

60 GHz Wireless Data Transfer for Tracker Readout Systems - First Studies and Results

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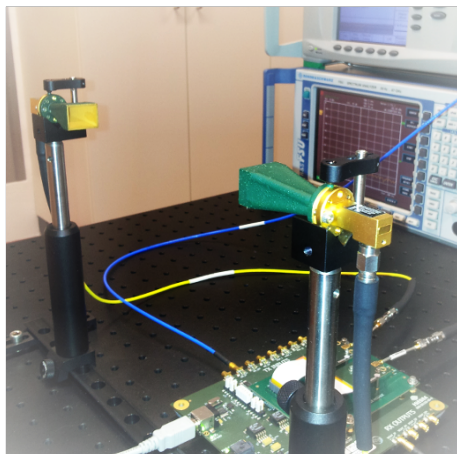
Physikalisches Institut, University of Heidelberg

Workshop on Intelligent Trackers 2014
14-16 May 2014



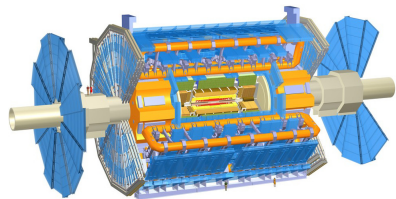
Outline

- 1 Introduction
- 2 Quality of data transmission
- 3 Crosstalk in detectors
- 4 Conclusion



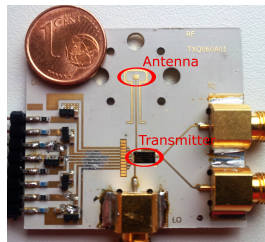
Data readout of particle detectors

- Huge data rates of HEP detectors at frontend, e.g. upgraded ATLAS tracker ~ 100 Tb/s
- Fast readout system with $\sim 20\,000$ links at ~ 5 Gb/s ?
- Approach: wireless data transmission at 60 GHz



Why to use this frequency band?

- License-free band 57 - 66 GHz (Europe)
- Large bandwidth \rightarrow data rates of 10's Gb/s
- Wavelength $\lambda \approx 5$ mm \rightarrow small form factor



Gotmic TXQ060A01 development board

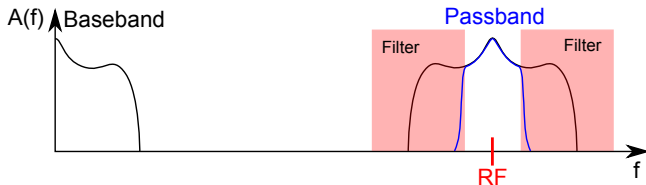
Signal transmission

Wired Baseband data

Optical Modulation of light

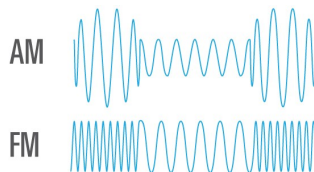
Wireless Modulation onto a carrier (60 GHz)

Filtered frequency range around the carrier: passband



Modulation

- Spectral efficiency (bit/s/Hz) depends on modulation scheme



General modulation schemes: analog and digital

AM Amplitude modulation

FM Frequency modulation

PM Phase modulation

ASK Amplitude shift keying

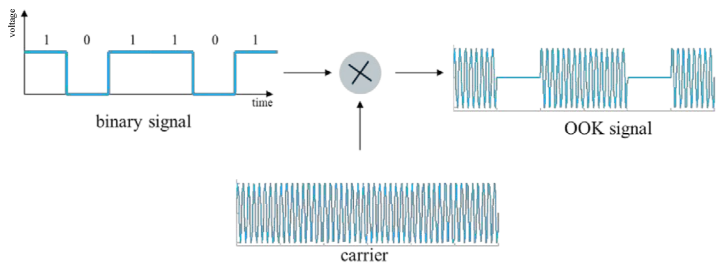
FSK Frequency shift keying

PSK Phase shift keying

QAM Quadrature amplitude modulation

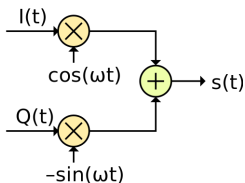
Modulation schemes: On-Off-Keying

- Simple implementation
- Non-coherent demodulation: no Tx phase information required at Rx
- No large baseband circuitry required, little power consumption
- Spectral efficiency ≤ 0.5 bit/s/Hz



Modulation schemes: IQ Modulation

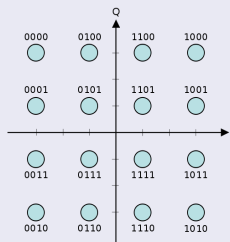
- In-phase and Quadrature components



- Requirements:
 - Higher signal to noise
 - Coherent demodulation
 - ADC baseband circuitry

Example for IQ-modulation

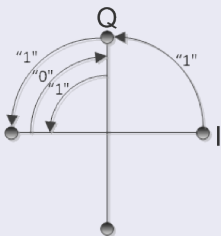
16-Quadrature amplitude modulation
(16-QAM, 4 bit/s/Hz)



Modulation schemes: Minimum Shift Keying

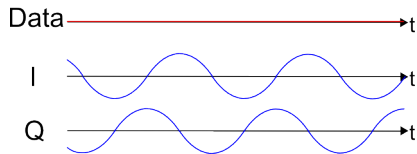
Constellation Diagram: MSK

Continuous Phase Frequency Shift Keying (CPFSK, 1 bit/s/Hz)



Non-coherent demodulation

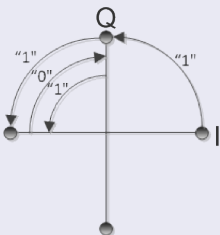
Data information in phase rotation of baseband I/Q sinusoids



Modulation schemes: Minimum Shift Keying

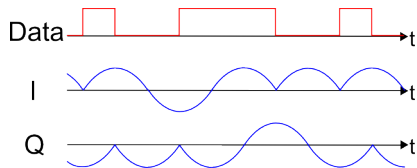
Constellation Diagram: MSK

Continuous Phase Frequency Shift Keying (CPFSK, 1 bit/s/Hz)



Non-coherent demodulation

Data information in phase rotation of baseband I/Q sinusoids



Current 60 GHz transceivers

- Two examples with separate Tx and Rx

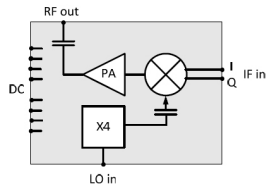
Gotmic TX/RXQ060A01

- 8 GHz IF bandwidth
- IQ-modulator

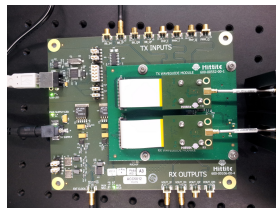
Hittite HMC6000/6001

- 1.8 GHz IF bandwidth
- IQ-modulator with MSK
- AM-, FM-detector

- Commercial chips are not designed for harsh detector environment
- Using full bandwidth, we could use a simple modulation scheme



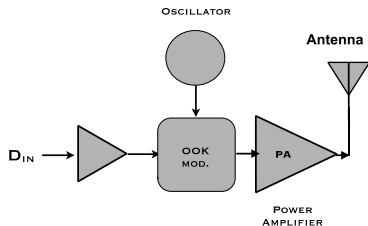
Schematic of Gotmic TXQ060A01



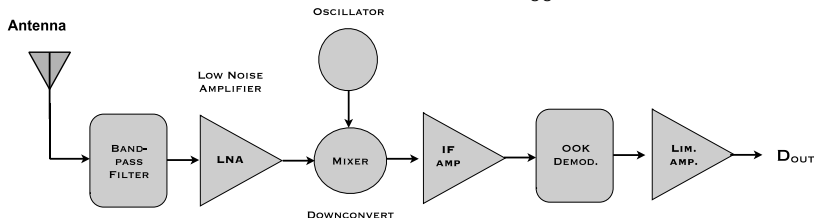
Hittite Transceiver Evaluation Kit

Heidelberg 60 GHz transceiver

- Under development by Hans Kristian Soltveit
- Radiation hard transceiver
- 130 nm SiGe BiCMOS 8HP technology
- Aims at 4.5 Gb/s using 9 GHz bandwidth with OOK
- Power consumption $\approx 0.25 - 0.5 W$

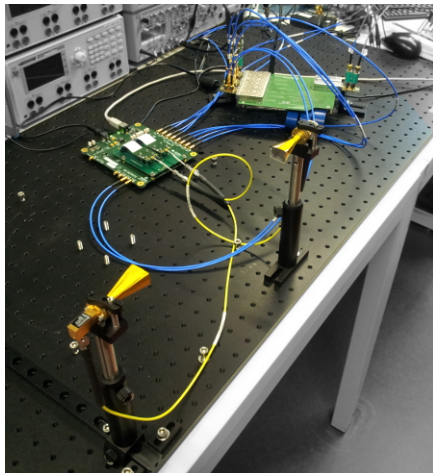


H K Soltveit et al 2012 JINST 7 C12016
 doi:10.1088/1748-0221/7/12/C12016
 Multi-gigabit wireless data transfer at 60 GHz



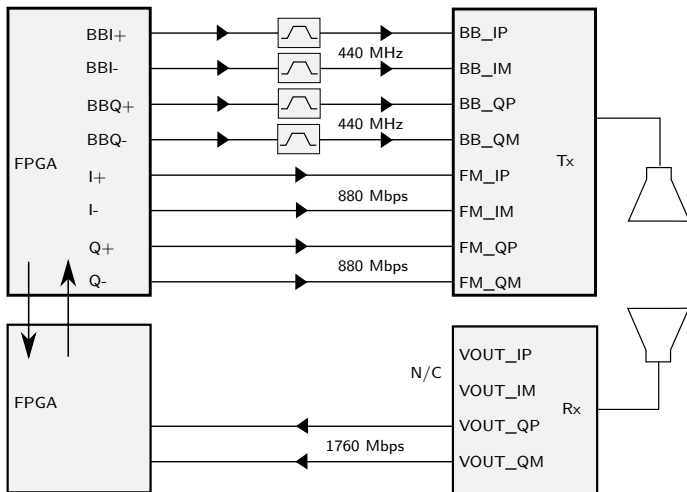
Test bench: bit error rate test using MSK

- For efficient testing: operating setup required
- Test bench built using commercial transceiver by Hittite
- Tx: MSK-modulation
- Rx: internal FM-demodulator
- Implemented with Stratix V GS FPGA

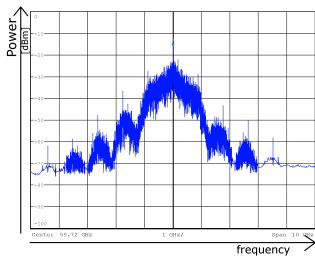


Radio link with Kapton/aluminium horn antennas over a distance of 22 cm

Setup of wireless data transmission

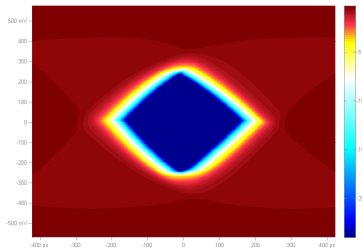
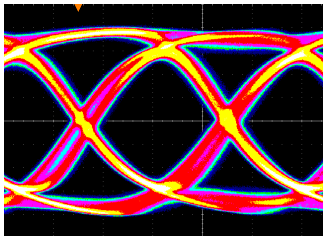


Results with Hittite transceiver

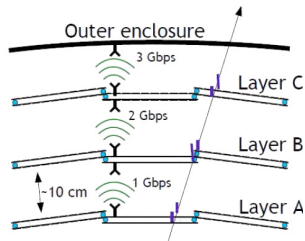


Measured bit error rate

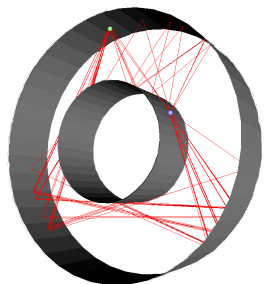
at 1.76 Gbps: $BER < 4.0 \times 10^{-15}$
fast and stable data transmission



Crosstalk problem in detectors



- Possible radial readout direction
- What are the challenges?



Simulations with ray tracer

- Assumption: signal cannot penetrate tracker layer
- Reflections introduce crosstalk
- Absorbing reflections would be a solution

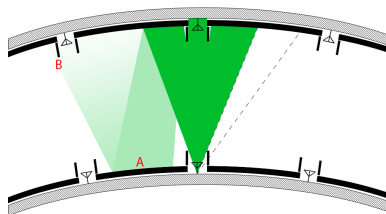
Bachelor thesis by Thomas Hugle, Heidelberg 2013

Approaches to reduce crosstalk

Two types of crosstalk: direct and induced by reflections

How to control crosstalk

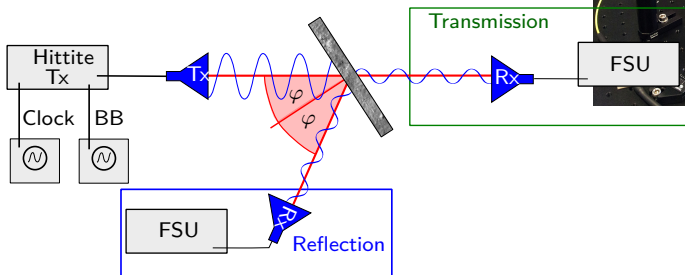
- Antennas with high directivity
- Polarisation of antennas
- Channelling of frequency band
- Absorbing material to reduce reflections



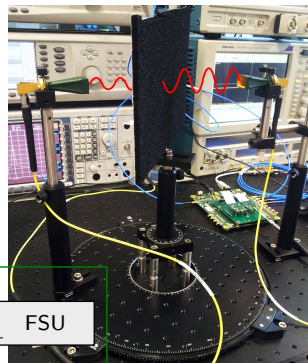
- Low density materials required (multiple scattering)
- Aluminized Kapton foil for antennas
- Graphite foam as absorber (density: $\rho \approx 50 - 74 \text{ kg/m}^3$)

Material studies: setup

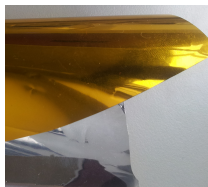
- Characterize reflection and transmission of
 - Common detector materials to estimate crosstalk signal power
 - Graphite foam as a candidate to be used as absorber



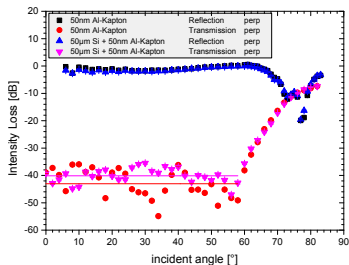
FSU = spectrum analyzer by R&S



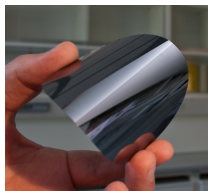
Material properties at 60 GHz



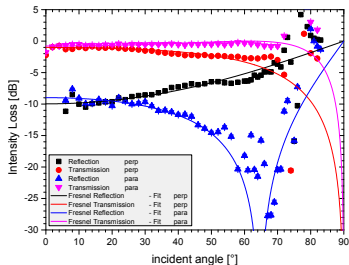
25 μm Kapton foil with 50 nm aluminum



50 nm Al
perfect reflector



50 μm silicon



50 μm Si (unprocessed)
almost transparent
transmitted intensity
reduced by ≈ 1 dB
at small angles

Properties of graphite foam

- Under test: four different samples of foam
- Measure reflectivity R and transmittance T
→ absorbance $A = 1 - T - R$
- Pores of foam $\ll \lambda \approx 5$ mm
→ assumption: flat surfaces

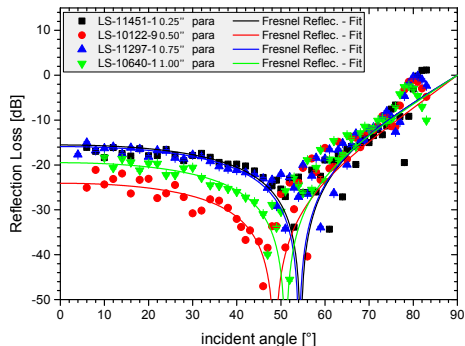
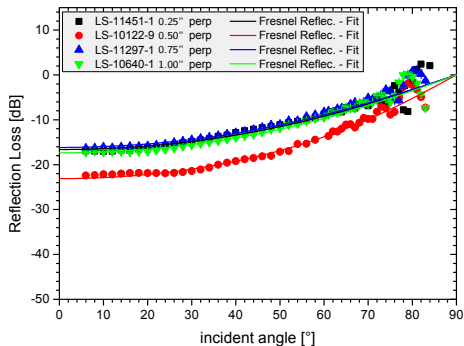


graphite foam sample

Foam	Thickness [mm]	ρ [mg/cm ³]	X_0 [m]	1 cm/ X_0 [%]
LS-11451-1	6.35	73.8 ± 0.7	6.06 ± 0.06	0.165 ± 0.002
LS-10122-9	12.70	54.0 ± 1.0	8.29 ± 0.15	0.121 ± 0.002
LS-11297-1	19.05	58.8 ± 0.5	7.61 ± 0.06	0.138 ± 0.001
LS-10640-1	25.40	50.7 ± 0.5	8.83 ± 0.08	0.113 ± 0.001

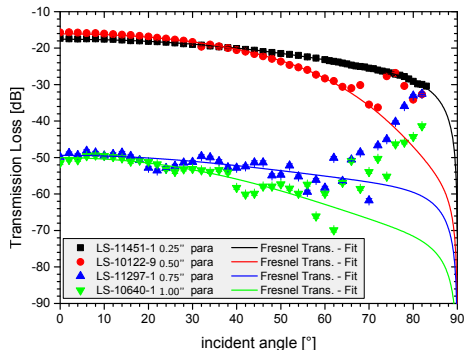
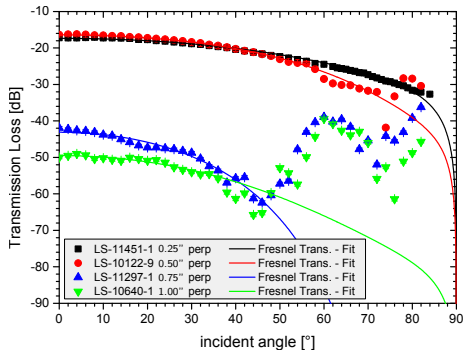
Reflectivity of graphite foam

- Reflections are reduced by more than 10 dB up to large incident angles
- These are single reflection measurements
- Crosstalk by multiple reflections can be highly suppressed



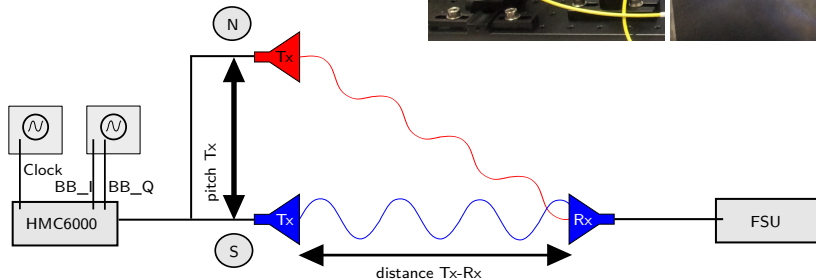
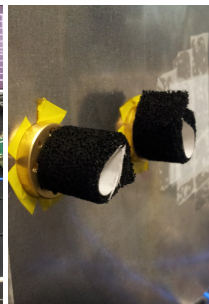
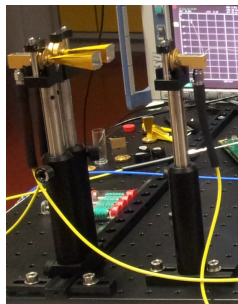
Transmittance of graphite foam

- Insertion loss strongly dependent on thickness
- Samples show large differences
- High absorbance: $A \approx 13 - 27$ dB/cm



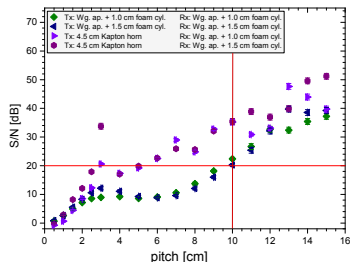
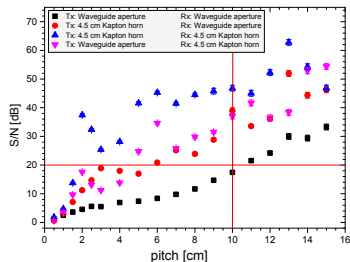
Quantify direct crosstalk - without reflections

- Compare direct transmission to displaced transmitter
- Under test
 - Kapton horn antennas
 - Graphite foam cylinders
- Distance 10 cm



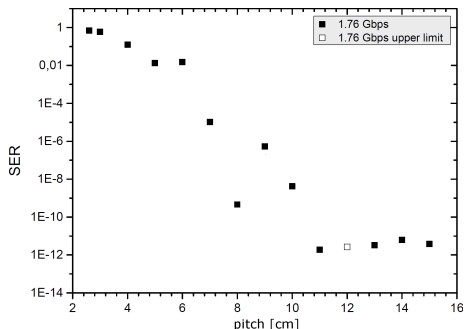
Direct crosstalk suppression

- Requirement: $S/N > 20$ dB
- Even with low directivity sufficient S/N possible at a pitch of 10 cm
- At larger pitches directive antennas are more efficient
- Graphite foam “antennas” reduce direct crosstalk if links are close



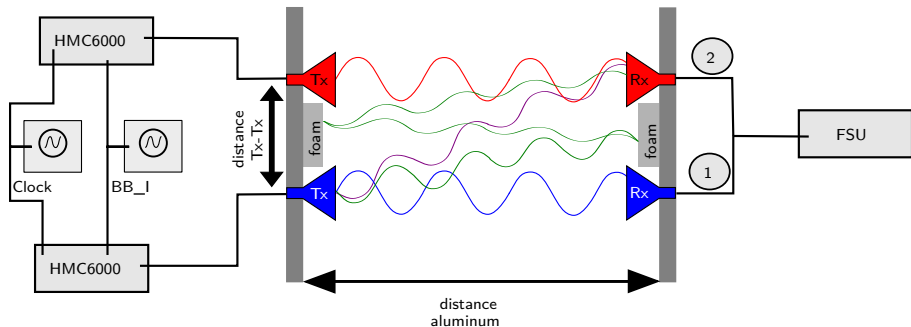
Direct crosstalk suppression: Bit error rates

- 1 cm long hollow graphite foam cylinder shielding
- Links over 10 cm distance
- $\text{BER} < 10^{-12}$ for pitch between links > 10 cm



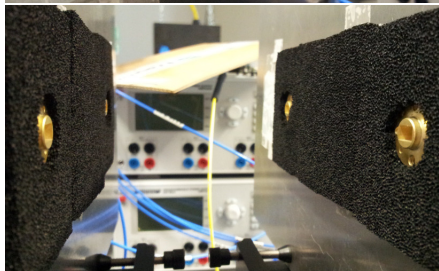
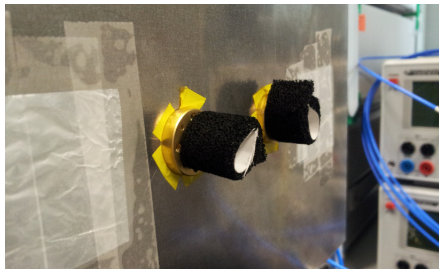
Multi-path crosstalk setup

- Study reflections in multi-path setup
- Measured signal to noise in RF
- Different transmitter pitches: 5 cm, 10 cm and 15 cm

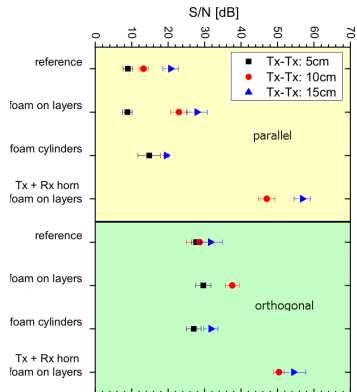
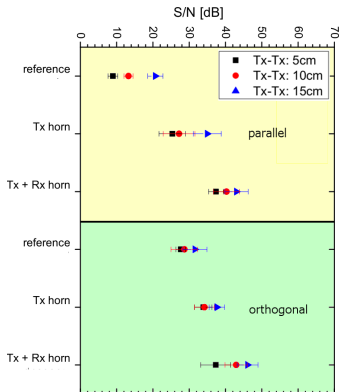


Multi-path crosstalk setup in the lab

- Under test
 - Kapton horn antennas
 - Graphite foam cylinders
 - Graphite foam coating of Al board
- Tested with parallel and orthogonally polarised transmitters



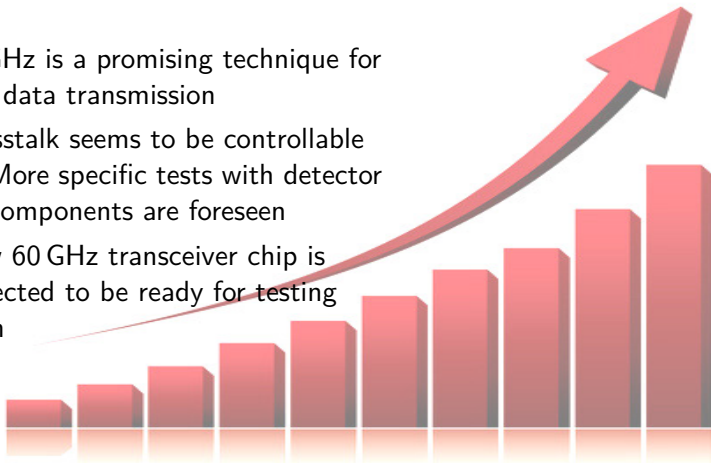
Multi-path signal to noise ratios



- Polarisation has huge effect for close-by links
- High directivity also decreases multi-path crosstalk strongly
- Foam on layers can reduce noise additionally

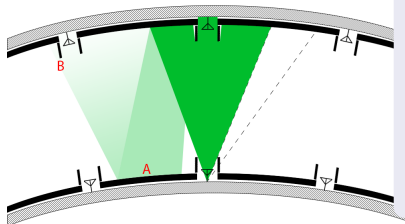
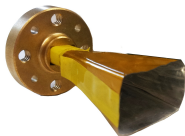
Conclusion

- 60 GHz is a promising technique for fast data transmission
- Crosstalk seems to be controllable
→ More specific tests with detector components are foreseen
- New 60 GHz transceiver chip is expected to be ready for testing soon



Backup

Material budget calculation



Material budget per layer (Tx + Rx)

Assumption: 10 cm between links

- Transceiver (Si):
 $2 \times (5 \times 5) \text{ mm}^2 \times 100 \mu\text{m} \rightarrow 5.35 \times 10^{-4} \% X_0$
 - Foam:
 $2 \times (5 \times 5) \text{ cm}^2 \times 1 \text{ cm} \rightarrow 7.15 \times 10^{-2} \% X_0$
 - Kapton:
 $2 \times 15.2 \text{ cm}^2 \times 35 \mu\text{m} \rightarrow 2.68 \times 10^{-6} \% X_0$
 - Aluminium:
 $2 \times 15.2 \text{ cm}^2 \times 25 \mu\text{m} \rightarrow 4.44 \times 10^{-6} \% X_0$
- Total: 0.072 % X_0 per layer

Wireless signal transmission

What limits the bandwidth?

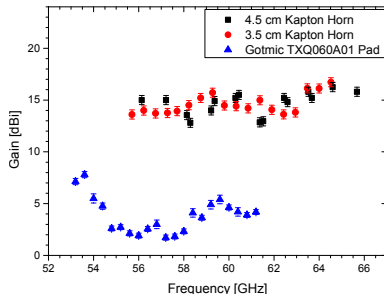
- Antennas: gain and efficiency
 - depend on frequency
 - determined by dimensions
- Free space path loss depends on frequency

$$FSPL = \left(\frac{4\pi df}{c} \right)^2$$

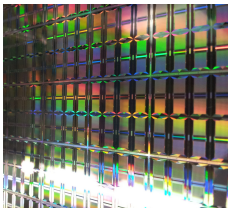
- Passband defines bandwidth $<$ carrier frequency



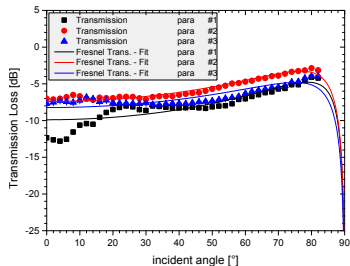
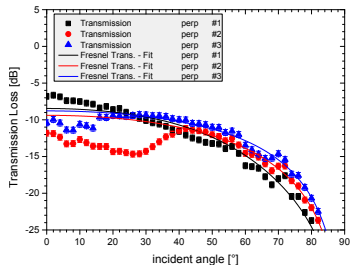
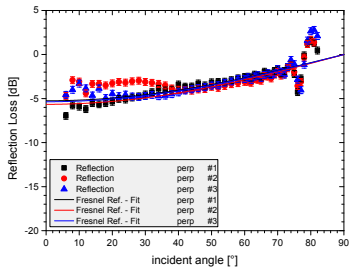
Pad antenna for 60 GHz
size about 1 mm^2



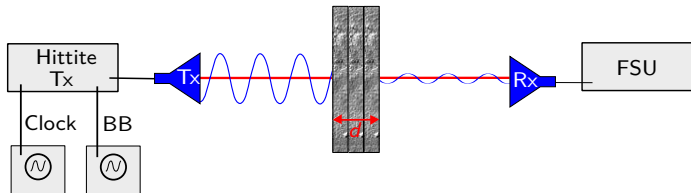
Detector materials' properties at 60 GHz



730 μm silicon wafer with test structures

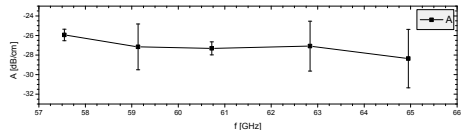
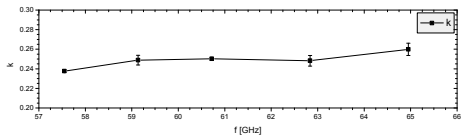
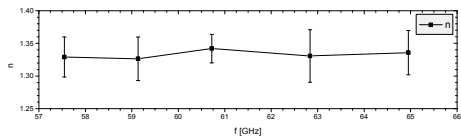


Frequency dependent insertion loss measurement setup

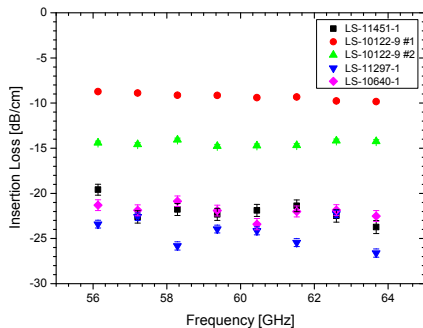
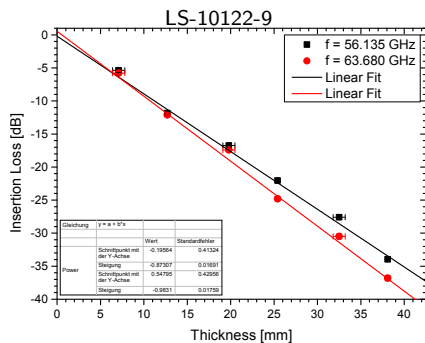


Frequency dependence of graphite foam's properties

- Properties of graphite foam stable in the whole 60 GHz band



Transmission loss of graphite foam at 0° incident angle

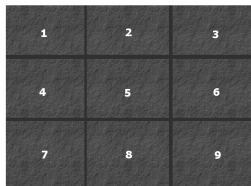


Results

- 1 Differences between foams
- 2 Large variations in material LS-10122-9
→ in other foams as well?

Inhomogeneity of graphite foam

- $61 \times 61 \text{ cm}^2$ foam pads segmented into 9 pieces
- Compared transmission loss of each piece at $f = 56.134 \text{ GHz}$
- Minimum and maximum peak values



$T :=$ Transmission loss

Material	T_{min} [dB/cm]	T_{max} [dB/cm]
LS-11451-1	-19.6 ± 0.6	-29.2 ± 1.0
LS-10122-9	-8.7 ± 0.2	-17.5 ± 1.0
LS-11297-1	-23.4 ± 0.4	-28.7 ± 1.0
LS-10640-1	-20.0 ± 1.0	-24.7 ± 1.0

\Rightarrow 5 - 10 dB variations in all foams!

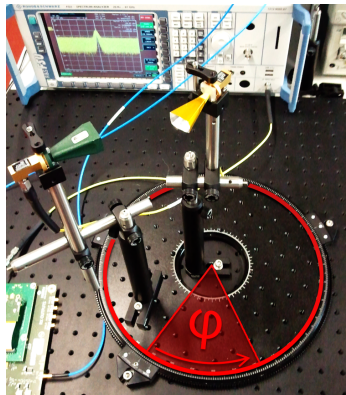
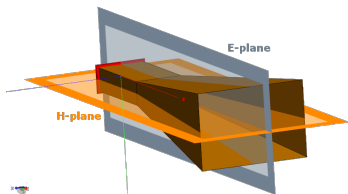
Characterisation of antennas

Goal

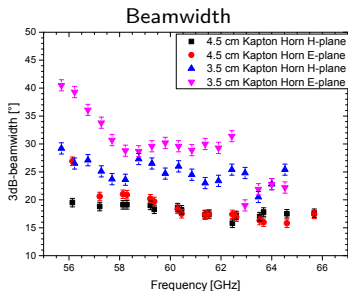
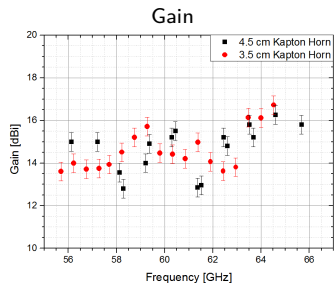
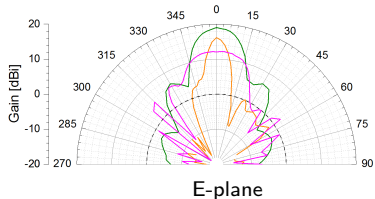
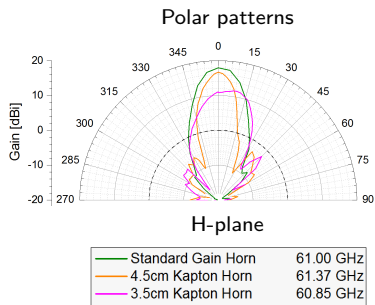
Reduce crosstalk: high directivity, little material

→ Horn antennas made from aluminized Kapton foil

- Characteristics:
 - Polar pattern
 - Gain [dBi]
 - 3dB-beamwidth [$^{\circ}$]
 - Polarisation



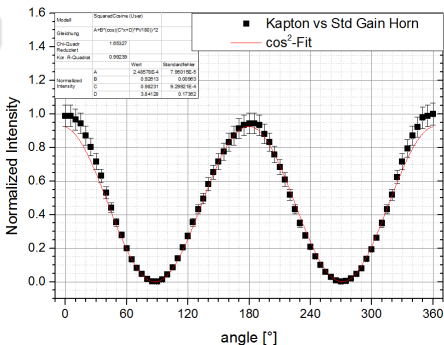
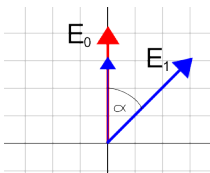
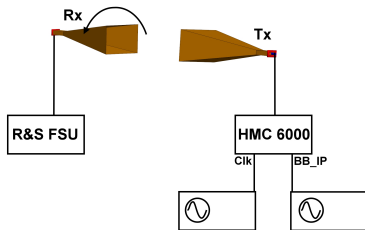
Comparison of directivity



Linear polarised waves: Measurements

Use polarisation to place transceivers in closer proximity to each other

Rotating Rx antenna

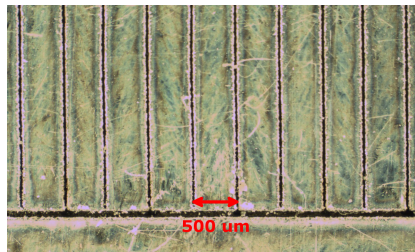
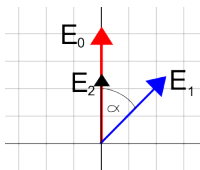
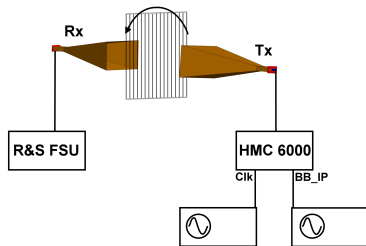


- Suppression of orthogonal polarisation component ≈ 30 dB by rotating Kapton antenna

Linear polarised waves: Measurements

Use polarisation to place transceivers in closer proximity to each other

Rotating polarising filter

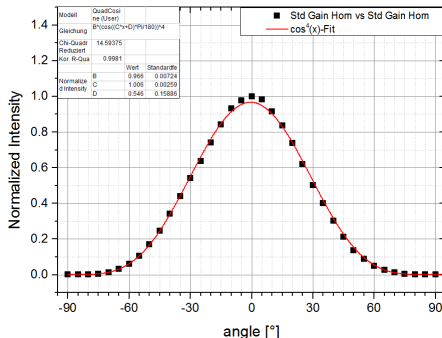
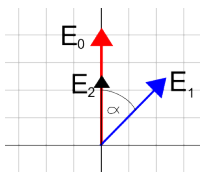
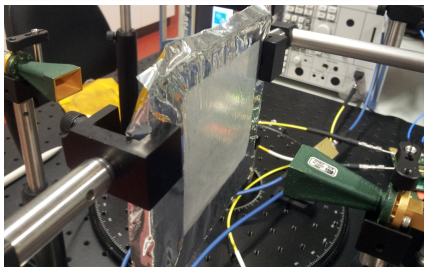


- Wire grid polariser:
500 μm pitch, 450 μm width
Al strips on kapton foil
- Strips have been cut out using laser

Linear polarised waves: Measurements

Use polarisation to place transceivers in closer proximity to each other

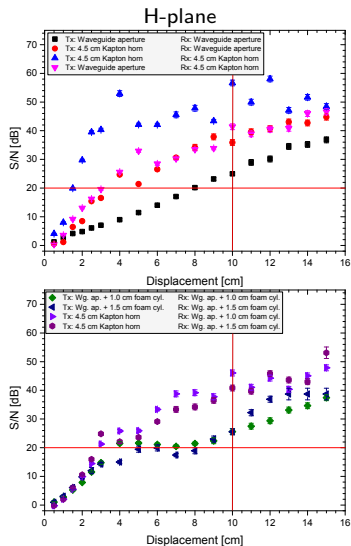
Rotating polarising filter



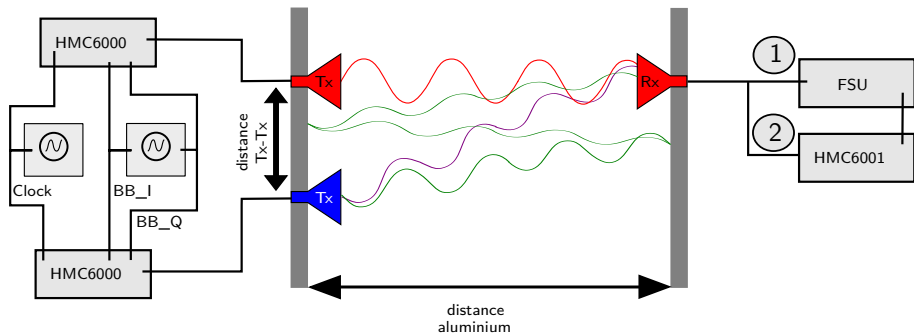
- Wire grid polariser:
500 μm pitch, 450 μm width
Al strips on kapton foil
- Stips have been cut out using laser

Direct crosstalk suppression

- Signal is more focused in H-plane
- Even with low directivity we can reach a sufficient signal to noise
- Graphite foam reduces direct crosstalk in close vicinity of the link
- At larger displacement directive antennas are more efficient

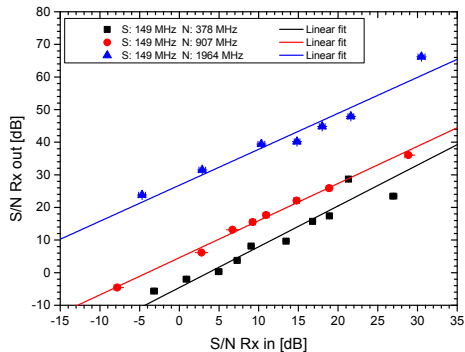


Compare S/N Receiver input and output

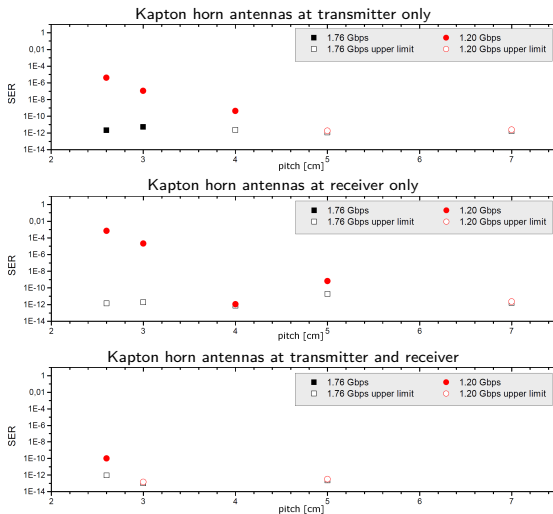


Compare S/N Receiver input and output

- Linear dependence
- Bandwidth affects the offset between S/N output to input

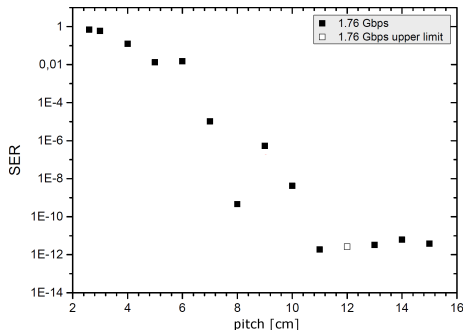


Crosstalk: 2 links with horn antennas



Crosstalk

Cylindrical foam shielding (1 cm)



Channeling : 2 links in immediate vicinity ($pitch < 3$ cm)

