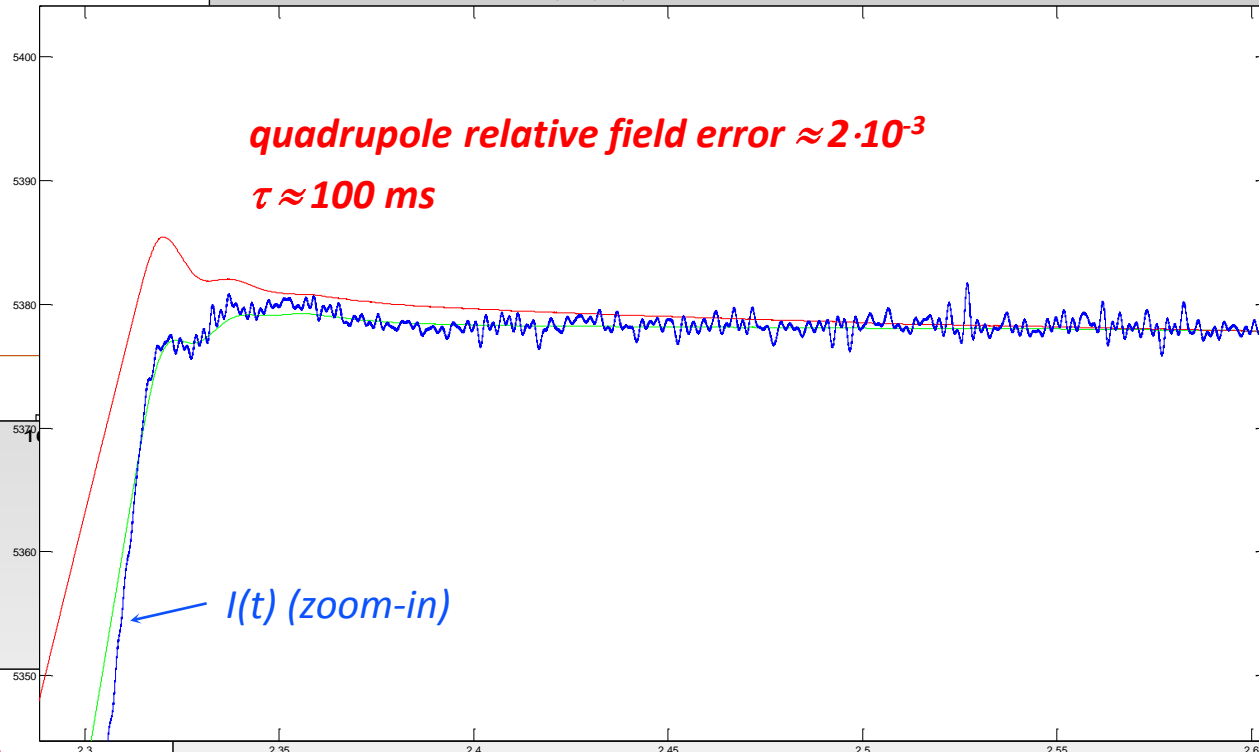
$$\nabla^2 B = \frac{\mu}{\rho} \frac{\partial B}{\partial t} \quad \rightarrow \quad \tau \propto \frac{\mu}{\rho} \ell^2$$

- Respect to intermediate field, the time constant and effect amplitude on the quadrupole are $\sim 2 \times$
- No appreciable change on the dipole
- Dimensional considerations based on magnetic field diffusion equations \rightarrow time constant at high field expected to drop by 20% due to saturation (low μ) only \rightarrow dominant effect appears to be increase of magnetic circuit length due to flux leakage

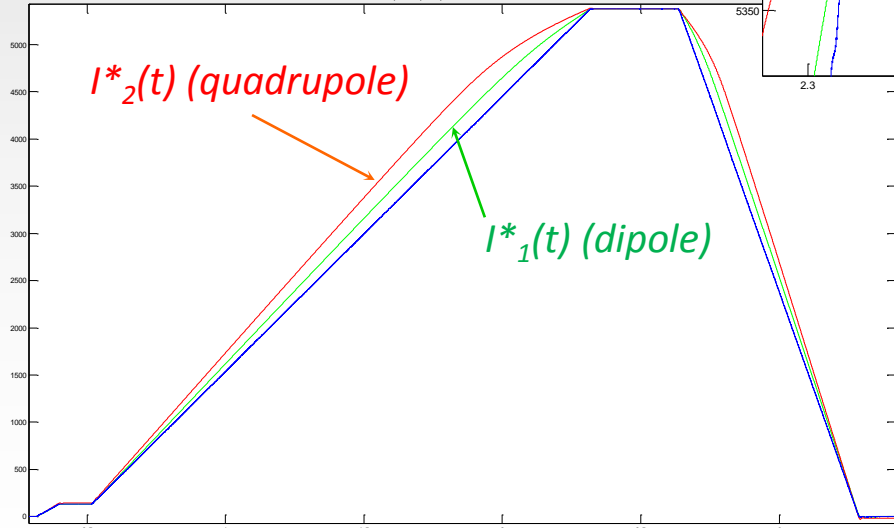
Dipole & quadrupole in D half (T @ 30 mm)



Dipole & quadrupole scaled in D half & I main

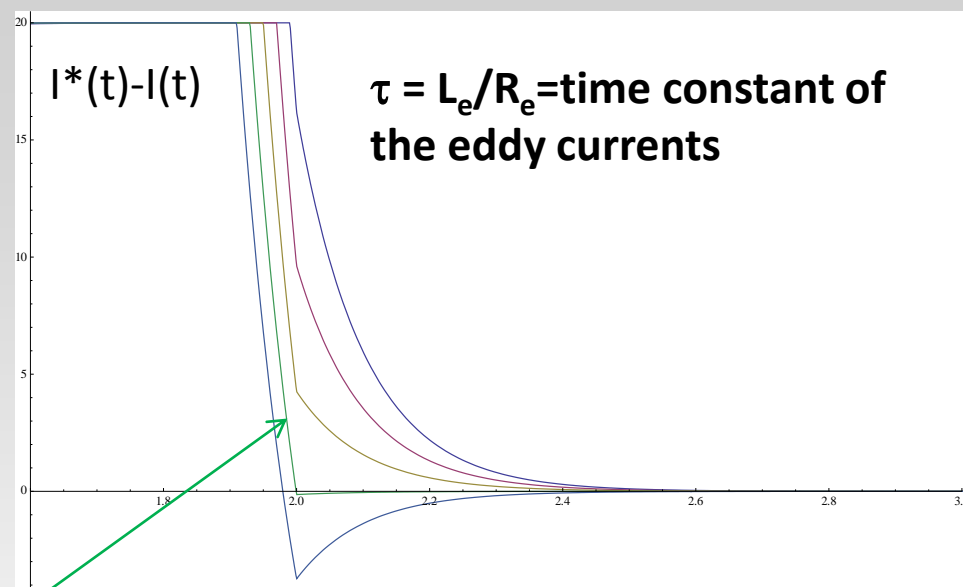
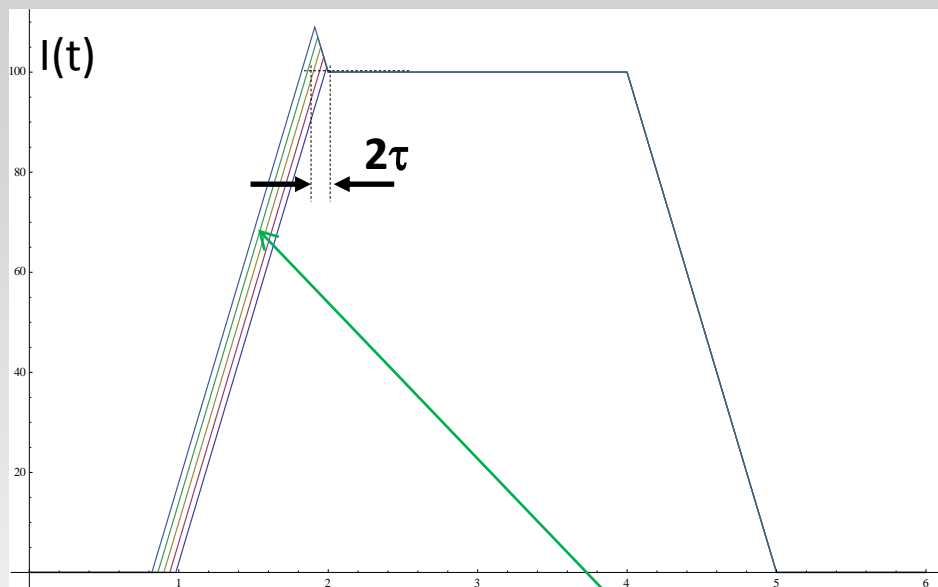


Dipole & quadrupole scaled in D half & I main



- Results at high field with reduced ramp rate show smaller relative field error → behaviour consistent with eddy currents
- Equivalent current plot indicates much higher saturation effects on quadrupole than on dipole

Compensation of eddy currents by controlled current overshoot:



Ideally, a linear overshoot of length 2τ can cancel out completely eddy currents on the flat-top

Can this be done with I_{main} or I_{F8L} ?

The new B-train system

functional specifications

Idea conceived in 2007 and green-lighted by IEFC in 2011.

Main goals evolved over time:

1. Ensure long-term reliability

concerns about spares: peaking strips (PS), old electronics (SPS)

2. Improve performance and flexibility as requested by users

higher resolution 0.05 G, higher dB/dt for future machines

remove operation constraints and downtime: <50 G flat-bottom due to peaking strips, constraints on cycle sequencing, recurrent timing and synch errors ...

3. Correct/extend measurements to counter beam instability

(e.g. instrument fully F half to cure injection instabilities observed in since 2010)

Applies to PS in priority; progressively to be extended to SPS, AD and ELENA, PSB (upgraded) and LEIR

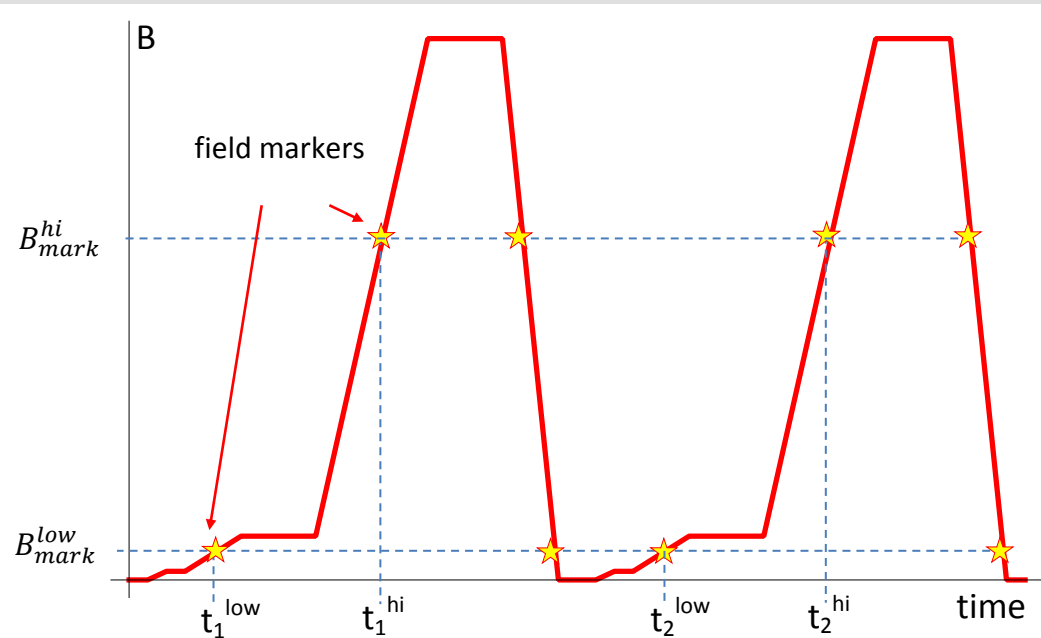
- All technical solutions discussed in detail with users: operation/power converters/RF/beam diagnostics + control
- Progressive upgrade in parallel with existing system
(for cross-check and as a fallback solution)

Phase	Objectives	Time frame
I – Component test	<ul style="list-style-type: none">• Test flux integration and correction algorithms• Test individual electronic cards and components• Compute offline $B(t)$ and validate vs. beam behaviour	2012
II – System test	Minimal viability test: broadcast simplified $B(t)$ to POPS and/or RF <u>with beam</u>	Nov-Dec 2012
Deployment	Replace current system	LS1

New concept: continuous digital distribution of B(t) on a fast serial line to replace old pulse train

- Output stream continuously updated with current B(t) values
→ no need for complex triggering and reset logic; no B₀ initialization burst
- Adaptive auto-calibration of gain and offset using field markers on flat-top and flat-bottom
→ fully transparent correction of measurement error

fast and robust transmission, no more timing conflicts
but: drift correction must be applied continuously on-the-fly



ADC offset (→ integrator drift)
updated every time the same marker level is reached

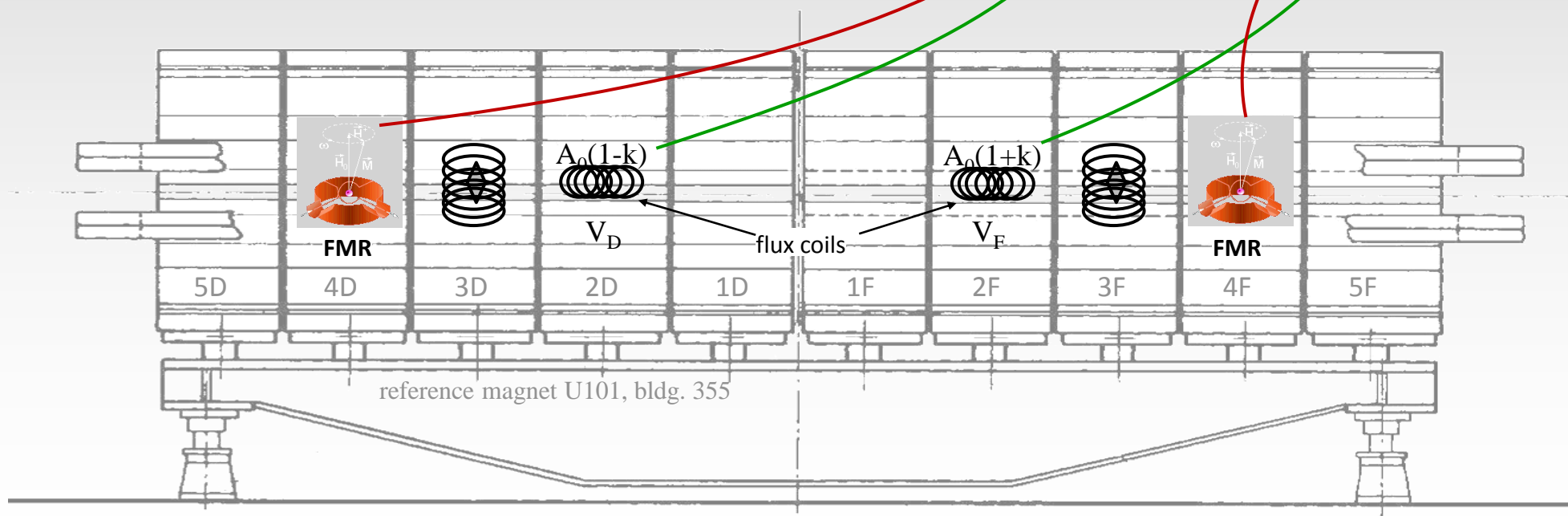
$$B = B_0 + a \int_0^t (1 + \varepsilon(t))(V(t) - V_0(t))dt$$

Integrator gain
updated every time a different marked level is reached

N different marker levels → up to 4 N corrections per cycle
Corrections smeared over a certain Δt to avoid sudden jumps

- Independent high-field markers and field integration in F/D halves
- No complicated marker interpolation required: B_{avg} is constantly updated as $t \rightarrow \infty$ marker triggers transparent internal correction in F/D halves independently
- **Higher ~600 G flat-bottom level** to improve magnetic reproducibility and allow the utilization of new FMR markers (NB: a stopgap solution using existing peaking strips was discussed and found to be technically as complex as switching over to FMR, minus all the advantages)

$$B_{avg} = \frac{1}{2} ((1-k)B_D + (1+k)B_F) = 60 \text{ mT} + \frac{1}{A_0} \int_{t_1}^t V_D dt + \frac{1}{A_0} \int_{t_2}^t V_F dt$$



Prototype harmonic coil array (*in operation*)

- $B_n(I_k)$ relationships
- dynamic effects on multipoles
- provide multipoles to operation

Second harmonic coil array (symmetric)

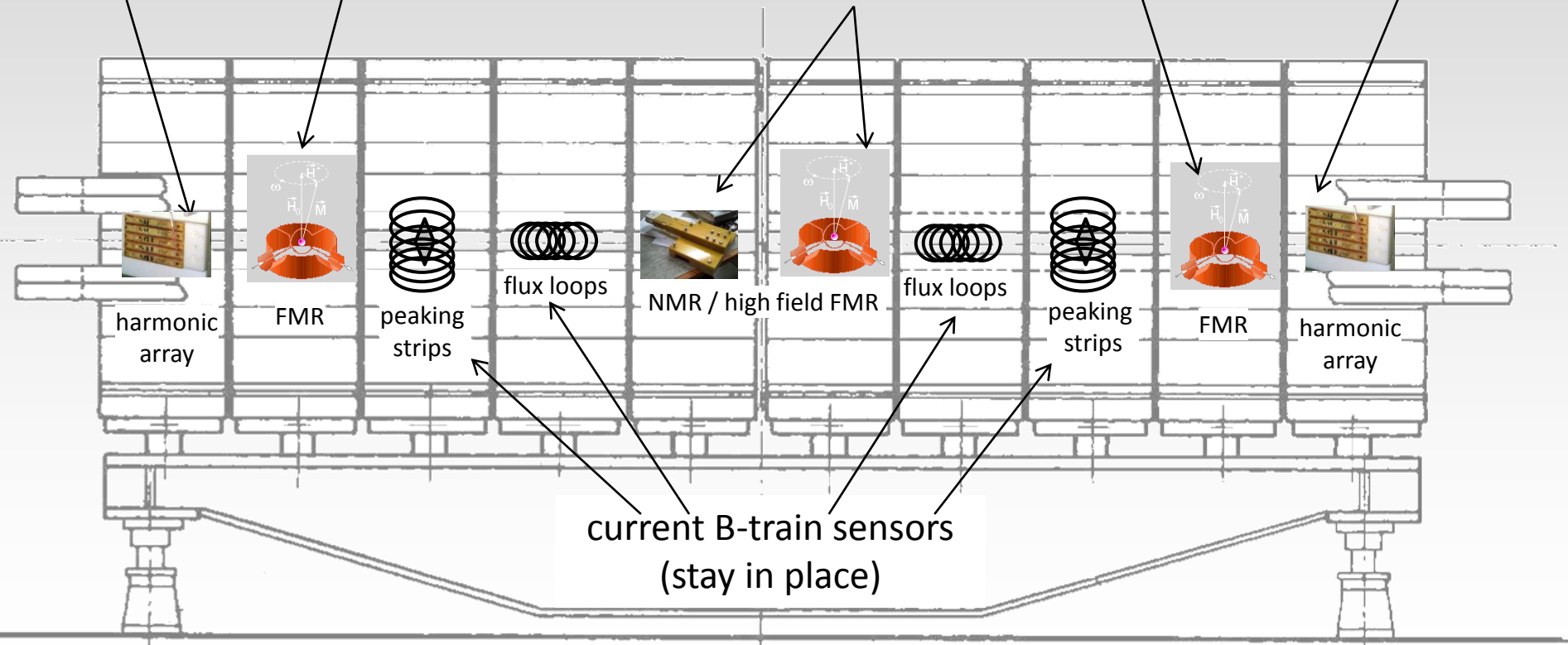
Prototype single-crystal FMR (*due in March*)

- stability vs. peaking strips/beam
- precision of trigger generation
- test 600 G flat-bottom mode

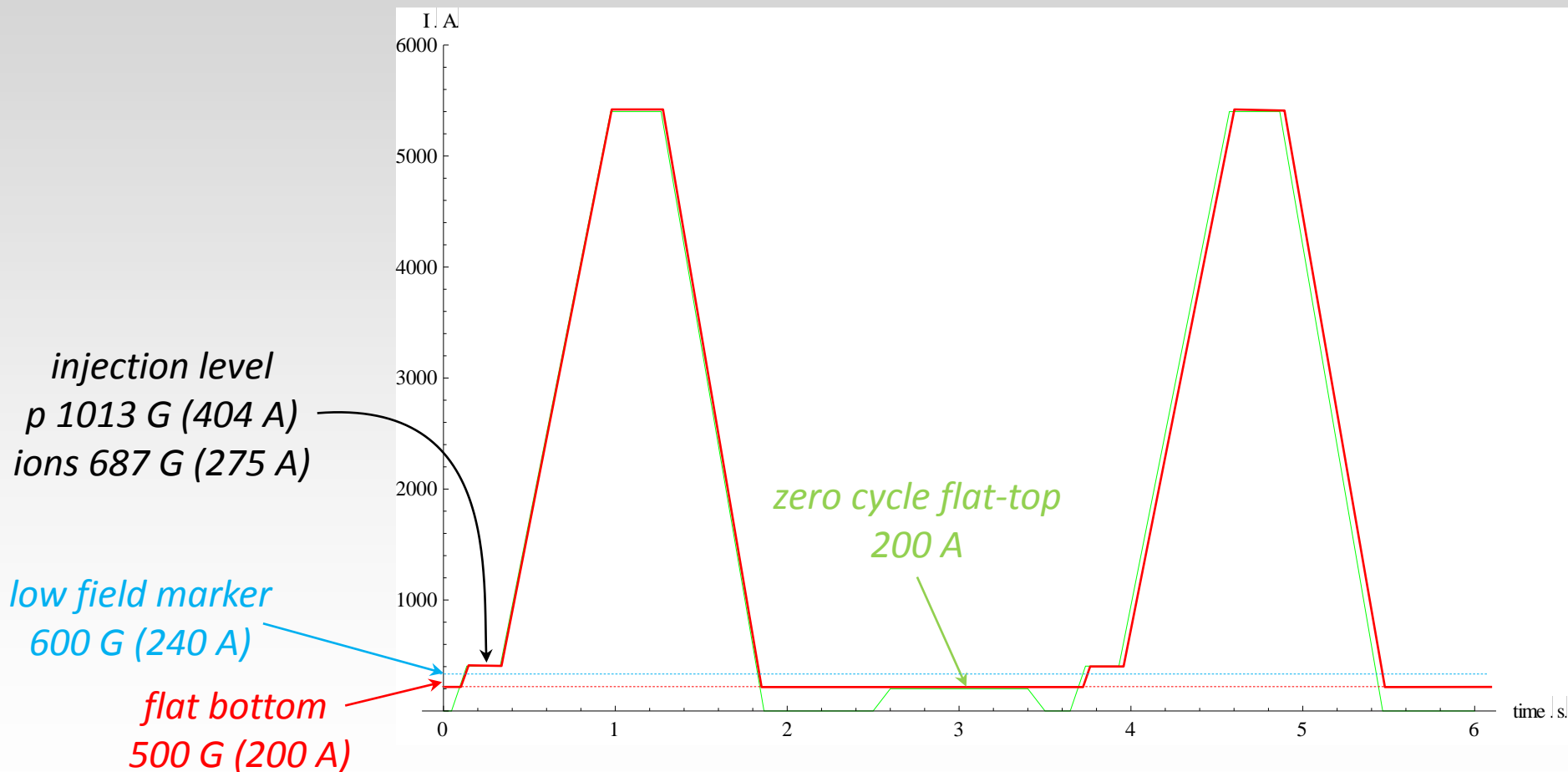
Final FMR (single or poly-crystal)

- symmetric placement
- generate full integral

High field marker for on-the-fly gain correction

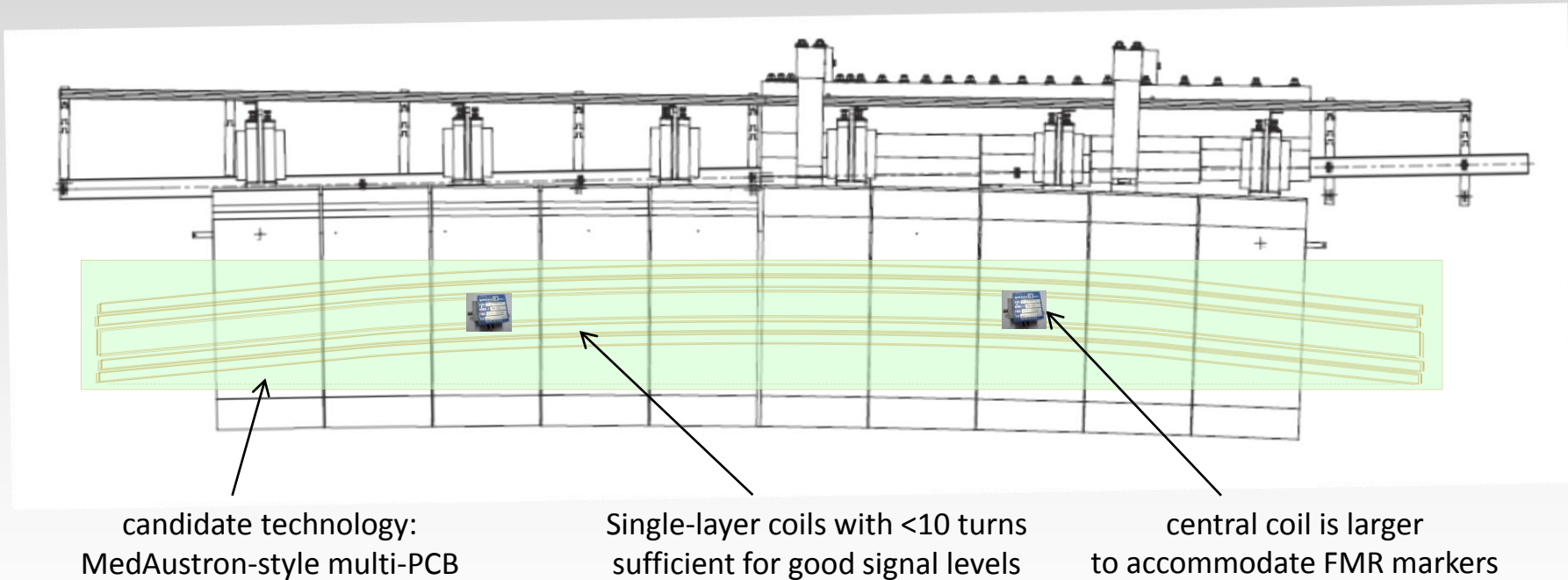


- descent below 50 G no longer needed when FMR replaces peaking strip
- **10× flat bottom=500 G** → better magnetic stability (should be tested shortly !)



Integral fluxmeter

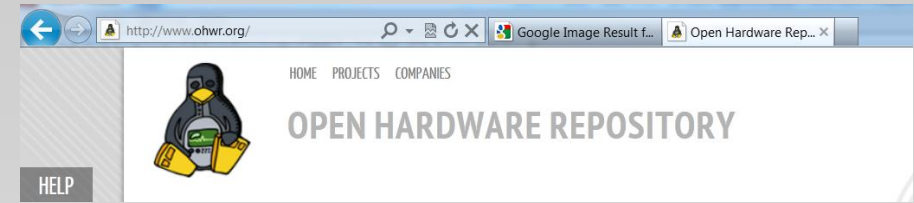
- near-ideal sensor: integral coil array for $\int \mathbf{B}_n d\ell$
- removes one major uncertainty source i.e. extrapolation from local to integral measurement
→ future-proof against more surprises
- no design available to date, but unlikely to be mechanically compatible with existing sensors
can we dare to give it a try with beam before LS1 ?



The new B-train system

hardware & software

Standard CERN HW/SW platform, mostly based on the Open Hardware concept strongly endorsed and fully supported by BE-CO in view of integration in the control infrastructure:



- **PICMG1.3** – (Open Modular Computing Specification) bus standard (industrial PC)
- **SPEC** (Simple PCIe FMC carrier) + **FMC** (FPGA mezzanine cards) for modular custom electronics, easily portable to other bus standards e.g. VME64x
- **Wishbone** serializer bridge FPGA core (used e.g. for DMA access)
- **White Rabbit** serial link
- **Scientific Linux** with C/C++, optional real-time kernel





New CERN fieldbus standard = Ethernet + ns-accurate timing

- 😊 inexpensive, robust, extremely fast (Gbit/s) and accurate optical link
- 😊 Ethernet standard guarantees maintainability and interoperability with future control systems
- 😊 64 byte standard data frames allow to pack in extra data for free (e.g. dB/dt)
- 😞 New electronic cards needed for compatibility with POPS (with support from BE/CO)
- 😞 The impact of switches on latency and jitter needs to be assessed

Proposed data format:

- **32-bit B(t) stream** → ppb-level resolution (e.g. 50 nT over a range of 20 T)

PS only needs 4 ppm (0.05 G), however the extra range is useful for:

- upcoming applications e.g. ELENA (very low field) will use same electronics and software
- reducing quantization noise in the computed derivative

- **1 MHz frame rate** → max. 0.02 G differences between updates

(even if the spectral content of the field is practically negligible above 10 kHz, high update rate comes at no cost and guarantees consistency for all users, which might be otherwise forced to use different interpolation methods)

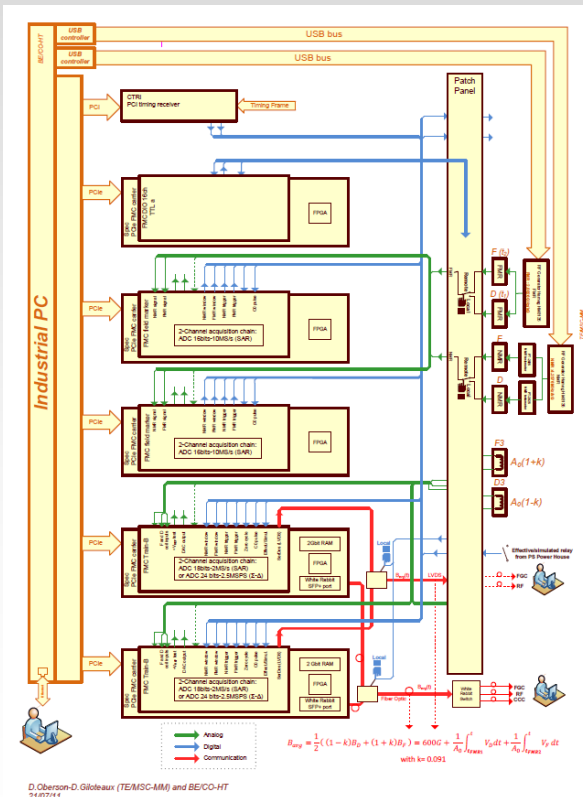
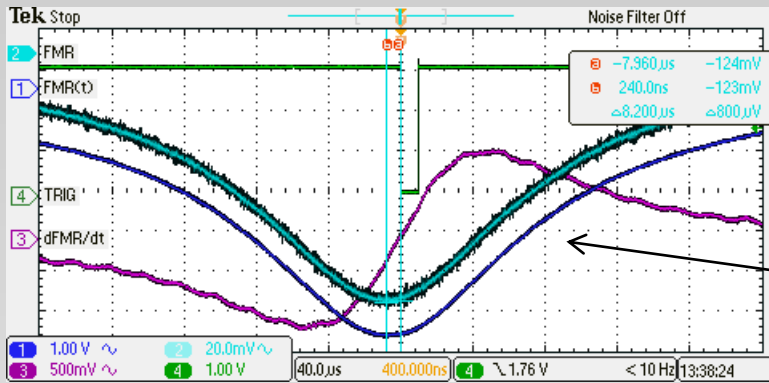
- No UTC referencing (1 μ s latency target, to be tested)

System based on industry standard BUT largely overspecified respect to actual needs
leaves headroom to scale down in case 1 MHz frame rate cannot be met in practical tests

NB: users wishing to maintain old system (i.e. Beam Current DCCT) will retrofit their HW with our help but under their responsibility

Two cards designed and prototyped by MSC-MM (no commercial equivalent exists):

- 1. FMR peak detection card:** based on 16-bit, 10 MHz ADC
 Function: produce a trigger corresponding to the FMR resonance peak via differentiation/threshold comparison. Compatible with NMR demodulated signal. Trigger validation (today done via error-prone synchronization with the timing system) could be accomplished downstream in software (e.g. by comparison with expected values $B(I)$)
- 2. Dual integrator card:** based on a 18-bit, 2 MHz ADC carries out the integration according to the model (page 3). Fully floating input for better CM rejection. Calibration coefficients are expected to be computed in software by the host controller (much less stringent timing needed, more flexibility to adapt software algorithms). General purpose ADC functions, card to be made available CERN-wide and beyond (OHCR community).

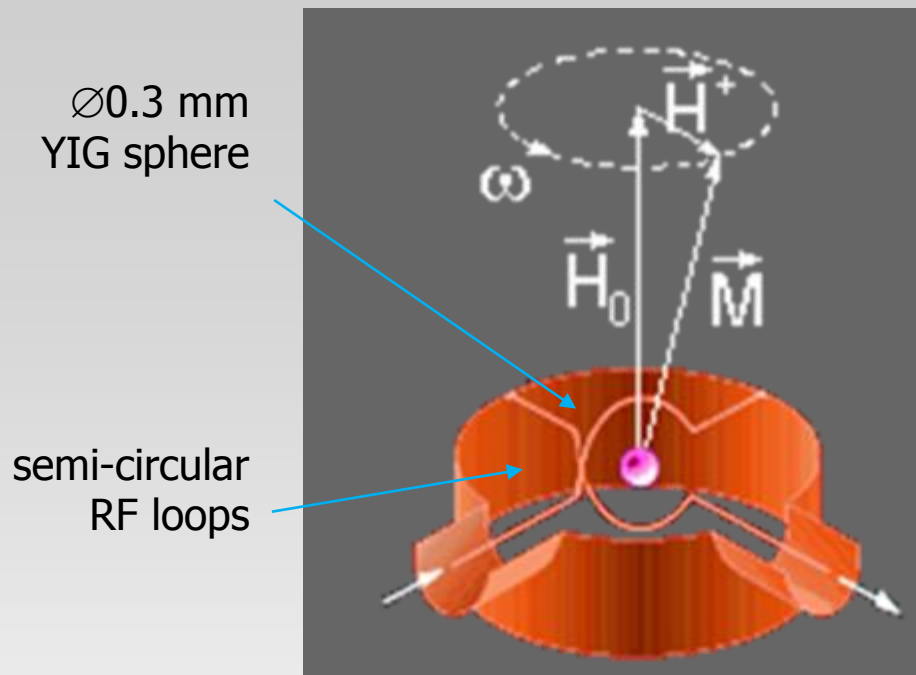


D. Oberson-D. Giloteaux (TE/MSC-MM) and BE/CO-HT 21/07/11

Overall system block diagram

FMR field marker

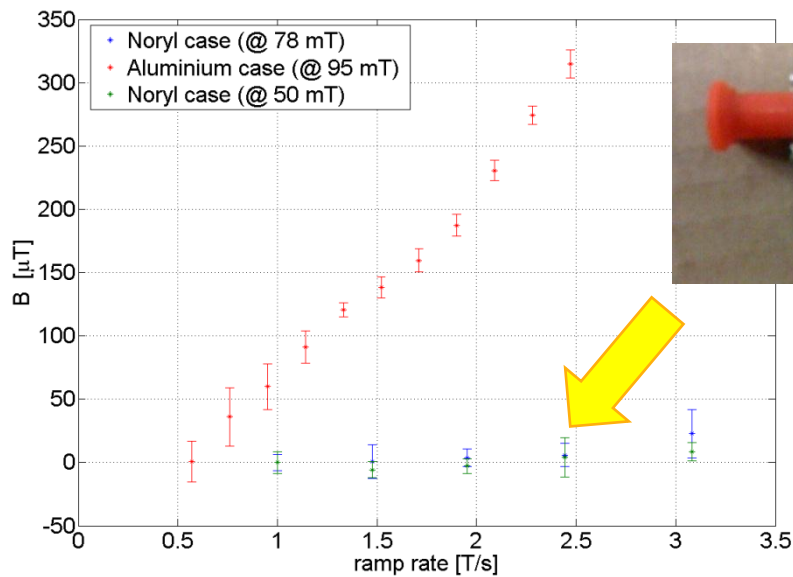




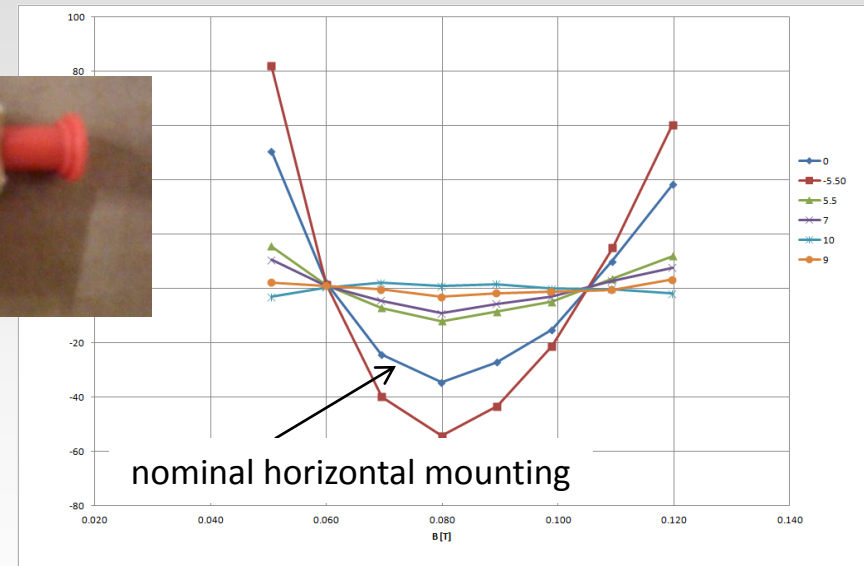
- Yttrium Iron Garnet (YIG) sphere coupling two orthogonal semi-circular RF loops
- Widespread commercial component used as bandpass RF filter
- Insensitive to field gradient due to small diameter
- Units tested at CERN:
 - old poly-crystalline YIG, installed in the PS since the '90s as a manually operated diagnostic tool low filter Q, but insensitive to temperature
 - new single-crystal YIG units. Higher Q $\approx 500-1000$, critical alignment and T dependence tested in 2010/2011: one standard commercial + one customized unit
 - *new poly-crystalline unit: awaited May*

Calibration of a factory-optimized Noryl-case resonator in a reference dipole vs. NMR:

- DC reproducibility: **0.04 G** over the range 550 to 1200 G
- Ramp reproducibility: **~0.1 G @ 600 G, 2.4 T/s**
- Temperature dependence: **0.04 G/°C** ($\pm 2^\circ\text{C}$ in U101 \rightarrow close to tolerance !)
- Roll angle dependence: **0.04 G/mrad @ 600 G**; non-linearity vs. B disappears at certain angles \rightarrow very difficult to use in teslameter mode; strict mechanical tolerance in marker mode
- Resonance spread due to ∇B : still very good **$Q \approx 1000$** at $\nabla B/B = 4.6 \text{ m}^{-1}$ (as in U101)

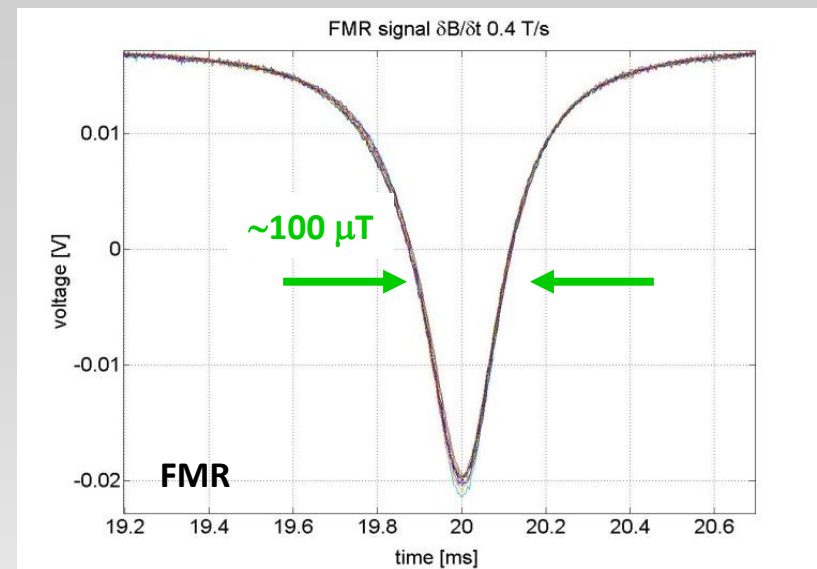
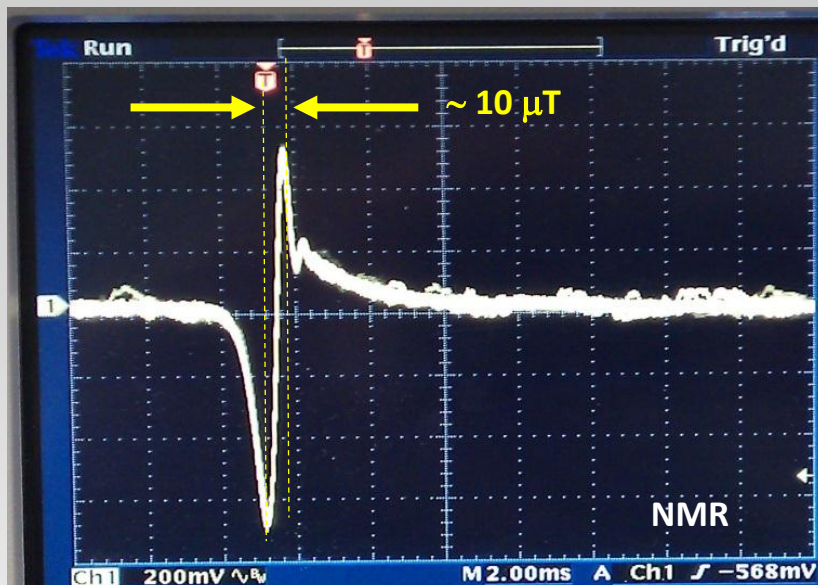


ramp-rate dependence \leq reproducibility



non. linearity error vs. B, roll angle

Sensor output during a field ramp in “marker mode” (fixed frequency excitation)



Marker	😊	😞
NMR	<ul style="list-style-type: none"> • absolute reference (metrological standard) • commercially available instrument 	<ul style="list-style-type: none"> • Requires ∇B compensation • limited ramp rate: 20 to 50 mT/s • $B \geq 43$ mT
FMR single-crystal	<ul style="list-style-type: none"> • works up to several T/s • simple direct acquisition of the resonance • commercially available sensor • larger dynamic range for a given sensor (current unit: 60-300 mT) 	<ul style="list-style-type: none"> • $B \geq 60$ mT • broader resonance peak • needs complex calibration • temperature-dependent (must be optimized for a target field)

Conclusions & outlook

Diagnostic Spy System:

- The system can be used to explain beam observations in terms of the magnetic response of U101. Indication obtained: the tune drift on the flat top can be mitigated by controlling current overshoot
- Systematic correlations will continue to be investigated at the demand of Operation and Beam Physics teams. Other tests will proceed in parallel when possible (e.g. impact of eddy currents in the vacuum chambers).

Upgraded B-train:

- Prototypes of sensors and electronic components and are being built and shall be tested in the next months. When the existing B-train appears to be inconsistent with the beam, we shall provide a simulation of the upgraded $B(t)$ to validate the measurement.
- Detailed schedule of tests and requests of MD for 2012 is in preparation. Online tests involving generation and distribution of $B(t)$ with both proton and ion beams should be carried out to validate the design before the 2013 shutdown. An extended period in which the old and new B-train work in parallel is foreseen in 2014.