



CMOS detectors for ILC: radiation hardness

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 CIEMAT

Avenida Complutense 40 Madrid 28040 España

Summary

- Hybrid vs CMOS detectors:
What's He Got That I Ain't Got?
- Depleted CMOS and strip readout
- Radiation hardness (beyond ILC needs) of depleted CMOS
- Conclusions

Hybrid detectors

Advantages:

Both parts (sensor and RO chip) can be **optimized separately**

Sensor: radiation hardness

RO chip: allows for **complex signal processing:**

Digest & process high rates (\sim Mhz/mm²)

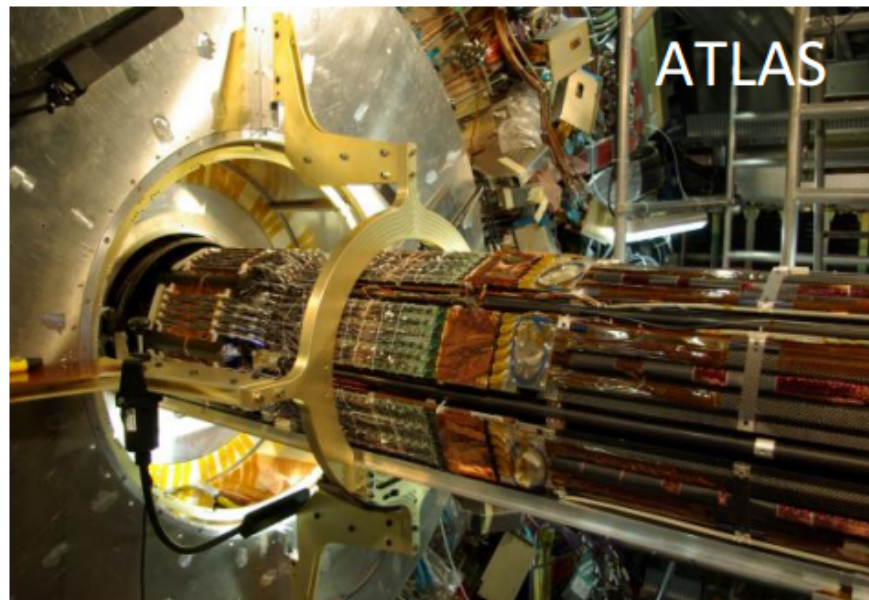
Zero suppression

Storage of hits during decision time (L1 latency)

Disadvantages:

1) Large **material budget** ($\sim 3\%$ X_0 in ATLAS& CMS) due to:
sensor, RO chip, flex capton, support, cooling, services.

2) **Module** production: bump bonding and flip chipping is complex and expensive



all based on
"Hybrid Pixels"

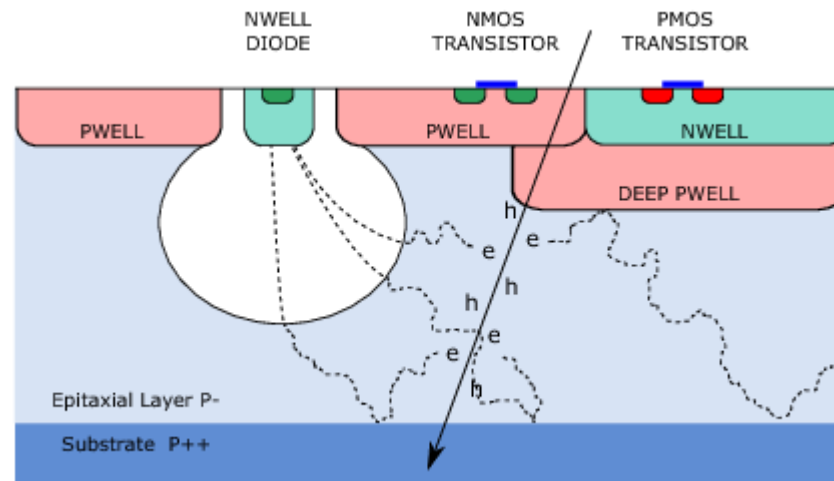
CMOS

Advantages

- 1) **Low cost**, as long as **standard commercial** process not modified
- 2) Monolithic: avoids hybridization, mechanically simpler, large **area (stitching)**
- 3) Smaller pixel size achievable (not limited by BB size)
- 4) They can be thinned down to 50 μm

Disadvantages (of MAPs, triple well MAPS, INMAPs...)

- 1) Industry does not care about particle detection (except photons)
- 2) Limited **radiation hardness** if charge is collected by **diffusion** (they use cheap low resistive substrates)
- 3) Normally “simpler” RO (MAPs do not allow PMOS transistors): signal is **multiplexed**



CMOS for ILC

- CMOS detectors are **no strangers** to ILD:
 - Present in **VTX** detector
 - Present in 2 first **endcaps** of forward tracking
 - They are not the only detectors considered: **FPCCD** and **DEPFETs** are other candidates
- **CMOS** and in general, undepleted MAPs, are **enough** for **ILC** needs:
 - Bunch structure (1 ms long bunch train at 5 Hz) is “slower” than charge collection by diffusion ($\sim\mu\text{s}$)
 - Very modest radiation environment (compared to LHC): **1kGy and $10^{11} n_{\text{eq}}/\text{cm}^2/\text{yea}$**
 - Sensors with similar needs already used in running experiments. Example:

STAR experiment (@RHIC)

AMS 350 nm, pitch 21 μm , 400 Ωcm , T=35 $^{\circ}\text{C}$, 200 μs ,

$3\cdot 10^{12} n_{\text{eq}}/\text{cm}^2$ and 150 kRad

ALICE ITS

Tower Jazz 0.18 μm , 20-40 μm epi layer, >1 $\text{k}\Omega\text{cm}$, **700 kRad/ $10^{13} n_{\text{eq}}/\text{cm}^2$**

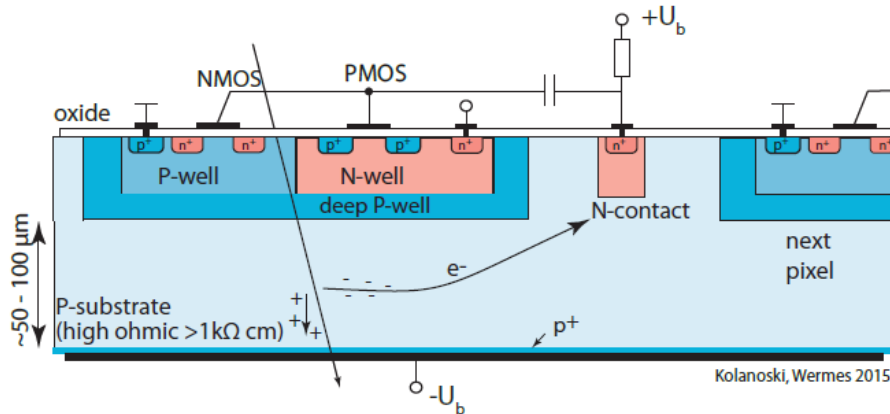
- These devices **collect charge by diffusion**.
- LHC upgrade** is working in a **new** (in HEP) family of **CMOS** devices that LC community could benefit from ...

(Partially) Depleted CMOS

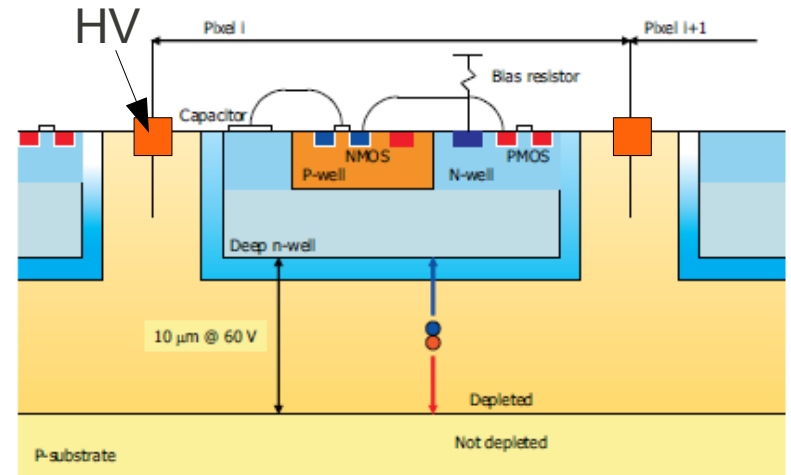
- **Radiation Hardness:** Drift as charge collection mechanism → deplete bulk
- **Cost:** Stick to commercial CMOS technology

$$d \sim \sqrt{\rho \cdot V}$$

HR-CMOS



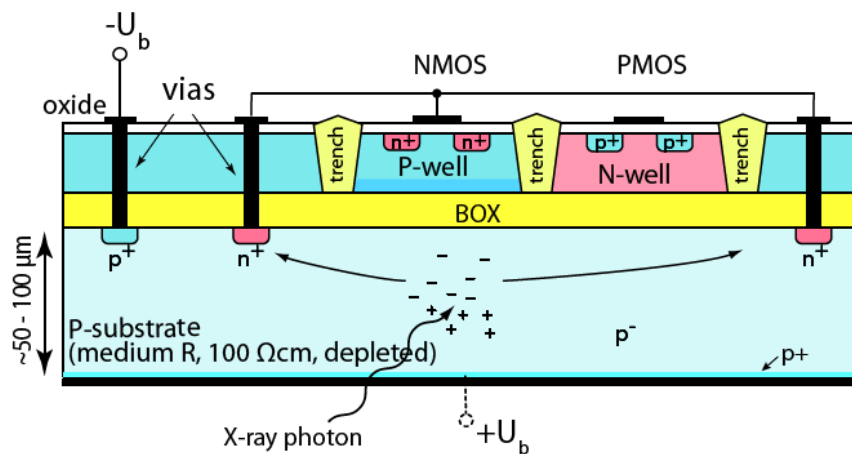
Kolanoski, Wermes 2015



HV-CMOS

Note: a bit “oldish” classification. Now measuring what is the optimal choice for $\rho \cdot V$

CMOS on SOI



ATLAS HV/HR-collaboration

HVCMOS: [2015 JINST 10 C05021]

HRCMOS: [JINST 10 (2015) 02, P02013]

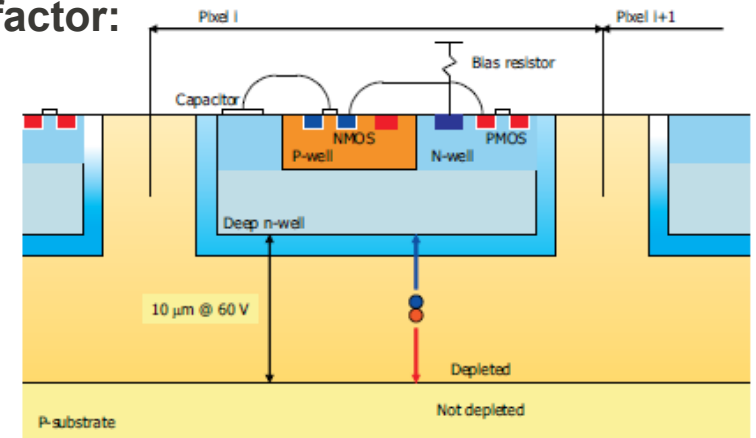
SOI: [NIM A796 (2015) 8]

HVCMOS

- IFCA (within RD50 collaboration) involved in radiation hardness studies of these devices
- Uses **commercial** high-voltage ($\leq 120\text{V}$) CMOS technology (used in industry as HV switches). This HV is applied to a **low resistivity substrate** ($\rho \sim 10 \Omega \cdot \text{cm}$)
 - Expected **10 μm depletion at 100 V** \rightarrow Charge can be collected **by drift**
 - Expected **900 e-h pairs** \rightarrow built-in **preamp** is needed
- To avoid damage to transistors both, **NMOS and PMOS** are “**embedded**” in a **Deep N Well (DNW)** . Any complex signal processing can be implemented inside. The DNW works both as a substrate for transistors and as the signal collection region. **Nearly 100% fill factor:**

– Charge carriers do not have to travel long before being collected \rightarrow **reduced trapping** impact.
We need to quantify this radiation hardness

– Capacitance due to big collecting region increases noise of CSA



- Two development strategies:

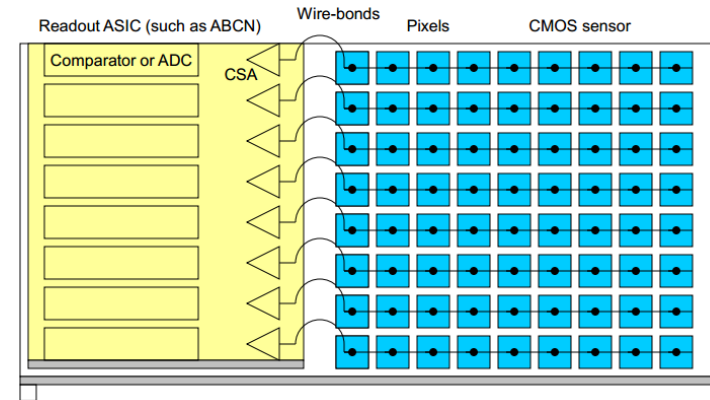
1) **Conservative:** drop in replacement of current sensors. Build preamplifier in sensor chip and capacitively couple to RO chip. CCPD (Capacitively Coupled Pixel Detector) concept

2) **Liberal:** Go fully monolithic and totally replace RO chip

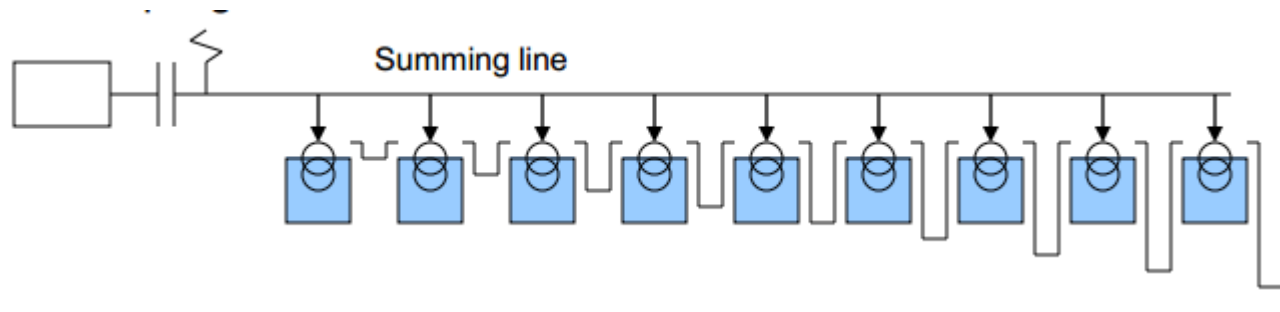


HVCMOS strip readout (pixelated strips)

- A standard **strip** sensor can be **replaced** with an HVCMOS detector segmented into **long pixels**, e.g., $40\ \mu\text{m} \times 800\ \mu\text{m}$ (ATLAS case) .
- Detector can be connected to **existing readout chip** (one-to-one strip sensor replacement)



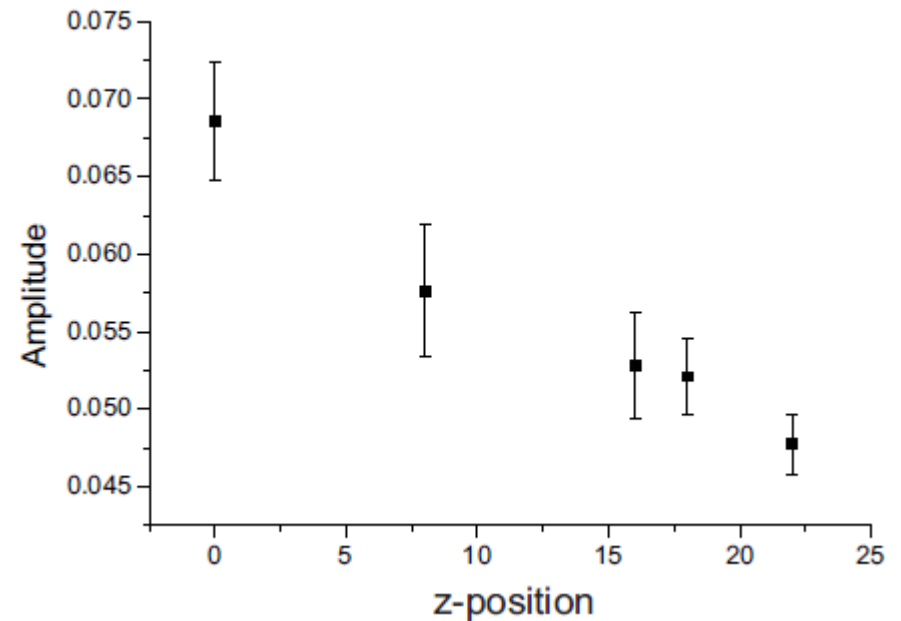
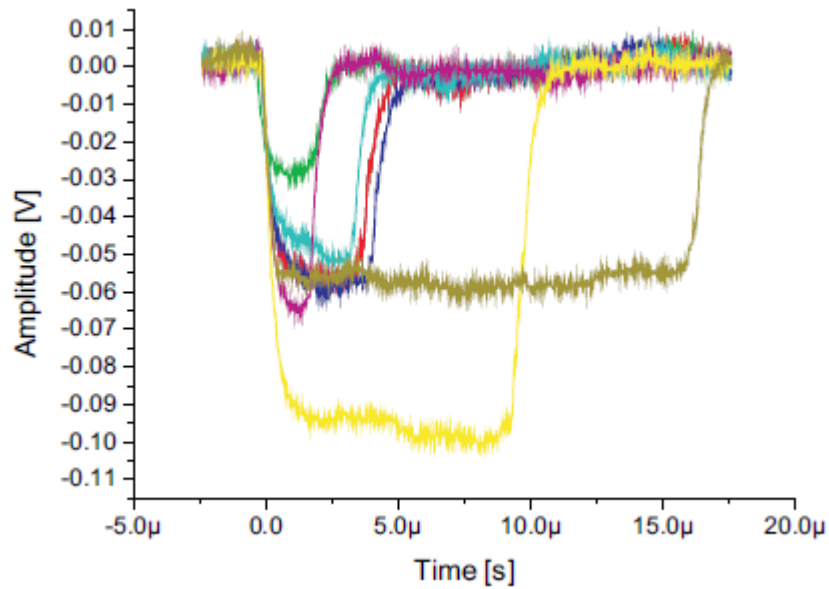
- Large area CMOS sensors can be produced by **stitching** several 2cm x 2cm wafer reticles
- **2D position information**: every pixel generates a digital current pulse with unique amplitude



- Prototypes:
 - **Demonstrator** (ams 180 nm) [I. Peric, PoS(Vertex 2012)021]
 - [2015] Small test detector HVStrip in 350nm AMS technology (ATLAS upgrade)

- **Power** consumption per pixel including a CSA and a comparator $\sim 7\ \mu\text{W}$

HVCMOS strip readout



Pulse height discrimination

No strip readout chip used. Measurements directly in an oscilloscope

⁵⁵Fe source measurements: **Z-dependent bias** is used.

Amplitude variations due to current source mismatch and noise

Pulse width proportional to the charge measured by the integrated CSA

Measured output signal amplitude as the function of pixel-position.

Signal amplitudes are Z-dependent and fluctuate from hit to hit.

HVCMOS radiation hardness

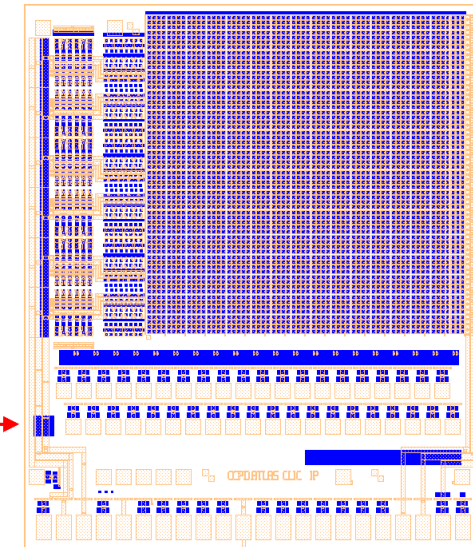
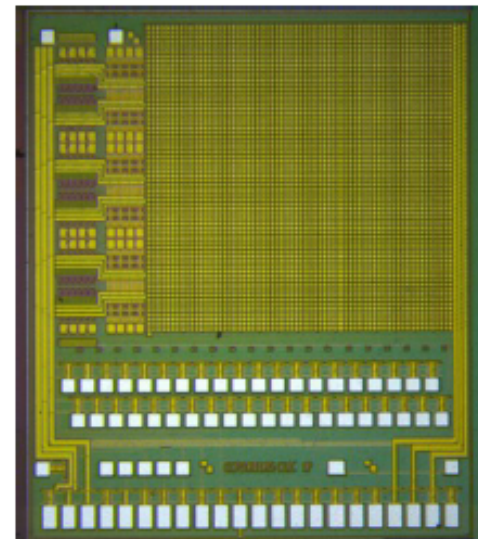
M. Fernandez et al, JINST paper
(approved for publication)

- Contributed by **IFCA-UC**, together with **CERN-SSD** within **RD50** collaboration
- Charge collection studied using dedicated diode conceived for **edge-TCT** measurements

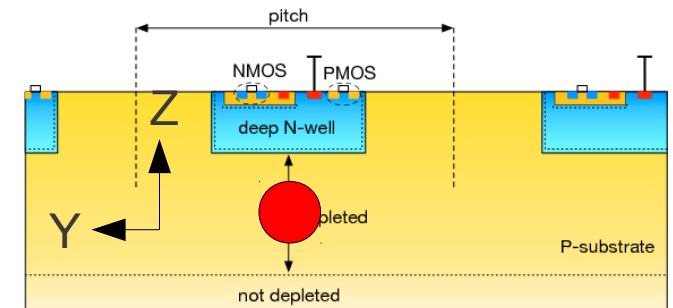
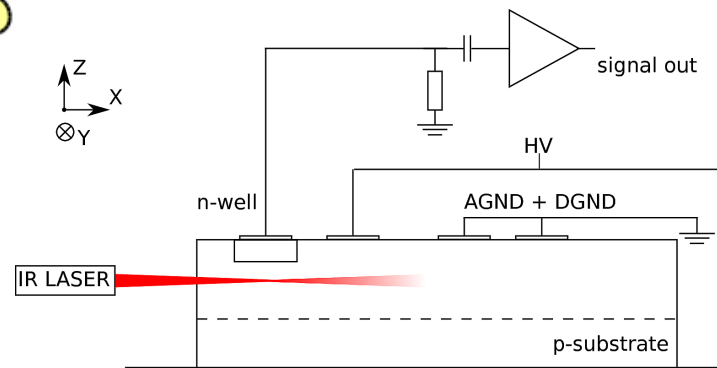
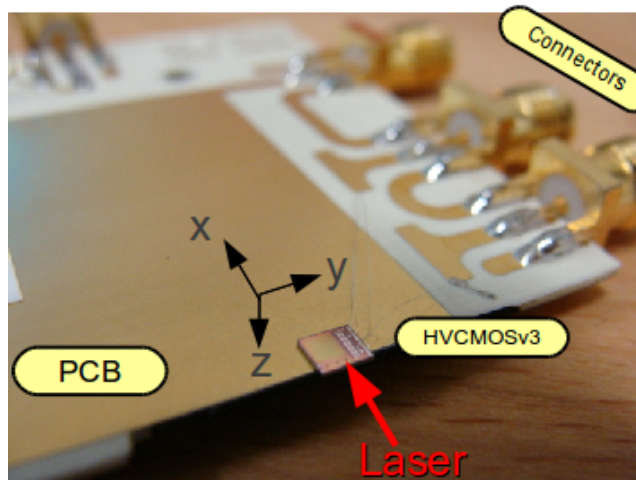
Neutron (Lbj)	1	7	20
Protons (PS)	1.4	3.7	6.9

$\times 10^{15} n_{eq}/cm^2$

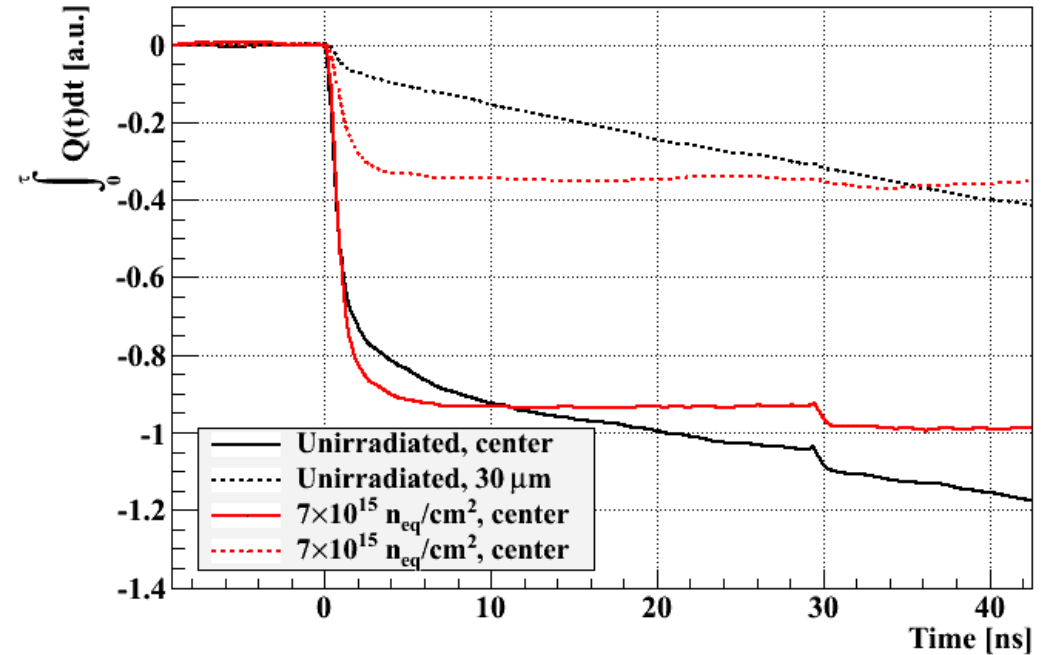
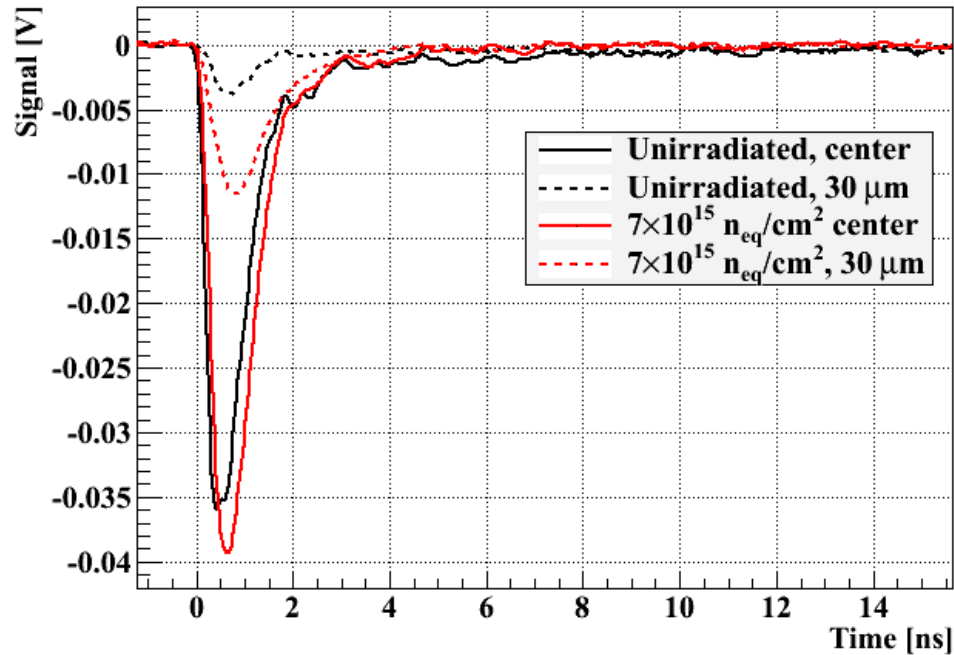
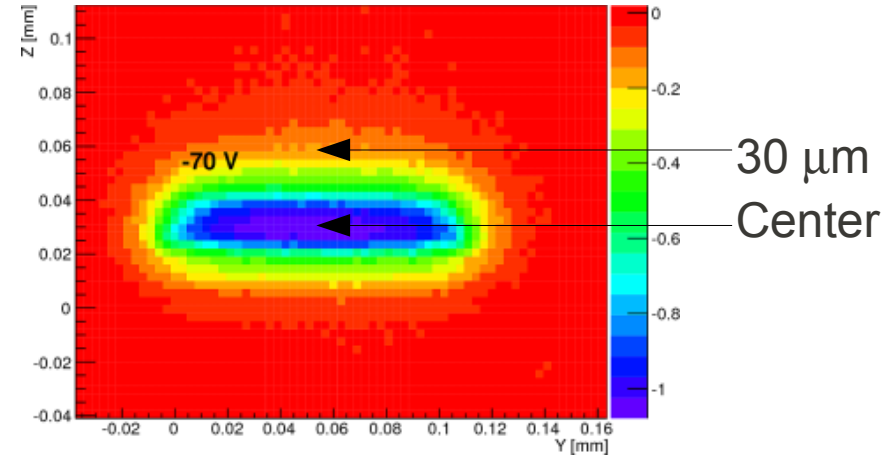
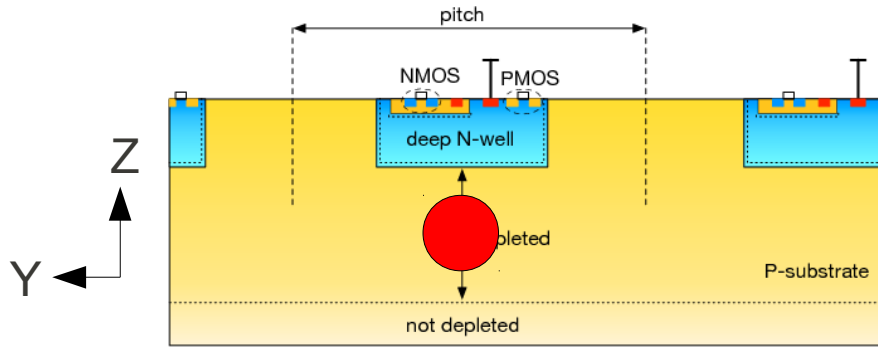
ILC needs: $10^{11} n_{eq}/cm^2/year$



- **Edge-TCT: PCB, HVCMOS biasing, measurement sketch**



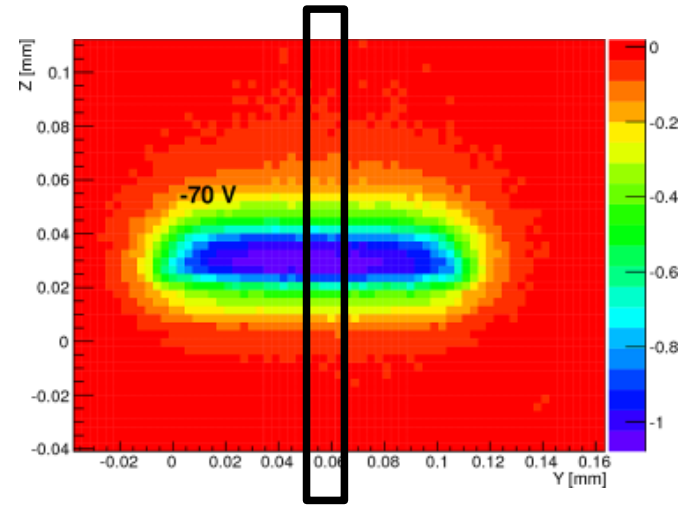
HVCMOS: radiation hardness



- Transients at the center of the diode and 30 μm below it.
- Fast collection → Seems like drift

- Running charge: accumulated charge as a function of integration time
- Distinct signature of diffusion is unveiled

HVCMOS: charge profiles



- Collected charge profiles $Q(z)$
FWHM of profiles \rightarrow Depletion depth

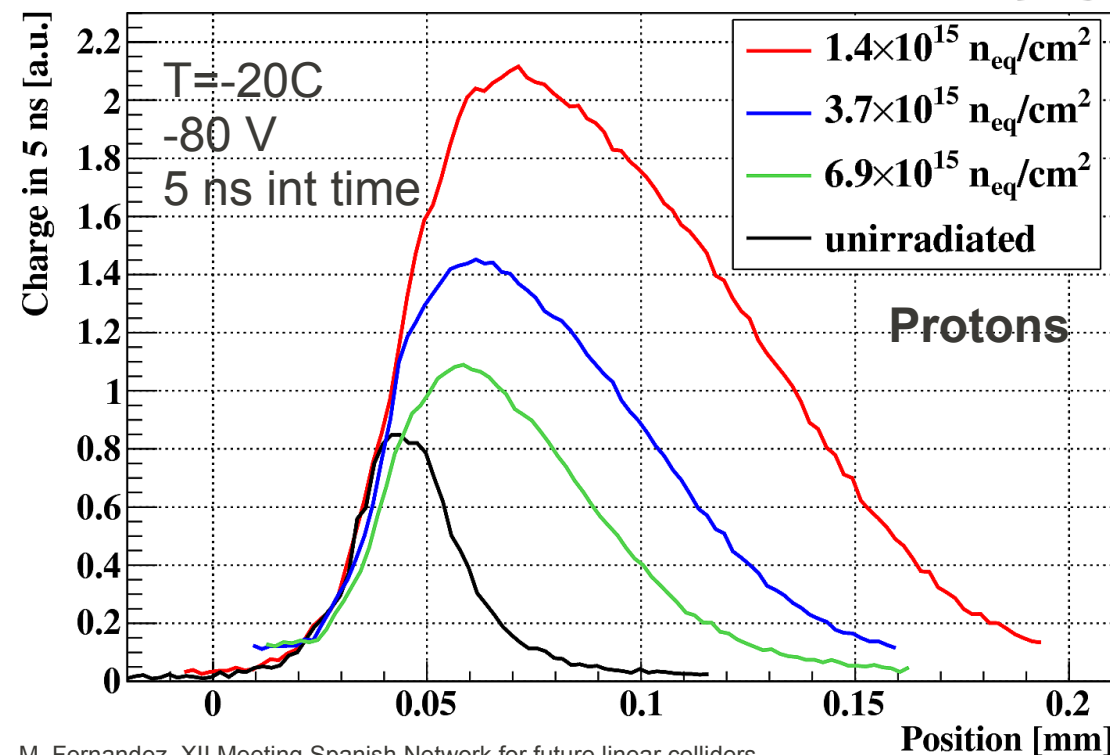
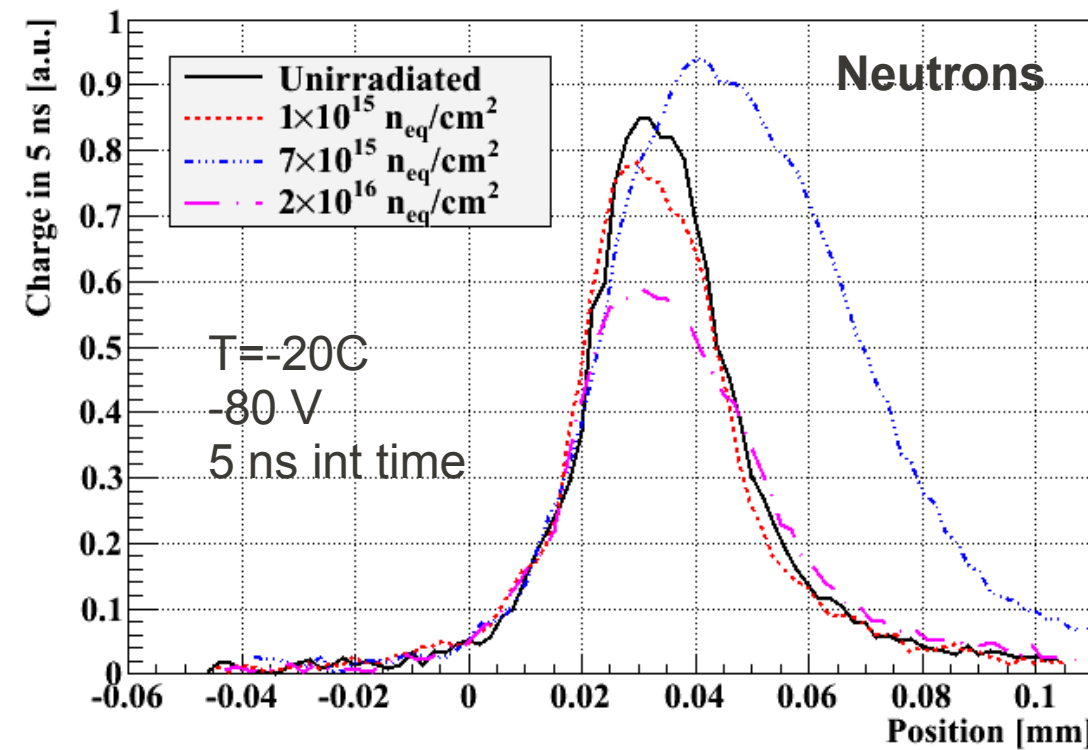
Neutrons:

Unirrad & $1e15$ very similar
Deepest depletion for $7e15 n_{eq}/cm^2$

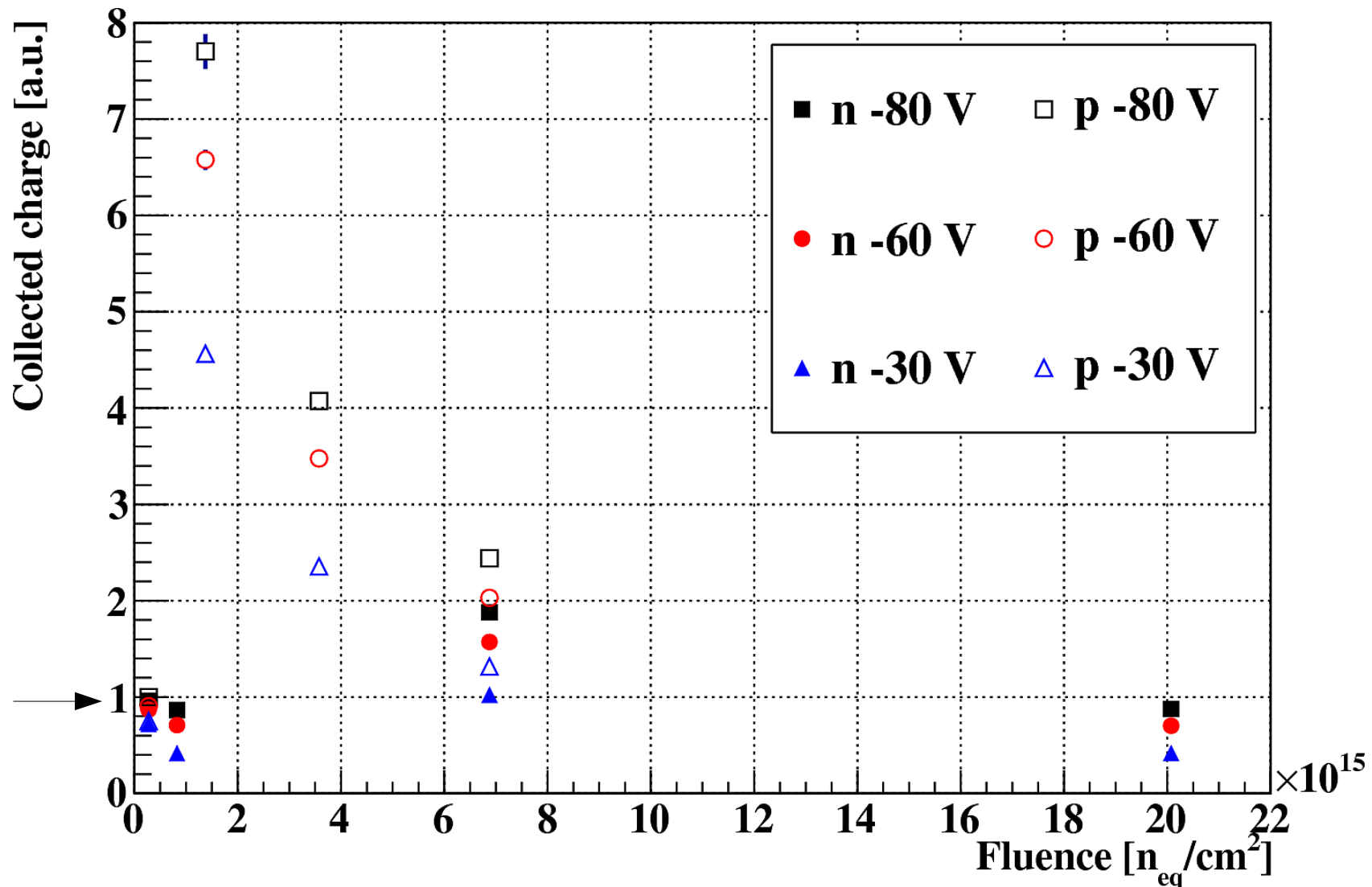
Protons:

More charge collected for any fluence
Deepest depletion for $1.5e15 n_{eq}/cm^2$

Note: Laser power variation 1% (RMS)₁₂
Mind different scales!



Collected charge (neutrons and protons)

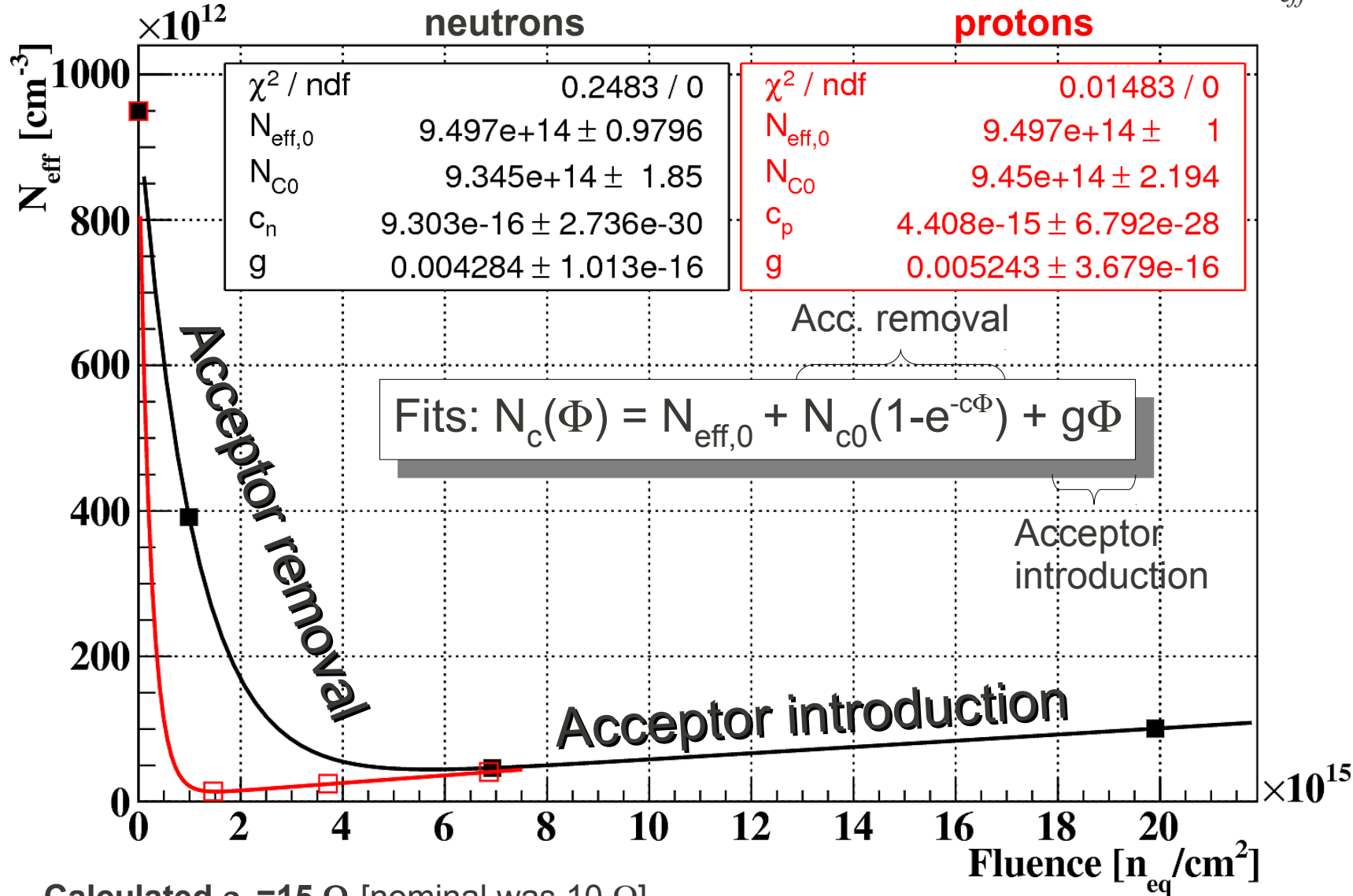


- **More charge** collected in **irradiated** detectors due to increased depletion depth
- Showing collected charge (at the center of the detector) in **5 ns**, over **200 μ m**
- **Fluence range** $1-1.5 \times 10^{15} n_{eq}/cm^2$, very fast increase of collected charge

Space charge (neutrons and protons)

- Doping concentration: $N_{\text{eff}} = N_{\text{donor}} - N_{\text{acceptor}} \sim - N_{\text{acceptor}}$

- It can be estimated (abrupt junction) from width of charge profiles: $N_{\text{eff}} = \frac{2\epsilon\epsilon_0}{q \cdot FWHM^2} V$



Calculated $\rho_0 = 15 \Omega$ [nominal was 10Ω]

Conclusions

- Growing surface of tracking systems, complexity of hybridization, cost of traditional hybrids, material budget... make monolithic solutions very attractive.
Large area+low cost → makes them excellent candidates to populate (outer strip) trackers

- Rad. hardness of vertex&tracker for ILC does not lead R&D.
Undepleted MAPS → slow
HVCMOS → Rad hard and fast (for ILC)

LC community can benefit from developments for HL-LHC

- HVCMOS ATLAS-collaboration demonstrated strip readout prototype
Strips mimicked daisy chaining “long pixels”
2D information by amplitude encoding

- HVCMOS radiation hardness tested until $2 \times 10^{16} n_{eq}/cm^2$
 - Unexpected effect observed: acceptor removal
 - For the first time observed, the collected charge bigger than unirradiated!
 - Irradiation of low resistivity substrates seem to have beneficial effect in:
 - Charge collection
 - Depletion depth
 - After $2 \times 10^{16} n_{eq}/cm^2$ detector performance very similar to unirradiated

BACKUP

HVCMOS

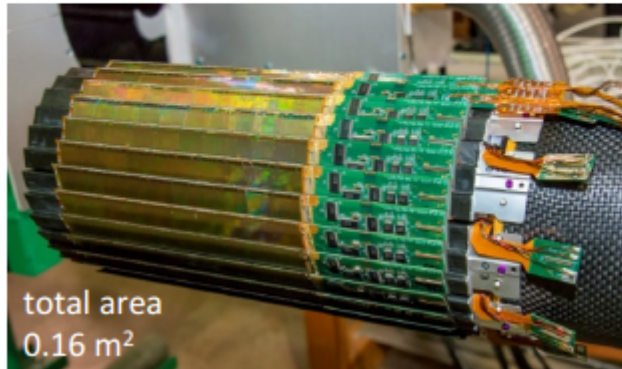
the deep n-well has two roles:
signal collecting region
substrate for PMOS transistors placed in it.

Usually the PMOS transistor substrate (n-well) is connected to the positive supply voltage which makes its potential stable. Here we cannot bias the n-well in this way because the signal charge would be absorbed by the positive supply. The n-well has to be "floating" (biased using high resistance). Electrons collected by the n-well cause a small voltage drop in it. The n-well is coupled to the input of the pixel amplifier. The amplifier amplifies the signal and restores the original n-well voltage.

- Power for VTX and tracking will benefit from low duty cycle (power pulsing possible).
Example: $130 \mu\text{W}/\text{mm}^2$ barrel of SiD
Air cooling envisaged (due to low mass needed for the system $0.15\%X_0/\text{layer}$)

(Semi)-Monolithic Pixel Detector Projects

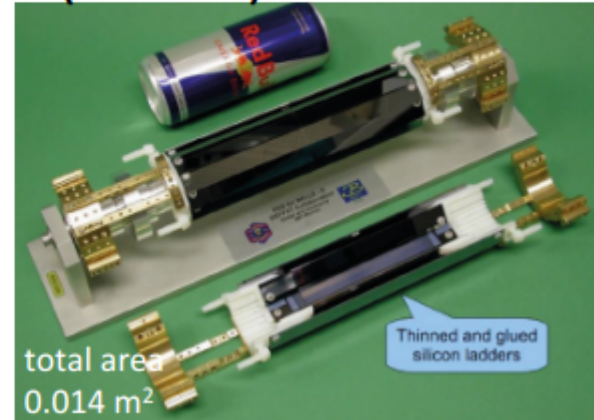
STAR / RHIC MAPS



total area
0.16 m²

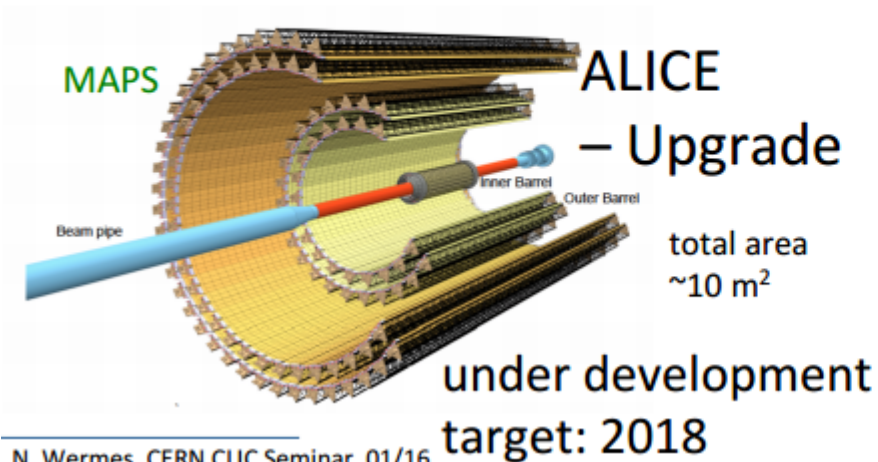
in operation since 2014

(Belle II) DEPFET



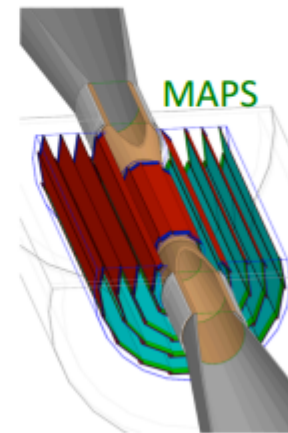
total area
0.014 m²

in production for 2017



total area
~10 m²

under development
target: 2018

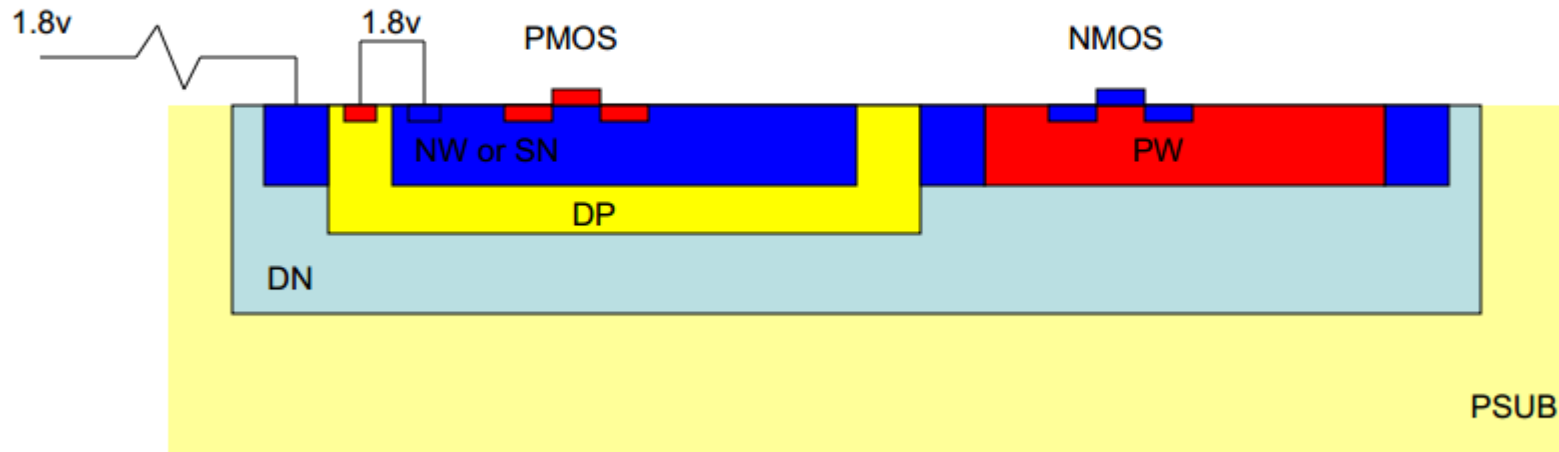


ILC

total area
? m²

current
baseline

- Capacitive feedback into the sensor (n-well)
- Many important circuits do not cause problems: charge sensitive amplifier, simple shaper, tune DAC, SRAM but...
- “Active” (clocked) CMOS logic gates and sometimes comparators cause large crosstalk
- Possibility 1: Implement the circuits only using NMOST: effects on radiation tolerance, layout area, power consumption, etc.
- **Possibility 2: Place the active digital circuits on the chip periphery or on separate chip.**
- Possibility 3: Isolate PMOST from n-well using an additional standard technology feature – the deep P-well – we still haven’t tested it...



SID Vertex requirements

ILC TDR, pg: 63 The time structure and low radiation background in the ILC provides an environment which allows us to consider very light, low power detector structures. The bunch structure, with a 1 ms long bunch train at 5 Hz, enables power pulsing of the electronics, providing a power saving of a factor of 50-100 for front-end analog power. Goals:

- Hit resolution better than 5 μm in the barrel
- Less than 0.3% radiation length per layer. For a device with less than 0.3% radiation length per layer air cooling appears

to be the only viable low-mass sensor cooling technique

- Average power less than 130 $\mu\text{W}/\text{mm}^2$ in the barrel
- Single bunch time resolution (\approx 300-700 ns)

These requirements then drive the design of the vertex system. The 5 μm resolution implies a pixel size of 17 μm , larger if charge sharing is used to improve the resolution. Some CMOS MAPS devices, which collect charge by diffusion rather than drift, can utilise larger pixels because diffusion naturally spreads the charge.

SID Vertex Baseline

the vertex detector for SiD is proposed to be an all-silicon structure in a barrel-disk geometry. The vertex barrel and inner endcaps have $\sim 20 \times 20 \mu\text{m}$ pixels. The pixel size increases to $\sim 50 \times 50 \mu\text{m}^2$ for the forward tracker disks. The total area of the vertex barrels is $1.63 \times 10^5 \text{ mm}^2$ ($1 \text{ m}^2 = 1 \text{e}6 \text{ mm}^2$) and is $0.59 \times 10^5 \text{ mm}^2$ for each set of 4 inner pixel disks and $1.96 \times 10^5 \text{ mm}^2$ for each set of 3 forward pixel disks.

SID Sensor technology

3D integrated sensors and readout chips [66], Silicon-on-insulator (SOI) [67], Monolithic Active Pixels (MAPS) [68, 69], hybrid pixels [70, 71], and DEPFETs [72]. All of these technologies have the capability of delivering sensors less than 75 μm thick with 5 μm hit resolution and low power consumption.

Comments on ILD Vertex Detector Requirements

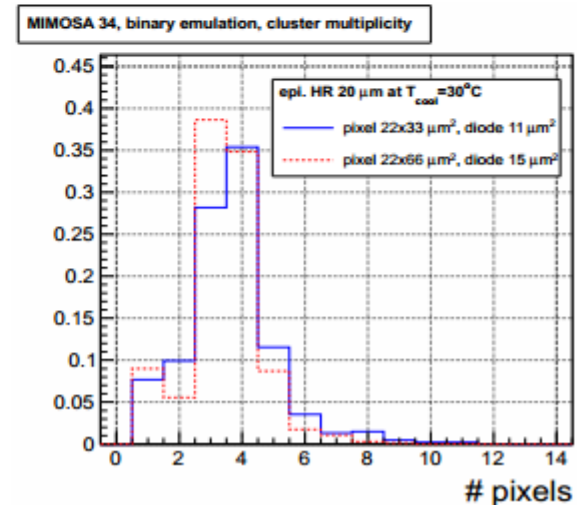
- Expected $N(\text{hits})/\text{cm}^2/\text{BX}$ at 500 GeV & 1 TeV with anti-DID, for each ILD-VXD layer (DBD) :

Layer	1	2	3	4	5	6
0.5 TeV	6.3 ± 1.8	4.0 ± 1.2	0.25 ± 0.11	0.21 ± 0.09	0.05 ± 0.03	0.04 ± 0.03
1 TeV	11.8 ± 1.0	7.5 ± 0.7	0.43 ± 0.13	0.36 ± 0.11	0.09 ± 0.04	0.08 ± 0.04

- Occupancy in L1 and L2 :

- * L1 : $17 \mu\text{m} \times 17 \mu\text{m}$ pixels (cluster mult. $\simeq 5$)
Pixel occ. at 500 GeV $\simeq 10^{-4}/\text{BX} \Rightarrow 10^{-2}/50 \mu\text{s}$
- * L2 : $17 \mu\text{m} \times 102 \mu\text{m}$ pixels (cluster mult. $\simeq 3$)
Pixel occ. at 500 GeV $\simeq 2 \times 10^{-4}/\text{BX} \Rightarrow 10^{-3}/2.5 \mu\text{s}$
- * 1 TeV : twice higher occupancy
- * Luminosity upgrade : 4 times higher occupancy
- * Large uncertainties (MC stat., anti-DID uncertainty, etc.)

\Rightarrow **Safety margins required on param. governing occupancy**



Submission Oct. 2015: AMS CMOS demonstrator chip (CCPD)
in 350 nm technology (different wafer resistivity: 20 – 1000 Ωcm)
received back: End of 2015

The vertex detector is realised as a multi-layer pixel-vertex detector (VTX), with three super-layers

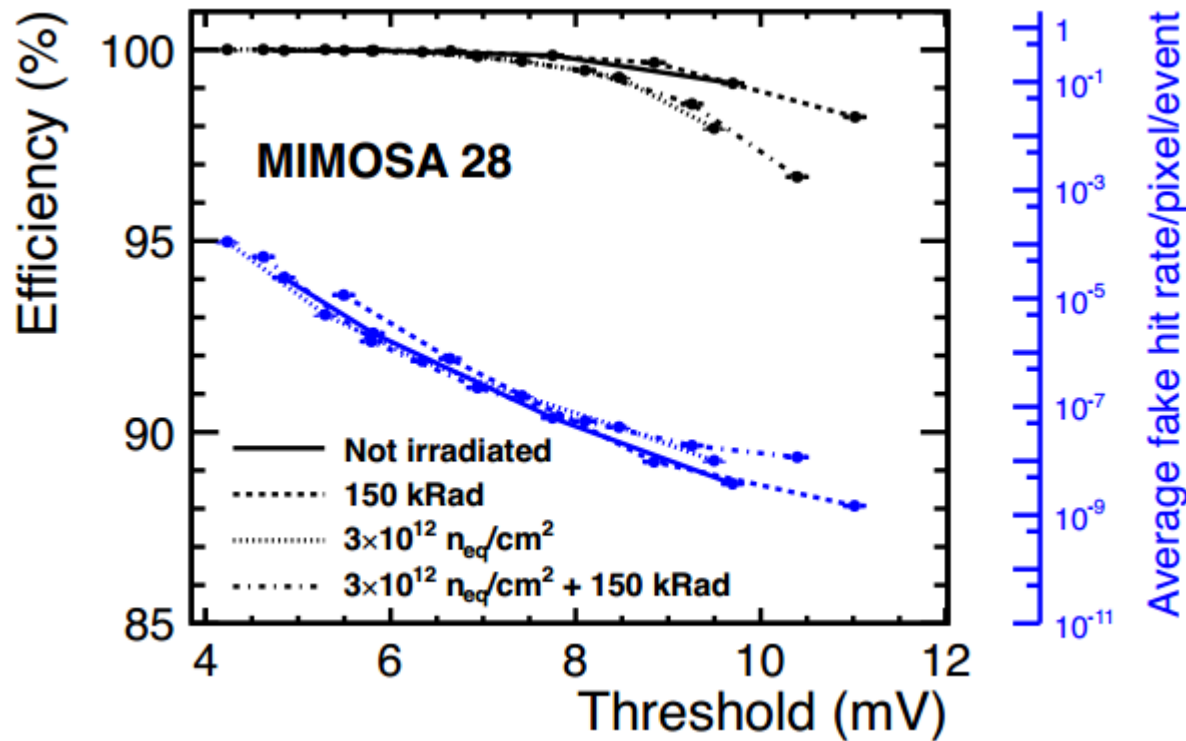
each comprising two layers, or a 5 layer geometry. Whilst the underlying detector technology has not yet been decided, the VTX is optimised for point resolution and minimum material thickness.

A system of silicon strip and pixel detectors surrounds the VTX detector. In the barrel, two layers of silicon strip detectors (SIT) are arranged to bridge the gap between the VTX and the TPC.

In the forward region, a system of two silicon-pixel disks and five silicon-strip disks (FTD) provides low angle tracking coverage. A distinct feature of ILD is a large volume time projection chamber (TPC) with up to 224 points per track.

Outside the TPC a system of Si-strip detectors, one behind the end-plate of the TPC (ETD) and one in between the TPC and the ECAL (SET), provide additional high precision space points

The required radiation tolerance follows entirely from the beam related background (i.e. beamstrahlung) (see section 5.5.6), which is expected to affect predominantly the innermost layer. The requirements for the total ionising dose and the fluence amount respectively to about 1kGy and 10^{11} neq/cm² per annum.



Measured variations of the m.i.p. detection efficiency and fake hit rate (fraction of pixel noise fluctuations above threshold) of the STAR sensor as a function of the discriminator threshold, before and after irradiation (150kRad, $3 \times 10^{12} \text{ neq}/\text{cm}^2$) at a coolant temperature of 30C.

Fpg. 205: TD using CMOS pixelated strips could avoid stereo-angle modules (radial resolution of $O(100 \mu\text{m})$). To achieve a precise measurement of the longitudinal impact parameter the radial segmentation of the innermost disks is crucial. Due to their higher occupancy, the first two disks are implemented using highly granular pixel detectors. three technologies are under consideration: CPS, CCD and DEPFET.

With the current design one disk of the FTD strip system will use less than 200 W of power during electronics-on time, or less than 40 W on average.