

# JUAS RF Course 2016 Tutorials

## Tutorial I

Some physical constants (in SI units, but without the units attached) :

$c = \text{QuantityMagnitude}[\text{UnitConvert}[\text{Quantity}["\text{SpeedOfLight}"]]]$   
299 792 458

$\mu = \text{N}[\text{QuantityMagnitude}[\text{UnitConvert}[\text{Quantity}["\text{VacuumPermeability}"]]]]$   
 $1.25664 \times 10^{-6}$

$\epsilon = \text{N}[\text{QuantityMagnitude}[\text{UnitConvert}[\text{Quantity}["\text{VacuumPermittivity}"]]]]$   
 $8.85419 \times 10^{-12}$

Other resources :

$\eta = \sqrt{\frac{\mu}{\epsilon}}$   
376.73

### ■ A) Design of a "pillbox" cavity

**Problem :** Design a simple cavity of the "Pillbox" type with the following parameters :

**frequency :**

$\lambda = 1.0;$

$f = \frac{c}{\lambda}$

$2.99792 \times 10^8$

$\omega = 2 \pi f$

$1.88365 \times 10^9$

**EngineeringForm[%]**

$1.88365 \times 10^9$

Wall material: copper

$\sigma_{\text{Cu}} = 5.8 \times 10^7$

$5.8 \times 10^7$

Skin depth: copper (see also page 60)

$$\delta = \sqrt{\frac{2}{\omega \sigma_{\text{Cu}} \mu}}$$

$$3.81677 \times 10^{-6}$$

axial length:

$$h = 0.2;$$

## Questions

AI : Find from the analytical formula

### cavity radius a

The fundametal mode is the  $\text{TM}_{010}$  mode, with

$$\mu_r = 1;$$

$$\epsilon_r = 1;$$

$$m = 0;$$

$$n = 1;$$

$$p = 0;$$

$$a = .$$

$$f_{\text{mnp}} = \text{FullSimplify}\left[\text{N}\left[\frac{c}{2 \pi \sqrt{\mu_r \epsilon_r}} \sqrt{\left(\frac{\text{BesselJZero}[m, n]}{a}\right)^2 + \left(\frac{p \pi}{h}\right)^2}, a > 0\right];\right.$$

with the zero of the Bessel function:

$$\text{N}[\text{BesselJZero}[m, n]]$$

$$2.40483$$

see also page 24

$$f_{010} = \%$$

$$1.14743 \times 10^8$$

$$a$$

$$\lambda_0 = \frac{c}{f_{010}}$$

$$2.61274 a$$

see also page 32

$$\text{result} = \text{Solve}[\lambda == \%, a]$$

$$\{\{a \rightarrow 0.38274\}\}$$

$$\text{result}[[1]]$$

$$\{a \rightarrow 0.38274\}$$

```
a /. result[[1]]
```

```
0.38274
```

```
a = % λ
```

```
0.38274
```

see also page 30

cavity quality factor Q

$$Q = \frac{a}{\delta} \left(1 + \frac{a}{h}\right)^{-1}$$

```
34 416.2
```

see also page 32

"geometry factor", also known as "characteristic impedance" R/Q

```
BesselJ[1, N[BesselJZero[0, 1]]]
```

```
0.519147
```

exact result:

```
RoverQ =
```

$$\frac{4 \eta}{\text{BesselJZero}[0, 1]^3 \pi \text{BesselJ}[1, \text{N}[\text{BesselJZero}[0, 1]]]^2} \frac{\text{Sin}\left[\frac{\text{BesselJZero}[0, 1] \frac{h}{a}}{2}\right]^2}{\frac{h}{a}}$$

```
84.6096
```

approximation:

$$\text{RoverQapprox} = 185 \frac{h}{a}$$

```
96.6714
```

see also page 33

Error due to  $\sin(x) \approx x$  approximation in %:

```
100 (1 - RoverQ / RoverQapprox)
```

```
12.4771
```

### Is the cavity completely determined?

With the 3 given parameters the intrinsic cavity is fully determined!

## A2 : Find the equivalent circuit of the intrinsic cavity.

```
rApprox = RoverQapprox Q
```

```
3.32706 × 106
```

```
lApprox =  $\frac{\text{RoverQapprox}}{\omega}$ 
```

```
5.13213 × 10-8
```

```
EngineeringForm[%]
```

```
51.3213 × 10-9
```

$$c_{\text{Approx}} = \frac{1}{R_{\text{over}Q_{\text{approx}}} \omega}$$

$$5.49163 \times 10^{-12}$$

$$r_{\text{Exact}} = R_{\text{over}Q}$$

$$2.91194 \times 10^6$$

$$l_{\text{Exact}} = \frac{R_{\text{over}Q}}{\omega}$$

$$4.49178 \times 10^{-8}$$

$$\text{EngineeringForm}[\%]$$

$$44.9178 \times 10^{-9}$$

$$c_{\text{Exact}} = \frac{1}{R_{\text{over}Q} \omega}$$

$$6.27451 \times 10^{-12}$$

see also page 48-50

A3 : Find the 3-dB bandwidth of the intrinsic cavity.

$$\Delta f = \frac{f}{Q}$$

$$8710.79$$

see also page 56

A4 : Calculate the necessary RF power for a gap voltage of:

$$v = 100 \times 10^3;$$

$$P = \frac{v^2}{2 r_{\text{Exact}}}$$

$$1717.07$$

see also page 50

A5 : The cavity shall be fed by an amplifier designed for a load impedance of

$$z_1 = 50;$$

**Determine:**

**the peak voltage at the cavity input.**

$$v_{\text{Peak}} = \sqrt{2 z_1 P}$$

$$414.375$$

the necessary transformer ratio  $k$  of the input coupler.

$$k = \sqrt{\frac{r_{\text{Exact}}}{Z_1}}$$

241.327

see also page 50

## ■ B) Waves on a transmission line $Z = 50 \Omega$

Problem : Convert the couple (voltage  $V$ , current  $I$ ) into the equivalent couple (forward wave  $a$ , reflected wave  $b$ ) and vice versa using the relations

$$a = \frac{V + IZ}{2} \quad V = a + b$$

$$b = \frac{V - IZ}{2} \quad IZ = a - b$$

## Questions

BI : In a  $50 \Omega$  system, a directional coupler measured the forward and reflected waves  $a$  and  $b$  at a certain plane as  $a = 100 \angle 0^\circ$  and  $b = 60 \angle 45^\circ$

Calculate the corresponding voltage  $V$  and current  $I$ .

$$a = 100;$$

$$b = \text{N}[\text{ExpToTrig}[60 \text{Exp}[I 45 \text{Degree}]]]$$

$$42.4264 + 42.4264 i$$

$$v = a + b$$

$$142.426 + 42.4264 i$$

$$\{\text{Abs}[v], \text{Arg}[v] / \text{Degree}\}$$

$$\{148.611, 16.5879\}$$

$$z = 50;$$

$$i = \frac{a - b}{z}$$

$$1.15147 - 0.848528 i$$

$$\{\text{Abs}[i], \text{Arg}[i] / \text{Degree}\}$$

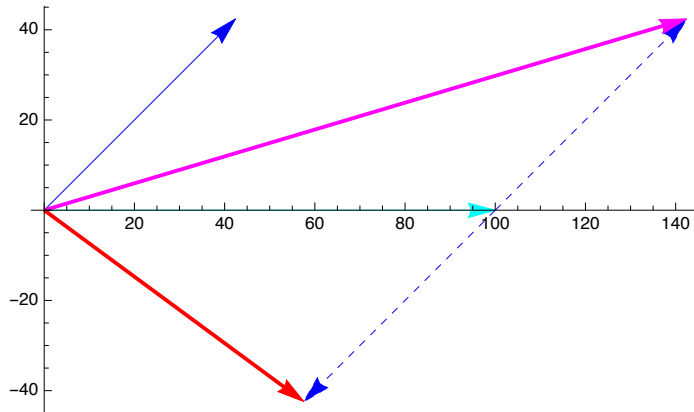
$$\{1.43035, -36.3868\}$$

$$i z$$

$$57.5736 - 42.4264 i$$

```
{Abs[i z], Arg[i z] / Degree}
{71.5173, -36.3868}
```

```
Graphics[
  {{Cyan, Arrow[{{0, 0}, {Re[a], 0}]}}, {Blue, Arrow[{{0, 0}, {Re[b], Im[b]}]}},
  {Blue, Dashed, Arrow[{{Re[a], 0}, {Re[a + b], Im[0 + b]}]}},
  {Magenta, Thick, Arrow[{{0, 0}, {Re[a + b], Im[0 + b]}]}},
  {Blue, Dashed, Arrow[{{Re[a], 0}, {Re[a - b], Im[0 - b]}]}},
  {Red, Thick, Arrow[{{0, 0}, {Re[i z], Im[i z]}]}},
  Axes → True, AspectRatio → Automatic]
```



B2 : At some plane of a  $50 \Omega$  system, a voltage of  $V = 100 \angle 0^\circ$  and a current of  $I = 1.0 \angle -45^\circ$  are measured.

**Calculate the corresponding forward and backward waves a and b.**

```
v = 100;
```

```
i = N[ExpToTrig[1.0 Exp[-I 45 Degree]]]
```

```
0.707107 - 0.707107 i
```

```
a =  $\frac{v + i z}{2}$ 
```

```
67.6777 - 17.6777 i
```

```
{Abs[a], Arg[a] / Degree}
```

```
{69.9483, -14.6388}
```

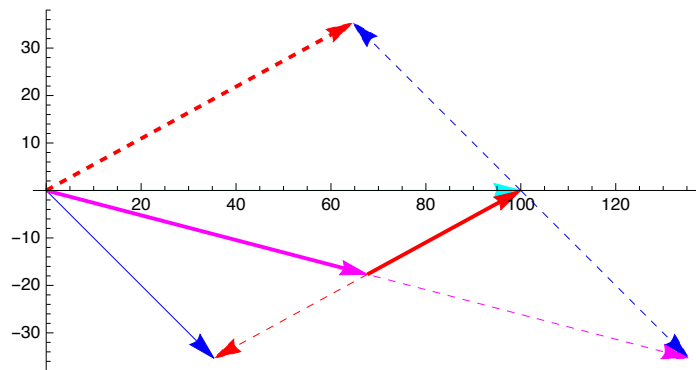
```
b =  $\frac{v - i z}{2}$ 
```

```
32.3223 + 17.6777 i
```

```
{Abs[b], Arg[b] / Degree}
```

```
{36.8406, 28.6751}
```

```
Graphics[{{Cyan, Arrow[{{0, 0}, {Re[v], 0}]}],
  {Blue, Arrow[{{0, 0}, {Re[i z], Im[i z]}]}],
  {Blue, Dashed, Arrow[{{Re[v], 0}, {Re[v] + Re[i z], 0 + Im[i z]}]}],
  {Blue, Dashed, Arrow[{{Re[v], 0}, {Re[v] - Re[i z], 0 - Im[i z]}]}],
  {Magenta, Dashed, Arrow[{{0, 0}, {Re[2 a], Im[2 a]}]}],
  {Magenta, Thick, Arrow[{{0, 0}, {Re[a], Im[a]}]}],
  {Red, Thick, Arrow[{{Re[a], Im[a]}, {Re[a + b], Im[a + b]}]}],
  {Red, Dashed, Arrow[{{Re[a], Im[a]}, {Re[a - b], Im[a - b]}]}],
  {Red, Dashed, Thick, Arrow[{{0, 0}, {Re[2 b], Im[2 b]}]}]},
  Axes → True, AspectRatio → Automatic]
```



## Tutorial 2

**Problem #1 to be treated as “CONTINUOUS ASSESMENT”**

- I) The following data has been determined on a cavity:

**Inductance:**

$$L = 15.915 \times 10^{-9};$$

**Capacitance:**

$$C = 1.5915 \times 10^{-12};$$

**3-dB bandwidth:**

$$\Delta f = 50 \times 10^3;$$

---

## Questions

**Determine**

- the resonance frequency

$$f_{\text{res}} = \frac{1}{2\pi} \frac{1}{\sqrt{LC}}$$

$$1.00003 \times 10^9$$

see also page 49

- the characteristic impedance  $R/Q$

$$R/Q = \sqrt{\frac{1}{C}}$$

100.

see also page 50

- the quality factor  $Q$

$$Q = \frac{f_{res}}{\Delta f}$$

20 000.6

see also page 56

- the time constant  $\tau$

$$\tau = \frac{Q}{\pi f_{res}}$$

$6.3662 \times 10^{-6}$

see also page 63

- the peak induced voltage immediately after the passage of a short particle bunch with charge

$$q = 15.916 \times 10^{-9};$$

$$V_{step} = \text{Abs} \left[ 0 - \frac{q}{C} \right]$$

10 000.6

see also page 65

- the remaining cavity voltage  $10 \mu s$  after the bunch passage

$$t = 10 \times 10^{-6};$$

$$V_{end} = V_{step} e^{-\frac{t}{\tau}}$$

2078.93

see also page 65

■ 2) A cavity is constructed from material with:

**thermal expansion coefficient:**

$$\Delta l / l = 20 \times 10^{-6};$$



**thermal resistivity coefficient:**

$$\frac{\Delta\rho}{\rho} = 4 \times 10^{-3};$$

**At room temperature the cavity resonates at a frequency**

$$f_1 = 100 \times 10^6;$$

**with a bandwidth of**

$$\Delta f_1 = 100 \times 10^3;$$

**Under RF power, the cavity temperature is increased by**

$$\Delta t = 100;$$

**(subscripts 2 apply)**

## Questions

**Determine**

- the ratio  $\lambda_2/\lambda_1$

The wavelength scales proportional with the cavity dimension  $d$ , as  $\lambda$  is inverse proportional to  $f$

$$\frac{\lambda_2}{\lambda_1} = \frac{d_2}{d_1} = \frac{d_1(1+\Delta l/\Delta T)}{d_1} = 1 + \frac{\Delta l}{l} \Delta T$$

$$\lambda_2/\lambda_1 = N[1 + \Delta l/\Delta T]$$

$$1.002$$

see also page 69

- the ratio  $L_2/L_1$

The characteristic impedance  $R/Q$  stays constant, therefore:

$$\frac{L_2}{L_1} = \frac{R/Q}{\omega_2} \frac{\omega_1}{R/Q} = \frac{\omega_1}{\omega_2} = \frac{\lambda_2}{\lambda_1}$$

$$L_2/L_1 = \lambda_2/\lambda_1$$

$$1.002$$

see also pages 50 and 69

- the ratio  $C_2/C_1$

The characteristic impedance  $R/Q$  stays constant, therefore:

$$\frac{C_2}{C_1} = \frac{1}{R/Q \omega_2} \frac{R/Q \omega_1}{1} = \frac{\omega_1}{\omega_2} = \frac{\lambda_2}{\lambda_1}$$

$$C_2/C_1 = \lambda_2/\lambda_1$$

$$1.002$$

see also pages 50 and 69

- the ratio  $Q_2/Q_1$

The quality factor  $Q \propto \frac{\lambda_0}{\delta}$ , and the skin depth follows  $\delta \propto \sqrt{\frac{\rho}{f}}$ , there-

fore:

$$\frac{Q_2}{Q_1} = \frac{\text{const} \lambda_2}{\delta_2} \frac{\delta_1}{\text{const} \lambda_1} = \frac{\lambda_2}{\lambda_1} \frac{\delta_1}{\delta_2} = \frac{\lambda_2}{\lambda_1} \sqrt{\frac{\rho_1 f_2}{\rho_2 f_1}} = \frac{\lambda_2}{\lambda_1} \sqrt{\frac{\rho_1 \lambda_1}{\rho_2 \lambda_2}} = \sqrt{\frac{\rho_1 \lambda_2}{\rho_2 \lambda_1}}$$

$$\rho_1 \text{over} \rho_2 = \mathbf{N}[(1 + \Delta \rho \text{over} \rho \Delta t)]$$

1.4

$$\mathbf{q_2 \text{over} q_1} = \sqrt{\rho_1 \text{over} \rho_2 \lambda_2 \text{over} \lambda_1}$$

1.1844

see also pages 34 and 69

- the resonance frequency  $f_2$

$$\frac{f_2}{f_1} = \frac{\lambda_1}{\lambda_2}$$

$$\mathbf{f_2} = \frac{\mathbf{f_1}}{\lambda_2 \text{over} \lambda_1}$$

$9.98004 \times 10^7$

**EngineeringForm[%]**

$99.8004 \times 10^6$

- the bandwidth  $\Delta f_2$

$$\Delta f_1 = \frac{f_1}{Q_1}, \Delta f_2 = \frac{f_2}{Q_2} \implies \frac{\Delta f_2}{\Delta f_1} = \frac{f_2}{Q_2} \frac{Q_1}{f_1}$$

$$\Delta \mathbf{f_2} = \frac{\mathbf{f_2}}{\mathbf{f_1}} \frac{1}{\mathbf{q_2 \text{over} q_1}} \Delta \mathbf{f_1}$$

84 262.5

**EngineeringForm[%]**

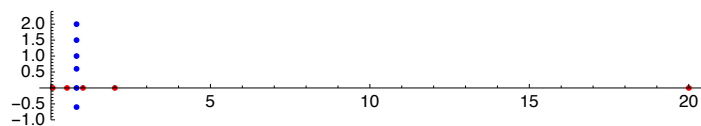
$84.2625 \times 10^3$

### ■ 3) Plot the following impedances:

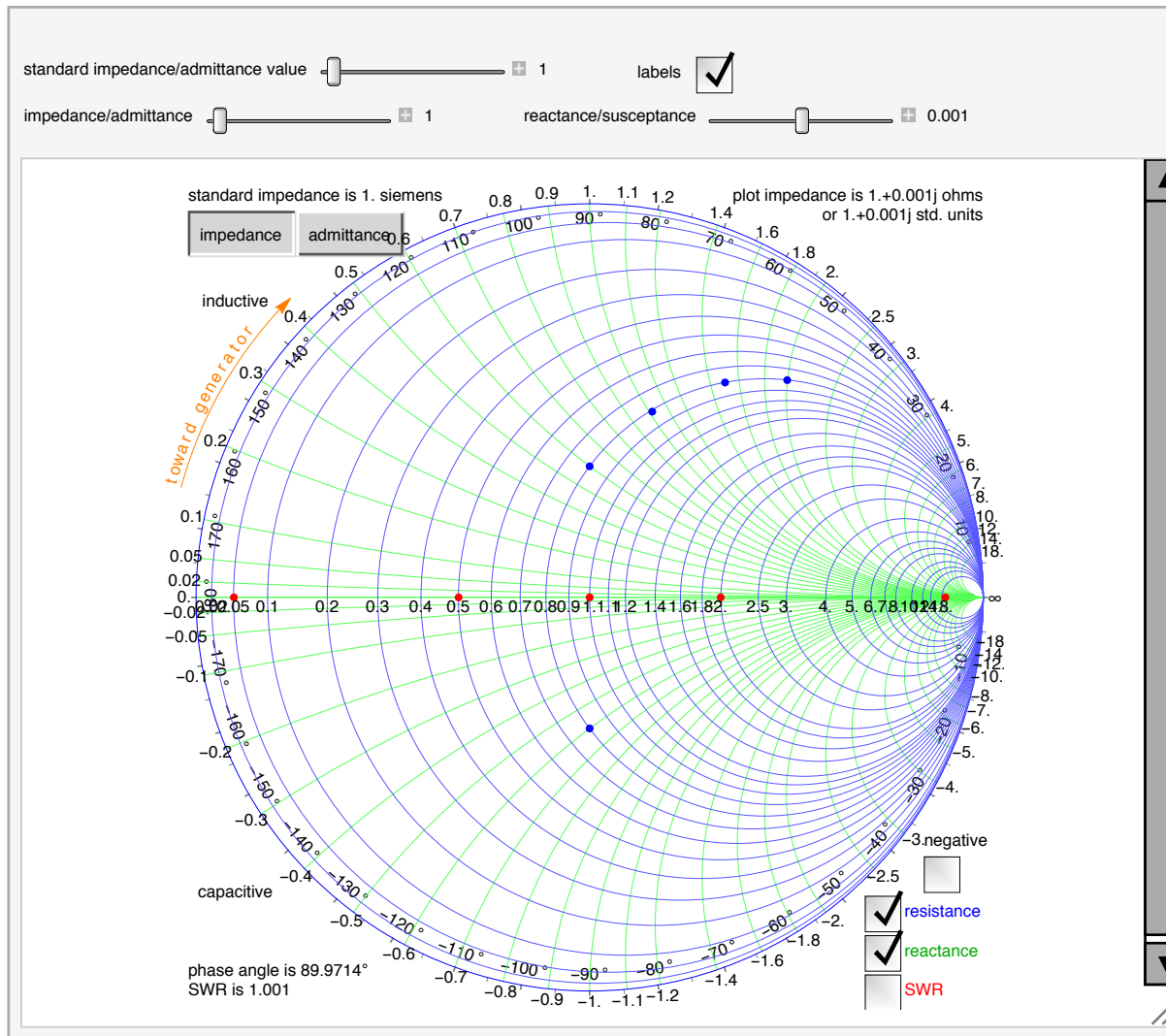
## Questions

- in “normal” cartesian coordinates

```
Graphics[{{Red, Point[{0.05, 0}], Point[{0.5, 0}], Point[{1, 0}], Point[{2, 0}],
  Point[{20, 0]}], {Blue, Point[{0.8, 0}], Point[{0.8, 0.6}],
  Point[{0.8, 1}], Point[{0.8, 1.5}], Point[{0.8, 2}], Point[{0.8, -0.6}]}}],
  Axes -> True, AspectRatio -> Automatic]
```



- as reflection factors in the SMITH chart



## Tutorial 3

### ■ I) Cavity optimization (counted for the “continuous assessment”)

The following parameters of a 100-MHz cavity have been evaluated by a cavity design program as a function of the gap width  $g$ :

**R/Q (characteristic impedance),**  
**Q (quality factor)**

$$f = 100 \times 10^6;$$

The cavity is connected to an amplifier delivering a power of 1 kW

$$p = 1 \times 10^3;$$

The cavity beam has a relative velocity of

$$\beta = 0.15;$$

## Questions

Calculate for each gap width:

- shunt impedance R

$$g = \{0.1, 0.2, 0.3\};$$

$$R_{\text{over}Q} = \{100, 150, 200\};$$

$$q = \{5000, 7000, 9000\};$$

$$r = R_{\text{over}Q} q$$

$$\{500\,000, 1\,050\,000, 1\,800\,000\}$$

- intrinsic cavity voltage  $V_{\text{cav}}$  for 100 kW power

$$p = 100 \times 10^3;$$

$$V_{\text{cav}} = N[\sqrt{2 r p}]$$

$$\{316\,228., 458\,258., 600\,000.\}$$

see also page 50

- phase angle  $\Theta$  of the passage through the gap

Velocity  $v$  of the beam through the cavity gap  $g$ :

$$v = \beta c_0 = \frac{g}{t}$$

Time  $t$  through the gap:

$$t = \frac{g}{\beta c_0} = \frac{g}{\beta f \lambda}$$

The total phase angle accounts for  $\Theta = 2\pi$ ,

the phase  $\Theta$  is given by multiplying both sides by  $2\pi f$  (remember:  $f t = 1$ ):

$$2\pi f t = 2\pi \frac{g f}{\beta f \lambda} \implies 2\pi = \Theta = 2\pi \frac{g}{\beta \lambda} = \frac{2\pi f g}{\beta c_0}$$

$$\Theta = \frac{2\pi f g}{\beta c}$$

$$\{1.39723, 2.79446, 4.19169\}$$

- transit time factor T

The transient time factor T is related to the phase angle  $\Theta$ :

$$\frac{g \omega}{2\beta c_0} = \frac{2\pi f g}{2\beta c_0} = \frac{\Theta}{2}$$

$$T = \frac{\sin[\theta / 2]}{(\theta / 2)}$$

{0.920618, 0.704948, 0.412864}

see also page 79

- beam voltage  $V_{\text{beam}}$  maximally seen by the beam taking the transit time factor  $T$  into account

$$V_{\text{beam}} = V_{\text{cav}} T$$

{291 125., 323 048., 247 719.}

Which of the three versions gives the highest beam voltage?

Version 2 with a gap of  $g = 200$  mm gives the highest beam voltage  $V_{\text{beam}} = 323$  kV

Supposing a field enhancement factor of  $\sim 1.5$  in the gap region, evaluate the danger of voltage breakdown for each design using the Kilpatrick limit as a yardstick.

$$1.5 V_{\text{cav}}$$

{474 342., 687 386., 900 000.}

$V_{\text{Kilpatrick}} (@ 100 \text{ MHz}) = \{\sim 1.15 \text{ MV}, \sim 2 \text{ MV}, \sim 3.3 \text{ MV}\}$   
 For all gap configurations there is no danger of a voltage breakdown, the cavity voltage is always well below the Kilpatrick limit.

see also page 94

## ■ 2) A higher order-mode (HOM) in a cavity

**A RF cavity has an unwanted high-order mode (HOM) at**

$$f_{\text{HOM}} = 600 \times 10^6;$$

**with a shunt impedance of**

$$r_{\text{HOM}} = 6 \times 10^6;$$

**a 3-dB bandwidth of**

$$\Delta f_{\text{HOM}} = 15 \times 10^3;$$

**and a transit time factor of**

$$T = 1;$$

**The beam consists of very short bunches, following each other at interval of**

$$t_{\text{bunch}} = 20 \times 10^{-6};$$

**The circulating current is**

$$i_{\text{beam}} = 0.1;$$

(Remember:  $I = \text{charge per time}$ )

## Questions

Calculate:

-  $Q$ ,  $R/Q$ , and  $C$  at the HOM frequency

$$Q_{\text{HOM}} = \frac{f_{\text{HOM}}}{\Delta f_{\text{HOM}}}$$

40 000

see also page 56

$$R_{\text{over}Q} = \frac{r_{\text{HOM}}}{Q_{\text{HOM}}}$$

150

$$C_{\text{HOM}} = N \left[ \frac{1}{2 \pi f_{\text{HOM}} R_{\text{over}Q}} \right]$$

$1.76839 \times 10^{-12}$

see also page 50

- HOM voltage induced by a single bunch

Assuming a single bunch in the ring :

$$q_{\text{bunch}} = i_{\text{beam}} t_{\text{bunch}}$$

$2. \times 10^{-6}$

$$\Delta V = \frac{q_{\text{bunch}}}{C_{\text{HOM}}}$$

$1.13097 \times 10^6$

see also page 64

- The time constant  $\tau$  of the cavity

$$\tau = 2 r_{\text{HOM}} C_{\text{HOM}}$$

0.0000212207

$$\text{EngineeringForm}[\%]$$

$21.2207 \times 10^{-6}$

see also page 64

- The HOM voltage at the arrival of the next bunch

$$V_{\text{end}} e^{-t/\tau} = V_{\text{step}} \implies V_{\text{next}} = \Delta V e^{-t/\tau}$$

$$V_{\text{next}} = \Delta V e^{-t_{\text{bunch}}/\tau}$$

440 696.

- The total HOM voltage in steady state, after the passage of an infinite number of bunches, supposing that the HOM resonance lies at an exact multiple of the beam revolution frequency

$$V_{\text{end}} = \frac{q_{\text{bunch}}}{C_{\text{HOM}} (1 - e^{-t_{\text{bunch}}/\tau})}$$

$1.85303 \times 10^6$

### ■ 3) Lossless low-pass filter

A lossless low-pass filter is inserted between 50 Ohm load and 50 Ohm generator. The measured input impedance is shown below on a very rudimentary SMITH chart (only circles of constant magnitude of the reflection factor are shown). The arrows indicate the direction of increasing frequency.

## Questions

**Determine:**

- the relative power transmission at point  $P_1$   
(normalized to the power at  $P_0$ , 100%)

At point  $P_1$  the magnitude of the reflection coefficient reads :

$$\text{abs}\Gamma = 0.5;$$

In a lossless filter terminated equally at input and output, the output power equals the input power :

$$P_{\text{out}} = P_{\text{in}} = |a|^2 + |b|^2 = |a|^2 (1 - |\Gamma|^2) = P_0 (1 - |\Gamma|^2)$$

$$\frac{P_{\text{out}}}{P_0} = 1 - \text{abs}\Gamma^2$$

0.75

$$10 \text{ Log}_{10} [\%]$$

-1.24939

see also page 176

- the relative voltage transmission at point  $P_1$   
(normalized to the power at  $P_0$ , 100%)

$$V \propto \sqrt{P} \Rightarrow \frac{V_{\text{out}}}{V_0} = \sqrt{\frac{P_{\text{out}}}{P_0}}$$

```
vOutoverv0 =  $\sqrt{\text{pOutoverp0}}$   
0.866025
```

```
20 Log10 [%]  
-1.24939
```