

# Losses Scaling Law for Superconducting Filaments Magnetized by an Elliptical Field.

Clement Lorin,  
Denis Netter,  
Philippe J. Masson.

University of Houston  
Department of Mechanical Engineering  
Texas Center for Superconductivity

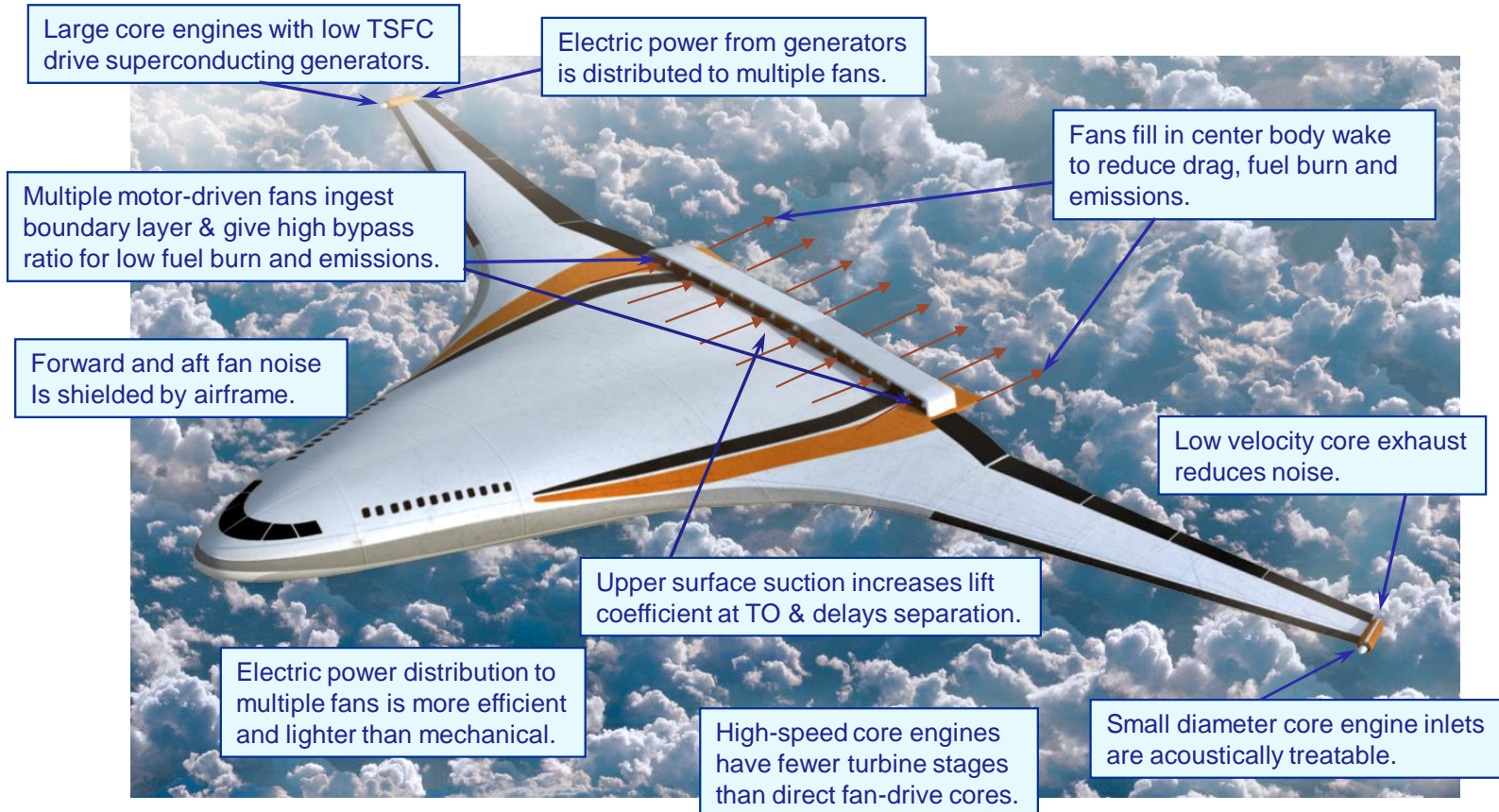
October 11<sup>th</sup>, 2013



National Aeronautics and Space Administration



## BENEFITS

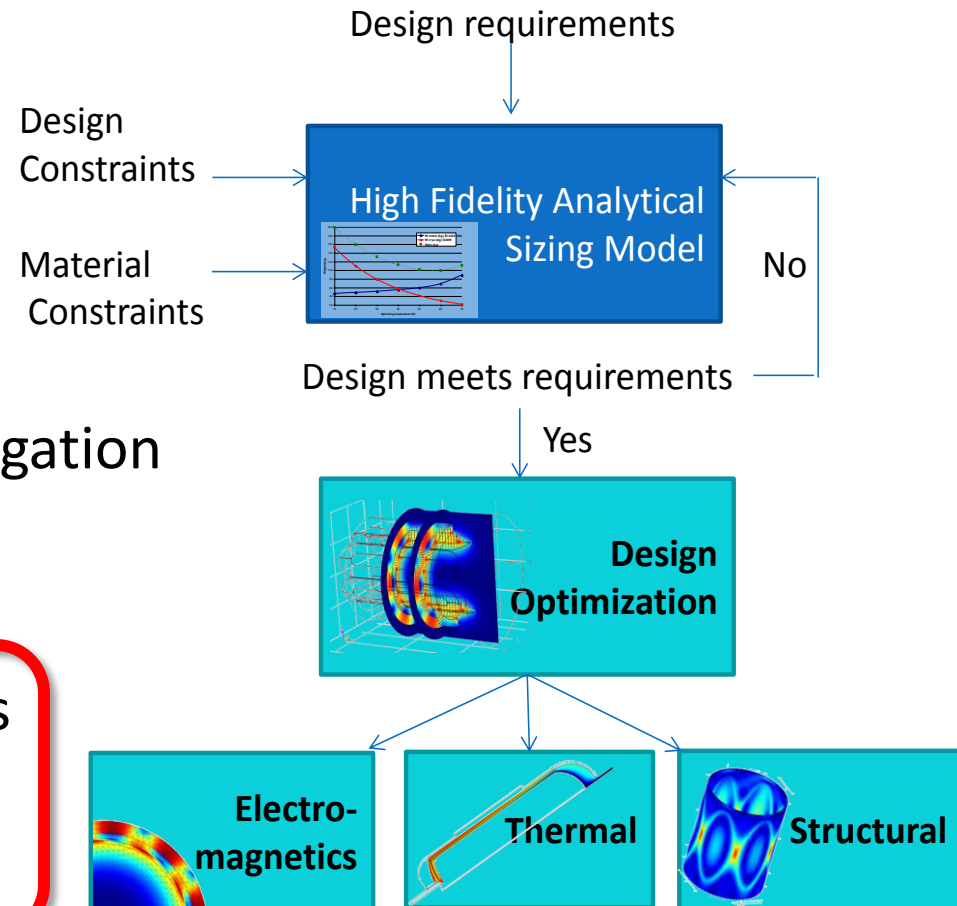


**THE TURBOELECTRIC APPROACH CONTRIBUTES TO EVERY CORNER OF THE SFW TRADE SPACE**

# Project Objectives

- Develop a **high fidelity** sizing tool for **fully superconducting rotating machines**.

- Accurate 3D geometry represented
  - Electromagnetics, mechanical and thermal
- Portable code in Python and C



- Develop model for quench propagation
  - Address detection and protection

- Develop **new model** for AC losses for superconducting stators
  - Based on FEA simulations

- Validate AC losses model **experimentally**

# Frequency Dependence of AC Losses

- 3 sources of AC losses in superconductors:

- Magnetization losses

$$P_{\text{mag}} \propto B, f, J_c$$

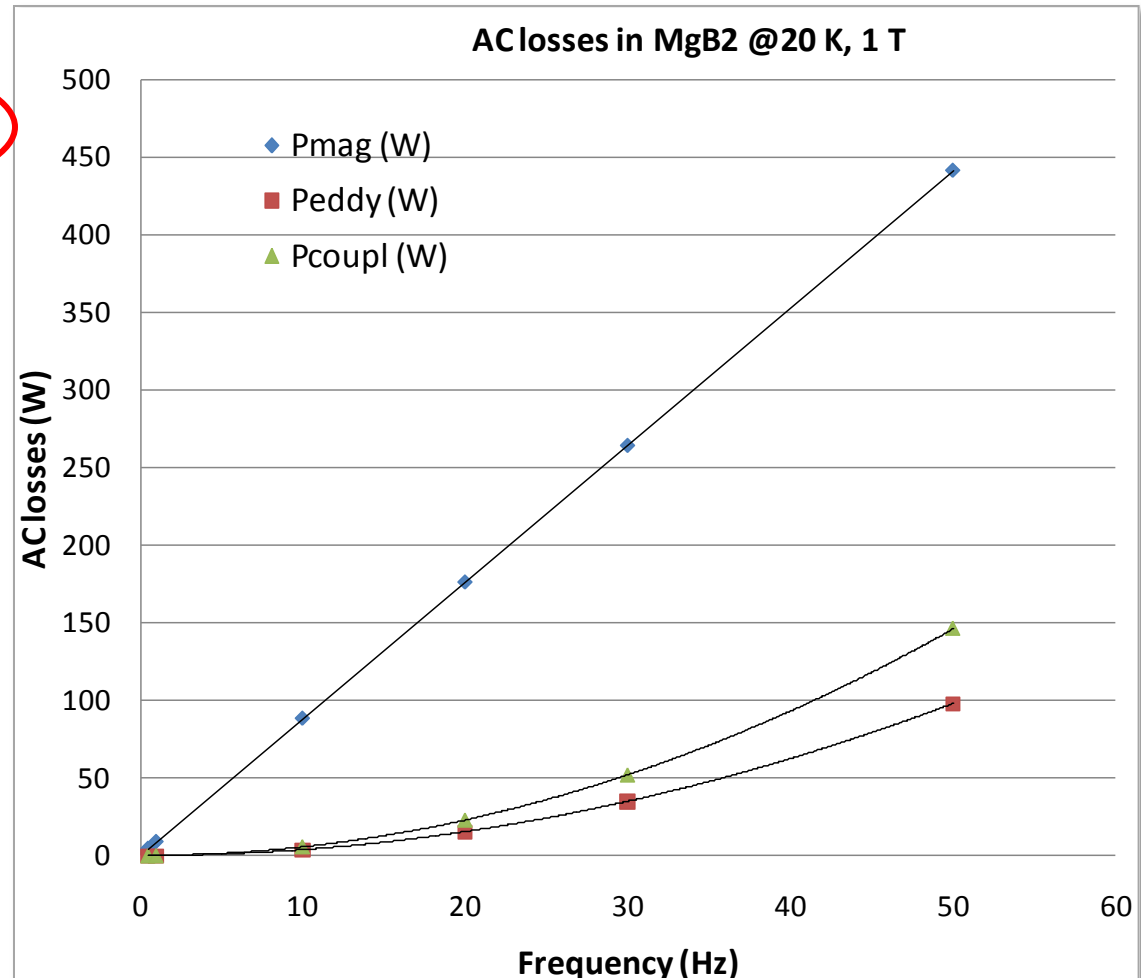
- Coupling losses

$$P_{\text{coupl}} \propto B^2, f^2$$

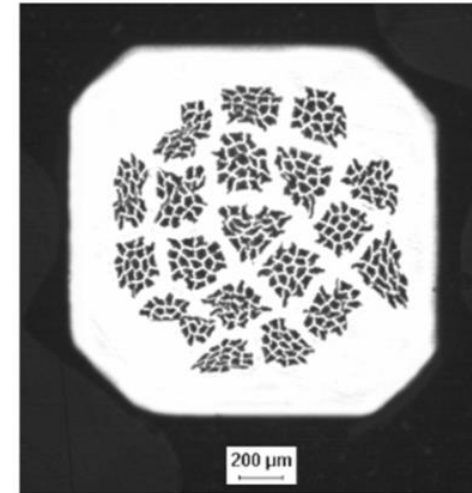
- Eddy current losses

$$P_{\text{Eddy}} \propto B^2, f^2$$

- **Magnetization losses are dominating at low frequency**



- Assumes AC field (alternating)
  - Limited to self field
- Transport current and applied field are in phase



$$P_m = \left( \frac{8}{3\pi} J_c \lambda_s d B_0 \right) (1 + F^2) \quad [W / m^3]$$

$J_c$ : filament critical current density

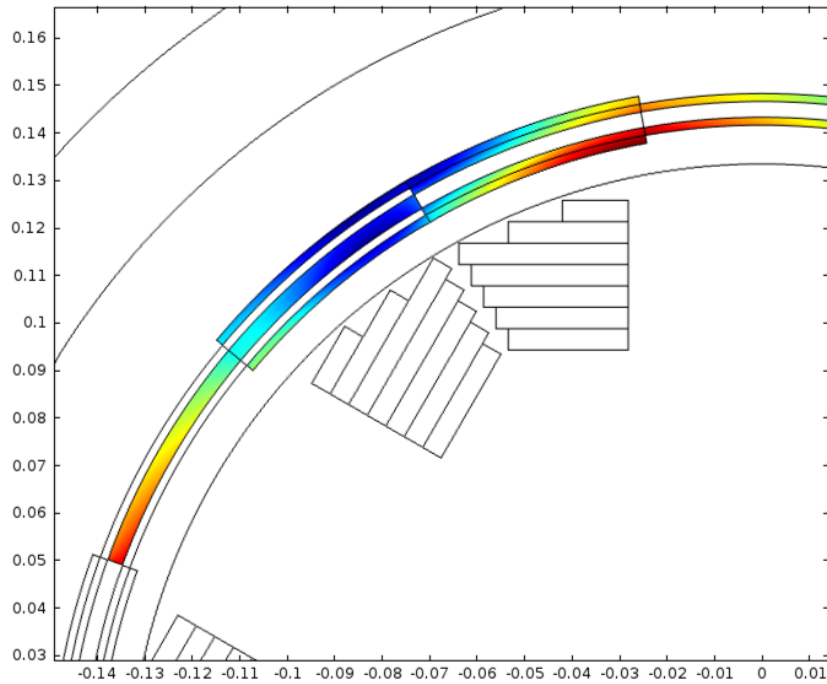
$\lambda_s$ : fraction of superconductor

$d$ : filament size

$B_0$ : flux density variation

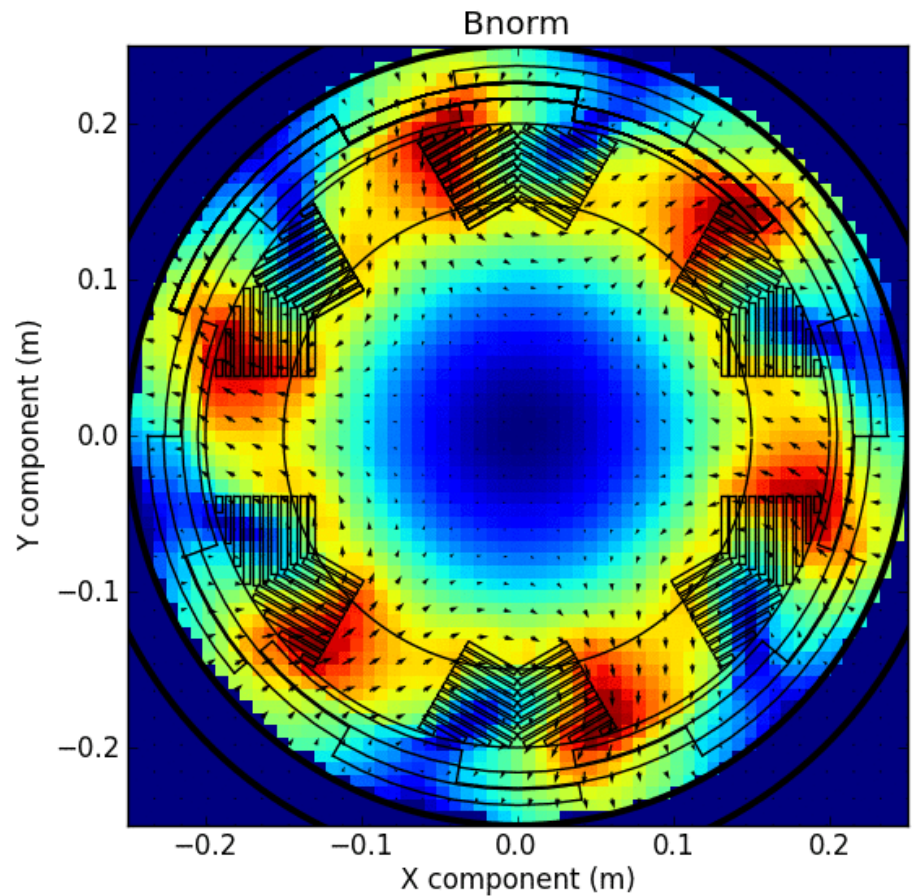
$F$ :  $I/I_c$

# Stator Peak Field at Full Load



- Non uniform flux density distribution

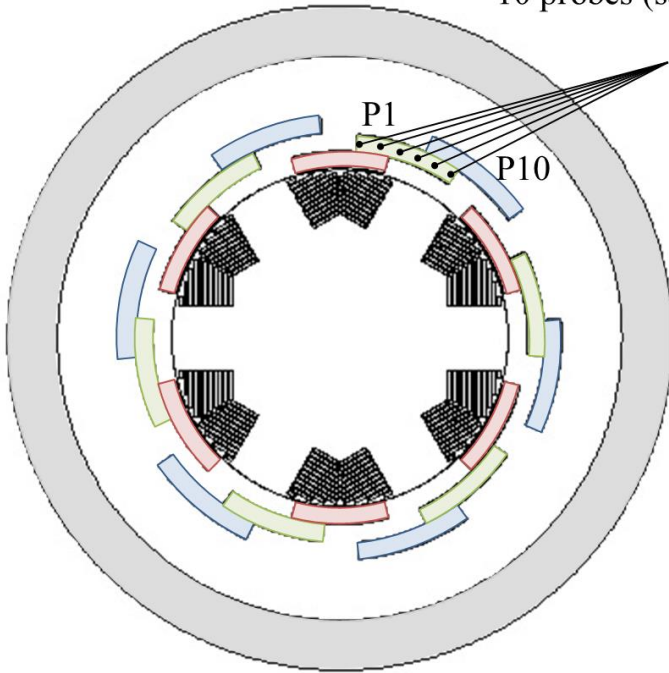
- Flux density contribution from both rotor and stator



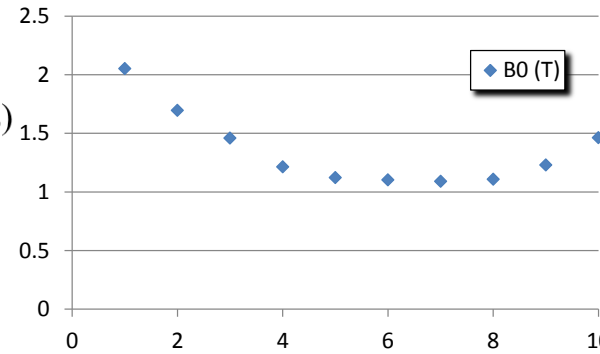
- Magnetization losses in stator wires of rotating machines:
  - **Stator**: **Rotating** field generated by **temporal variation** of currents inside the windings.
  - **Rotor**: **Rotating** and **Alternating** field produced by **rotation of the spatial distribution** of the currents.
- Local field = rotating field + alternating field

# Introduction (2/3)

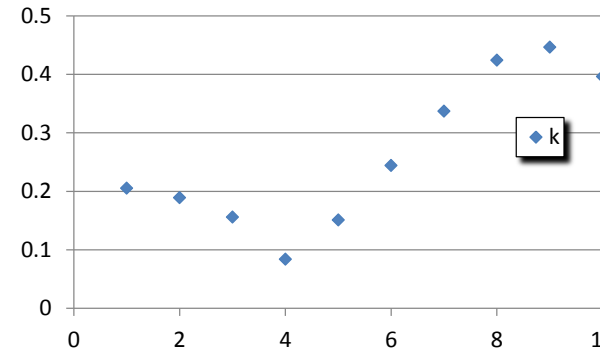
10 probes (same radius)



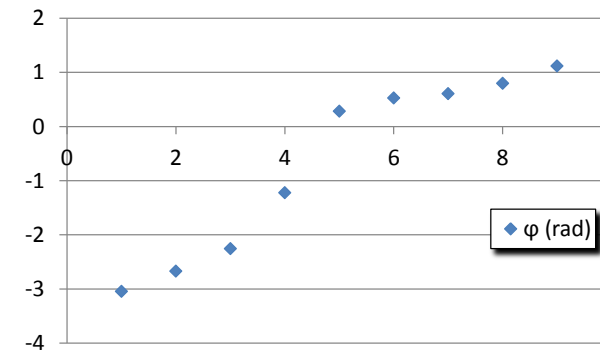
**In 1 pole, each stator conductor is in a unique magnetic configuration**



Peak field



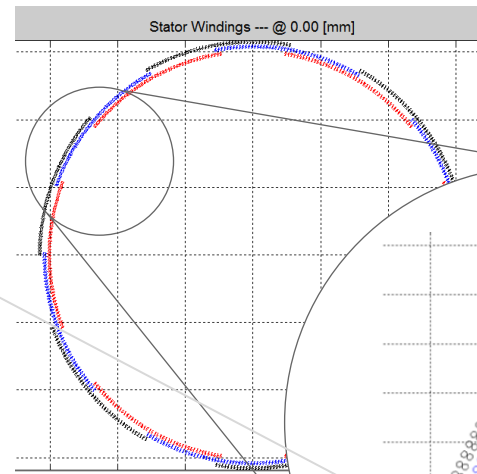
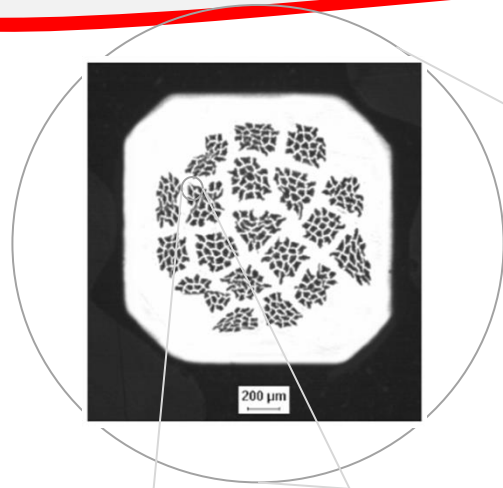
Rotating/pulsating



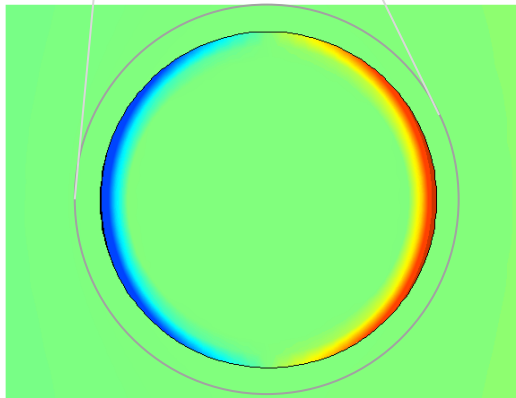
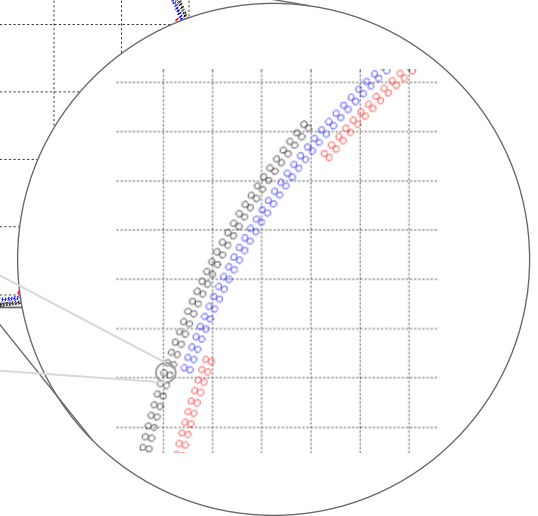
Phase angle rotating-pulsating

Probe #





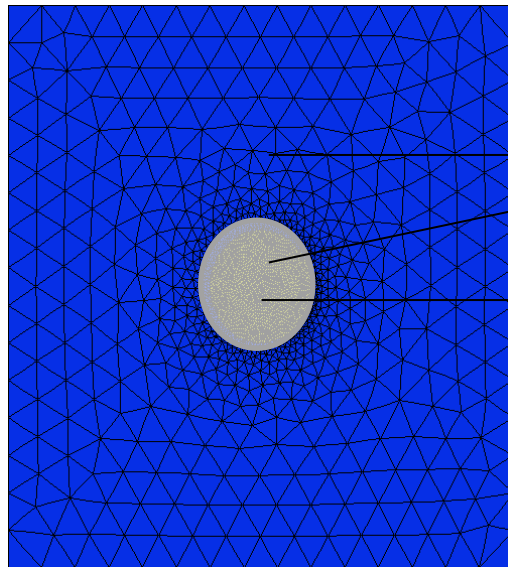
**Armature  
Cross-Section**



- Filaments are twisted and magnetically decoupled
  - Analysis can be done on 1 filament
  - Model implemented in COMSOL Multiphysics
  - Sc. Modeled using a power law
  - No current yet

# Model description (1/2)

- Geometry and model description:



$r$  filament radius

H-Formulation: 
$$\begin{cases} \overrightarrow{\text{rot}} \vec{E} = -\mu_0 \frac{\partial \vec{H}}{\partial t} \\ \vec{J} = \overrightarrow{\text{rot}} \vec{H} \end{cases}$$

Electrical behavior, power-law:

$$\vec{E} = \frac{E_c}{J_c} \cdot \left( \frac{|\vec{J}|}{J_c} \right)^{n-1} \cdot \vec{J}$$

$n$  n-value  
 $J_c$  critical current @  $E_c = 100 \mu\text{V/m}$

Boundary conditions:

$$\vec{B}_{rot} = \begin{cases} kB_0 \sin(2\pi ft) \\ kB_0 \cos(2\pi ft) \end{cases}$$

$$\vec{B}_{alt} = \begin{cases} 0 \\ (1-k)B_0 \cos(2\pi ft + \varphi) \end{cases}$$

$B_0$  field amplitude

$k = [0,1] \rightarrow$  field configuration: “*ellipticity*”

$f$  frequency

$\varphi = [-\pi, \pi]$

**7** variables

- Loss [J/m/cycle]:

$$Q = \oint dt \int_S E_z J_z dS$$

Over 1 cycle

Filament cross-section ( $\pi r^2$ )

# Dimensional Analysis (1/2)

- Dimensional Analysis:

| Variable      | Usual unit        | [kg] | [m] | [s] | [A] |    |
|---------------|-------------------|------|-----|-----|-----|----|
| $J_c$         | A.m <sup>-2</sup> | 0    | -2  | 0   | 1   | n1 |
| $\mu_0$       | H.m <sup>-1</sup> | 1    | 1   | -2  | -2  | n2 |
| $r$           | m                 | 0    | 1   | 0   | 0   | n3 |
| $E_c$         | V.m <sup>-1</sup> | 1    | 1   | -3  | -1  | n4 |
| Q (per cycle) | J.m <sup>-1</sup> | 1    | 1   | -2  | 0   | n5 |
| $B_0$         | T                 | 1    | 0   | -2  | -1  | n6 |
| $f$           | Hz                | 0    | 0   | -1  | 0   | n7 |
| $\varphi$     | rad               | 0    | 0   | 0   | 0   |    |
| $k$           | adim              | 0    | 0   | 0   | 0   |    |
| $n$           | adim              | 0    | 0   | 0   | 0   |    |

$$\left( \frac{r^2 f J_c \mu_0}{E_c} \right)^{n_7} \left( \frac{B_0}{J_c \mu_0 r} \right)^{n_6} \left( \frac{Q}{J_c^2 \mu_0 r^4} \right)^{n_5}$$

3 dimensionless numbers

$f^*$  ,  $b^*$  ,  $q^*$

Remark: adding a current  $I_t$  would lead to an additional dimensionless number.

$$\underbrace{[J_c]^{n_1} [\mu_0]^{n_2} [r]^{n_3} [E_c]^{n_4} [Q]^{n_5} [B_0]^{n_6} [f]^{n_7}} = [0]$$

$$\left( \frac{I_t}{J_c r^2} \right)^{n_8}$$

This number must be dimensionless.

- Our parameters:

$$b^* = \frac{B_0}{B_p}$$

$$\text{with } B_p = \frac{2J_c \mu_0 r}{\pi}$$

$$f^* = \frac{f}{f_c}$$

$$\text{with } f_c = \frac{E_c}{r^2 J_c \mu_0}$$

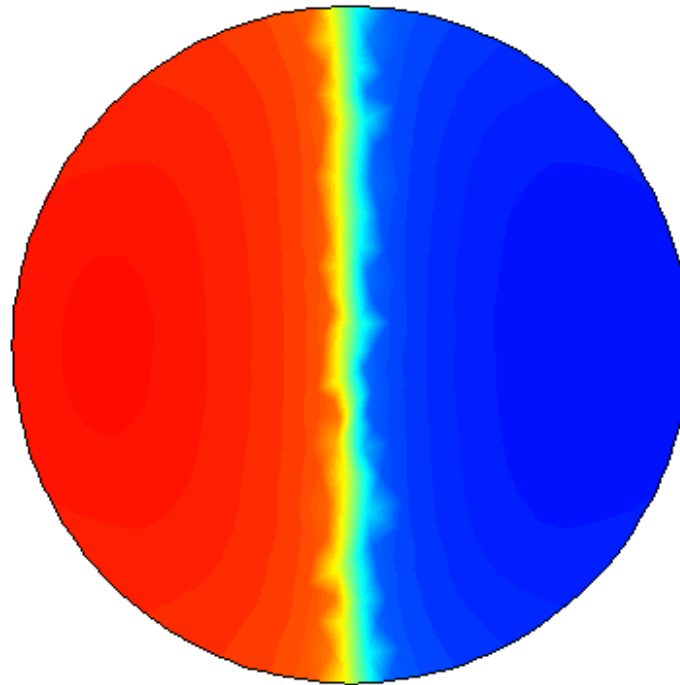
and  $n, k, \varphi$  5 variables instead of 7

Those parameters can be called invariants since the reduced losses  $q$  do not change if they do not change.

Reduced losses:

$$q^* = \frac{Q}{Q_c} \quad \text{with} \quad Q_c = \mu_0 J_c^2 r^4 = r B_p I_c \quad \text{with} \quad I_c = J_c \pi r^2$$

$b^*$   
 $f^*$   
 $k=0$   
 $n$

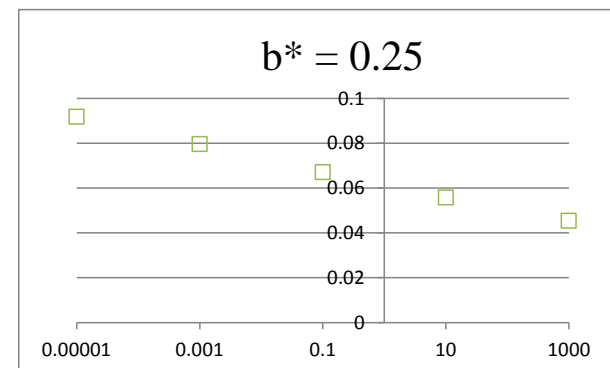
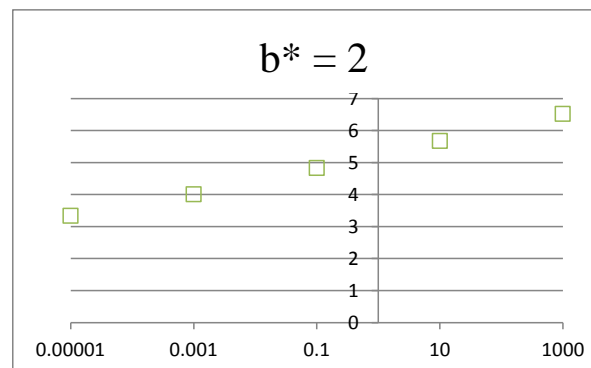
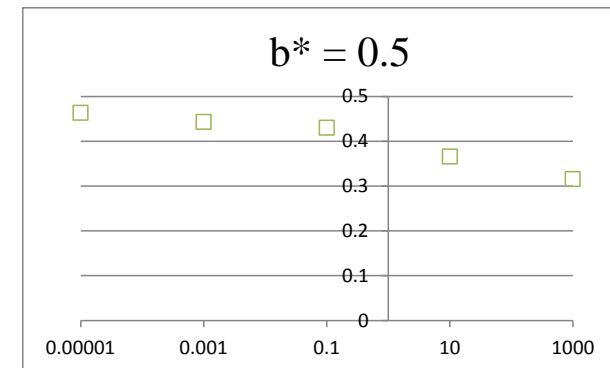
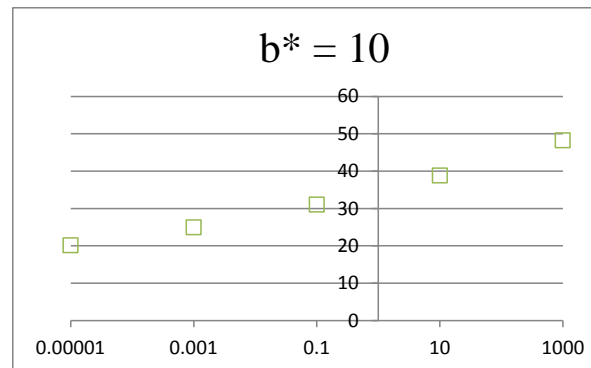
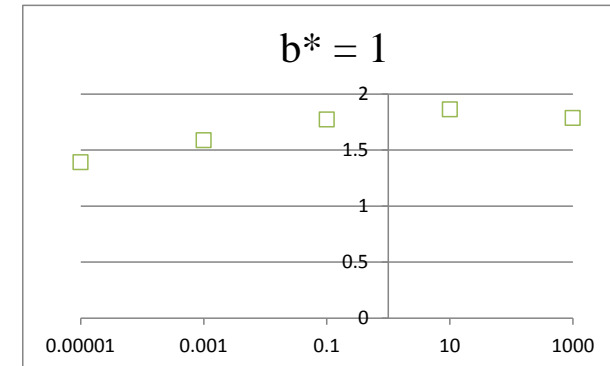
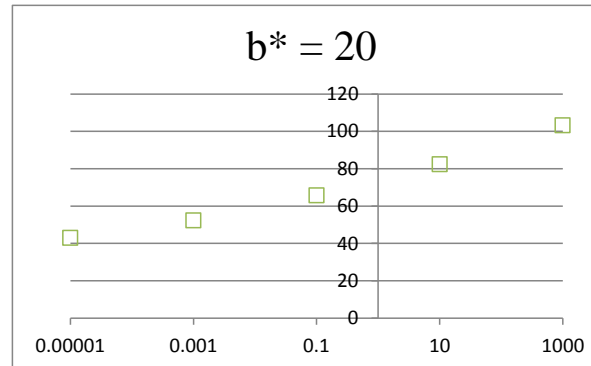


# Alternating field losses – SL (1/5)

- $n = 20$

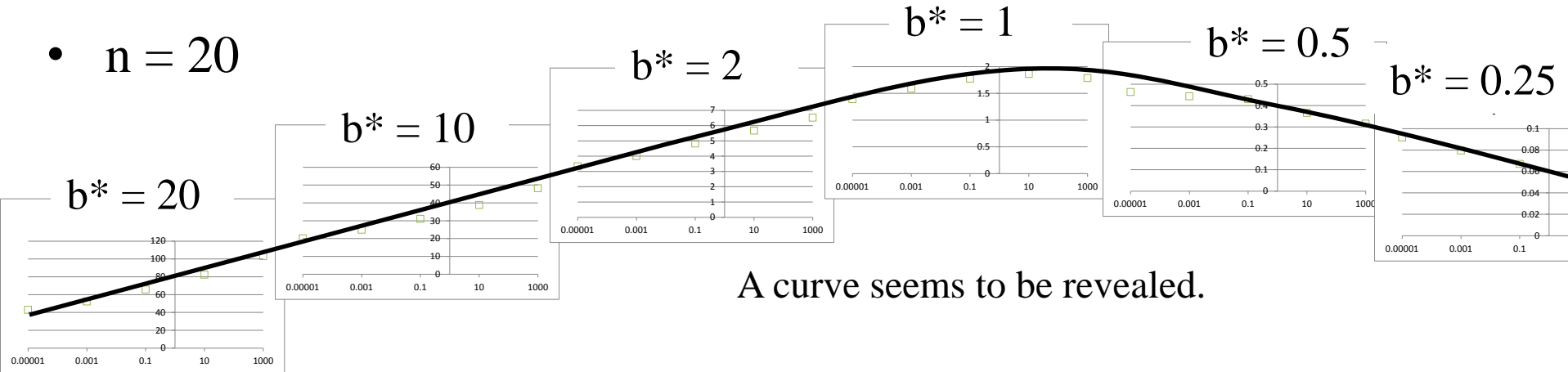
$q^*$  versus  $f^*$  for various  $b^*$

We need to  
**reorder the charts**  
 to explain the  
**thought process**  
 which led to a  
 semi-analytical  
 formula for AC  
 losses under  
 alternating field or  
 rotating field



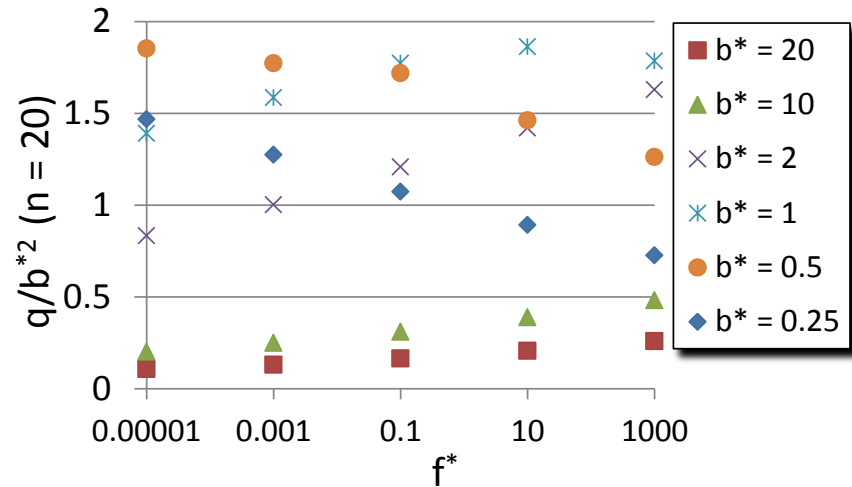
# Alternating field losses – SL (2/5)

- $n = 20$



A curve seems to be revealed.

→ We try to scale the ordinate to get reduced losses values in the same order of magnitude, to do so we divide  $q^*$  by  $b^{*2}$



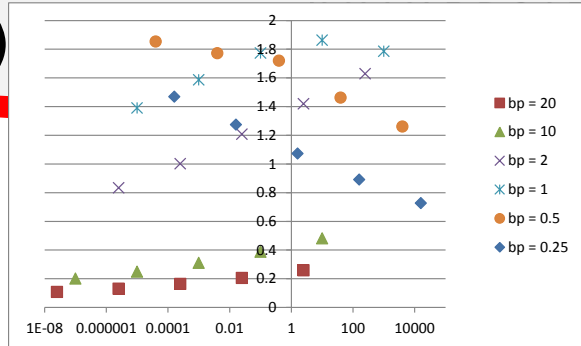
$q^*/b^{*2}$  versus  $f^*$  for multiple  $b^*$   
 ( $b_p = b^*$  former notation)

→ Then we try to adjust the abscissa values for each values of  $b^*$  so as to create a curve with all the dots looking like the black curve above.

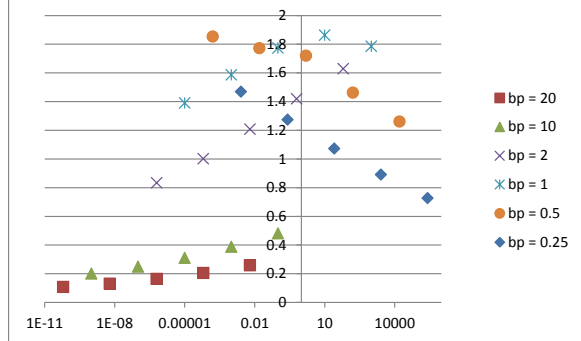


# Alternating field losses (n=20) – SL (3/5)

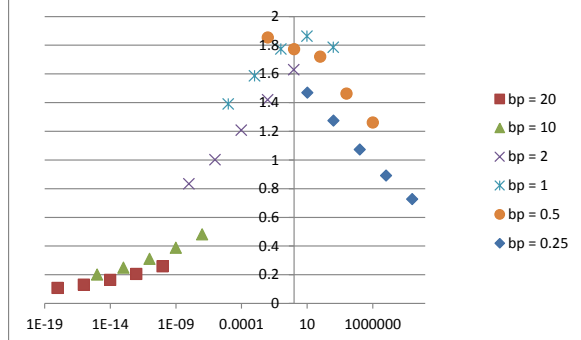
If we divide the abscissa by  $b^{*2} \Rightarrow$



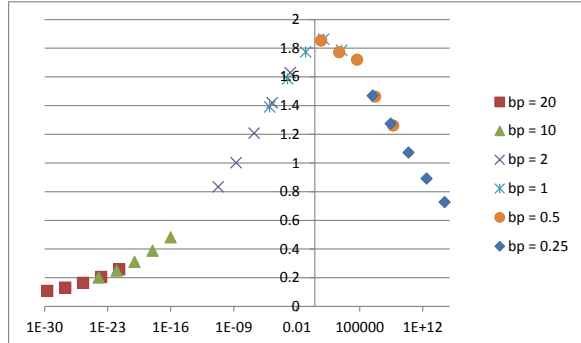
If we divide the abscissa by  $b^{*4} \Rightarrow$



If we divide the abscissa by  $b^{*10} \Rightarrow$

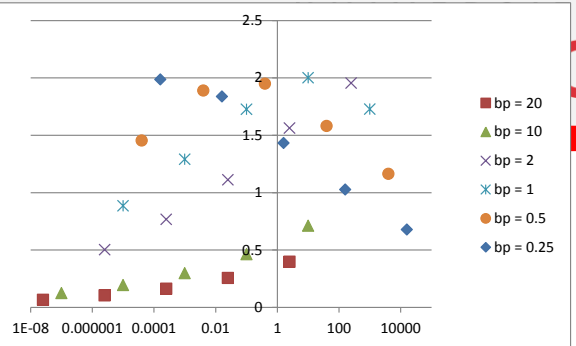


If we divide the abscissa by  $b^{*19} \Rightarrow$

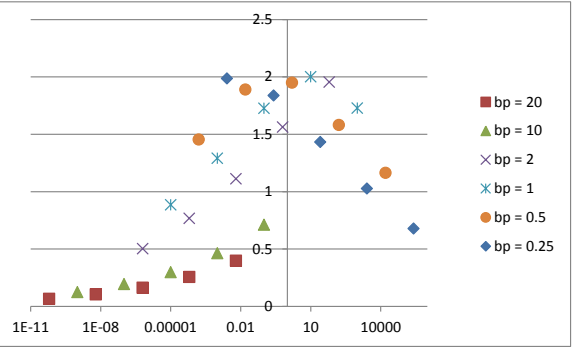


# Alternating field losses (n=10) – SL (4/5)

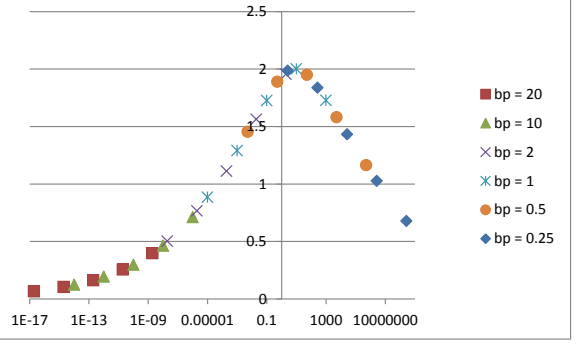
If we divide the abscissa by  $b^{*2} \Rightarrow$



If we divide the abscissa by  $b^{*4} \Rightarrow$



If we divide the abscissa by  $b^{*9} \Rightarrow$



We did the same for  $n = 2, n = 3, n = 5, n = 40, n = 50$

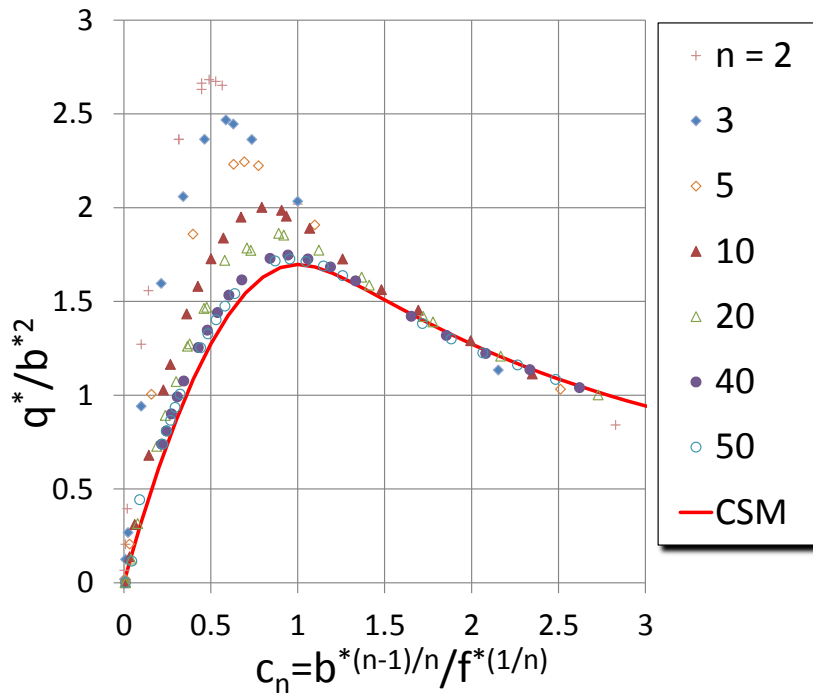
Dividing by  $b^{*n-1}$  aligns the data points. Instead of an abscissa  $f^{*}/b^{*n-1}$  we can use  $(f^{*}/b^{*n-1})^{1/n}$  to match **Critical State Model** (i.e. **no frequency dependency when  $n \rightarrow \infty$** ).



# Alternating field losses – Scaling Law (5/5)

$$\frac{q^*}{b^{*2}} = SL(c_n)$$

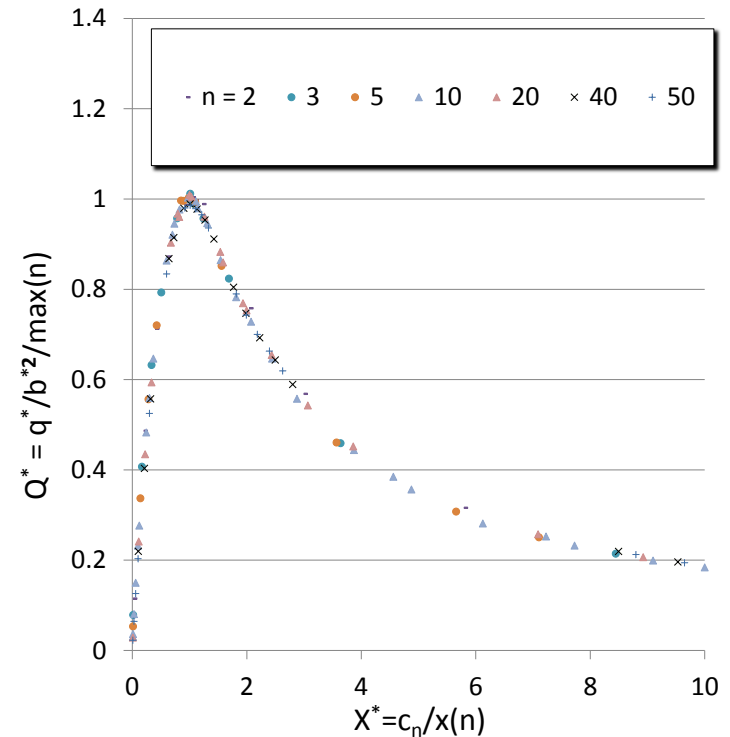
with  $c_n = \left( \frac{b^{*n-1}}{f^*} \right)^{\frac{1}{n}}$



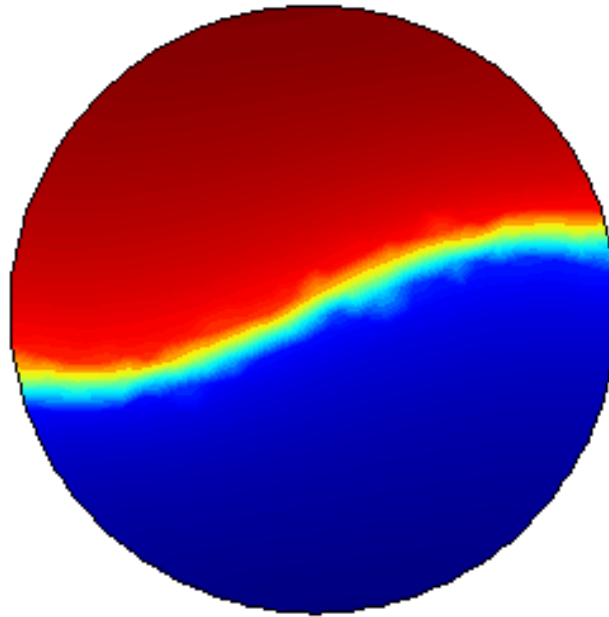
- Data can be normalized with respect to a function of n

$$X^* = \frac{c_n}{x(n)} \quad \text{with} \quad x(n) = 1 - \frac{\alpha_1}{n^{\alpha_2}}$$

$$Q^* = \frac{q^*}{b^{*2} \max(n)} \quad \text{with} \quad \max(n) = \alpha_3 + \frac{\alpha_4}{n^{\alpha_5}}$$



$b^*$   
 $f^*$   
 $k=1$   
 $n$



# Losses in Rotating field Scaling Law

Rotating  
field:  $k = 1$

$$\vec{B}_{rot} = \begin{cases} kB_0 \sin(2\pi ft) \\ kB_0 \cos(2\pi ft) \end{cases}$$

$$\frac{q^*}{b_p^2} = SL_{rot}(c_n)$$

Same method

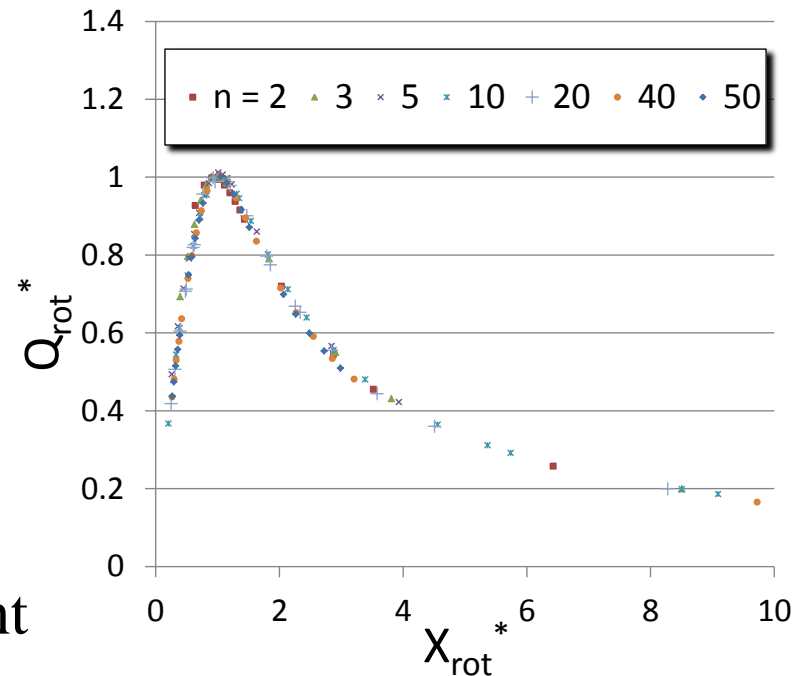
with  $c_n = \left( \frac{b^{*n-1}}{f^*} \right)^{\frac{1}{n}}$

- Data can be normalized with respect to a function of n

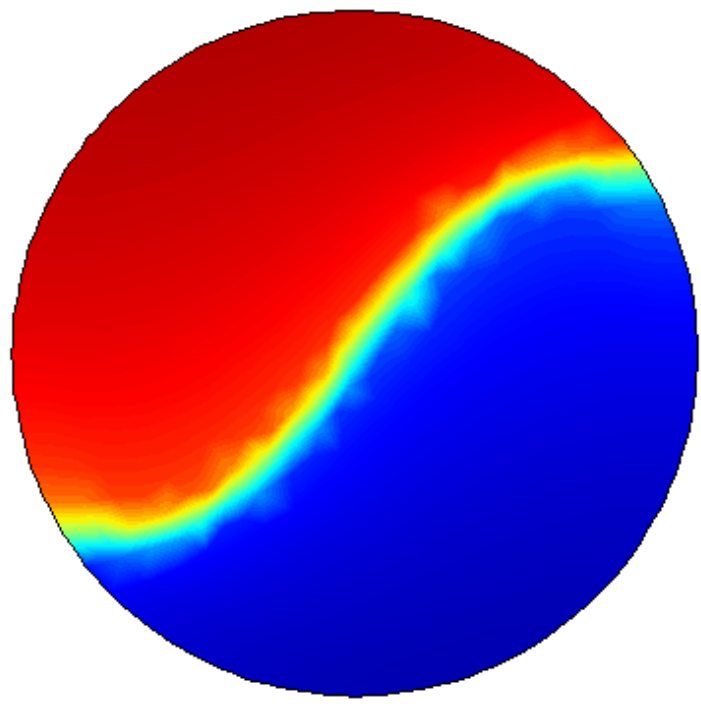
$$X_{rot}^* = \frac{c_n}{x_{rot}(n)} \quad \text{with} \quad x_{rot}(n) = 1 - \frac{\alpha_1}{n^{\alpha_2}}$$

$$Q_{rot}^* = \frac{q^*}{b^{*2} \max_{rot}(n)} \quad \text{with} \quad \max_{rot}(n) = \alpha_3 + \frac{\alpha_4}{n^{\alpha_5}}$$

Coefficients  $\alpha_1 \alpha_2 \alpha_3 \alpha_4 \alpha_5$  are different

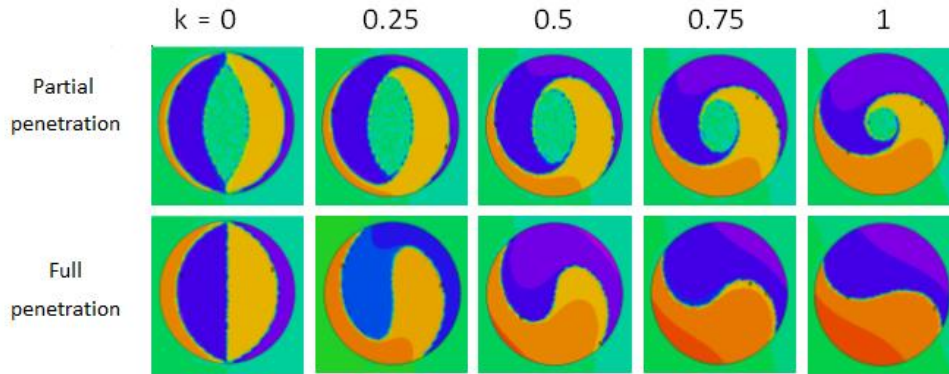


$b^*$   
 $f^*$   
 $k$   
 $n$



# Elliptical field

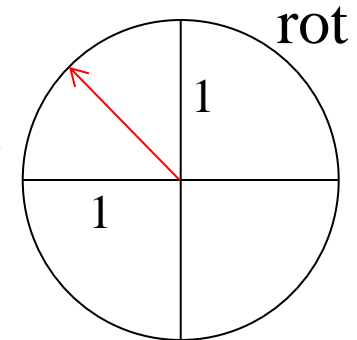
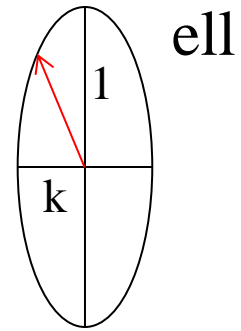
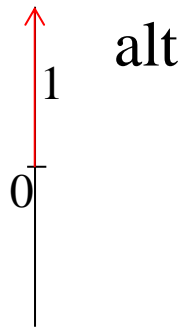
- $k = 0.25, 0.5, 0.75$



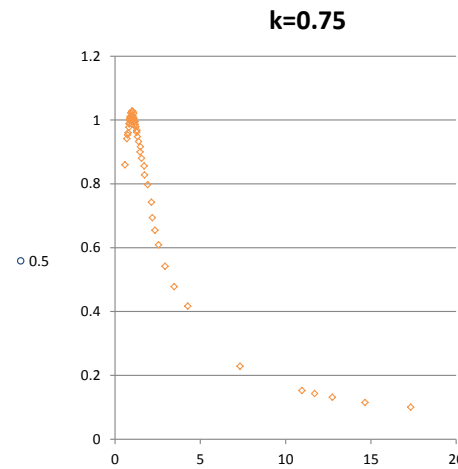
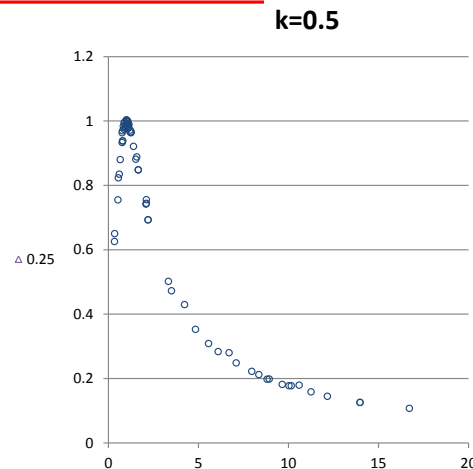
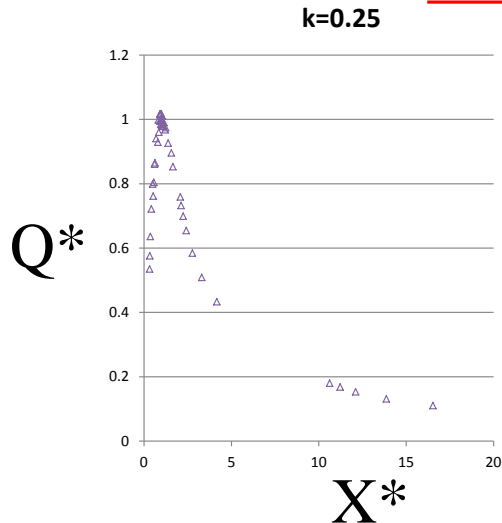
$$\vec{B}_{rot} = \begin{cases} kB_0 \sin(2\pi ft) \\ kB_0 \cos(2\pi ft) \end{cases}$$

$$\vec{B}_{alt} = \begin{cases} 0 \\ (1-k)B_0 \cos(2\pi ft + \varphi) \end{cases}$$

$$\vec{B}_{total} = \vec{B}_{rot} + \vec{B}_{alt}$$



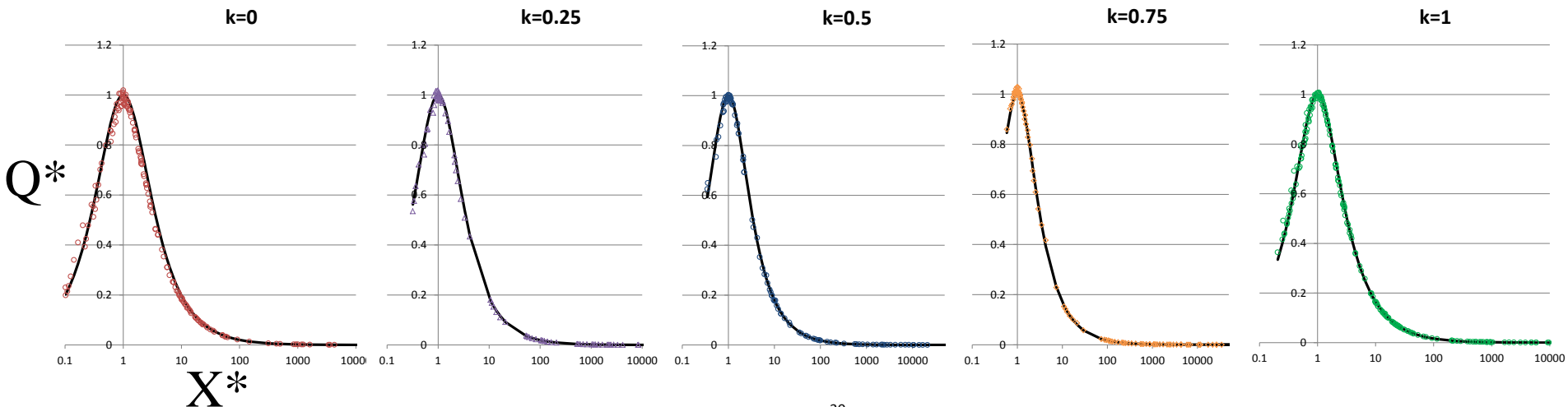
Same method



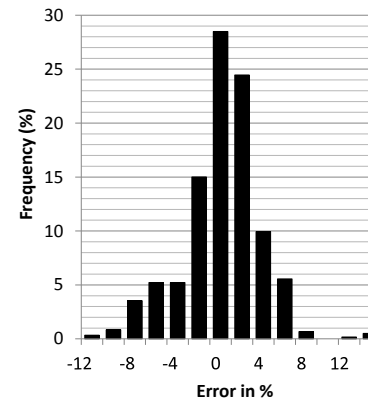
# Fit( $b^*, f^*, n, k$ ) or fit ( $c_n, n, k$ )

- Fit curves

General expression: 
$$Q^* = \frac{X^*}{\left[0.5(1 + X^* \beta_{in})\right]^{\beta_{out}}}$$

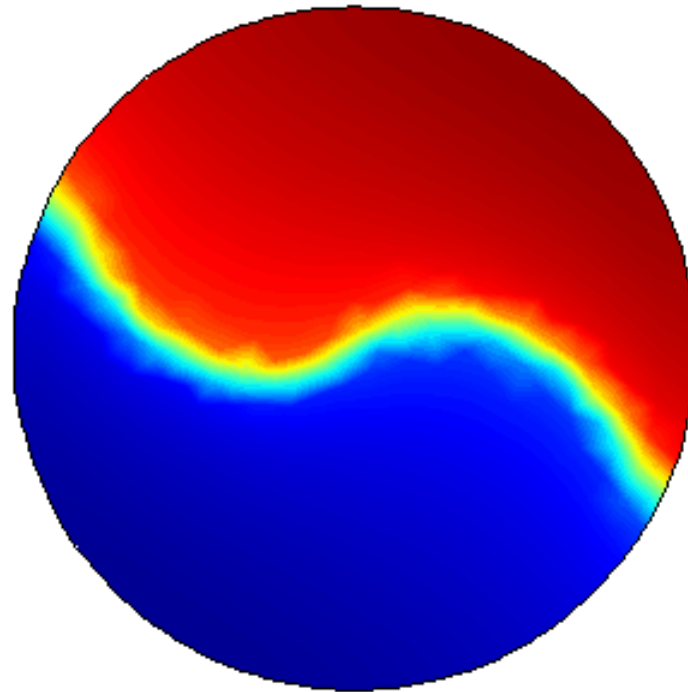


Error on data used to  
 get the fit (~600):





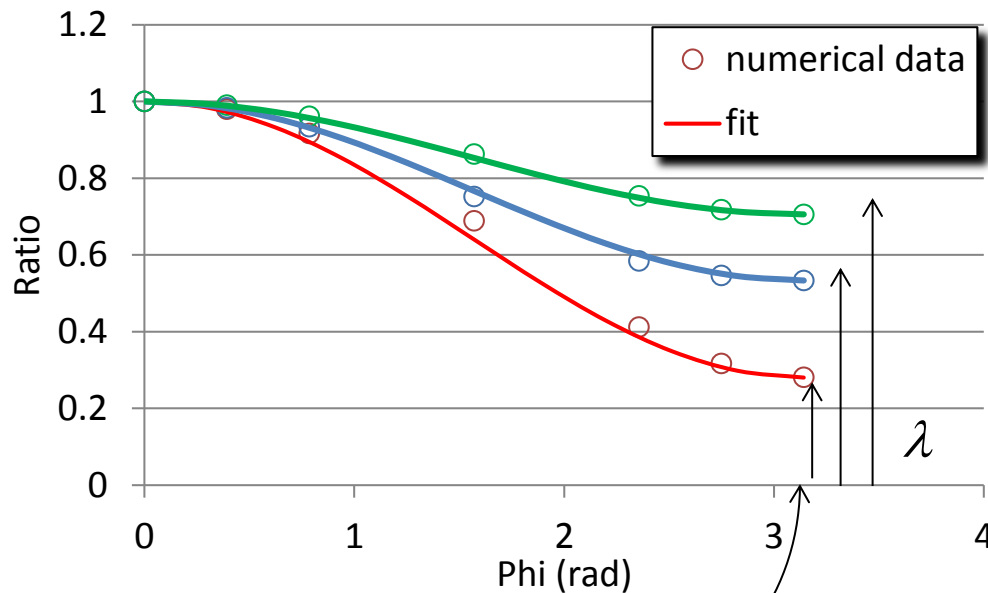
$b^*$   
 $f^*$   
 $k$   
 $n$   
 $\varphi$



# Elliptical field – with phase angle

- Ratio between losses with and without phase angle can be approximate by:

$$\frac{Q^*(k, n, b^*, f^*, \varphi)}{Q^*(k, n, b^*, f^*, \varphi = 0)} = \frac{(1 - \lambda)}{2} (1 + \cos \varphi) + \lambda$$



$$\lambda = \frac{Q^*(k, n, b^*, f^*, \varphi = \pi)}{Q^*(k, n, b^*, f^*, \varphi = 0)}$$

$$Q^*(k, n, b^*, f^*, \varphi = \pi)$$

# Elliptical field – with phase angle

- Transform<sup>☺</sup>  $Q^*(k, \varphi = \pi, B_0, \dots)$  to  $Q^*(k_1, \varphi = 0, B_1, \dots)$
- $k_1$  must be in  $[0, 1]$  to use our previous fit.

TABLE I  
 LOSSES EQUIVALENCE BETWEEN PI- AND ZERO-LOAD ANGLE

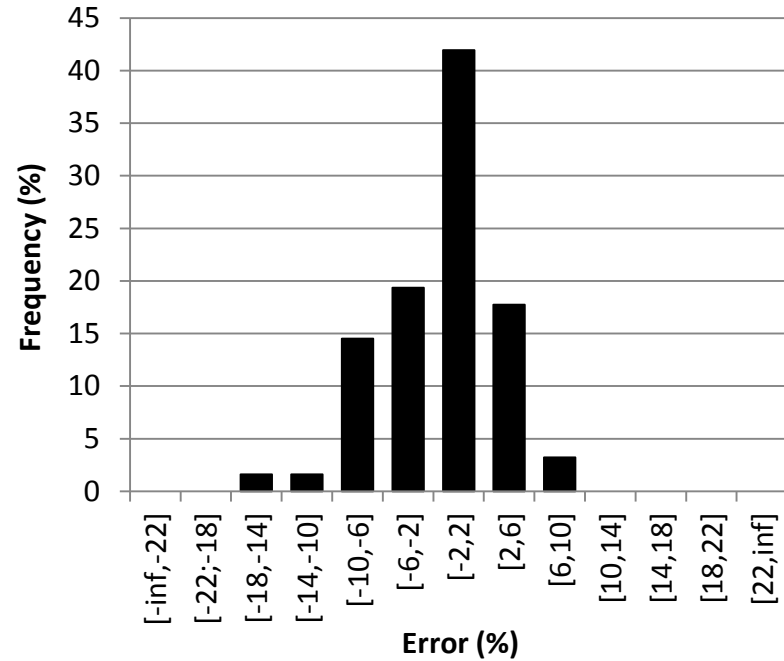
| k range                            | $k_1$            | $B_1$       |
|------------------------------------|------------------|-------------|
| $0 < k < \frac{1}{3}$              | $\frac{k}{1-2k}$ | $(1-2k)B_0$ |
| $\frac{1}{3} \leq k < \frac{1}{2}$ | $\frac{1-2k}{k}$ | $kB_0$      |
| $\frac{1}{2} \leq k \leq 1$        | $\frac{2k-1}{k}$ | $kB_0$      |

Based on geometrical considerations on the applied field.

$$\text{So } \lambda = \frac{W(k, \varphi = \pi, B_0)}{W(k, \varphi = 0, B_0)} = \frac{W(k_1, \varphi = 0, B_1)}{W(k, \varphi = 0, B_0)} \text{ is known.}$$

- Results

60 simulations  
randomly taken in  
the whole domain



- Investigation of elliptical field losses in smooth power-law electrical behavior round-shaped filaments:

- Reveal a fundamental variable:

$$c_n = \frac{b_0 \pi}{2 \mu_0 j_c r} \left( \frac{2 E_c}{b_0 f \pi^2 r} \right)^{\frac{1}{n}}$$

- Semi-analytical formulae give *fast* and reliable magnetization losses values depending on:

**r** filament radius

**n** n-value

**J<sub>c</sub>** critical current @ E<sub>c</sub> = 100 μV/m

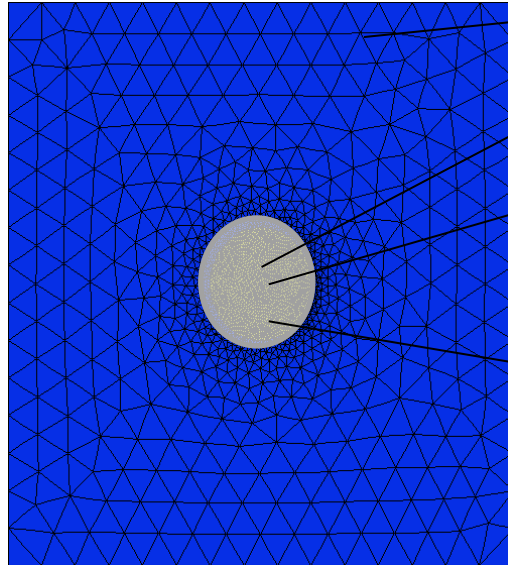
**B<sub>0</sub>** field amplitude

**k** = [0,1] → field configuration: “*ellipticity*”

**f** frequency

**φ** = [-π,π]

# Next Steps (1/2)



H-Formulation: 
$$\begin{cases} \text{rot} \vec{E} = -\mu_0 \frac{\partial \vec{H}}{\partial t} \\ \vec{J} = \text{rot} \vec{H} \end{cases}$$

Electrical behavior, power-law:

$$\vec{E} = \frac{E_c}{J_c} \left( \frac{|\vec{J}|}{J_c} \right)^{n-1} \vec{J} \quad \text{where} \quad \begin{aligned} J_c &= 100 \text{ A/mm}^2 \\ E_c &= 100 \mu\text{V/m} \\ n &= 20 \end{aligned}$$

Transport current constrain:

$$\iint_A j dA = I(t) \quad \text{where } A = \text{filament area}$$

Boundary conditions:

$$\frac{\vec{B}_{rot}}{B_p} = \begin{cases} kb_p \sin(\omega t) \\ kb_p \cos(\omega t) \end{cases}$$

$$\frac{\vec{B}_{alt}}{B_p} = \begin{cases} 0 \\ (1-k)b_p \cos(\omega t + \varphi) \end{cases}$$

$$\frac{I}{I_p} = i_p \cos(\omega t + \theta)$$

**bp** penetration field ratio

**k** = [0,1] → field configuration

**φ** = [-π,π] → rotor load angle

**ip** = [0,1] → penetration current ratio

**θ** = [0,2π] → current/field phase angle

## Next Steps (2/2)

- Implement model in Amber HTS Machine Sizing Code
- Perform experimental validation
  - Tests will start in January 2014
    - AC field
    - Transport current
  - Upgrade to include elliptical field scheduled for 2015

