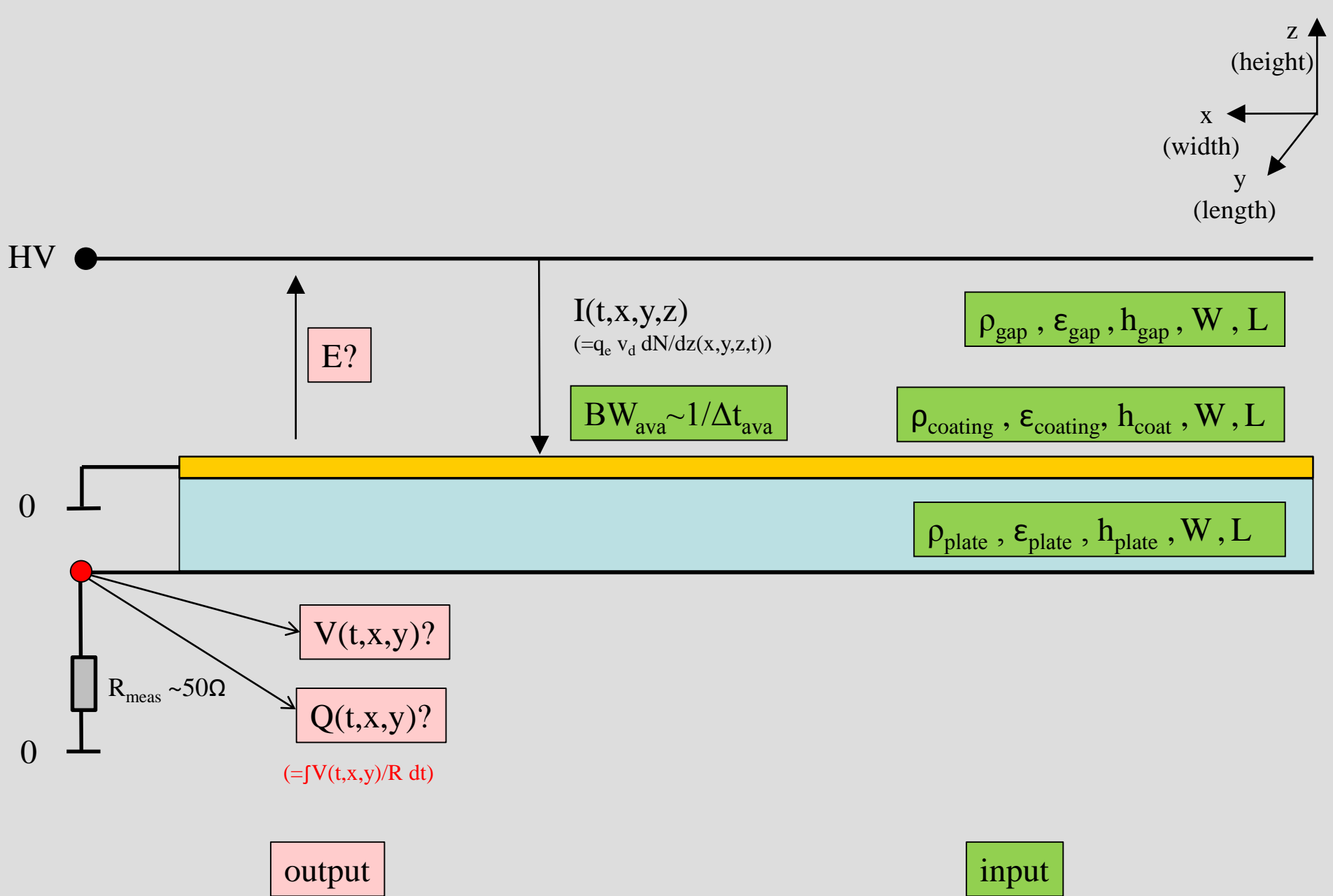


# useful analytical solutions in typical limiting cases

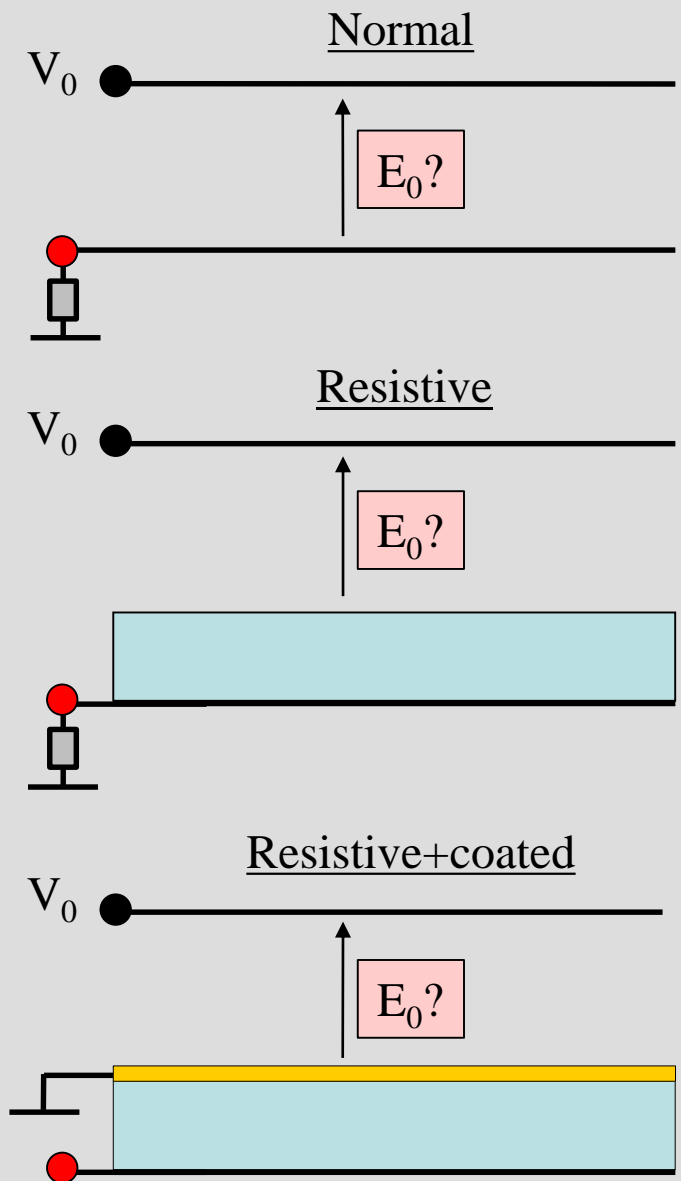
Diego González-Díaz  
(CERN & Uludag University)

# **1. Electric fields**



\*In the following, it is assumed that length and width are enough so that no edge effects are present.

1) Field distribution (low rates)



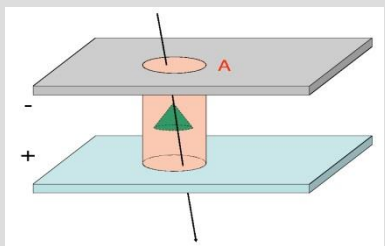
$R_{\text{gap}}$  → Refers to dark current in the gap, whatever the origin, so  $R_{\text{gap}} \sim V_0 / I_{\text{dark}}$ .

t	t=0	t=∞ ( $\gg R_{\text{plate}}(C_{\text{gap}} + C_{\text{plate}})$ )
	$E_0 = \frac{V_0}{h_{\text{gap}}}$	$E_0 = \frac{V_0}{h_{\text{gap}}}$
	$E_0 = \frac{C_{\text{plate}}}{C_{\text{plate}} + C_{\text{gap}}} \frac{V_0}{h_{\text{gap}}}$	$E_0 = \frac{R_{\text{gap}}}{R_{\text{gap}} + R_{\text{plate}}} \frac{V_0}{h_{\text{gap}}} \approx \frac{V_0}{h_{\text{gap}}}$
	$E_0 \approx \frac{C_{\text{plate}}}{C_{\text{plate}} + C_{\text{gap}}} \frac{V_0}{h_{\text{gap}}}$	$E_0 \approx \frac{R_{\text{gap}}}{R_{\text{gap}} + R_{\text{plate}}} \frac{V_0}{h_{\text{gap}}} \approx \frac{V_0}{h_{\text{gap}}}$

# 1) Field distribution (high rates)

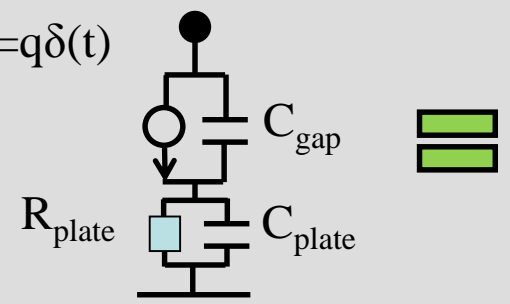
(Montecarlo results) [Gon06]

'cell model' [Abb04]

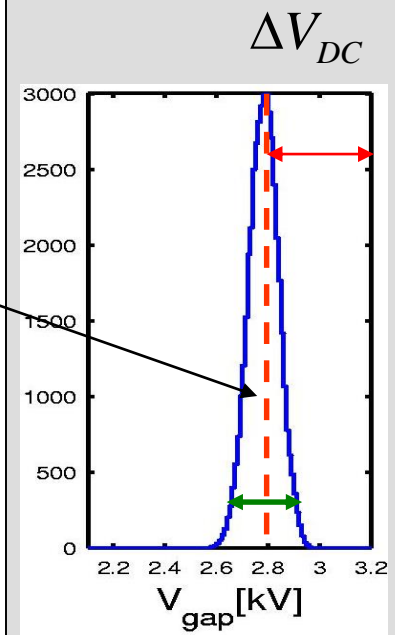
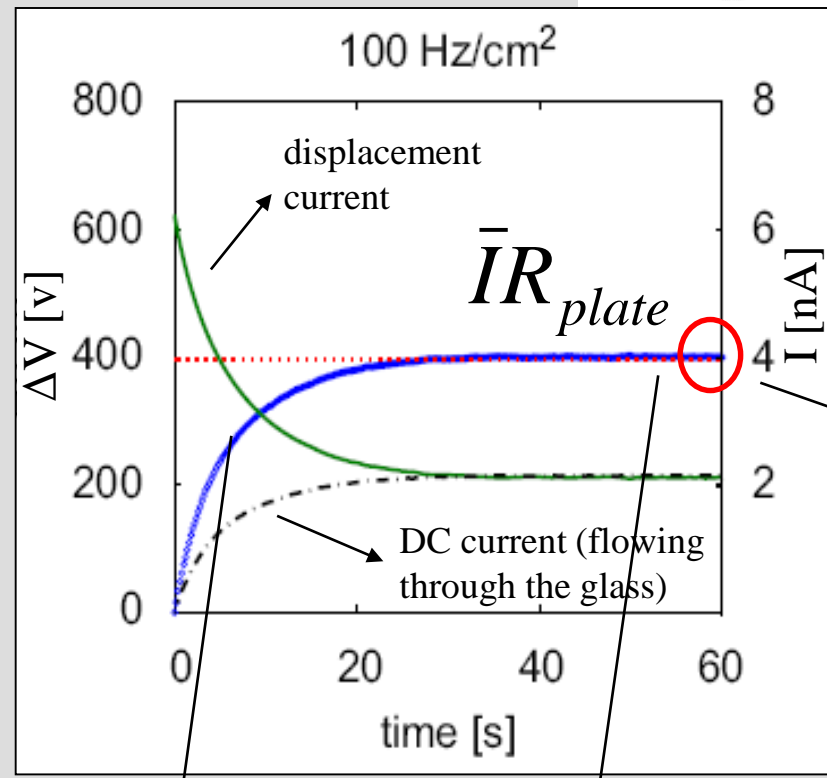
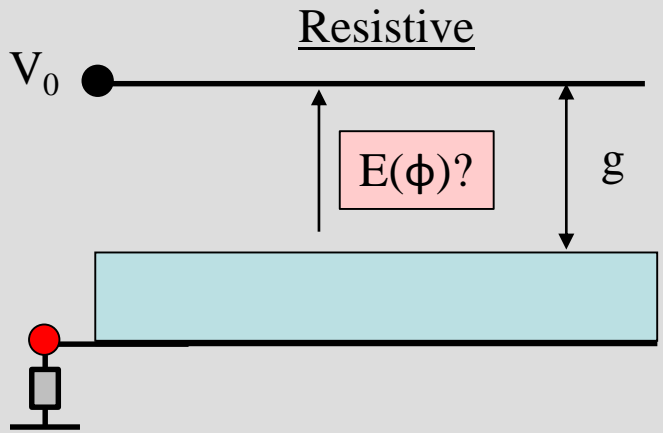
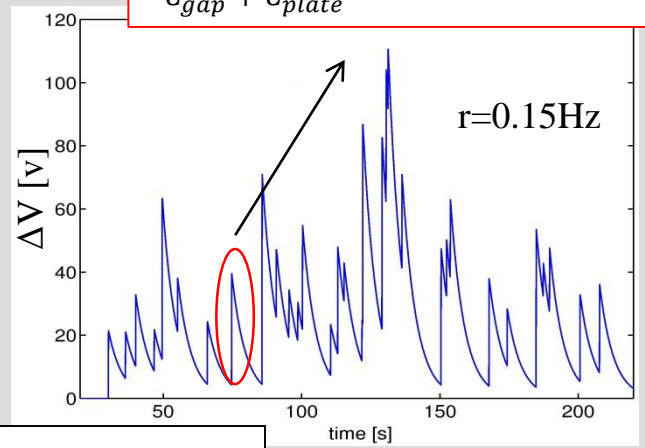


shot-noise circuit model

$$I(t) = q\delta(t)$$



$$\frac{q}{C_{gap} + C_{plate}} e^{-R_{plate}(C_{gap} + C_{plate})t}$$



transient behavior, 'charging-up time'

voltage drop in stationary conditions (DC limit)

voltage fluctuations for each cell

# 1) Field distribution (high rates)

(analytical results) [Gon06]

average field

fluctuations around the average field

equilibration time

the steady-state ('DC') limit must satisfy:

$$\begin{aligned} \bar{E}(\phi) &= \frac{1}{h_{gap}} (V_0 - \bar{I}R) \\ &= \frac{1}{h_{gap}} (V_0 - \bar{q} \phi \rho h_{plate}) \end{aligned}$$

if we further assume that:

$$\bar{q} \sim a(\bar{V} - V_{th})$$

~true for saturated-avalanche RPC



$$\bar{E}(\phi) = \frac{V_{th}}{h_{gap}} + \frac{(V_0 - V_{th})}{h_{gap} (1 + a \phi \rho h_{plate})}$$

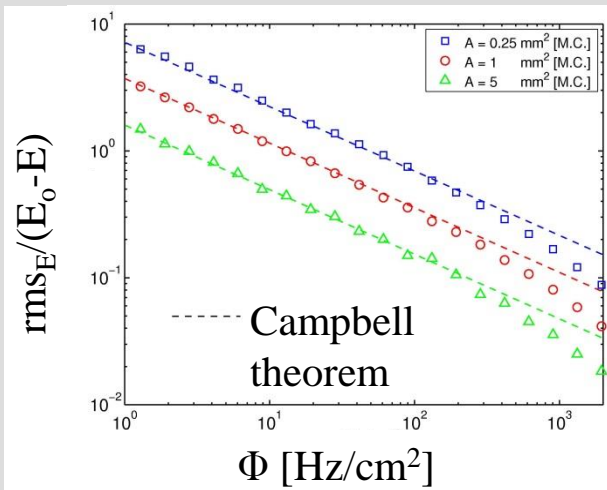
'DC model': detector response depends on average field. Good approximation for saturated-avalanche RPCs.

A solution for exponential multiplication in  $\mu$ -well given in [Mor15], and it also seems to work!. Also at this workshop.

Campbell theorem for shot noise (1910):

$$\frac{rms_E}{E_0 - \bar{E}(\phi)} = \sqrt{\frac{1 + \frac{rms_q^2}{\bar{q}^2}}{2\bar{N}}}$$

$$\bar{N} = A \tau \phi$$

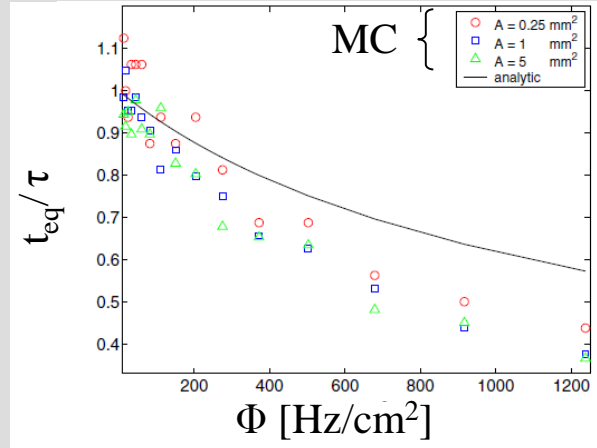


Attempts to generalize: [Lipp06]: avoid circuit, use quasi-static approximation, epsilon f-independent.

simple analytical estimate:

$$t_{eq} \sim \frac{\tau}{a \phi \rho h_{plate}} \ln(1 + a \phi \rho h_{plate})$$

$$\tau = R_{plate} (C_{gap} + C_{plate})$$

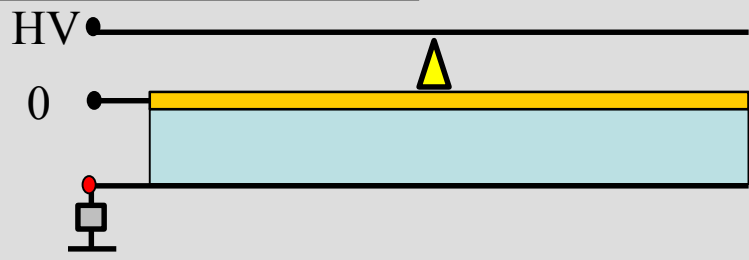


no dependence on the area A influenced by each avalanche!

More sophisticated attempts: [Bil09]: solve electrostatic problem. [Gon09]: use glass response function.

## **2. Induced signals**

## 2) Induced signals

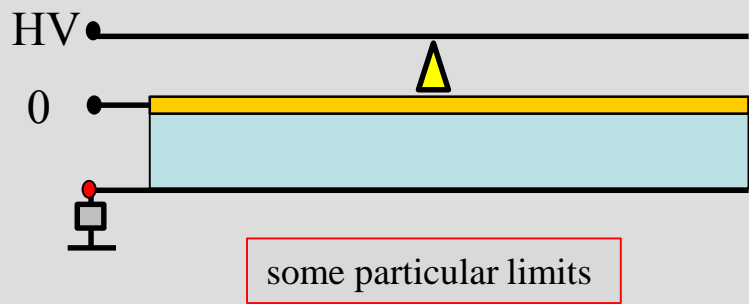


$$\rho_{\text{coating}} = \rho_{\text{plate}}$$

$$\Delta t_{\text{ava}} \ll \rho_{\text{plate}} \epsilon_{\text{plate}}$$

(RPC, well and  $\mu$ -well)

[Rie02a, 04]

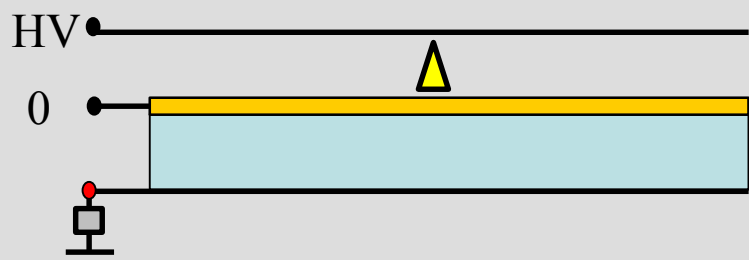


$$\rho_{\text{coating}} \neq \rho_{\text{plate}}$$

$$\Delta t_{\text{ava}} \sim R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$$

$$\Delta t_{\text{ava}} \ll \rho_{\text{plate}} \epsilon_{\text{plate}}$$

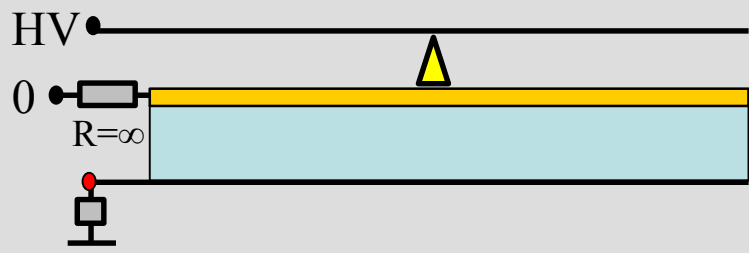
Resistive Micromegas



$$\Delta t_{\text{ava}} \ll R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$$

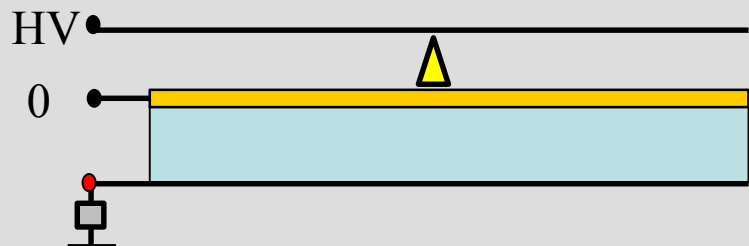
$$\Delta t_{\text{ava}} \ll \rho_{\text{plate}} \epsilon_{\text{plate}}$$

Direct induction



$$\Delta t_{\text{ava}} \gg R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$$

Capacitive coupling



$$\Delta t_{\text{ava}} \gg R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$$

No pick-up

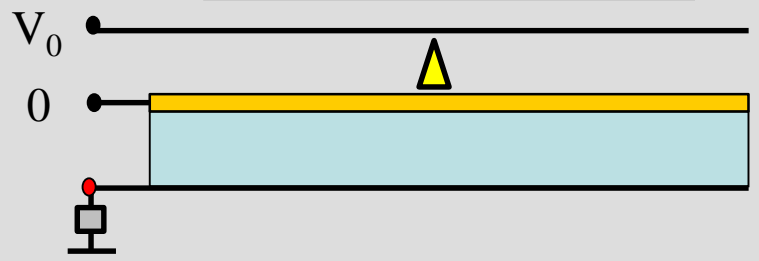
$$R_{\square} = \rho_{\text{coat}} / h_{\text{coat}} L/W$$



## 2) Induced signals. Limiting cases

Direct induction

$$\Delta t_{\text{ava}} \ll R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$$

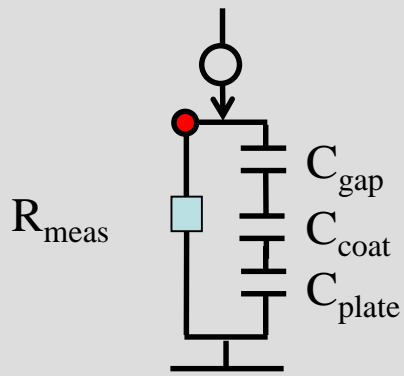
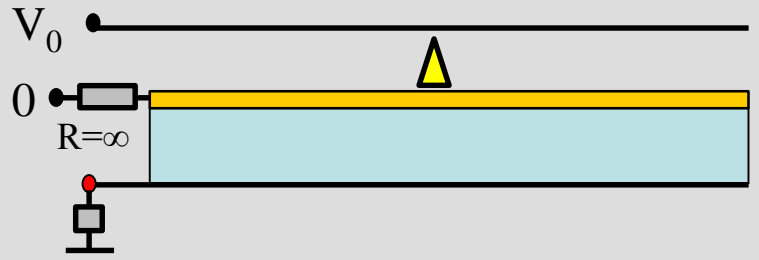


solutions for  $\Delta t_{\text{ava}} \sim R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$  are scarce: [Rie02a, 04, Dix06]

*Direct induction, resistive charge spread and capacitive coupling take place simultaneously. Can be used advantageously to increase position resolution (P. Colas at this workshop).*

Capacitive coupling

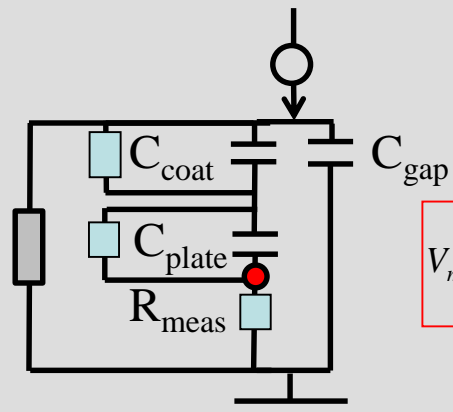
$$\Delta t_{\text{ava}} \gg R_{\square} \epsilon_0 (h_{\text{gap}} + h_{\text{plate}})$$



$$V_{\text{meas}}(t) = \frac{v_{\text{drift}}}{LW\epsilon_0} \int_0^t \exp\left[-\frac{t'-t}{R_{\text{meas}} C'}\right] N(t') q_e dt'$$

$$C' = \left( \frac{1}{C_{\text{gap}}} + \frac{1}{C_{\text{coat}}} + \frac{1}{C_{\text{plate}}} \right)^{-1}$$

position information preserved (e.g, if using strip readout)



$$V_{\text{meas}}(t) \cong \frac{v_{\text{drift}}}{LW\epsilon_0} \int_0^t \exp\left[-\frac{t'-t}{R_{\text{meas}} C_{\text{gap}}}\right] N(t') q_e dt'$$

position information destroyed

No net charge

### **3. Transmitted signals**

### 3) Signal transmission (long structures\*)

$$I_{meas,main}(t) \cong \left[ \frac{T}{2} \right] \frac{1}{h_{gap}} v_d N(t) + reflections$$

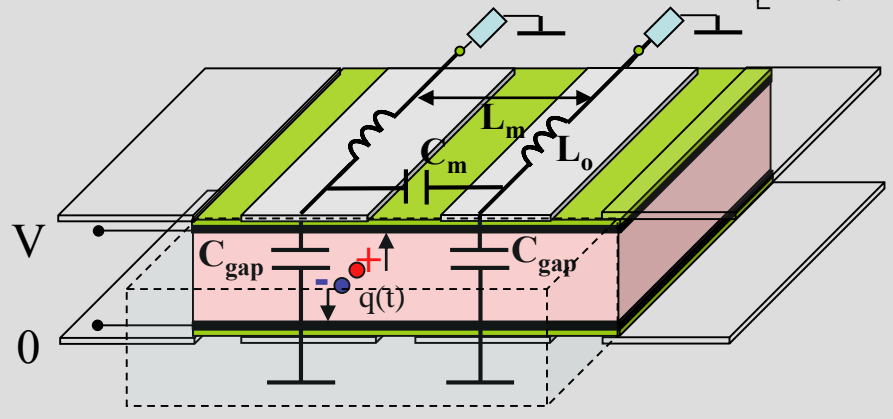
$$\Delta t_{ava} \ll R_{\square} \epsilon_0 h_{gap}$$

\*electrically long

$$\Delta t_{ava} \leq L/c$$

Normal

$$I_{meas.ct}(t) \cong \left[ \frac{T}{2} \frac{R}{Z_c + R} \frac{C_m}{C_{gap}} \right] \frac{1}{h_{gap}} v_d N(t) + reflections$$



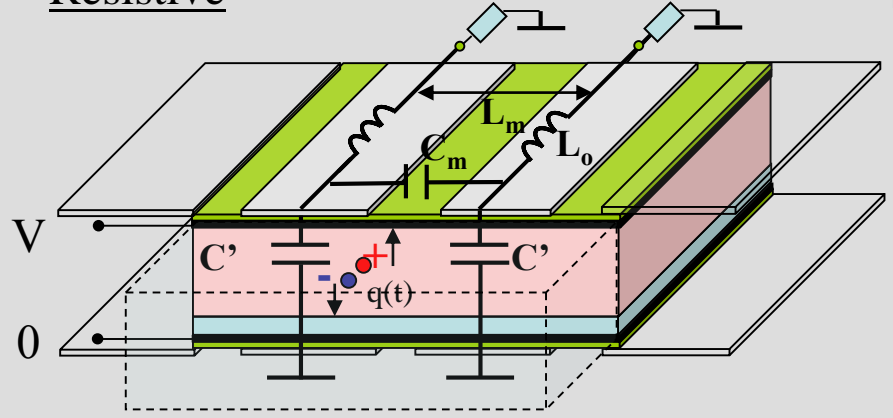
but compensation possible!

$$\frac{C_m}{C' + C_m} = \frac{C_m}{C' + C_m} \Big|_{vacuum} \equiv \frac{L_m}{L_o} = \frac{C_m}{C' + C_m}$$

Resistive

$$I_{means,main}(t) \cong \left[ \frac{T}{2} \frac{R}{Z_c + R} \frac{C_m}{C'} f(N(t)) + \frac{T}{2} \right] \frac{1}{h_{gap}} \frac{C'}{C_{gap}} v_d N(t) + reflections$$

$$I_{meas.ct}(t) \cong \left[ \frac{T}{2} \frac{R}{Z_c + R} \frac{C_m}{C'} + \frac{T}{2} f(N(t)) \right] \frac{1}{h_{gap}} \frac{C'}{C_{gap}} v_d N(t) + reflections$$



dispersive term due to the presence of insulators (modal dispersion).

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A popular way to make use of the  
‘DC model’: RPCs.

$$\sigma_T = \sigma_0 + K_T \bar{q} \phi \rho h_{plate}$$

$$\varepsilon = \varepsilon_0 - K_\varepsilon \bar{q} \phi \rho h_{plate}$$

