

# Lecture 3: Magnetization, cables and ac losses

## Magnetization

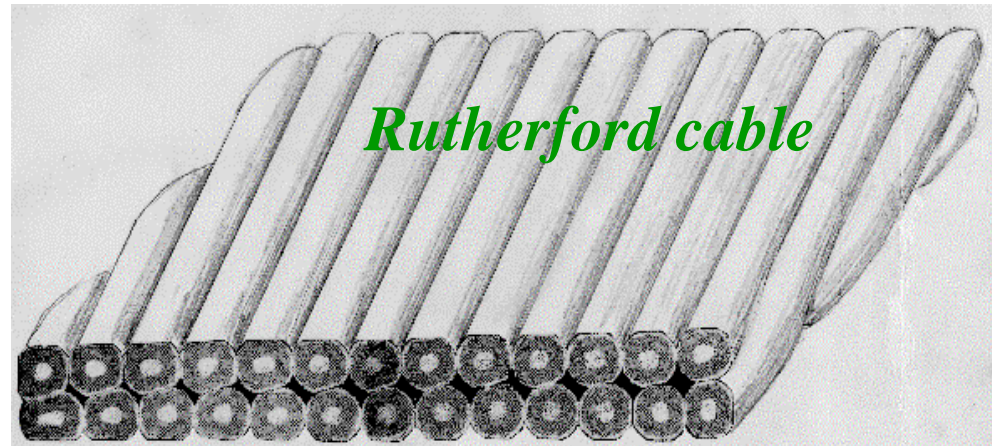
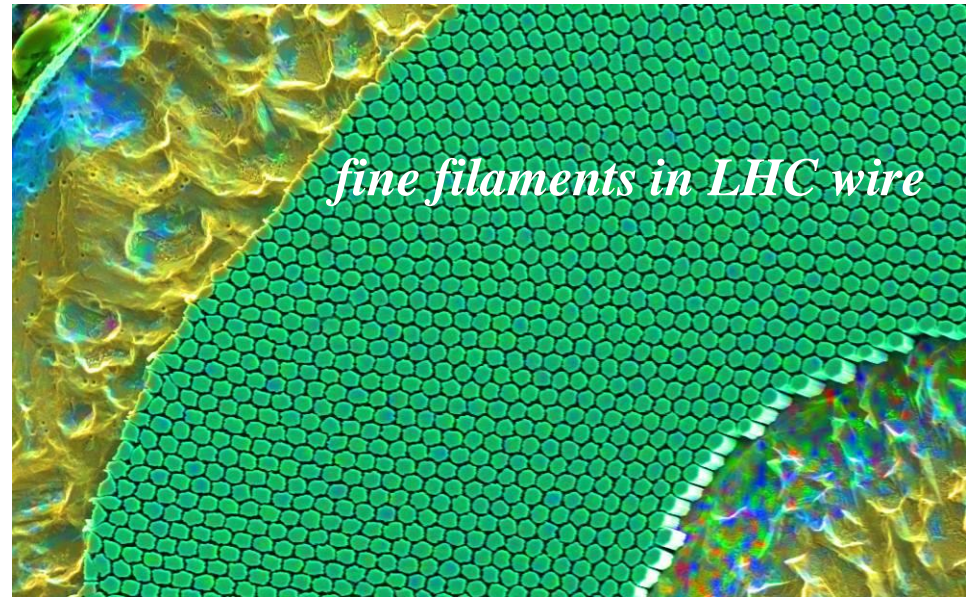
- magnetization of filaments
- coupling between filaments

## Cables

- why cables?
- coupling in cables
- effect on field error in magnets

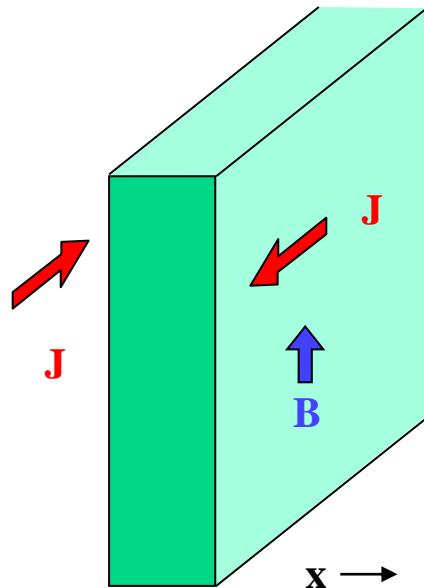
## AC losses

- general expression
- losses within filaments
- losses from coupling



# Recap: persistent screening currents

- **screening currents** are in addition to the **transport current**, which comes from the power supply
- like eddy currents but, because no resistance, they don't decay



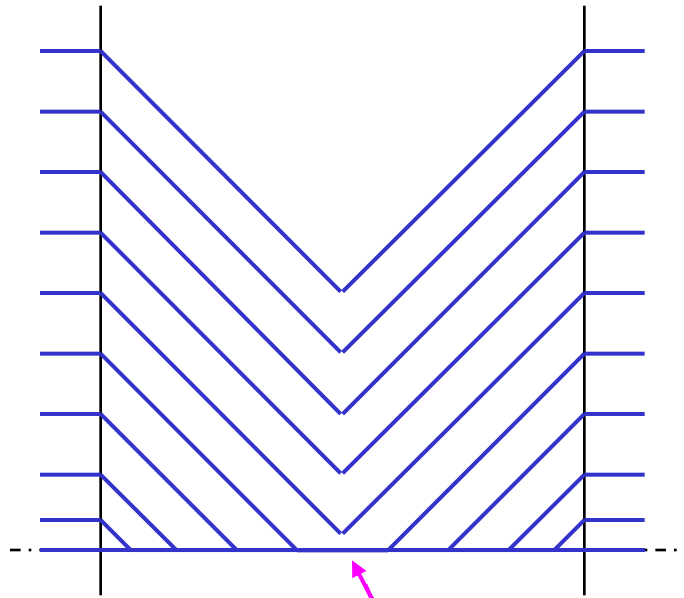
- $dB/dt$  induces an electric field **E** which drives the screening current up to critical current density  $J_c$
- so we have  $J = +J_c$  or  $J = -J_c$  or  $J = 0$  nothing else
- known as the **critical state model** or **Bean model**
- in the 1 dim infinite slab geometry, Maxwell's equation says

$$\frac{\partial B_y}{\partial x} = -\mu_0 J_z = \mu_0 J_c$$

- so uniform  $J_c$  means a constant field gradient inside the superconductor

# The flux penetration process

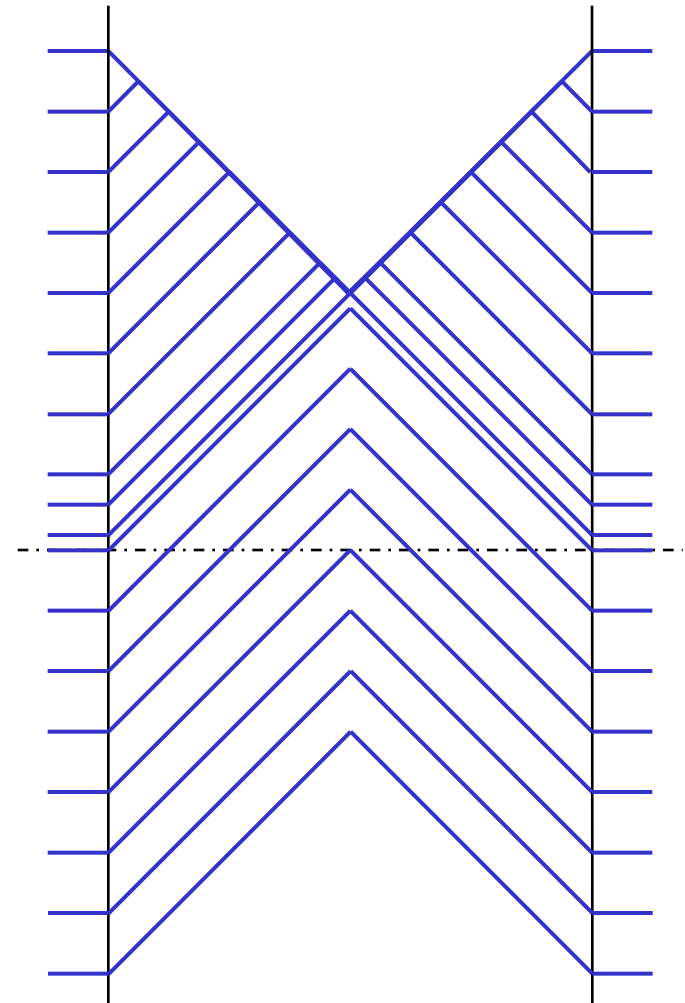
plot field profile across the slab



field increasing from zero

## Bean critical state model

- current density everywhere is  $\pm J_c$  or zero
- change comes in from the outer surface



field decreasing through zero

# Magnetization of the Superconductor

When viewed from outside the sample, the persistent currents produce a magnetic moment.

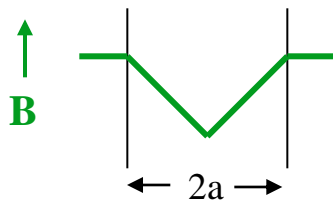
Problem for accelerators because it spoils the precise field shape

We can define a magnetization (magnetic moment per unit volume)

$$M = \sum_v \frac{I \cdot A}{V}$$

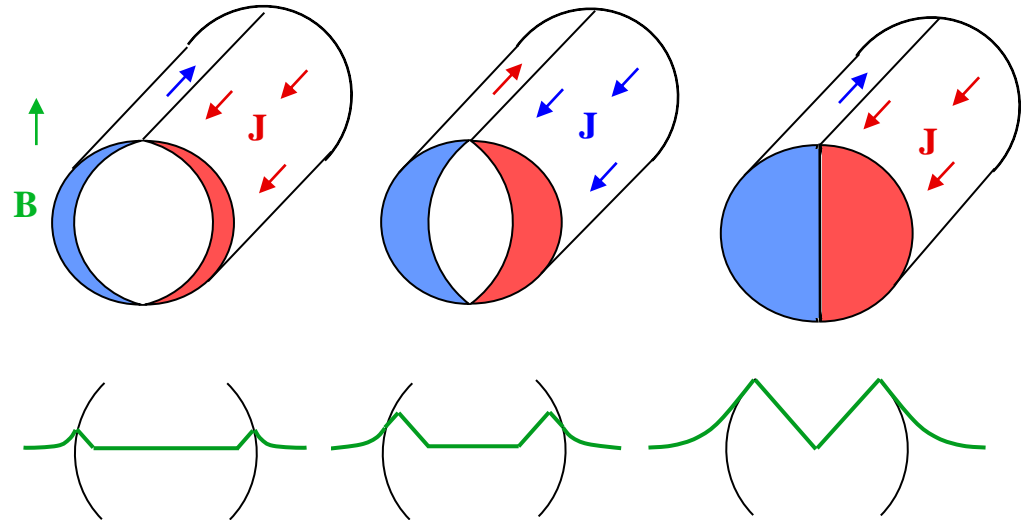
NB units of H

for a fully penetrated slab



$$M_s = \frac{1}{a} \int_0^a J_c x dx = \frac{J_c a}{2}$$

for **cylindrical** filaments the inner current boundary is roughly elliptical (controversial)



when fully penetrated, the magnetization is

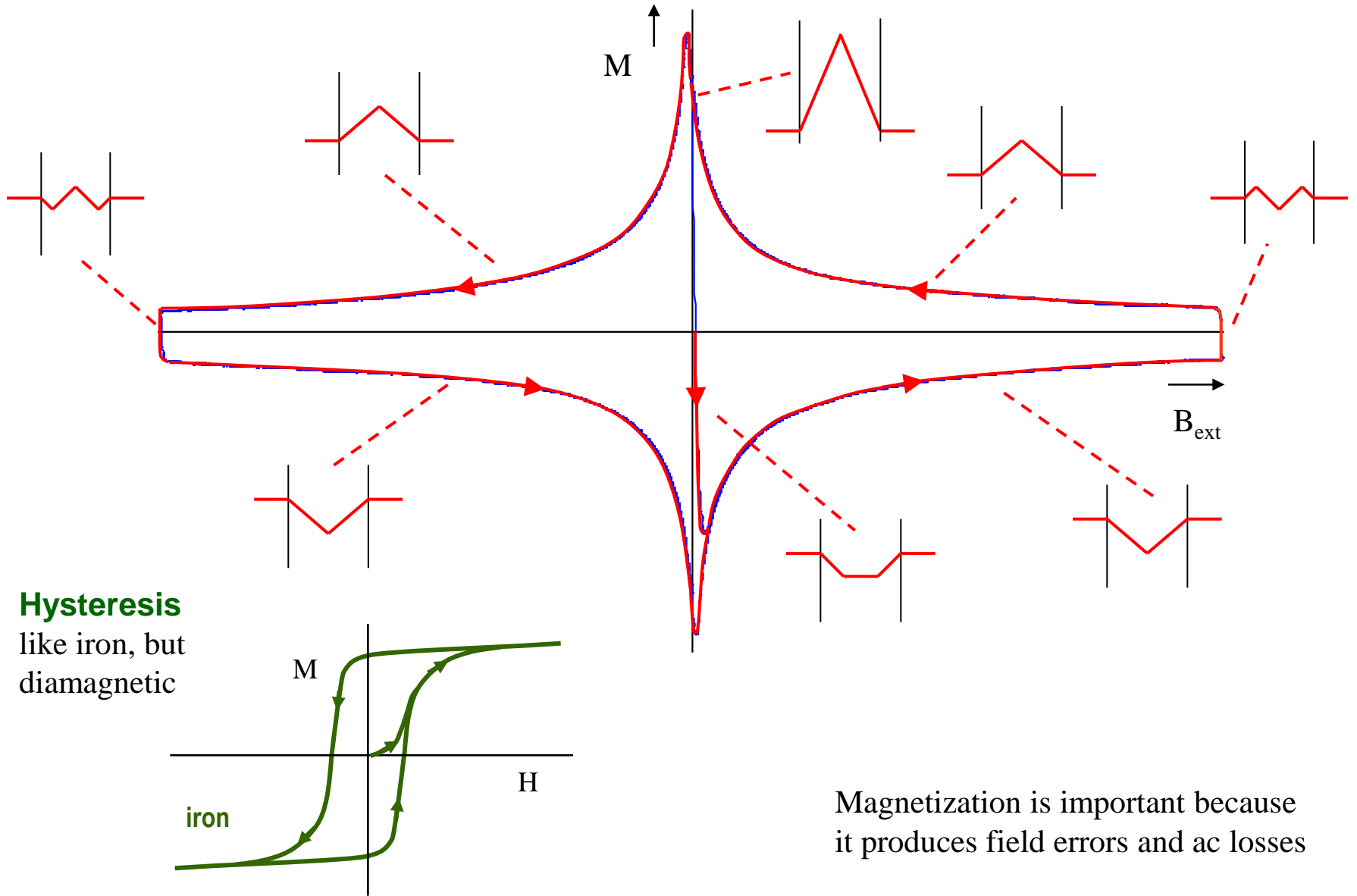
$$M_s = \frac{4}{3\pi} J_c a = \frac{2}{3\pi} J_c d_f$$

where  $a$ ,  $d_f$  = filament radius, diameter

Note:  $M$  is here defined per unit volume of NbTi filament

to reduce  $M$  need small  $d$  - fine filaments

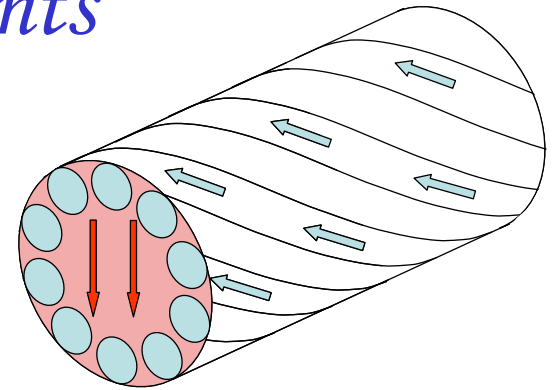
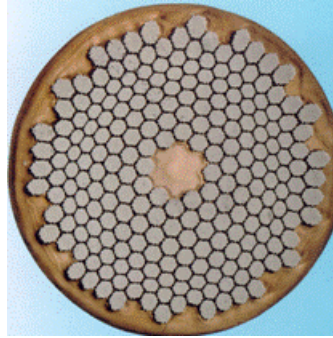
# Magnetization of NbTi



# Coupling between filaments

recap 
$$M_s = \frac{2}{3\pi} J_c d_f$$

- reduce M by making fine filaments
- for ease of handling, filaments are embedded in a copper matrix



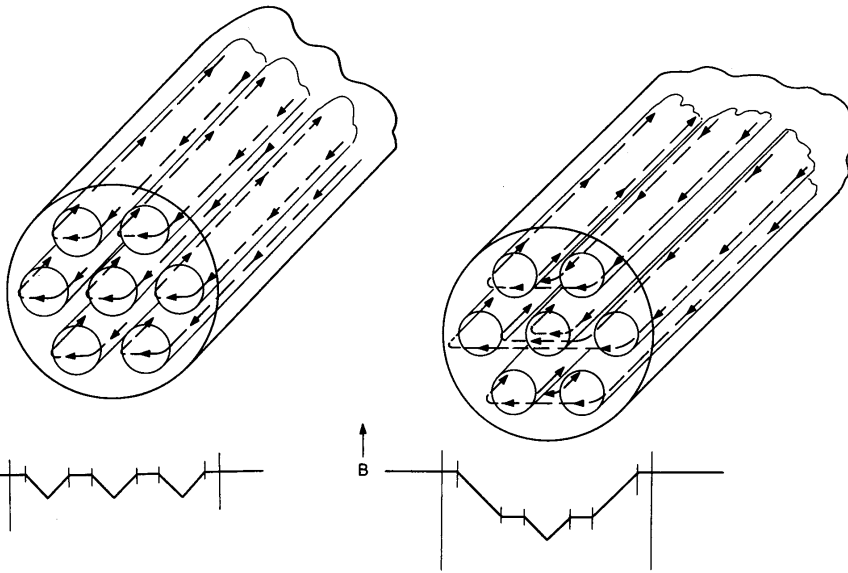
- coupling currents flow along the filaments and across the matrix
- reduce them by twisting the wire
- they behave like eddy currents and produce an additional magnetization

$$M_e = \frac{dB}{dt} \frac{1}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

$$M_e = \frac{2}{\mu_o} \frac{dB}{dt} \tau \quad \text{where} \quad \tau = \frac{\mu_o}{2\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

per unit volume of wire

$\rho_t$  = resistivity across matrix,  $p_w$  = wire twist pitch

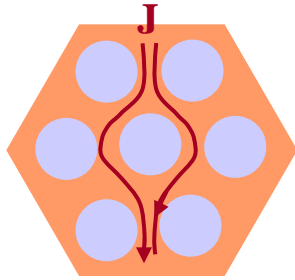


- but in changing fields, the filaments are magnetically coupled
- screening currents go up the left filaments and return down the right



# Transverse resistivity across the matrix

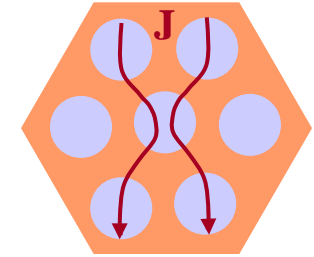
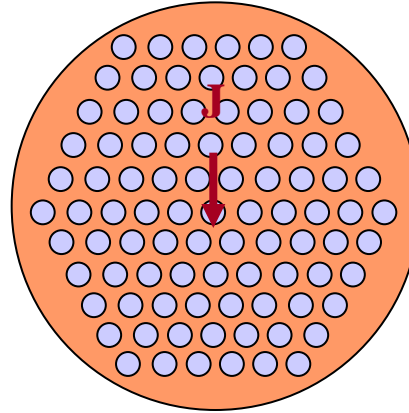
## Poor contact to filaments



$$\rho_t = \rho_{Cu} \frac{1 + \lambda_{sw}}{1 - \lambda_{sw}}$$

where  $\lambda_{sw}$  is the fraction of superconductor in the wire cross section (after J Carr)

## Good contact to filaments

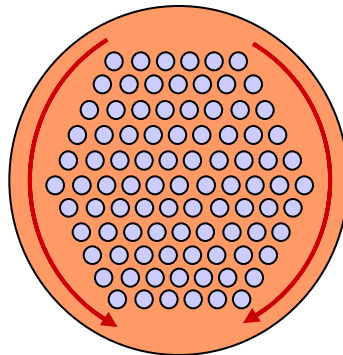


$$\rho_t = \rho_{Cu} \frac{1 - \lambda_{sw}}{1 + \lambda_{sw}}$$

## Some complications

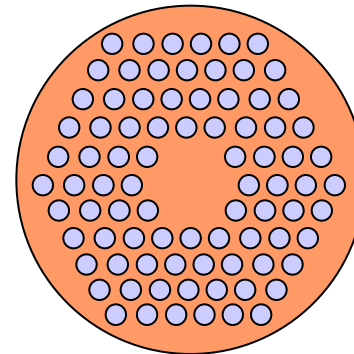
### Thick copper jacket

include the copper jacket as a resistance in parallel

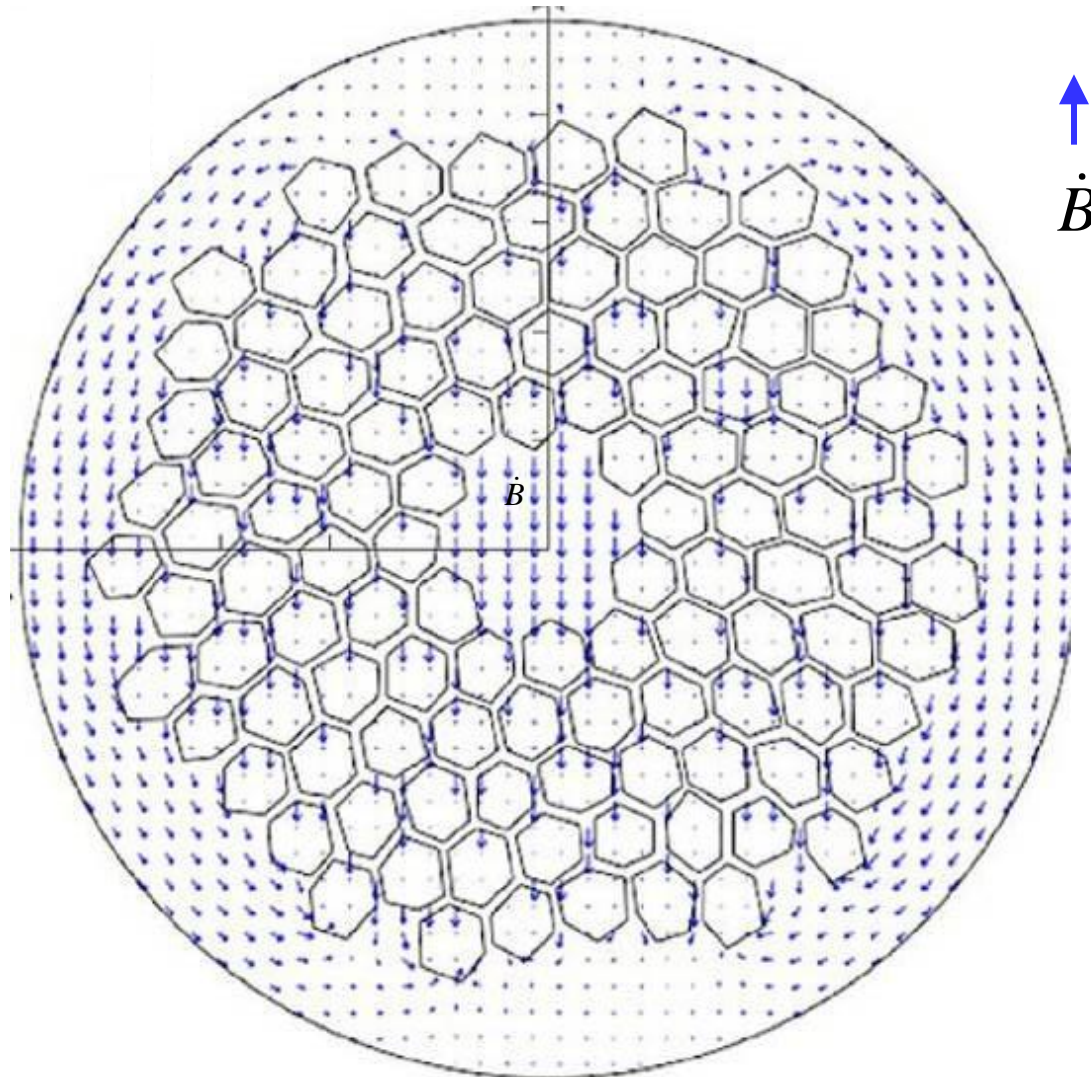


### Copper core

resistance in series for part of current path



# Computation of current flow across matrix



*calculated using  
the COMSOL  
code by  
P.Fabbricatore et  
al JAP, 106,  
083905 (2009)*



# Two components of magnetization

1) persistent current within the filaments

$$M_s = \lambda_{su} \frac{2}{3\pi} J_c(B) d_f$$

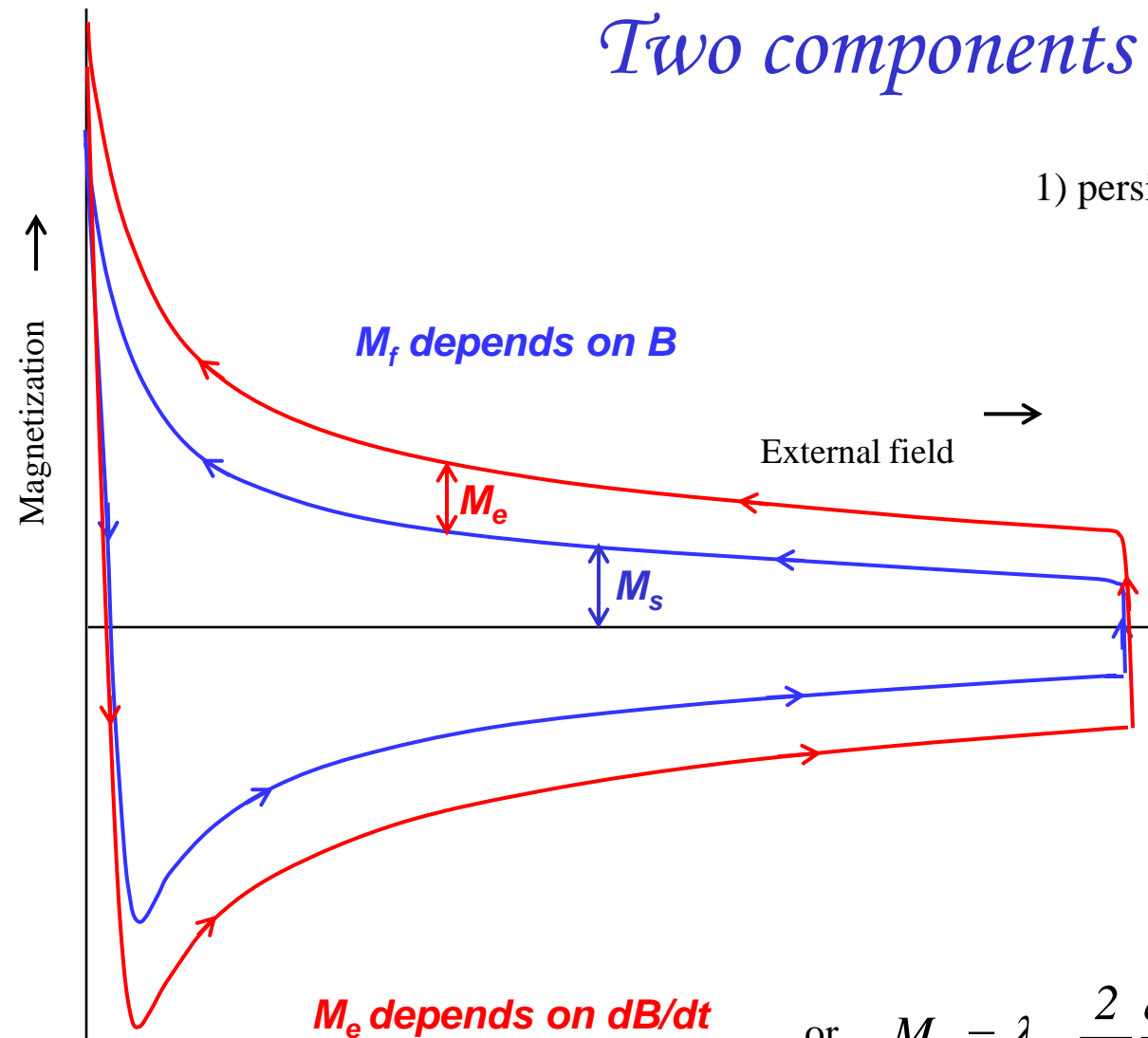
where  $\lambda_{su}$  = fraction of superconductor in the unit cell

2) eddy current coupling between the filaments

$$M_e = \lambda_{wu} \frac{dB}{dt} \frac{l}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

or 
$$M_e = \lambda_{wu} \frac{2}{\mu_o} \frac{dB}{dt} \tau \quad \text{where} \quad \tau = \frac{\mu_o}{2\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

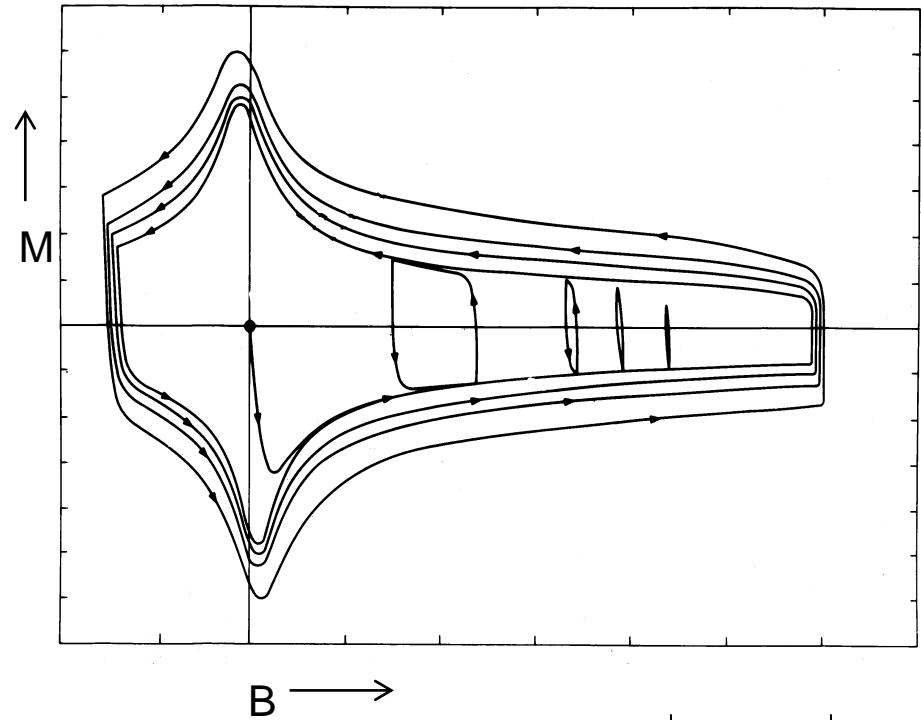
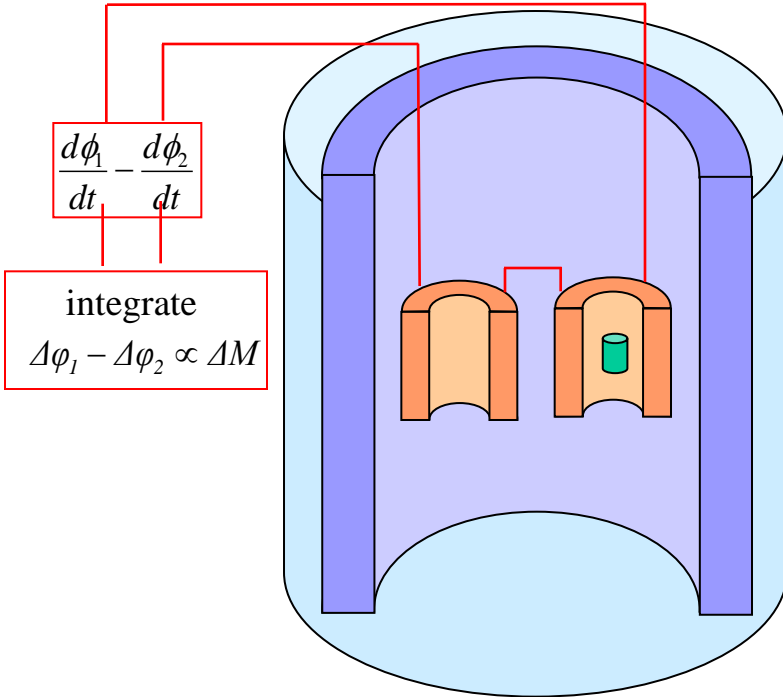
where  $\lambda_{wu}$  = fraction of wire in the section



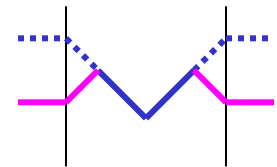
Magnetization is averaged over the unit cell

# Measurement of magnetization

In field, the superconductor behaves just like a magnetic material. We can plot the magnetization curve using a magnetometer. It shows hysteresis - just like iron only in this case the magnetization is both diamagnetic and paramagnetic.



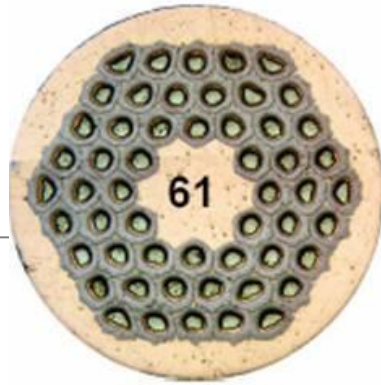
Note the minor loops, where field and therefore screening currents are reversing



Two balanced search coils connected in series opposition, are placed within the bore of a superconducting solenoid. With a superconducting sample in one coil, the integrator measures  $\Delta M$  when the solenoid field is swept up and down

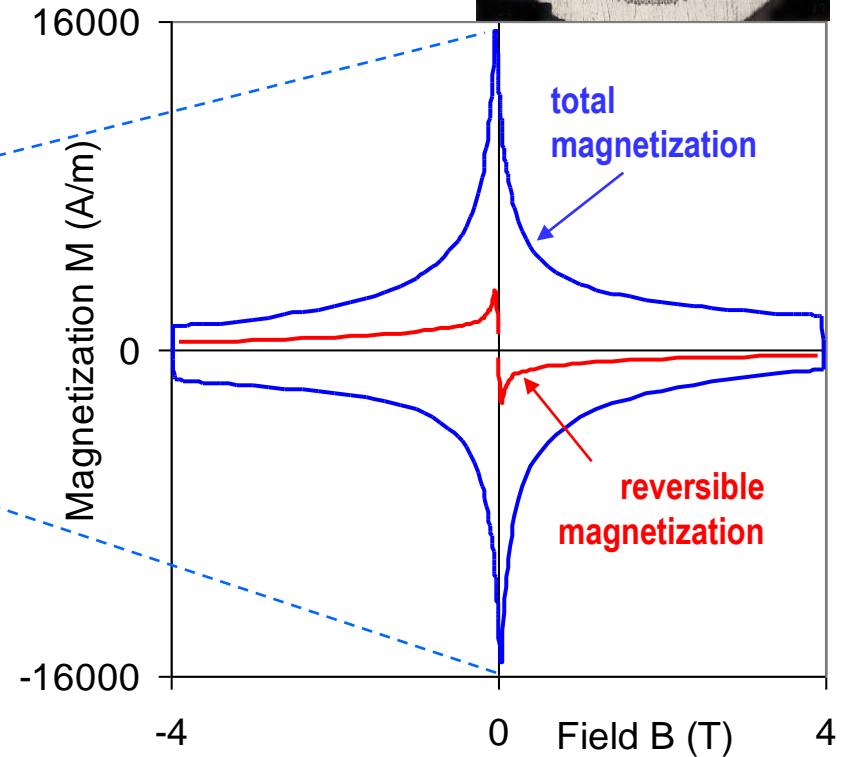
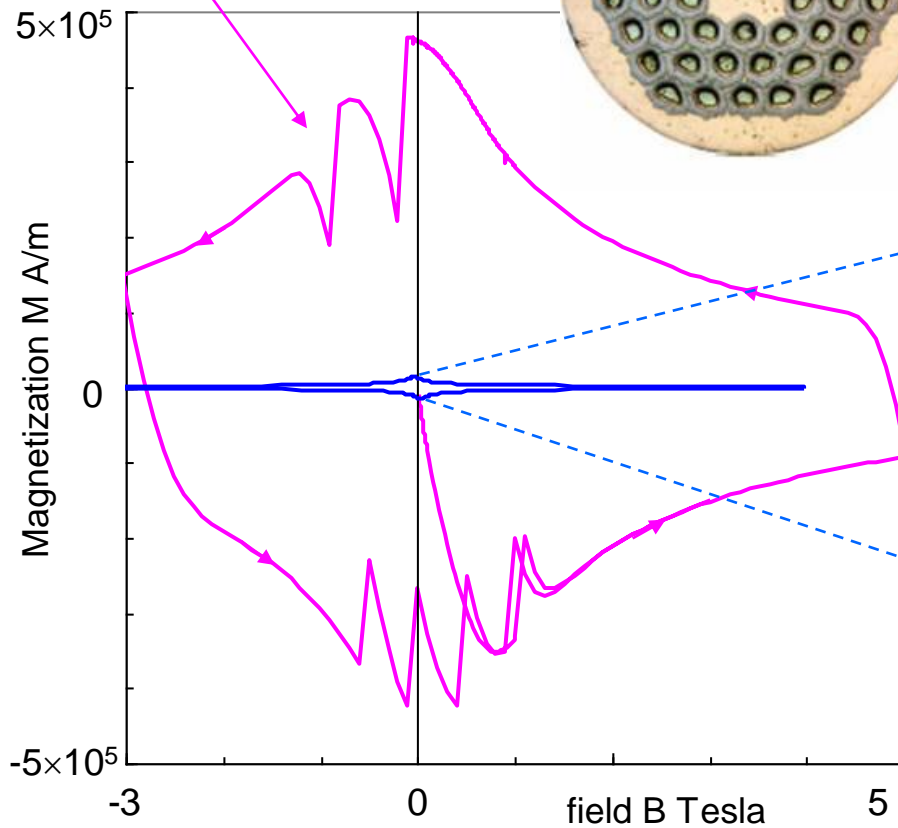
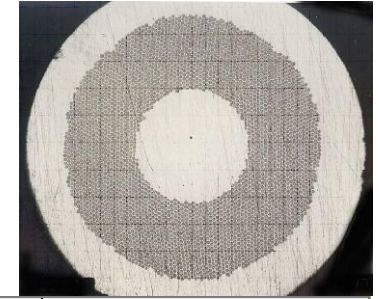
# Magnetization measurements

flux jumping at low field caused by large filaments and high  $J_c$



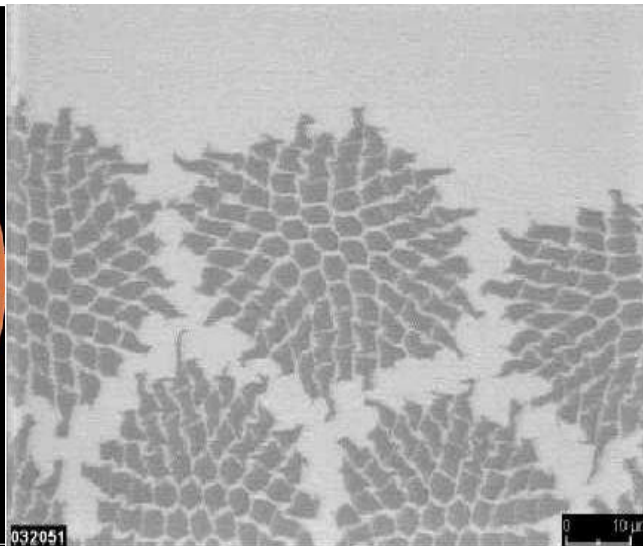
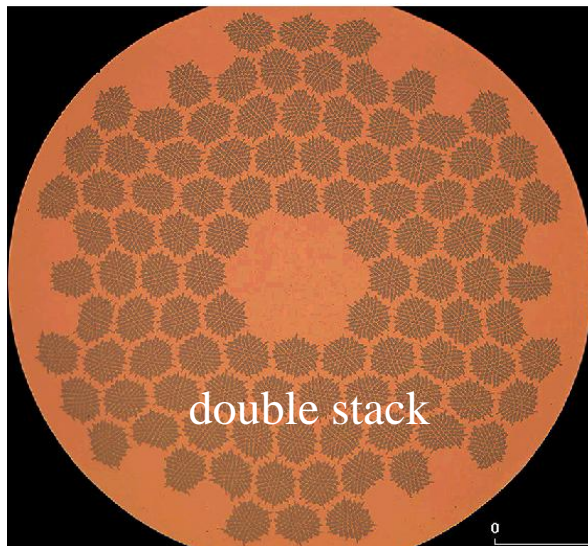
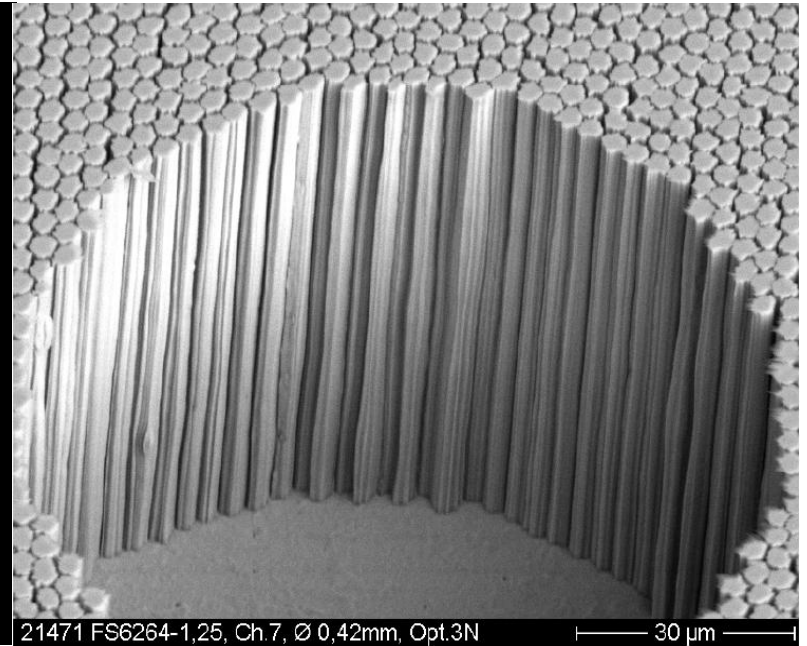
RRP  $Nb_3Sn$  wire with  $50\mu m$  filaments

NbTi wire for RHIC with  $6\mu m$  filaments



# Fine filaments for low magnetization

Accelerator magnets need the finest filaments - to minimize field errors and ac losses



Typical diameters are in the range 5 - 10µm. Even smaller diameters would give lower magnetization, but at the cost of lower  $J_c$  and more difficult production.



# Cables - why do we need them?

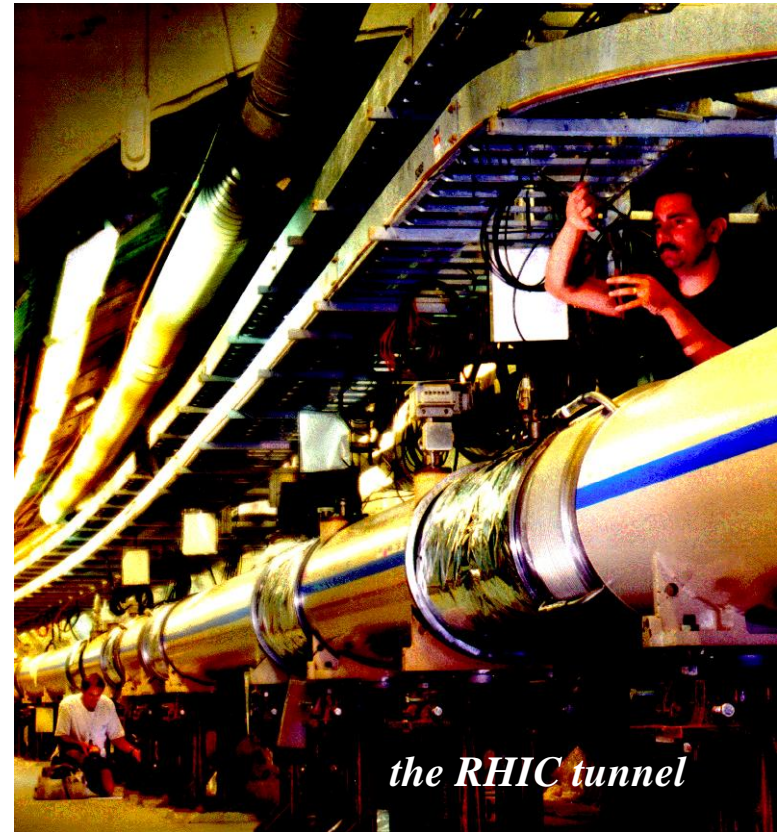
- for good tracking we connect synchrotron magnets in series
- if the stored energy is  $E$ , rise time  $t$  and operating current  $I$ , the charging voltage is

$$E = \frac{B^2}{2\mu_0} V = \frac{1}{2} LI^2 \qquad V = \frac{LI}{t} = \frac{2E}{It}$$

**RHIC**  $E = 40\text{kJ/m}$ ,  $t = 75\text{s}$ , 30 strand cable  
cable  $I = 5\text{kA}$ , charge voltage per km = **213V**  
wire  $I = 167\text{A}$ , charge voltage per km = **6400V**

**FAIR at GSI**  $E = 74\text{kJ/m}$ ,  $t = 4\text{s}$ , 30 strand cable  
cable  $I = 6.8\text{kA}$ , charge voltage per km = **5.4kV**  
wire  $I = 227\text{A}$ , charge voltage per km = **163kV**

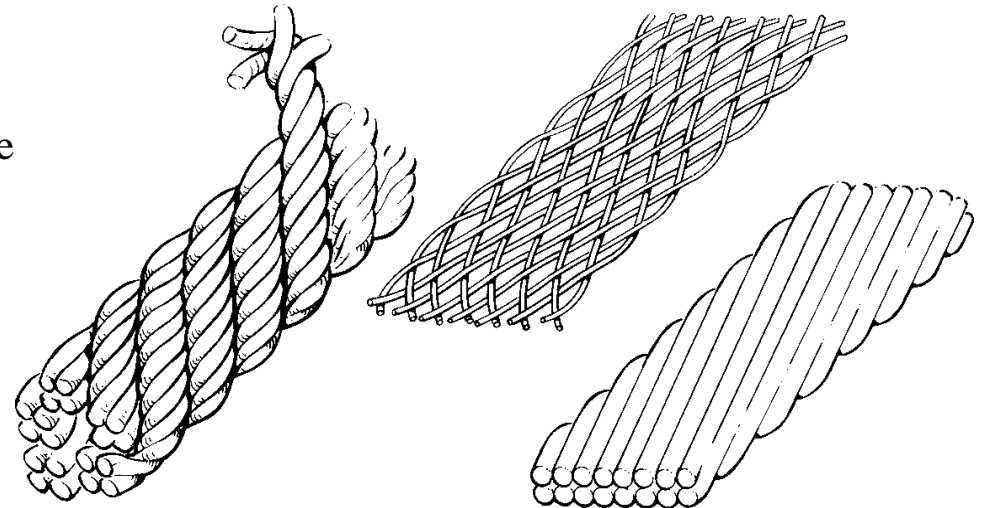
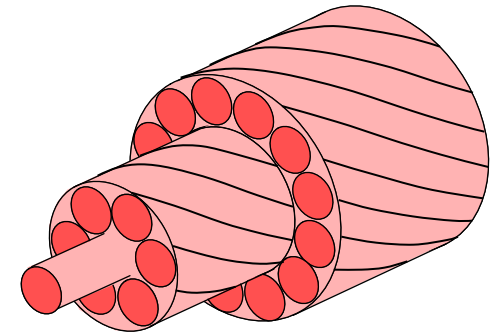
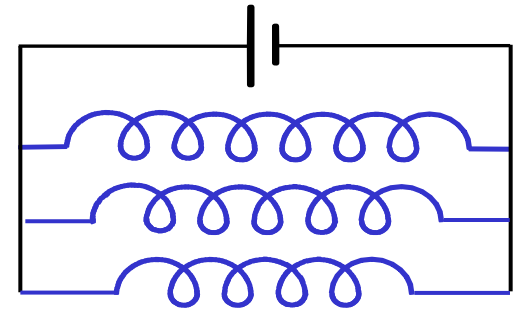
- so we need high currents!
- a single  $5\mu\text{m}$  filament of NbTi in 6T carries 50mA
- a composite wire of fine filaments typically has 5,000 to 10,000 filaments, so it carries 250A to 500A
- for 5 to 10kA, we need 20 to 40 wires in parallel - **a cable**



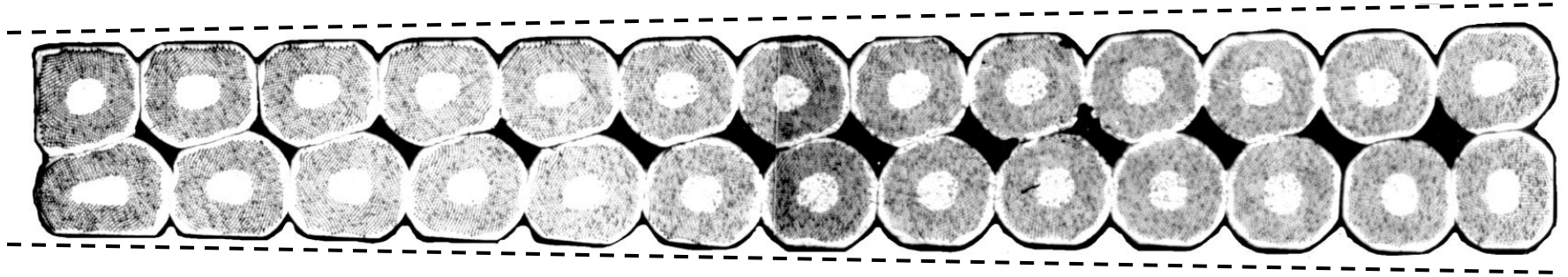


# Cable transposition

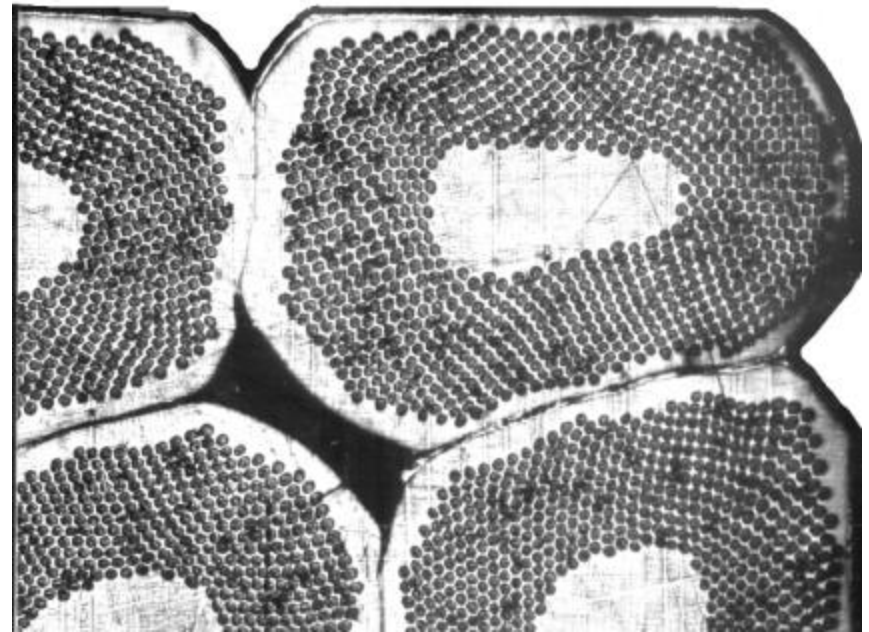
- many wires in parallel - want them all to carry same current  
zero resistance - so current divides according to inductance
- in a simple twisted cable, wires in the centre have a higher self inductance than those at the outside
- current fed in from the power supply therefore takes the low inductance path and stays on the outside
- so outer wires reach  $J_c$  while inner are still empty
- so the wires must be fully **transposed**, ie every wire must change places with every other wire along the length  
inner wires  $\Rightarrow$  outside    outer wire  $\Rightarrow$  inside
- three types of fully transposed cable have been tried in accelerators
  - rope
  - braid
  - Rutherford

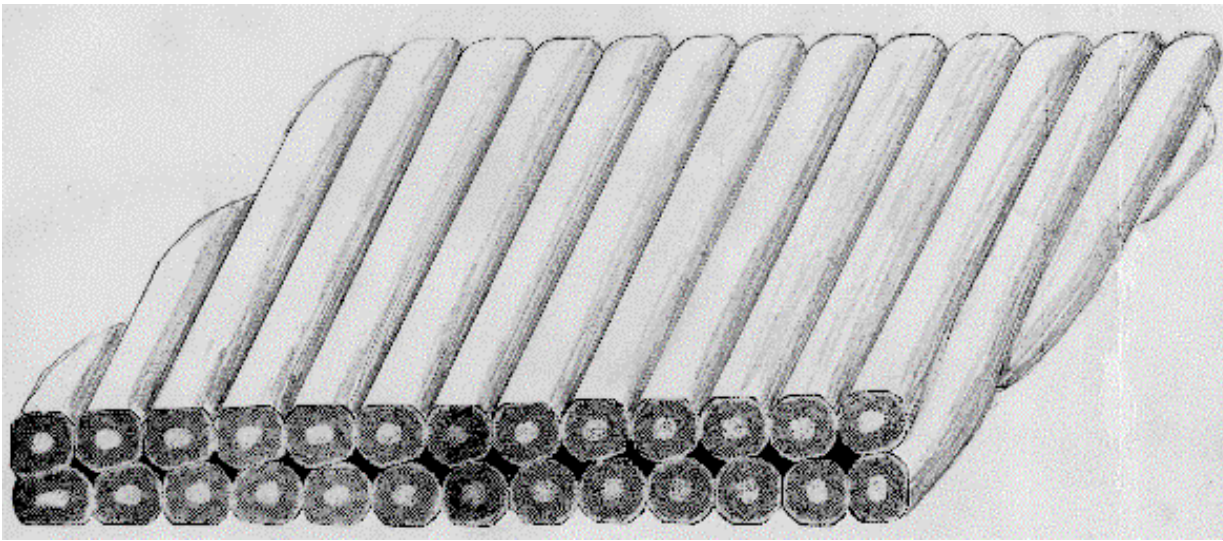


# Rutherford cable



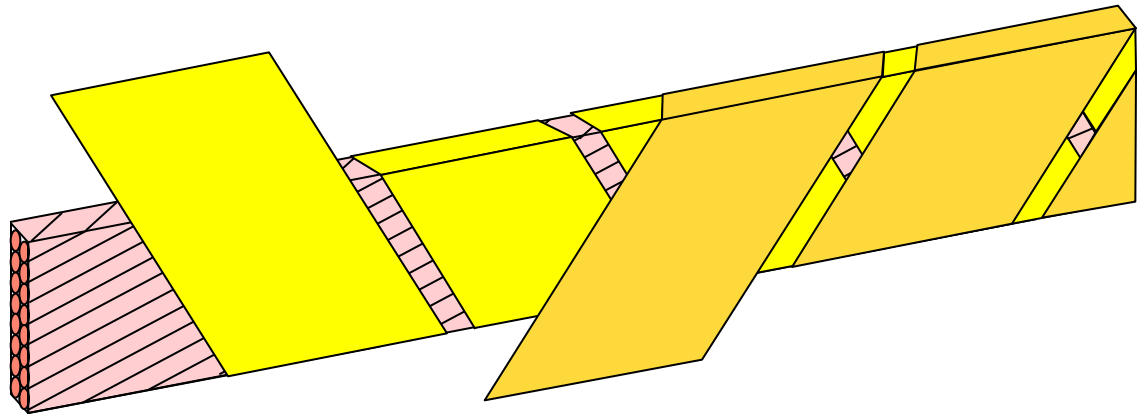
- Rutherford cable succeeded where others failed because it could be compacted to a high density (88 - 94%) without damaging the wires, and rolled to a good dimensional accuracy ( $\sim 10\mu\text{m}$ ).
- Note the 'keystone angle', which enables the cables to be stacked closely round a circular aperture





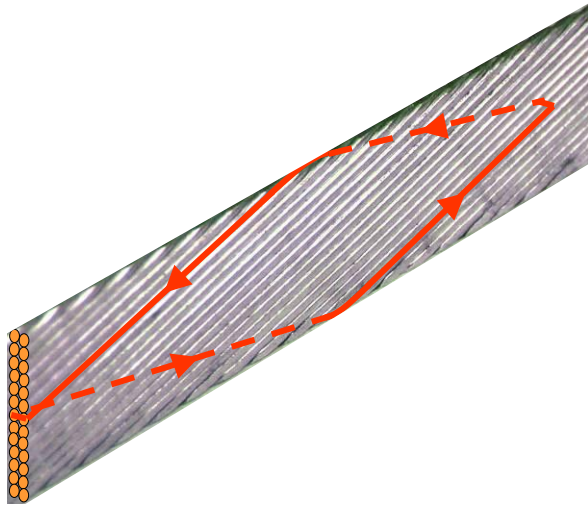
## Rutherford cable

- the cable is insulated by wrapping 2 or 3 layers of Kapton; gaps may be left to allow penetration of liquid helium; the outer layer is treated with an adhesive layer for bonding to adjacent turns.
- Recapitulate: the adhesive faces outwards, don't bond it to the cable (avoid energy release by bond failure)
- allow liquid helium to permeate the cable  
- increase the MQE

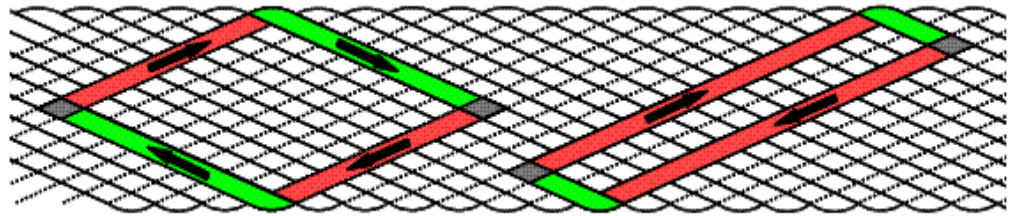
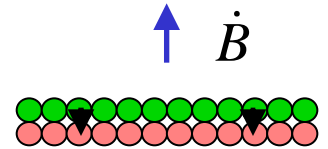




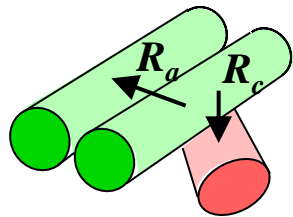
# Coupling in Rutherford cables



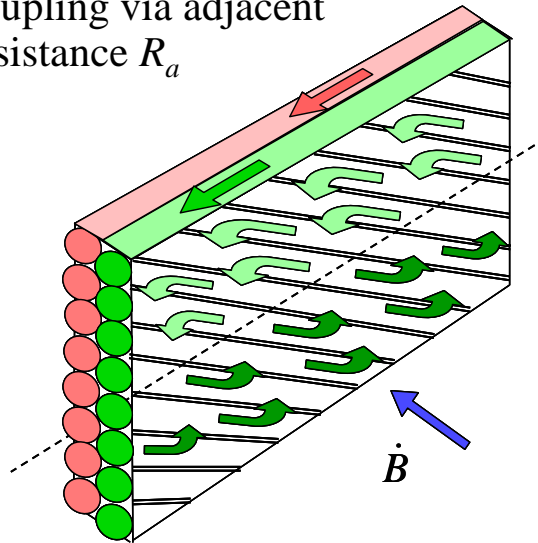
- Field transverse coupling via crossover resistance  $R_c$



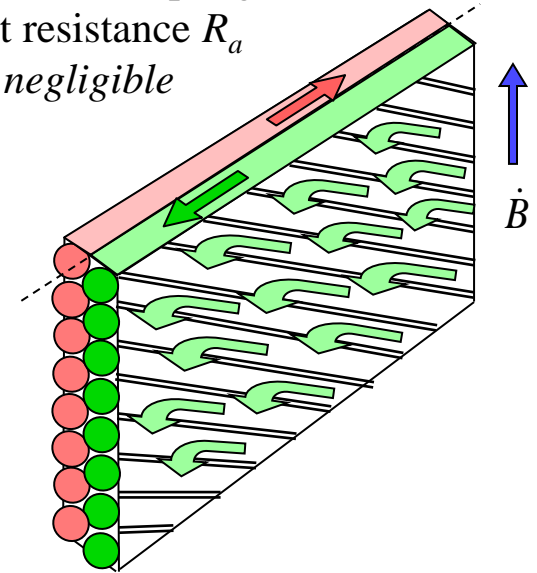
- Field transverse coupling via adjacent resistance  $R_a$



crossover resistance  $R_c$   
adjacent resistance  $R_a$



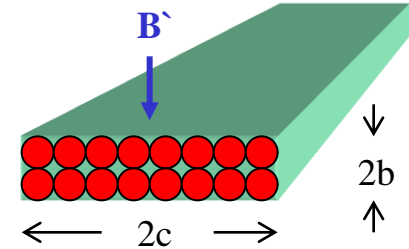
- Field parallel coupling via adjacent resistance  $R_a$   
*usually negligible*



# Magnetization from coupling in cables

- Field transverse  
coupling via  
crossover  
resistance  $R_c$

$$M_{tc} = \frac{1}{120} \frac{\dot{B}_t}{R_c} \frac{c}{b} p N(N-1) = \frac{1}{60} \frac{\dot{B}_t}{\rho_c} p^2 \frac{c^2}{b^2}$$



where  $M$  = magnetization *per unit volume of cable*,  $p$  = twist pitch,  $N$  = number of strands  
 $R_c$   $R_a$  = resistance per crossover  $\rho_c$   $\rho_a$  = effective resistivity between wire centres

- Field transverse

coupling via adjacent resistance  $R_a$

where  $\theta$  = slope angle of wires  $\cos \theta \sim 1$

$$M_{ta} = \frac{1}{6} \frac{\dot{B}_t}{R_a} p \frac{c}{b} = \frac{1}{48} \frac{\dot{B}_t}{\rho_a} \frac{p^2}{\cos^2 \theta}$$

- Field parallel

coupling via adjacent resistance  $R_a$

$$M_{pa} = \frac{1}{8} \frac{\dot{B}_p}{R_a} p \frac{b}{c} = \frac{1}{64} \frac{\dot{B}_p}{\rho_a} \frac{p^2}{\cos^2 \theta} \frac{b^2}{c^2}$$

(usually negligible)

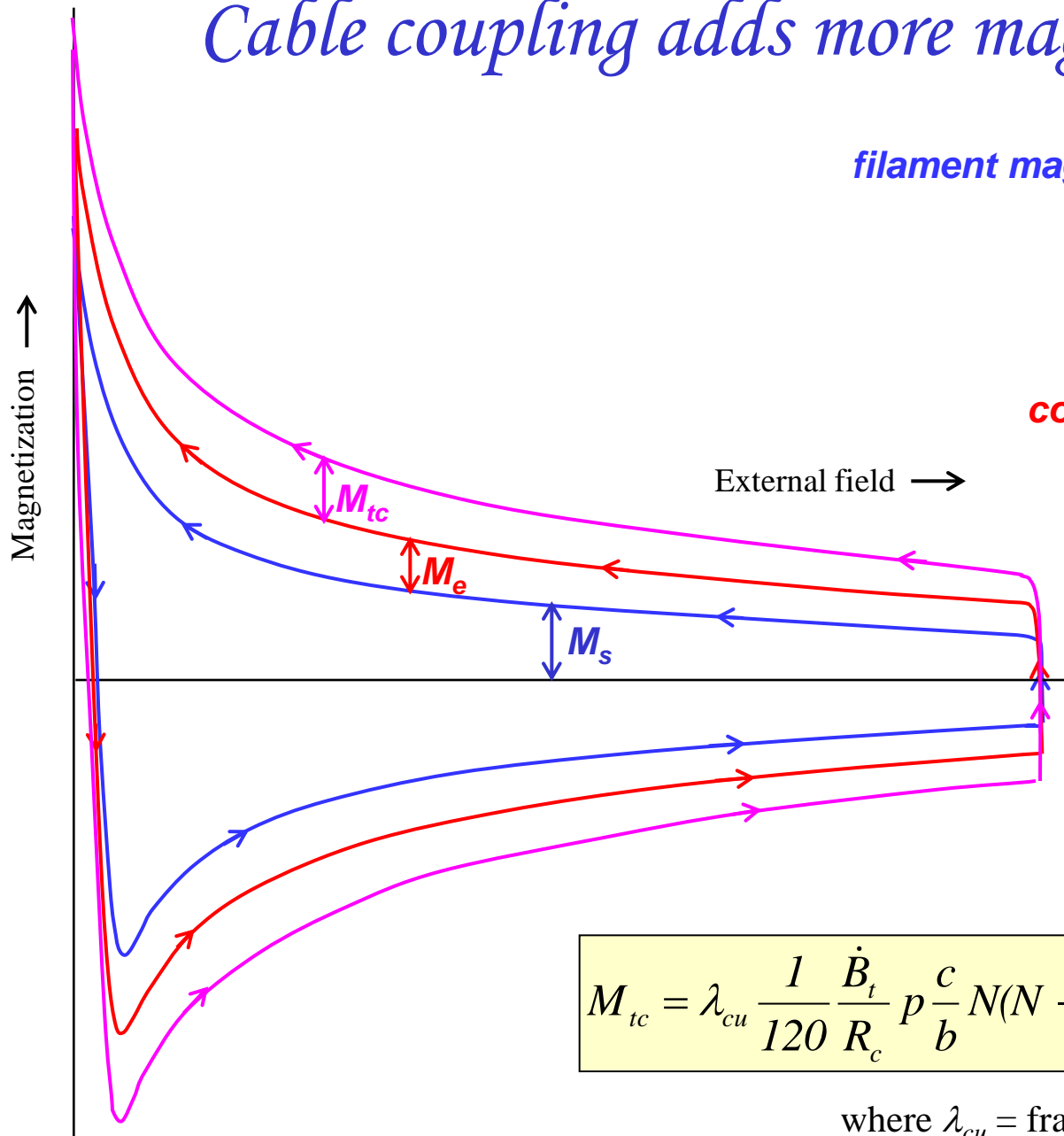
- Field transverse  
ratio crossover/adjacent

$$\frac{M_{tc}}{M_{ta}} = \frac{R_a}{R_c} \frac{N(N-1)}{20} \approx 45 \frac{R_a}{R_c}$$

So without increasing loss too much can make  $R_a$  50 times less than  $R_c$  - anisotropy



# Cable coupling adds more magnetization



filament magnetization  $M_f$  depends on  $B$

$$M_s = \lambda_{su} \frac{2}{3\pi} J_c(B) d_f$$

coupling between filaments  $M_e$  depends on  $dB/dt$

$$M_e = \lambda_{wu} \frac{dB}{dt} \frac{1}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

coupling between wires in cable depends on  $dB/dt$

$$M_{tc} = \lambda_{cu} \frac{1}{120} \frac{\dot{B}_t}{R_c} p \frac{c}{b} N(N-1)$$

$$M_{ta} = \lambda_{cu} \frac{1}{6} \frac{\dot{B}_t}{R_a} p \frac{c}{b}$$

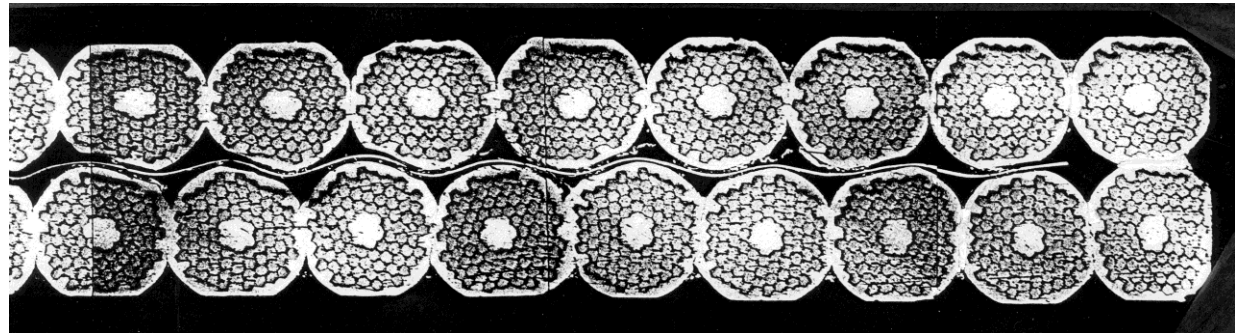
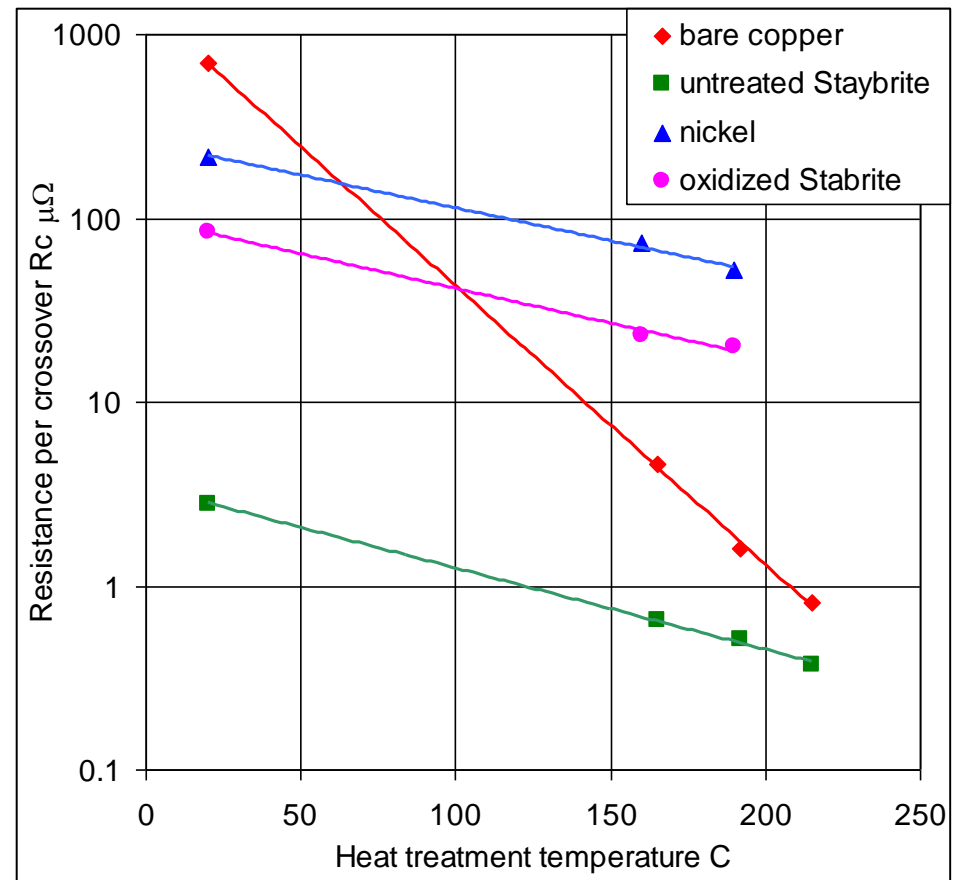
where  $\lambda_{cu}$  = fraction of cable in the section

# Controlling $R_a$ and $R_c$

- surface coatings on the wires are used to adjust the contact resistance
- the values obtained are very sensitive to pressure and heat treatments used in coil manufacture (to cure the adhesive between turns)
- *data from David Richter CERN*

## Cored Cables

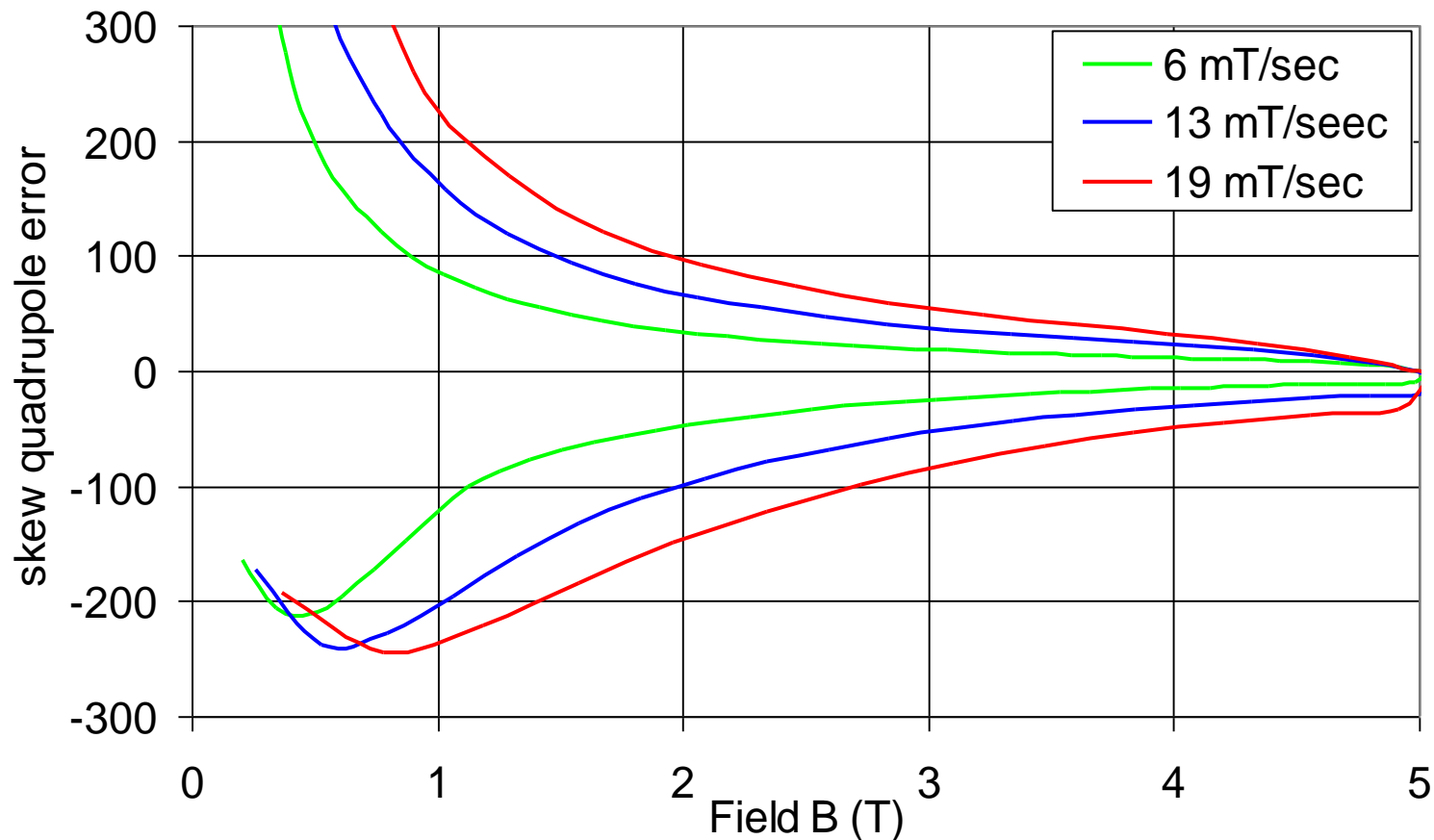
- using a resistive core allows us to increase  $R_c$  while keeping  $R_a$  the same
- thus we reduce losses but still maintain good current transfer between wires
- not affected by heat treatment



# Magnetization and field errors - extreme case

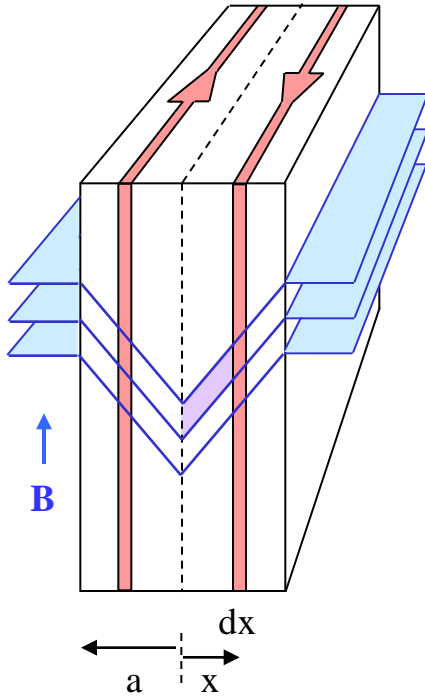
Magnetization is important in accelerators because it produces field error. The effect is worst at injection because

- $\Delta B/B$  is greatest
- magnetization, ie  $\Delta B$  is greatest at low field



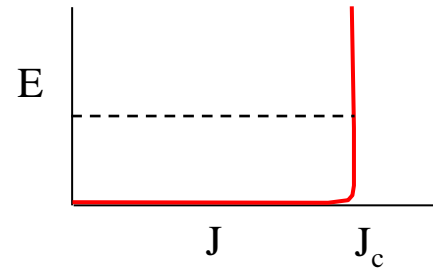
*skew  
quadrupole  
error in  
Nb<sub>3</sub>Sn dipole  
which has  
exceptionally  
large  
coupling  
magnetization  
(University of  
Twente)*

# AC loss power



Faraday's law of induction

$$\oint E dl = \frac{d}{dt} \int_a B da$$



loss power / unit length in slice of width  $dx$

$$p(x) = E J_c dx = \frac{dB}{dt} J_c dx$$

total loss in slab per unit volume

$$P = \dot{B} M$$

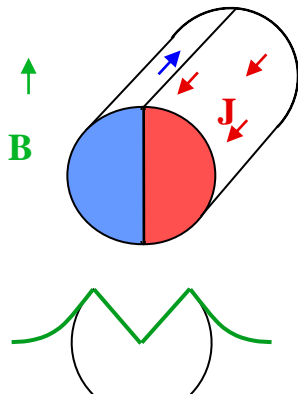
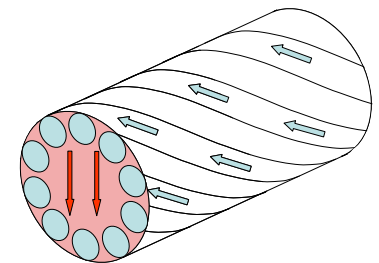
$$P = \frac{1}{a} \int_0^a p(x) dx = \frac{1}{a} \frac{dB}{dt} J_c \int_0^a x dx = \dot{B} J_c \frac{a}{2} = \dot{B} M$$

for round wires (not proved here)

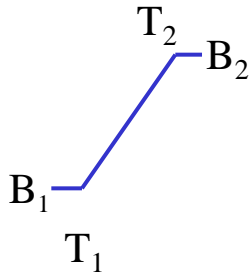
$$P = \dot{B} M = \frac{4}{3\pi} \dot{B} J_c a = \frac{2}{3\pi} \dot{B} J_c d_f$$

also for coupling magnetization

$$P_e = \dot{B} M_e = \dot{B}^2 \frac{l}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2 = \frac{\dot{B}^2}{\mu_o} 2\tau$$



# Hysteresis loss



loss over a field ramp

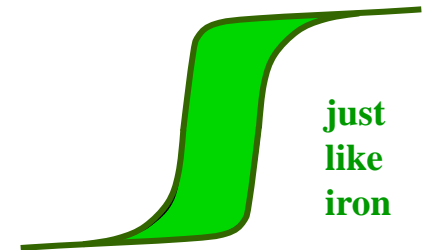
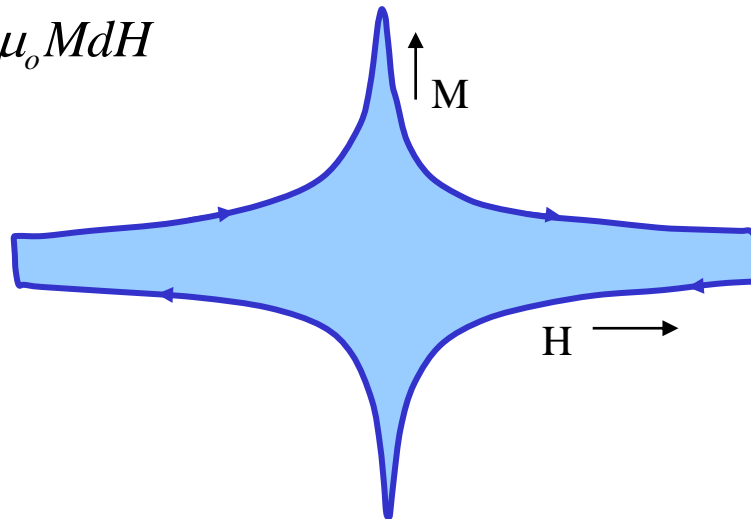
$$Q = \int_{T_1}^{T_2} M \frac{dB}{dt} dt = \int_{B_1}^{B_2} M dB$$

loss per ramp  
independent of  $\dot{B}$

- in general, when the field changes by  $\delta B$  the magnetic field energy changes by  $\delta E = H\delta B$  (see textbooks on electromagnetism)
- so work done by the field on the material  $W = \int \mu_o H dM$
- around a **closed loop**, this integral must be the energy dissipated in the material

$$Q = \int \mu_o H dM = \int \mu_o M dH$$

hysteresis loss  
per cycle  
(not per ramp)



just  
like  
iron



# Integrated loss over a ramp

## 1) Screening currents within filaments

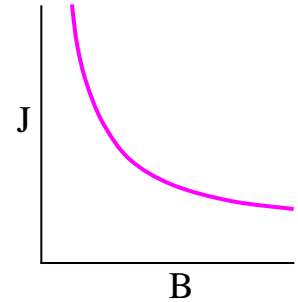
$$Q = \int_{B_1}^{B_2} M(B) dB$$

$J_c$  and  $M$  vary  
with field

Kim Anderson  
approximation

$$J_c(B) = \frac{J_o B_o}{(B + B_o)}$$

good at low field, less so at high field



round  
wire

$$M_s(B) = \frac{2}{3\pi} J_c(B) d_f$$

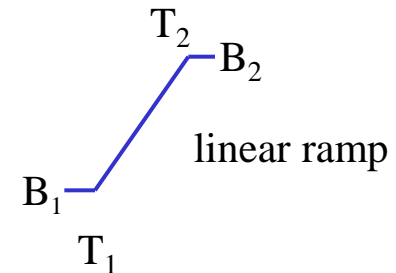
$$Q = \frac{2}{3\pi} \int_{B_1}^{B_2} \frac{J_o B_o}{(B + B_o)} d_f dB = \frac{2}{3\pi} d_f J_o B_o \ln \left\{ \frac{B_2 + B_o}{B_1 + B_o} \right\}$$

loss per ramp, independent of  $\dot{B}$

## 2) Coupling currents between filaments

$$M_e = \dot{B} \frac{1}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

$$Q = \int_{B_1}^{B_2} M(\dot{B}) dB = \frac{(B_2 - B_1)^2}{(T_2 - T_1)} \frac{1}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$



## 3) Coupling currents between wires in cable

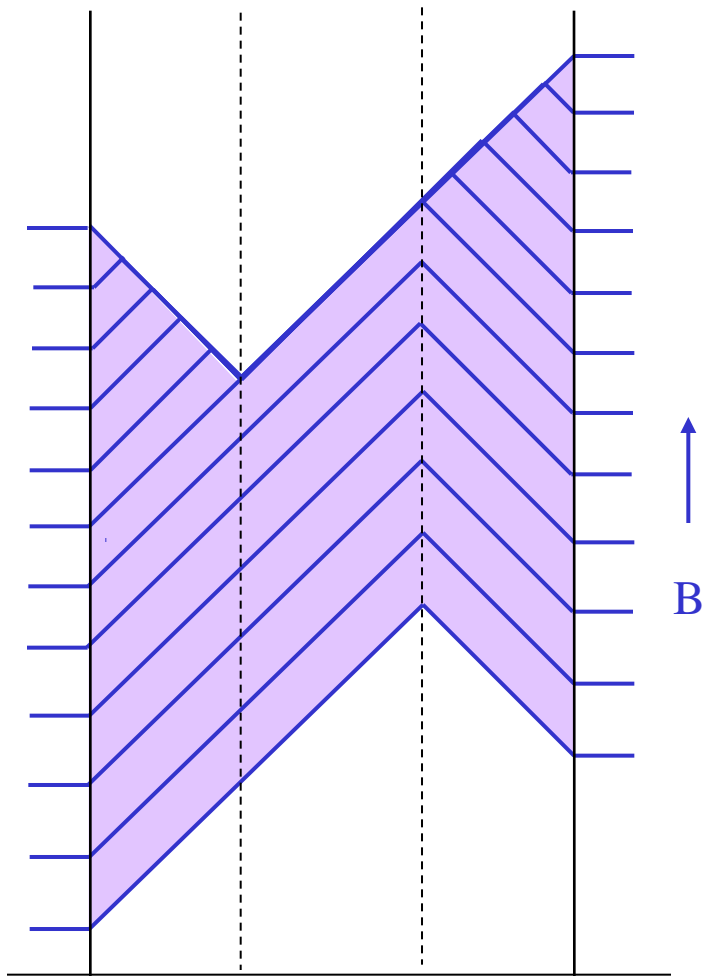
$$M_{tc} = \frac{1}{60} \frac{\dot{B}_t}{\rho_c} p_c^2 \frac{c^2}{b^2}$$

$$Q = \int_{B_1}^{B_2} M_{tc}(\dot{B}) dB = \frac{(B_2 - B_1)^2}{(T_2 - T_1)} \frac{1}{\rho_c} \frac{p_c^2}{60} \frac{c^2}{b^2}$$

loss per ramp  $\sim 1/\Delta T$

all per unit volume  
- volume of what?

# The effect of transport current



plot field profile across the slab

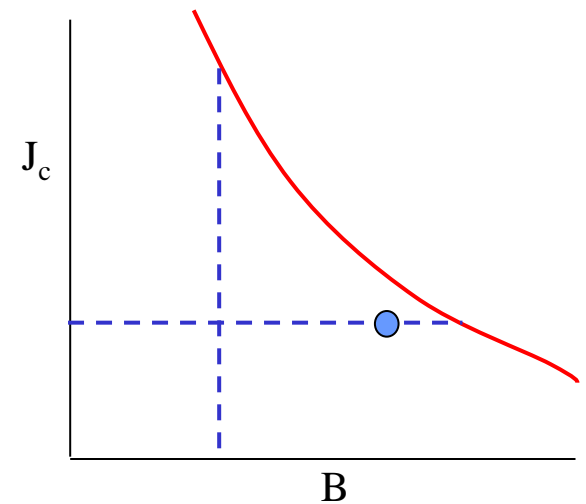
- in magnets there is a transport current, coming from the power supply, in addition to magnetization currents.
- because the transport current 'uses up' some of the available  $J_c$  the magnetization is reduced.
- but the loss is increased because the power supply does work and this adds to the work done by external field

total loss is increased by factor  $(1+i^2)$  where  $i = I_{max} / I_c$

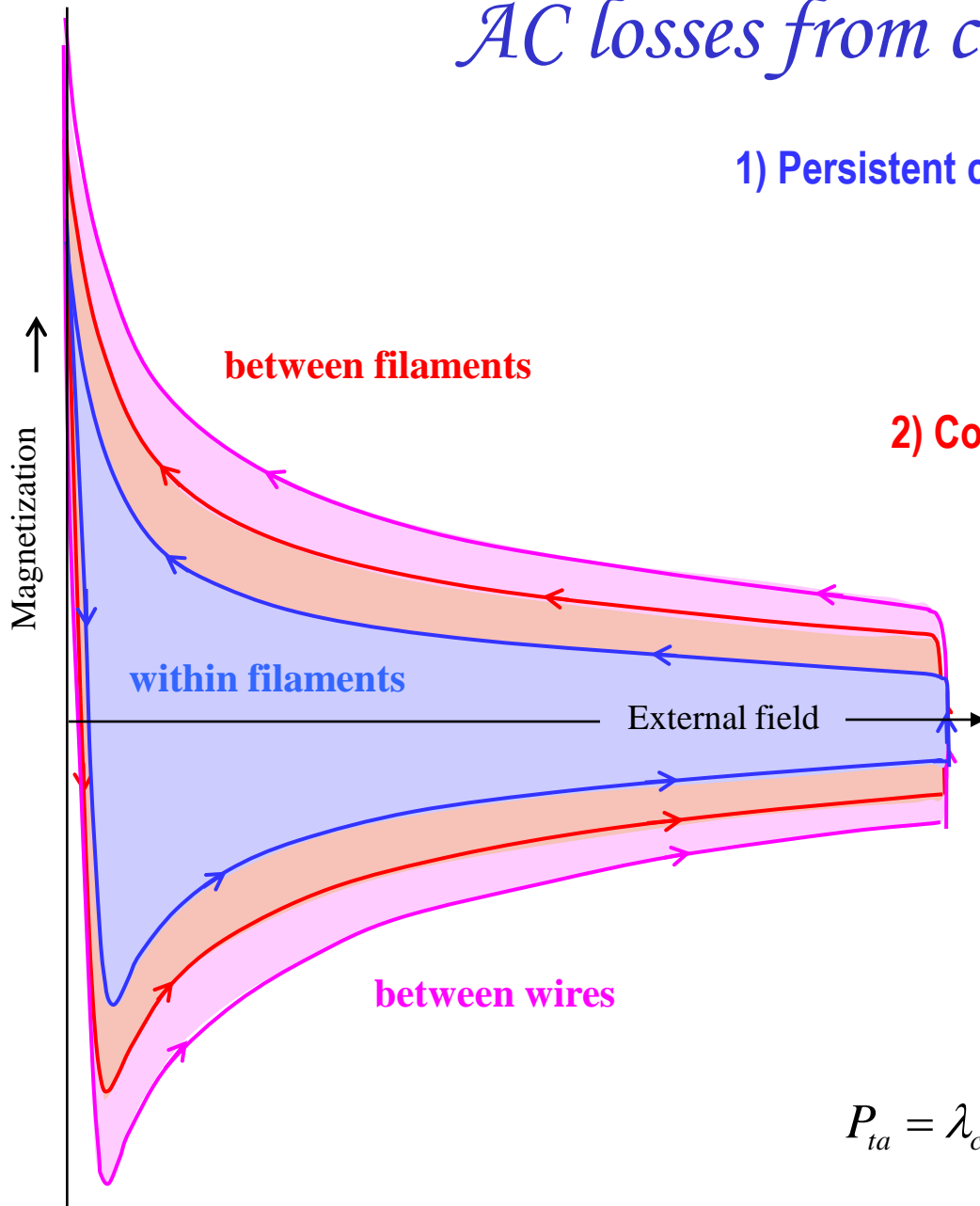
$$E = \frac{2}{3\pi} d_f J_o B_o \ln \left\{ \frac{B_2 + B_o}{B_1 + B_o} \right\} (1+i^2)$$

*usually not such a big factor because*

- *design for a margin in  $J_c$*
- *most of magnet is in a field much lower than the peak*



# AC losses from coupling



1) Persistent currents within filaments

$$P_s = \frac{2}{3\pi} \lambda_{su} \dot{B} J_c d_f$$

2) Coupling between filaments within the wire

$$P_e = \lambda_{wu} \dot{B}^2 \frac{1}{\rho_t} \left[ \frac{p_w}{2\pi} \right]^2$$

3) Coupling between wires in the cable

$$P_{tc} = \lambda_{cu} \frac{1}{120} \frac{\dot{B}_t^2}{R_c} \frac{c}{b} p_c N(N-1)$$

$$P_{ta} = \lambda_{cu} \frac{1}{6} \frac{\dot{B}_t^2}{R_a} p_c \frac{c}{b}$$

$$P_{pa} = \lambda_{cu} \frac{1}{8} \frac{\dot{B}_p^2}{R_a} p_c \frac{b}{c}$$

# Summary of losses - per unit volume of winding

## 1) Persistent currents in filaments

power  $\text{W.m}^{-3}$

$$P_s = \lambda_{su} M_f \dot{B} = \lambda_{su} \frac{2}{3\pi} J_c(B) d_f \dot{B}$$

loss per per ramp  $\text{J.m}^{-3}$

$$Q_s = \lambda_{su} \frac{2}{3\pi} d_f J_o B_o \ln \left\{ \frac{B_2 + B_o}{B_1 + B_o} \right\}$$

where  $\lambda_{su}$ ,  $\lambda_{wu}$ ,  $\lambda_{cu}$  = fractions of superconductor, wire and cable in the winding cross section

## 2) Coupling currents between filaments in the wire

power  $\text{W.m}^{-3}$

$$P_e = \lambda_{wu} M_e \dot{B} = \lambda_{wu} \frac{\dot{B}^2}{\rho_t} \left( \frac{p}{2\pi} \right)^2$$

## 3) Coupling currents between wires in the cable

transverse field crossover  
resistance power  $\text{W.m}^{-3}$

$$P_{tc} = \lambda_{cu} \frac{1}{120} \frac{\dot{B}_t^2}{R_c} p \frac{c}{b} N(N-1)$$

transverse field adjacent  
resistance power  $\text{W.m}^{-3}$

$$P_{ta} = \lambda_{cu} \frac{1}{6} \frac{\dot{B}_t^2}{R_a} p \frac{c}{b}$$

*don't forget the filling factors*

parallel field adjacent  
resistance power  $\text{W.m}^{-3}$

$$P_{pa} = \lambda_{cu} \frac{1}{8} \frac{\dot{B}_p^2}{R_a} p \frac{b}{c}$$

# Concluding remarks

- screening currents produce magnetization (magnetic moment per unit volume)
  - ⇒ lots of problems - field errors and ac losses
- in a synchrotron, the field errors from magnetization are worst at injection
- we reduce magnetization by making fine filaments - for practical use embed them in a matrix
- in changing fields, filaments are coupled through the matrix ⇒ increased magnetization
  - reduce it by twisting and by increasing the transverse resistivity of the matrix
- accelerator magnets must run at high current because they are all connected in series
  - combine wires in a cable, it must be fully transposed to ensure equal currents in each wire
- wires in cable must have some resistive contact to allow current sharing
  - in changing fields the wires are coupled via the contact resistance
    - different coupling when the field is parallel and perpendicular to face of cable
    - coupling produces more magnetization ⇒ more field errors
- irreversible magnetization ⇒ ac losses in changing fields
  - coupling between filaments in the wire adds to the loss
  - coupling between wire in the cable adds more

never forget that magnetization and ac loss are defined per unit volume - filling factors