

# PETIROC2A, a 32-channel readout ASIC for accurate time resolution and charge measurements

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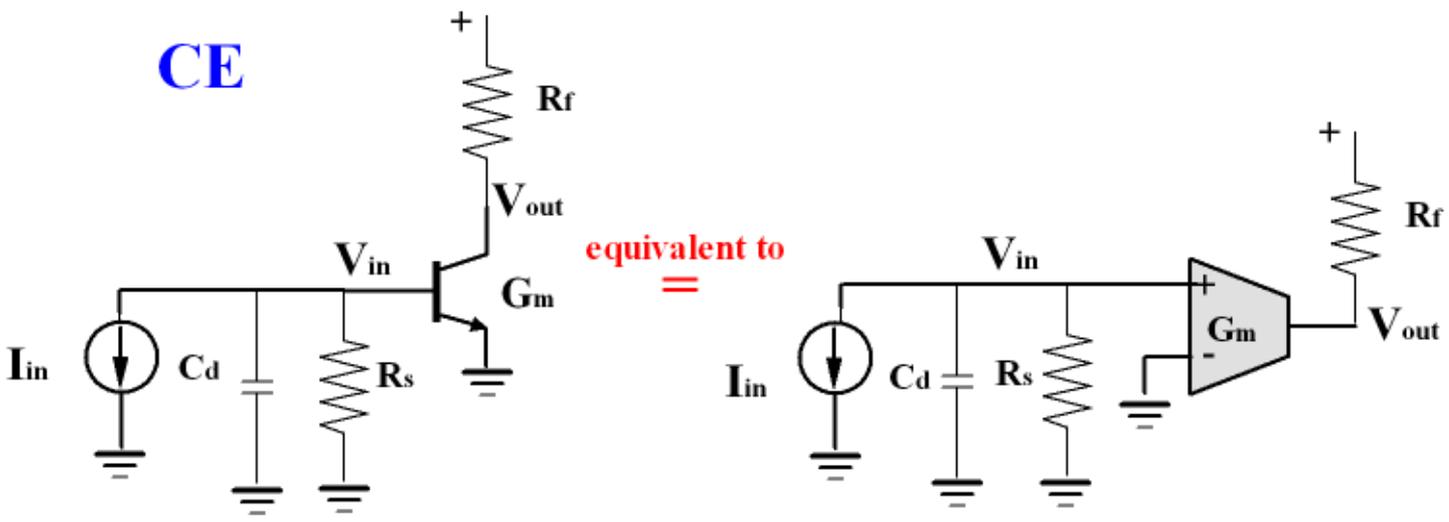
OMEGA microelectronics group  
Ecole Polytechnique & CNRS IN2P3  
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# HF amplifiers : BroadBand and Transimpedance

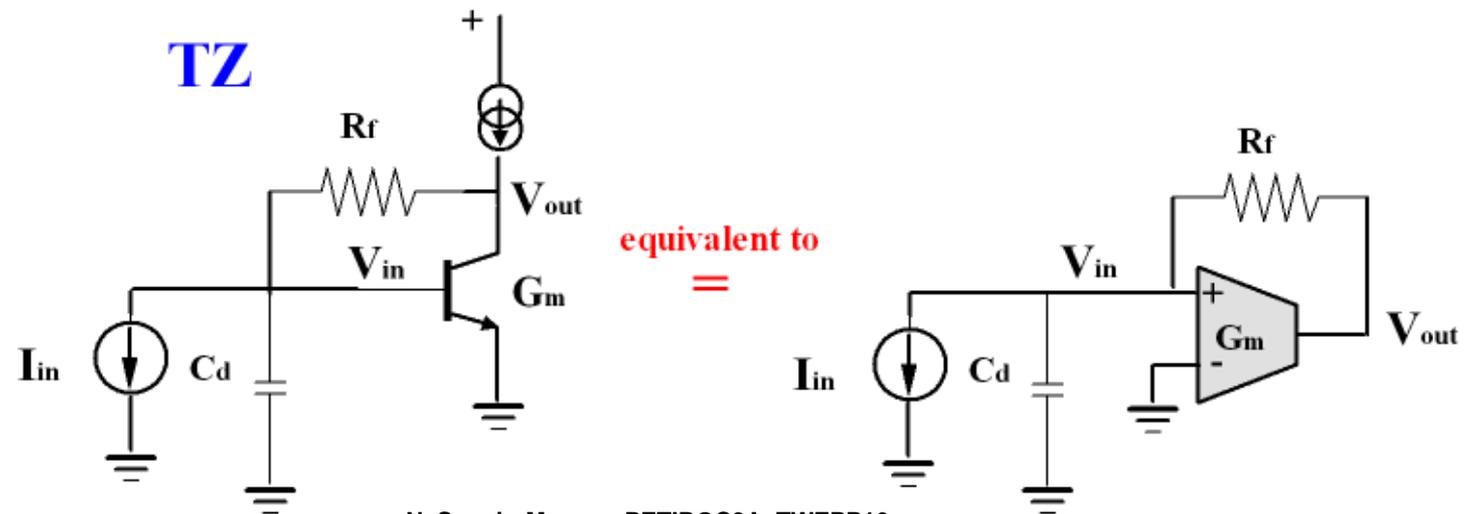


**CE**



Similar response at very high Frequency  $V_{OUT} = -g_m R_f \frac{Q_{IN}}{C_d}$

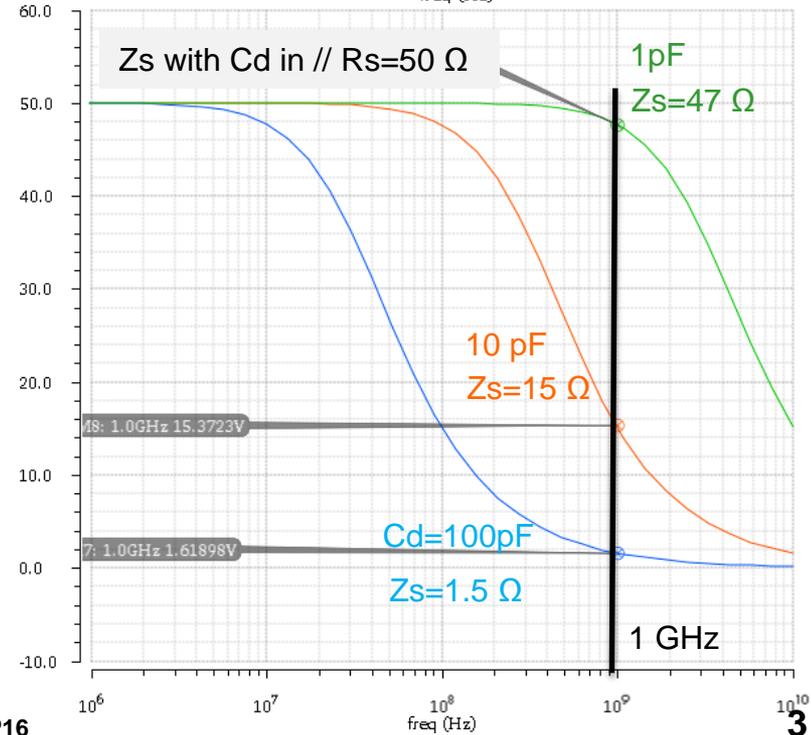
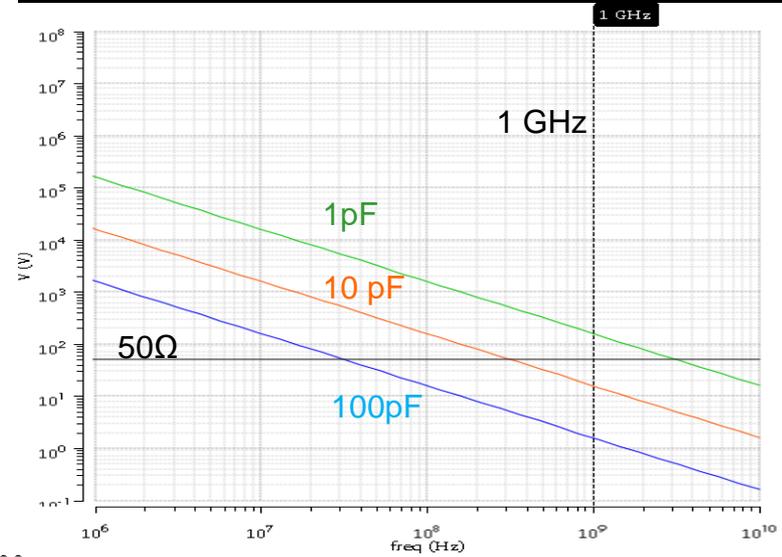
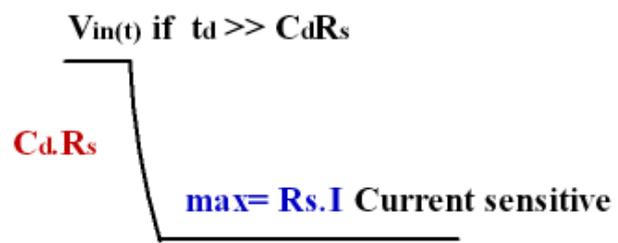
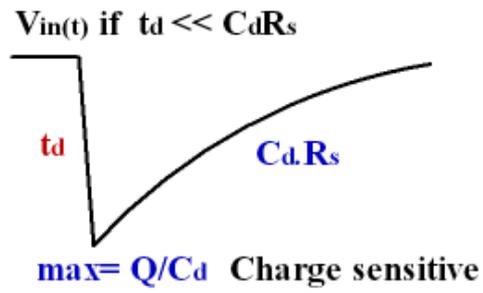
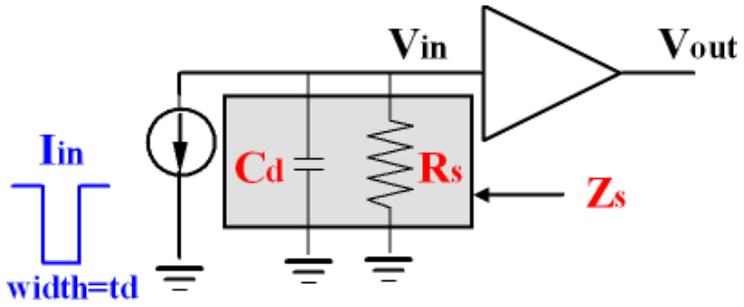
**TZ**



# Detector impedance and input voltage



- 1 GHz, Cd=few tens of pF, input signal <1ns
- Cd>1 pF, Zs@1GHz dominated by Cd
- Rise time: tr= td when td<<CdRs and tr=CdRs when td>>CdRs



# NOISE TZ and CE

- Similar noise for CE and TZ

$$S_{v_{MOS}}^{out}(\omega) = (g_m R_f)^2 \left( 4kTR_{GG'} + \frac{4kT\Gamma}{g_m} \right) + 4kTR_f + k \frac{I_D}{f}$$

- rms noise  $v_n$  depends on the  $\sqrt{BW}$
- Noise is independent of  $C_d$

$$N = v_n \approx R_f \sqrt{2kTg_m BW}$$

- Signal scales with  $1/C_d$

$$V_{OUT} = -g_m R_f \frac{Q_{IN}}{C_d}$$

- S/N depends on  $C_d$  and  $BW$

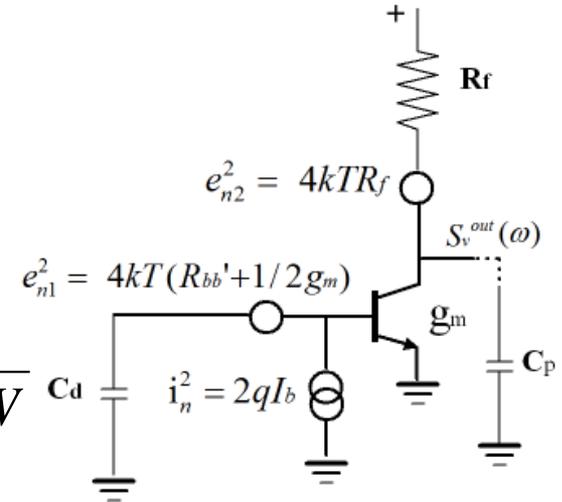
$$\frac{S}{N} \approx \frac{\alpha Q \sqrt{g_m}}{C_d \sqrt{BW}} \approx \alpha \frac{Q \sqrt{g_m} \sqrt{t_{r\_ampli}}}{C_d}$$

- Jitter :

$$\sigma_t^J = \frac{N}{dV/dt} = \frac{\sqrt{t_{r\_a}^2 + t_d^2}}{S/N} = \alpha \frac{C_d}{Q \sqrt{g_m}} \frac{\sqrt{t_{r\_a}^2 + t_d^2}}{\sqrt{t_{r\_a}}}$$

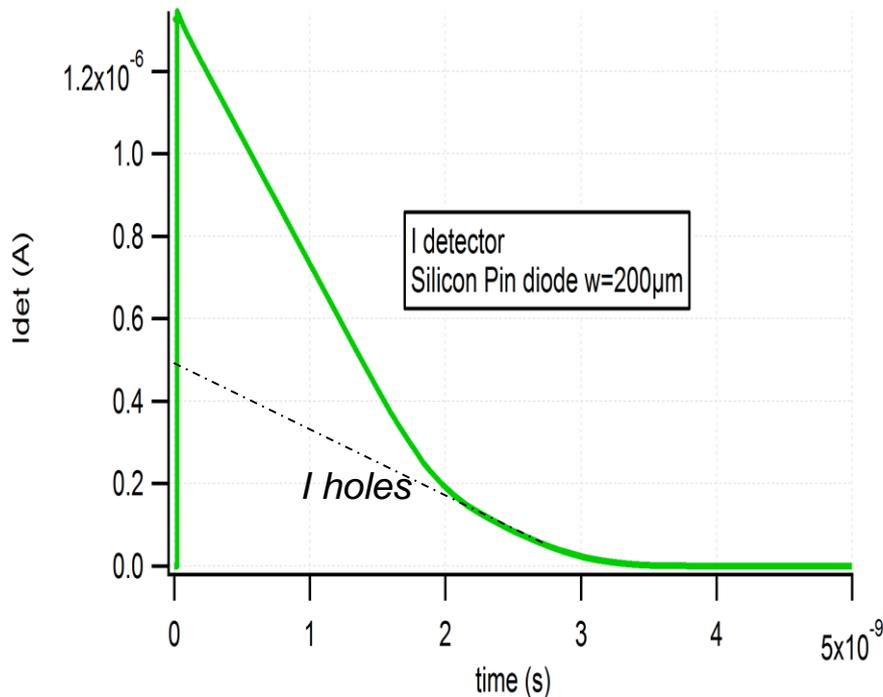
Optimum value:  $t_{r\_a} = t_d$  (current duration)

$$\sigma_t^J \approx \alpha \cdot \frac{C_d}{Q} \sqrt{\frac{t_d}{g_m}}$$

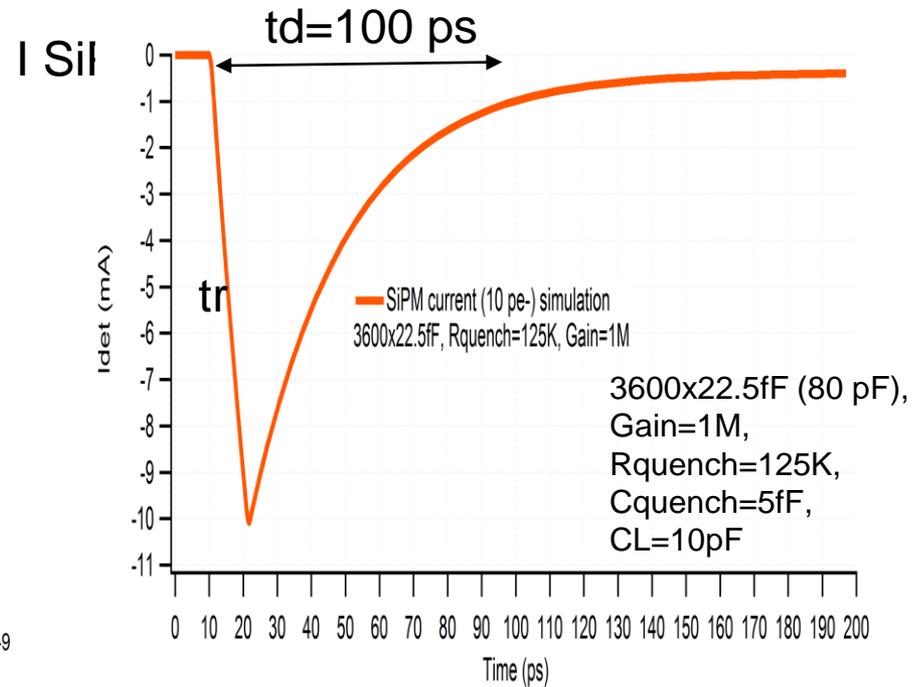


# Detector current

- PIN diode  $w = 200\mu\text{m}$
- Very short rise time :  $t_r \sim 10\text{ps}$
- Relatively long «drift time» :  $t_d \sim 2\text{ns}$



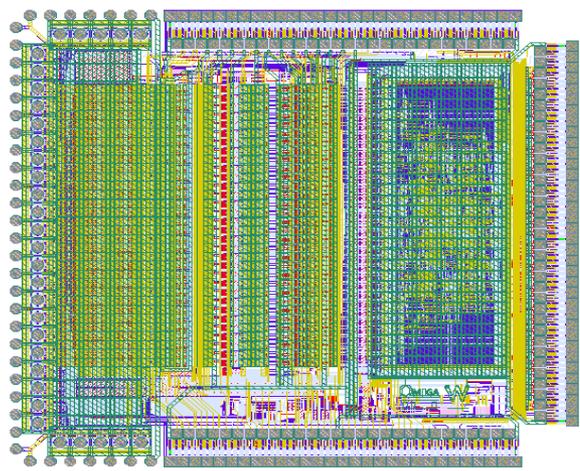
- SiPM detector (10pe-)
- very short rise time :  $t_r \sim 10\text{ps}$
- Short duration :  $t_d \sim 100\text{ps}$ ,



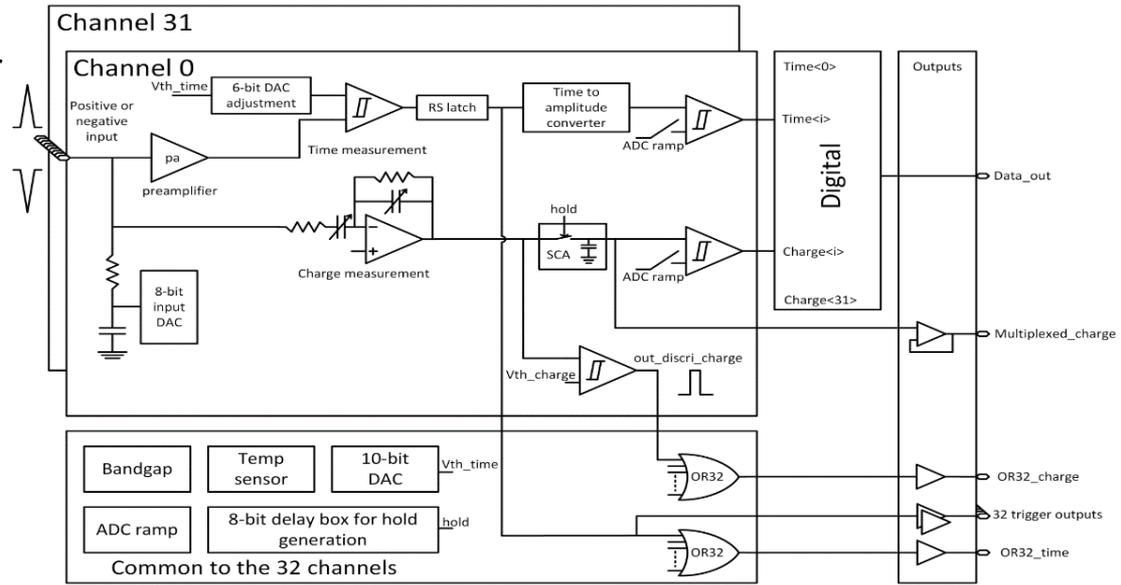
# PETIROC2 DESCRIPTION



AMS 0,35 $\mu$ m SiGe

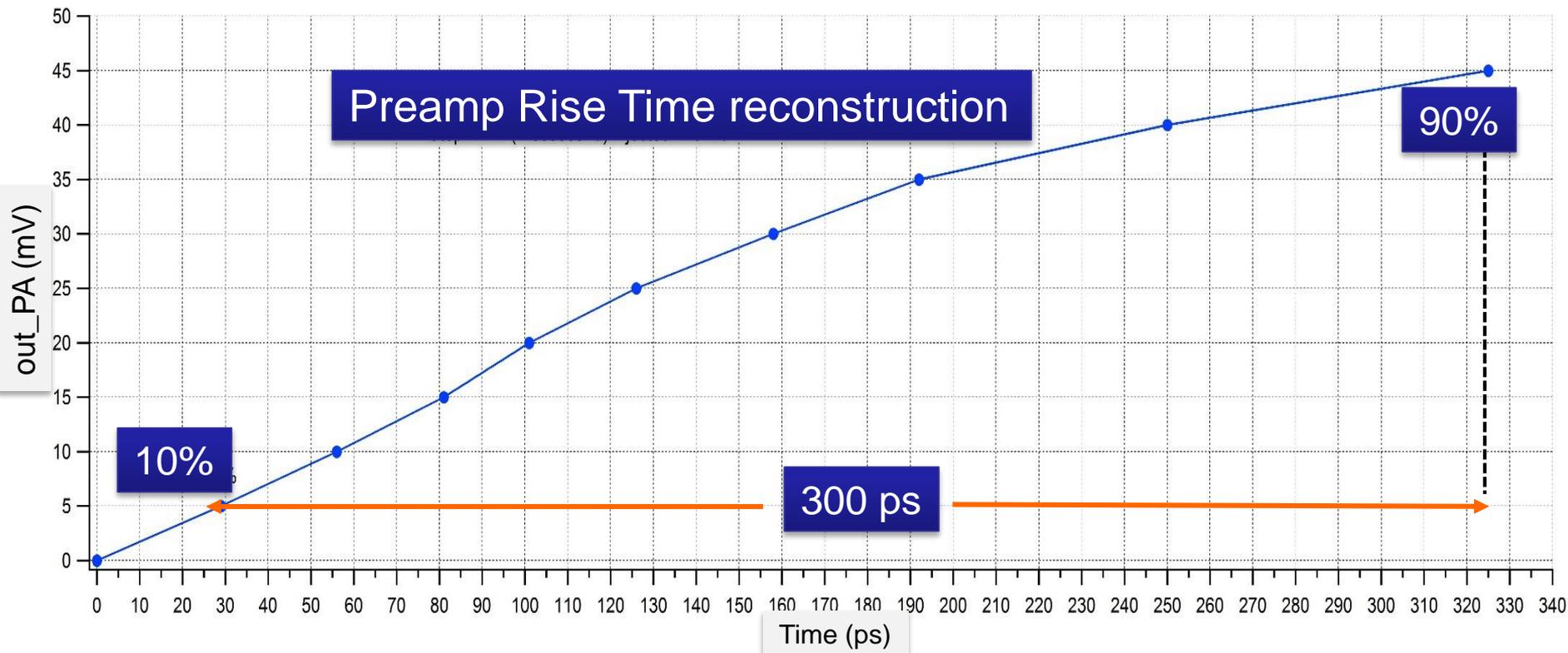
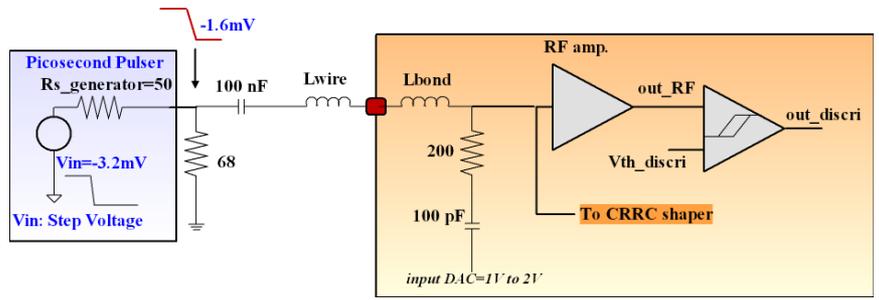


- Time of Flight read-out chip with embedded TDC (25 ps bin) and ADC
- Dynamic range: 160 fC up to 400 pC
- 32 channels (negative input)
  - 32 trigger outputs
  - NOR32\_chrage
  - NOR32 time
  - Charge measurement over 10 bits
  - Time measurement over 10 bits
  - One multiplexed charge output
- Common trigger threshold adjustment and 6bit-dac/channel for individual adjustment
- Variable shaping time of the charge shaper
- 32 8bit-input dac for SiPM HV adjustment
- Power consumption 6 mW/ch
- Front-end
  - Broad Band SiGe fast amplifier
  - Fast SiGe discriminator
  - 1 GHz overall bandwidth, gain = 25



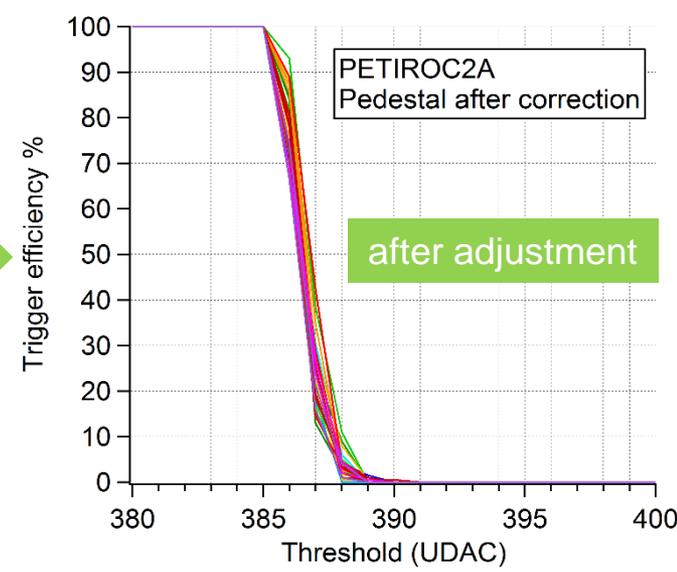
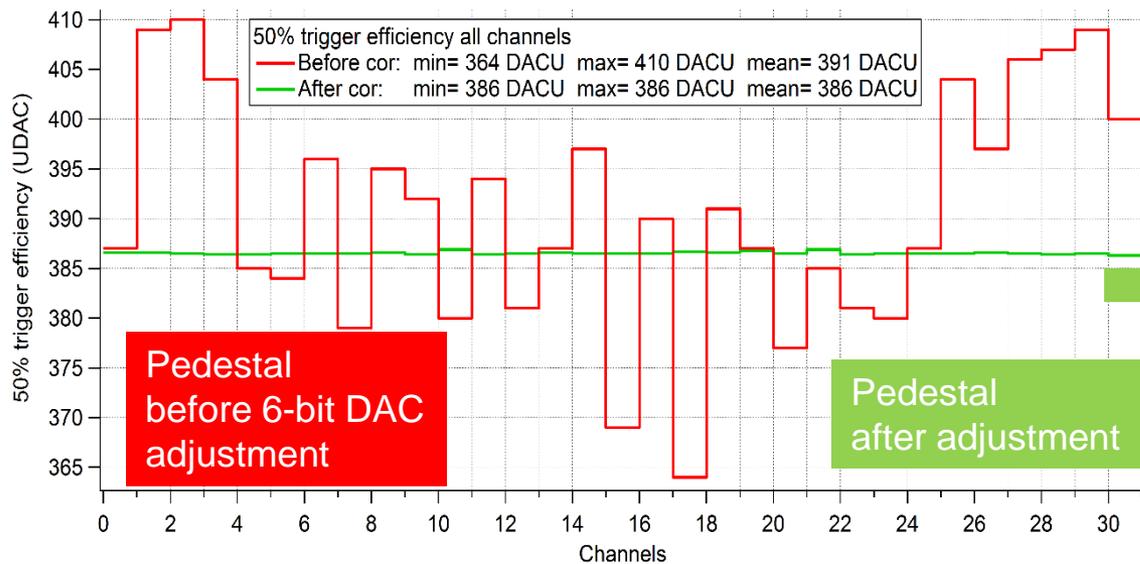
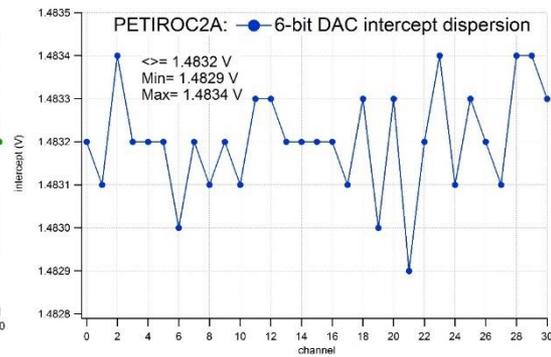
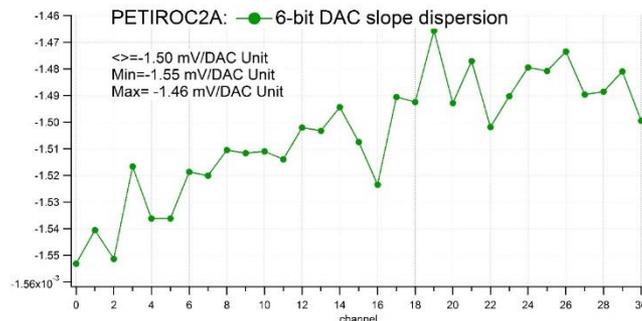
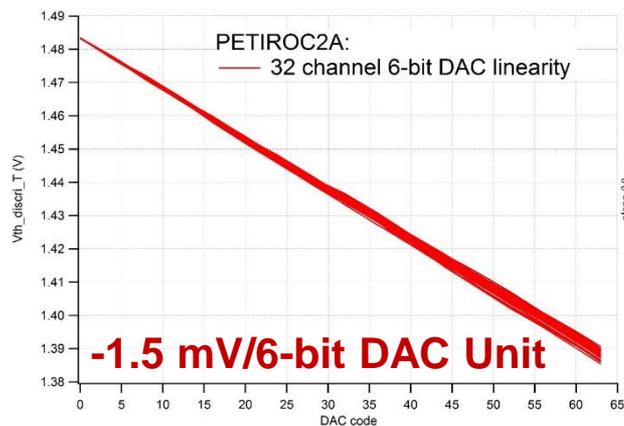
# PETIROC2A: Rise Time of the PA+discr

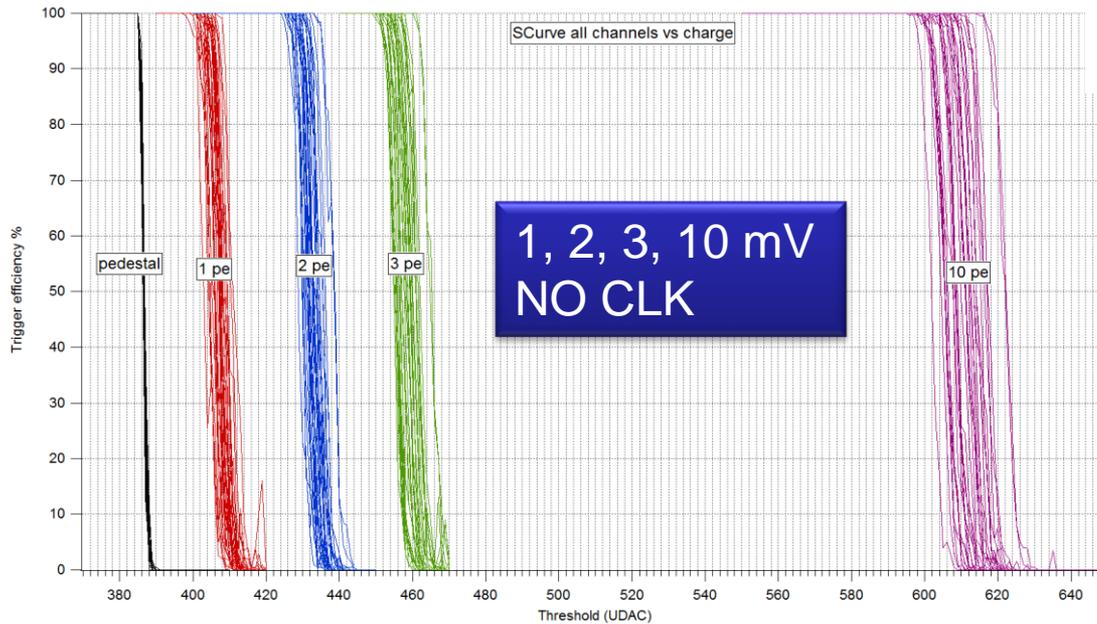
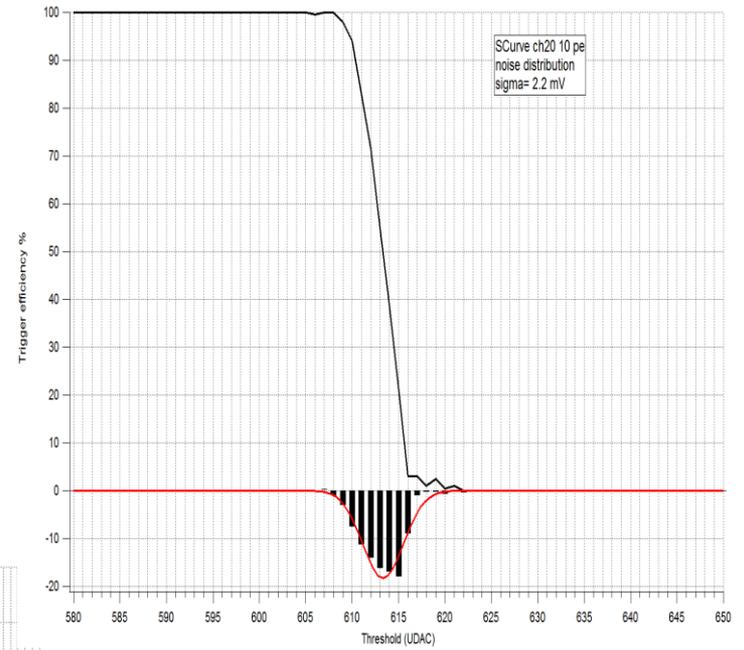
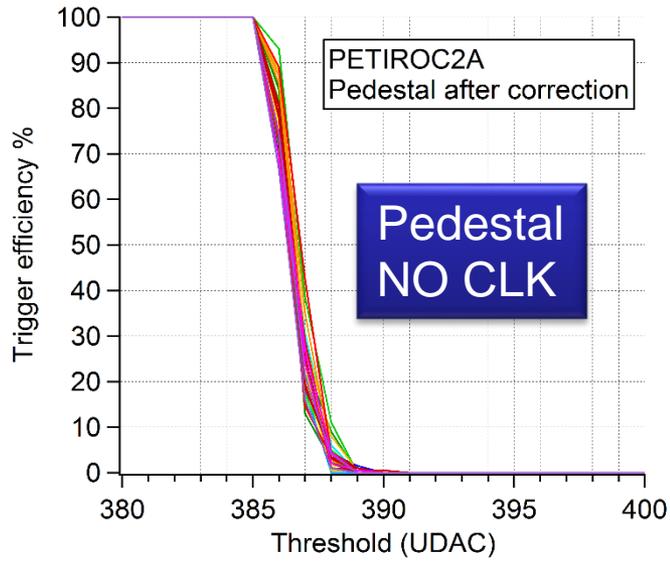
- Injection of a step of 2 mV in ch20,
- reconstruction of pulse by discriminator
- **tr 10%-90% = 300 ps,**
- **BW = 0.35 / tr ~ 0.9 GHz**
- can also use **BW = 0.1 / (t<sub>50%</sub> - t<sub>10%</sub>)**



# 6bit-DAC for pedestal adjustment

- Preamp: Open loop => dispersion of the pedestal
- 6 bit dac/channel for pedestal uniformization

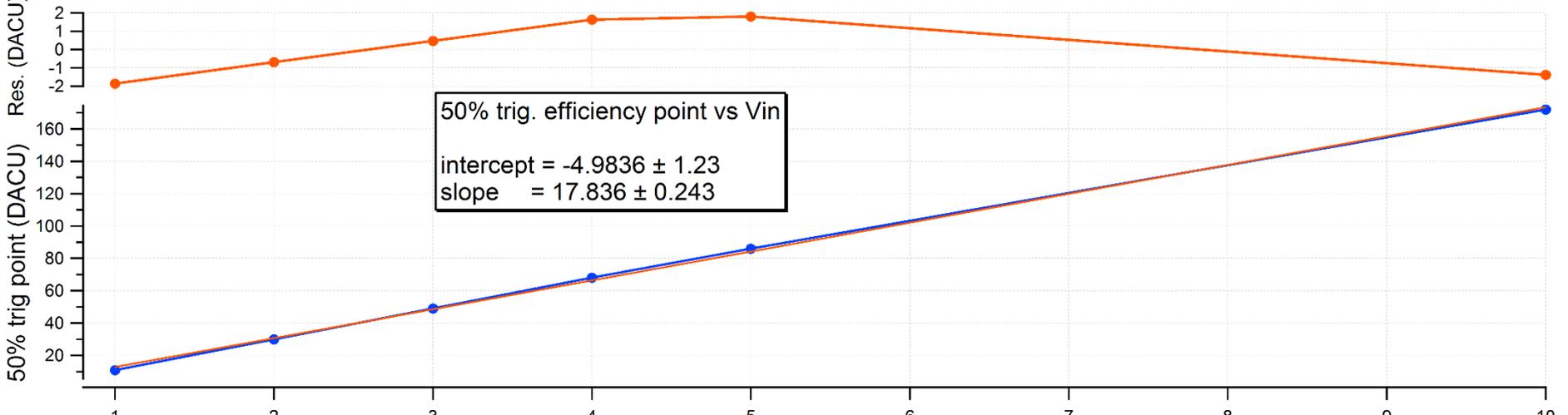
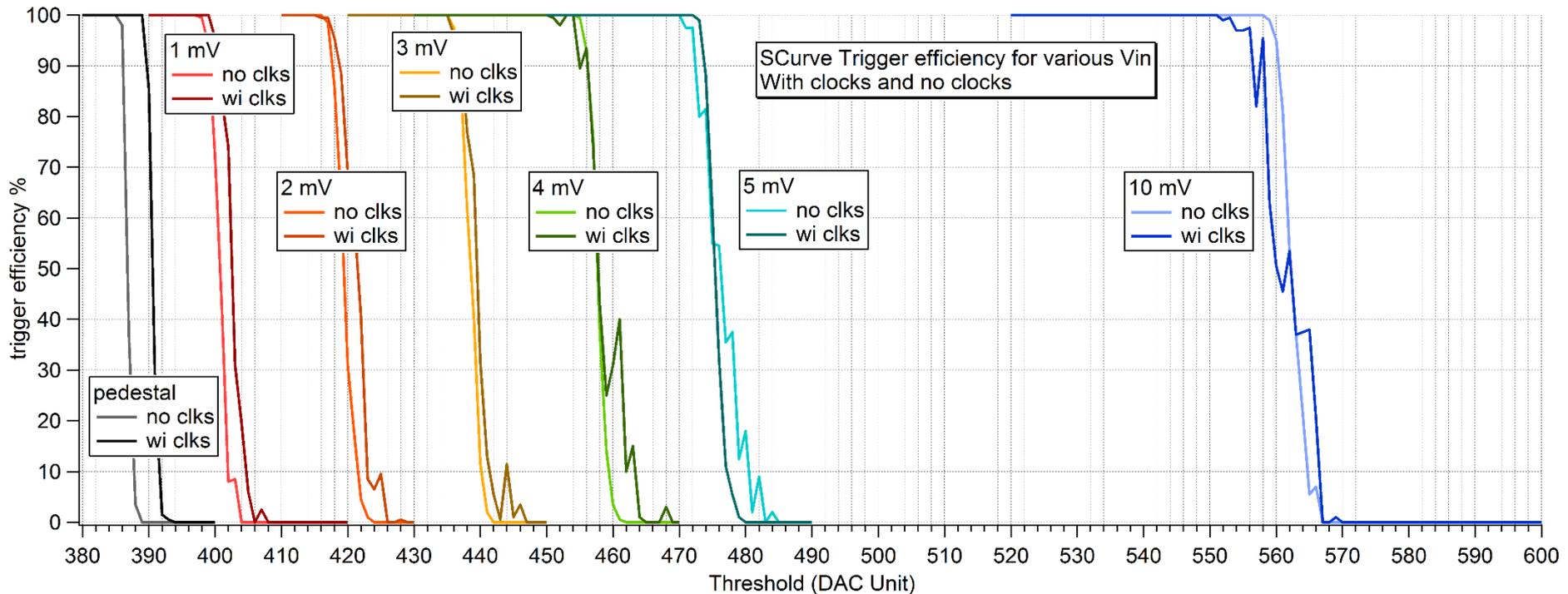




# SCURVES with and w/o clock

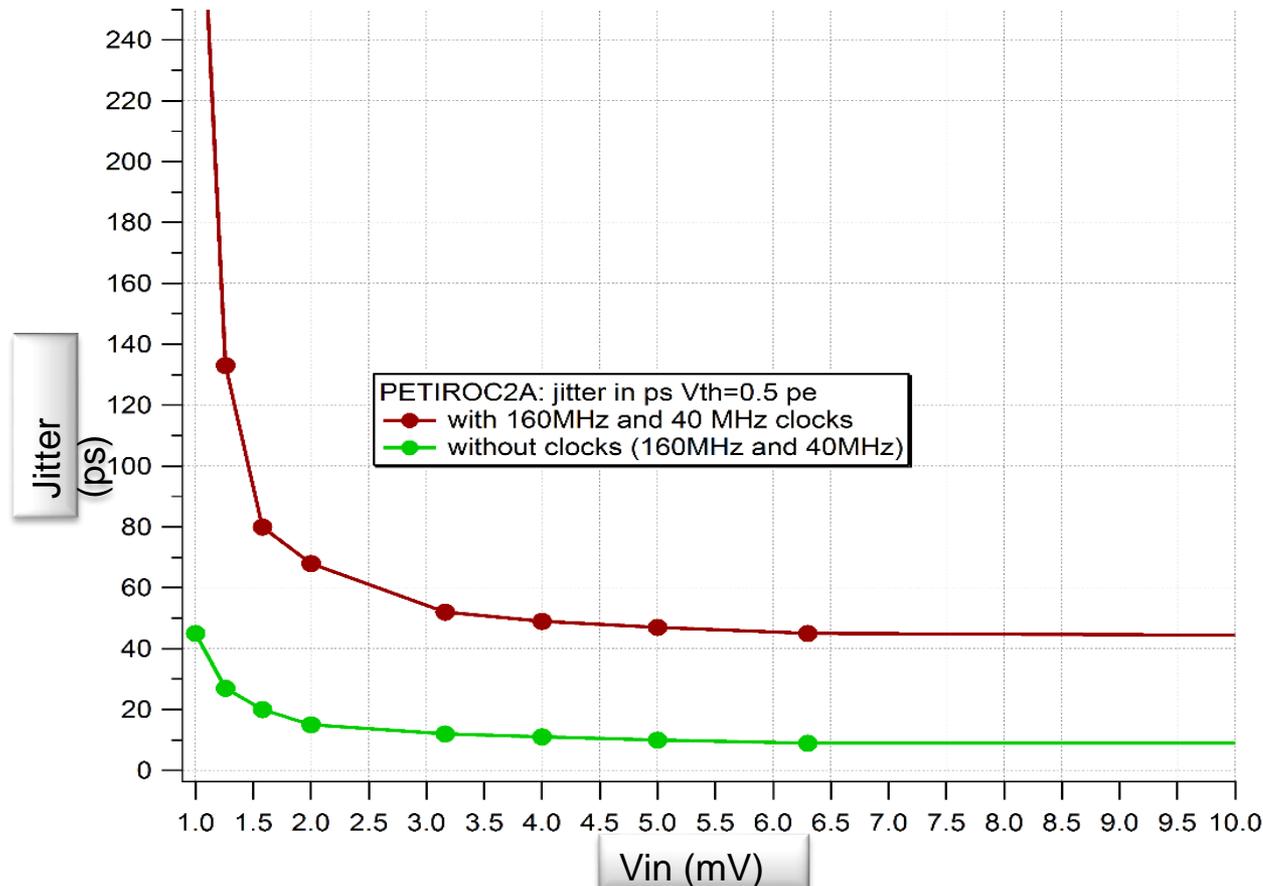


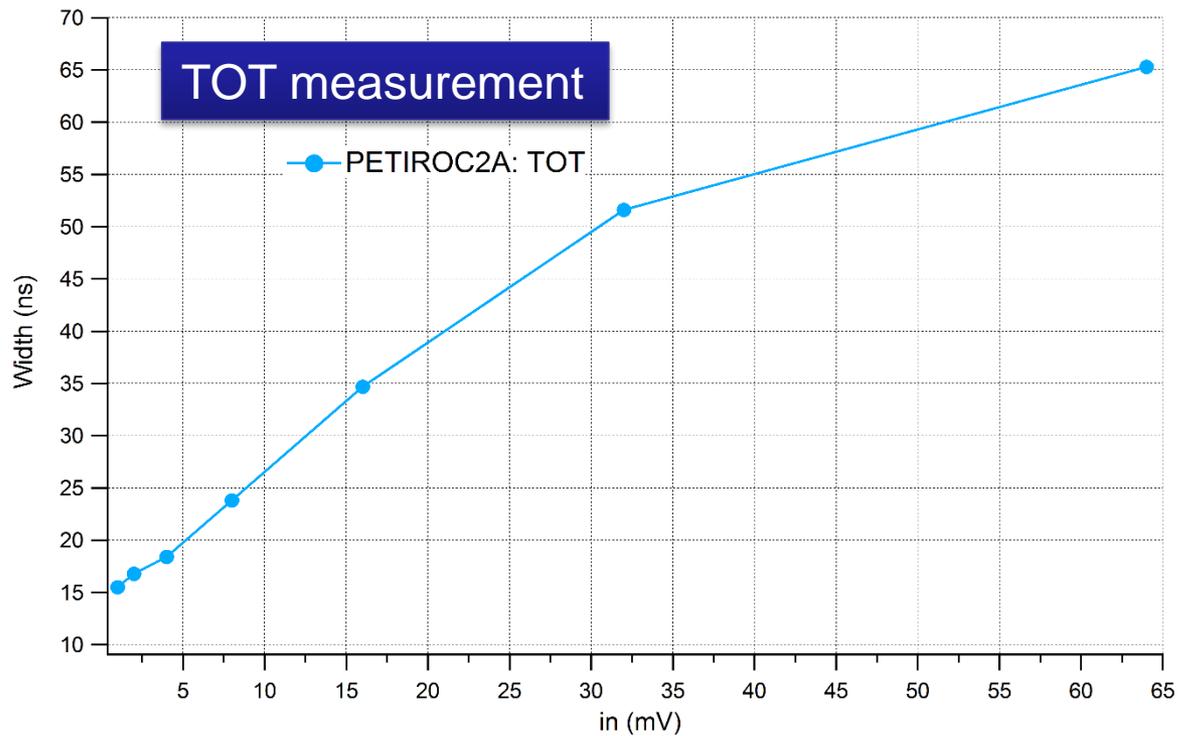
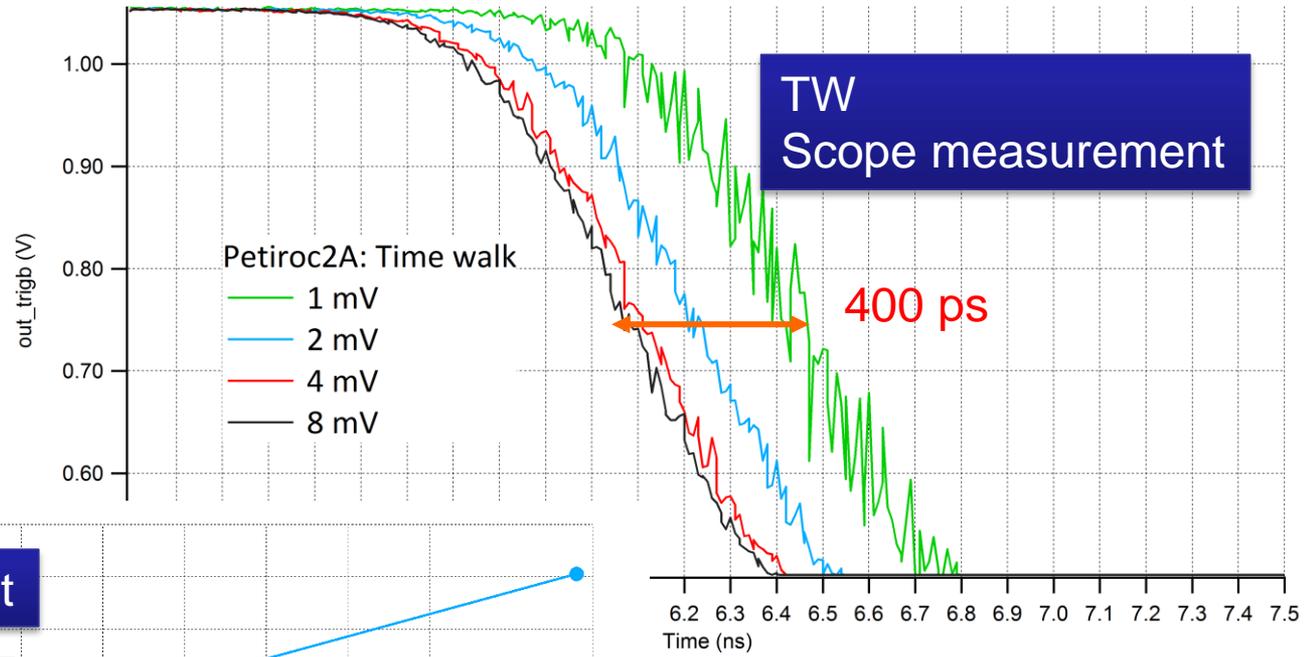
- Injection of a step (1mV, 2mV, 3mV, 4mV, 5mV and 10mV)



# JITTER MEASUREMENT

- Jitter vs threshold & injection, Jitter improves with signal
- Clock couplings: through substrate, better results expected with technologies that offer triple well trans (TSMC 130nm)
- Jitter < 40 ps for input pulse > 1mV

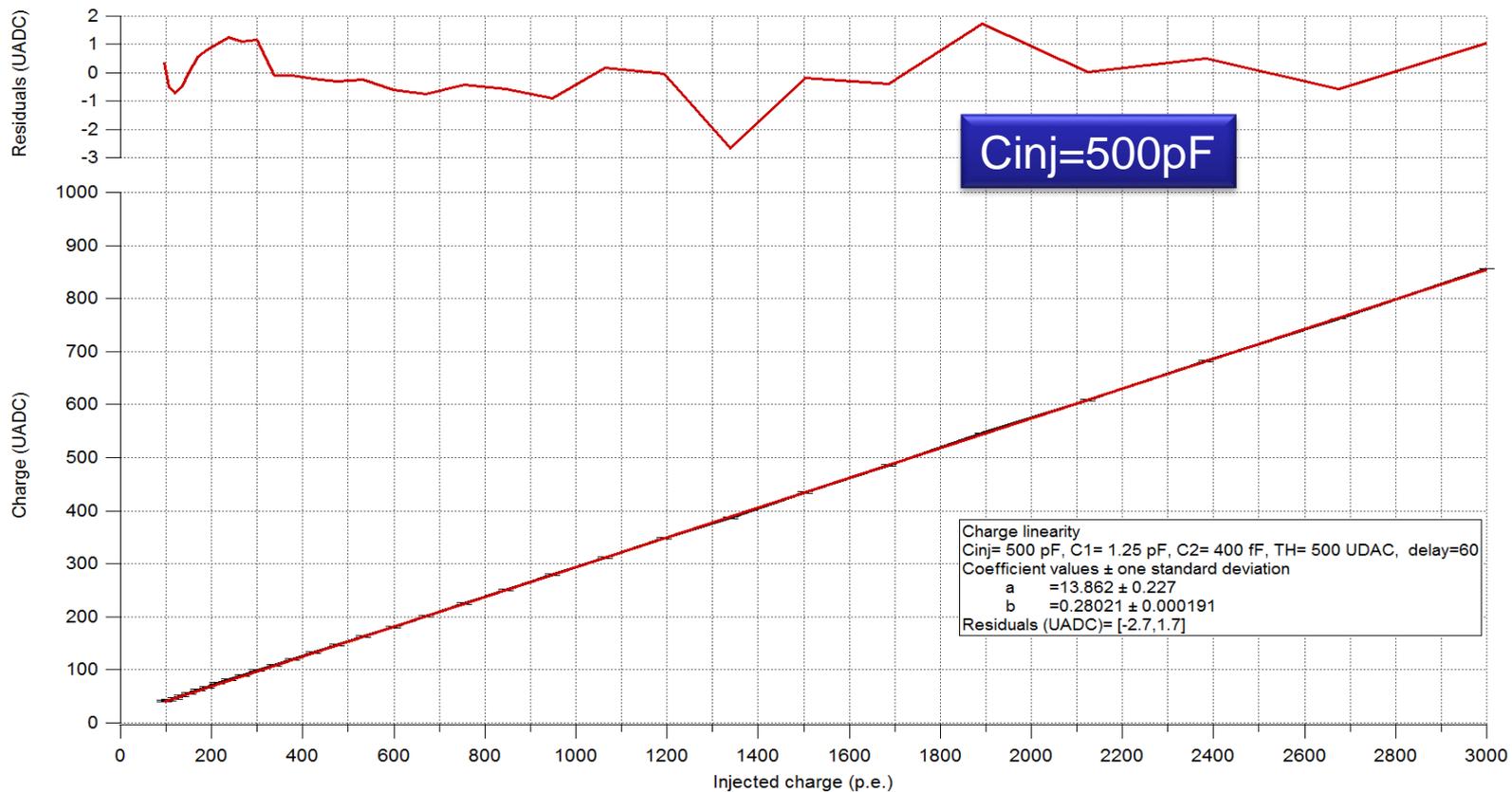




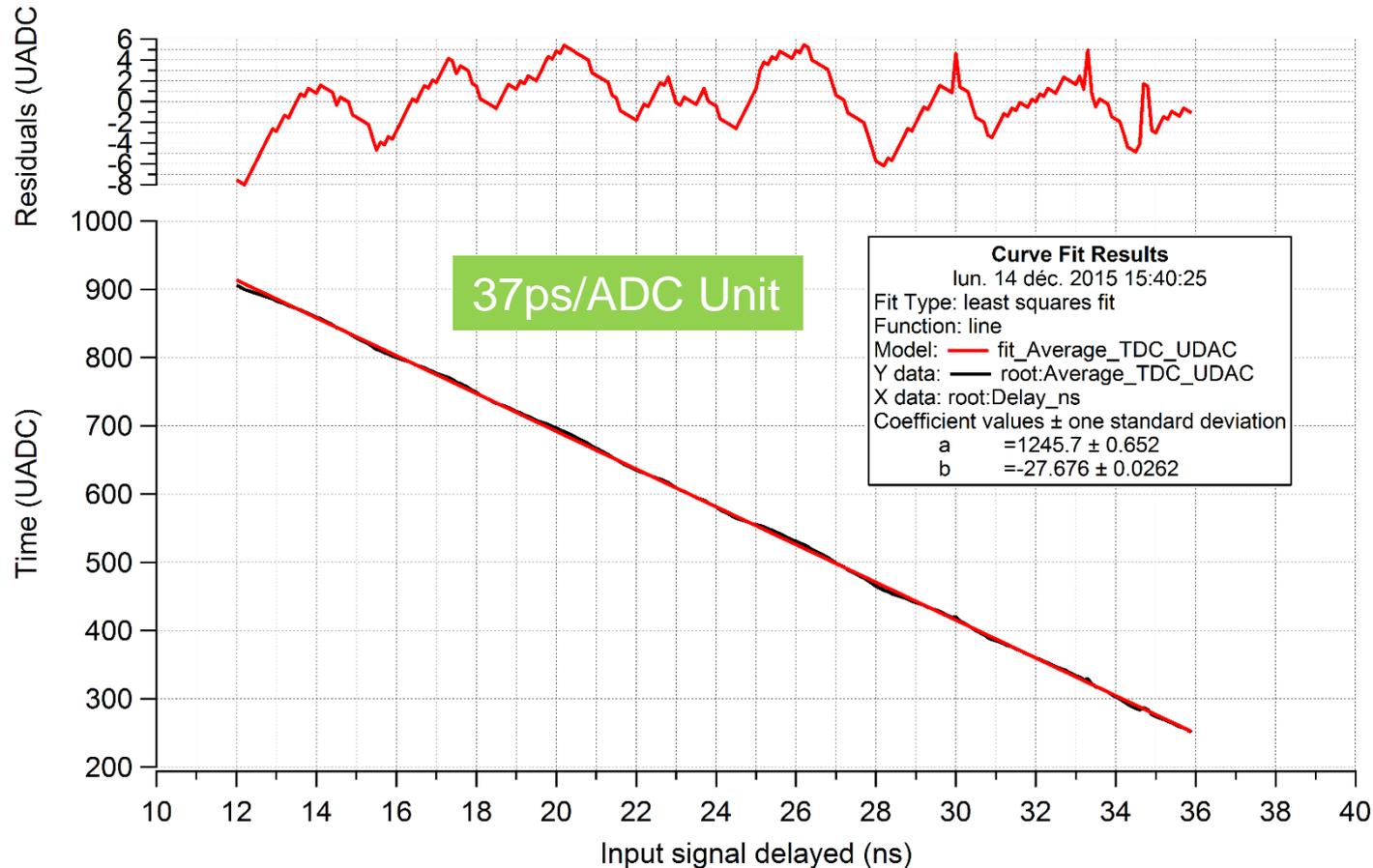
# CHARGE MEASUREMENT



- Linearity within  $\pm 3$  UADC up to 3000 p.e. (500 pC)
- Using the internal ADC, C1=1.25pF, C2=400 fF
- measurements made with C<sub>inj</sub> = 500 pF

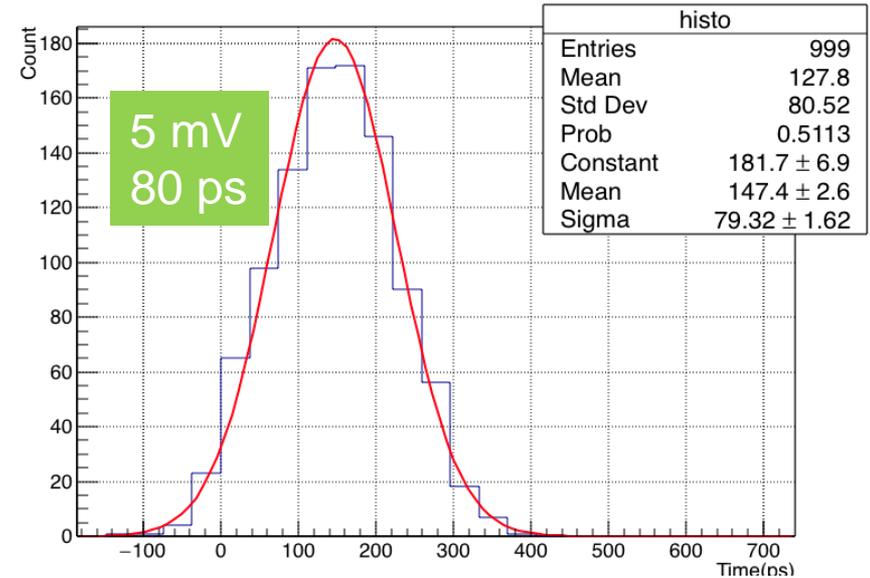


- 160 MHz clock seen on the TDC (residuals)
  - *rms* of the residuals: 2.6 ADC Unit
- ⇒ Time resolution:  $2.6 \times 37\text{ps}$  (step) = 100 ps

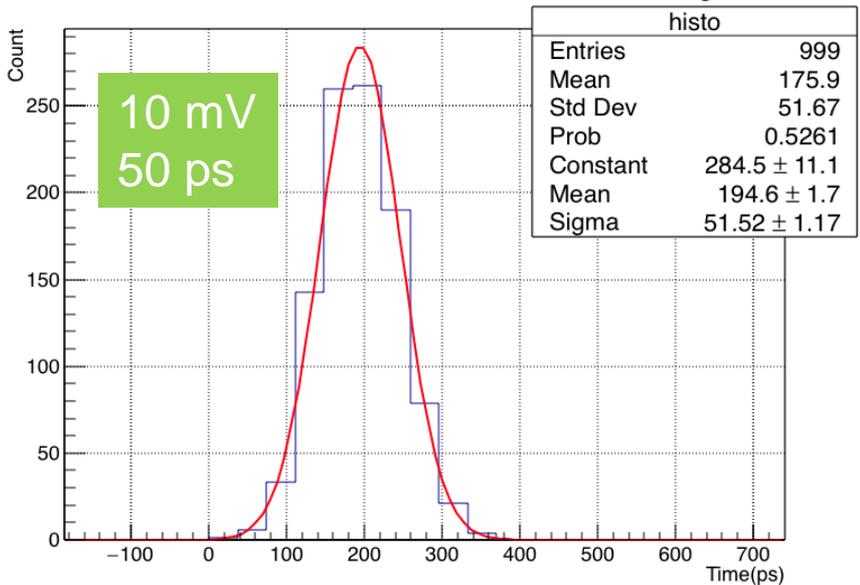


- Jitter measurement with innternal TDC
- Synchronous pulse injection
- Remove TDC non linearity
- Constant term :  $\sim 50$  ps

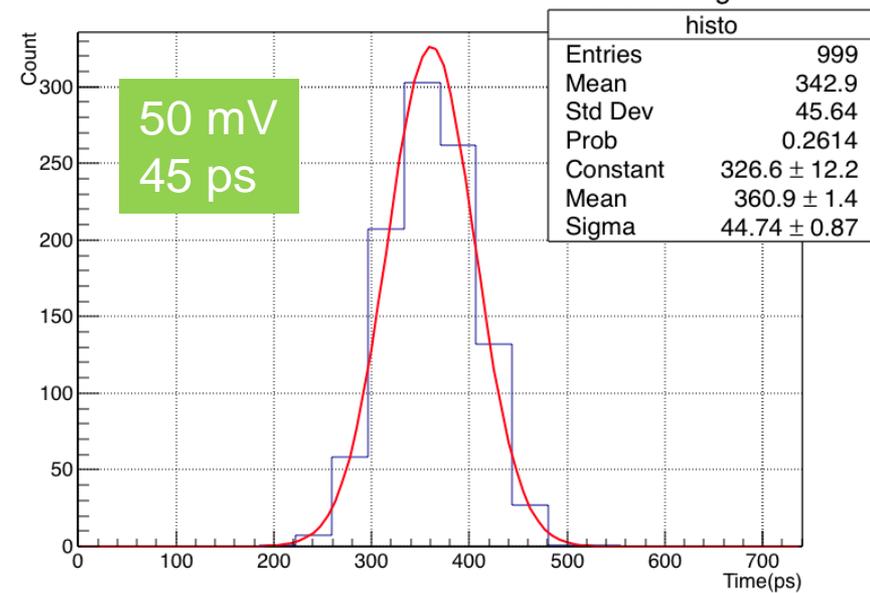
Ch4 & Ch6 - difference in coincidence timing



Ch4 & Ch6 - difference in coincidence timing



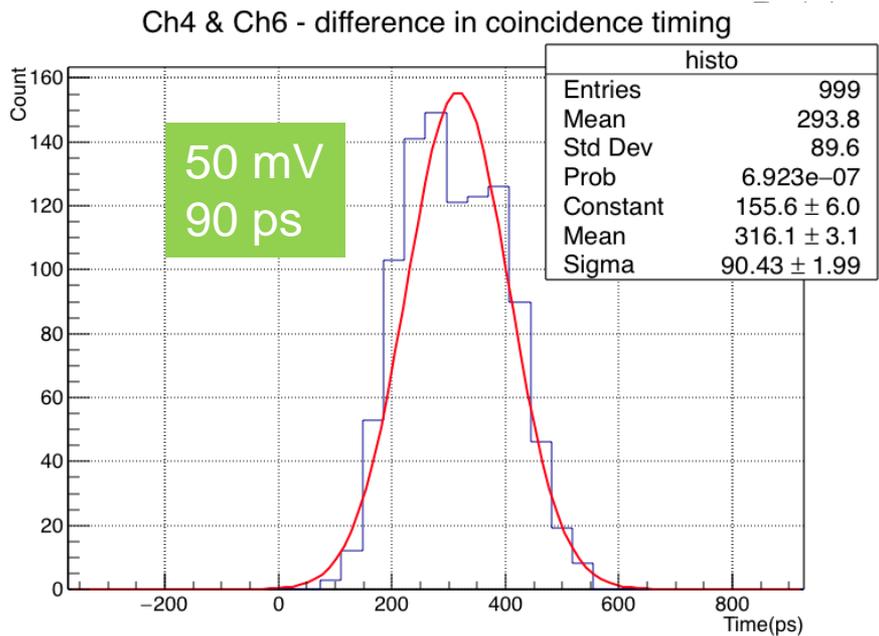
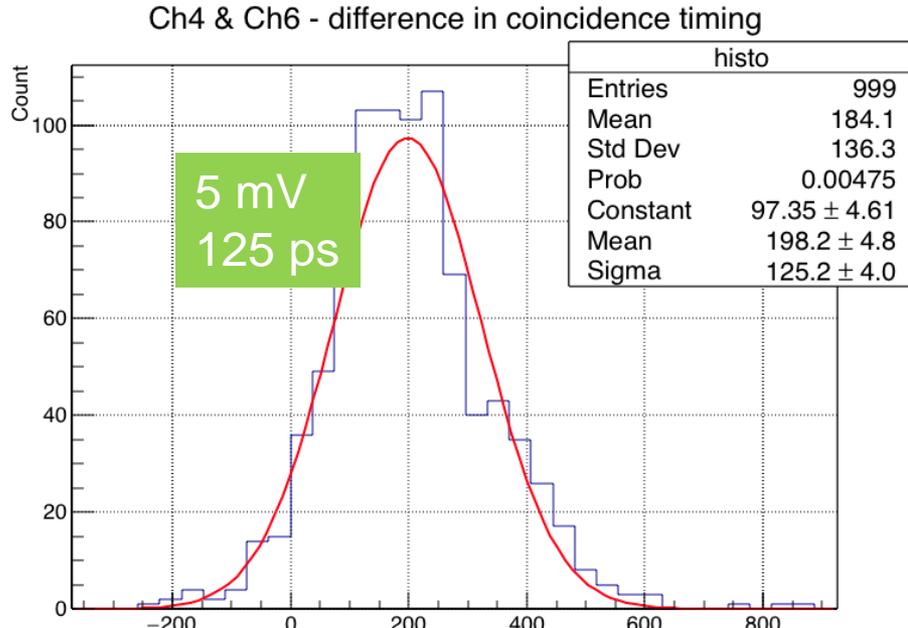
Ch4 & Ch6 - difference in coincidence timing



# COINCIDENCE with Asynchronous CK



- Jitter measurement with innternal TDC
- Asynchronous pulse injection
- Suffers from TDC non linearity
- Constant term :  $\sim 90$  ps



@ I. Laktineh PNL Lyon

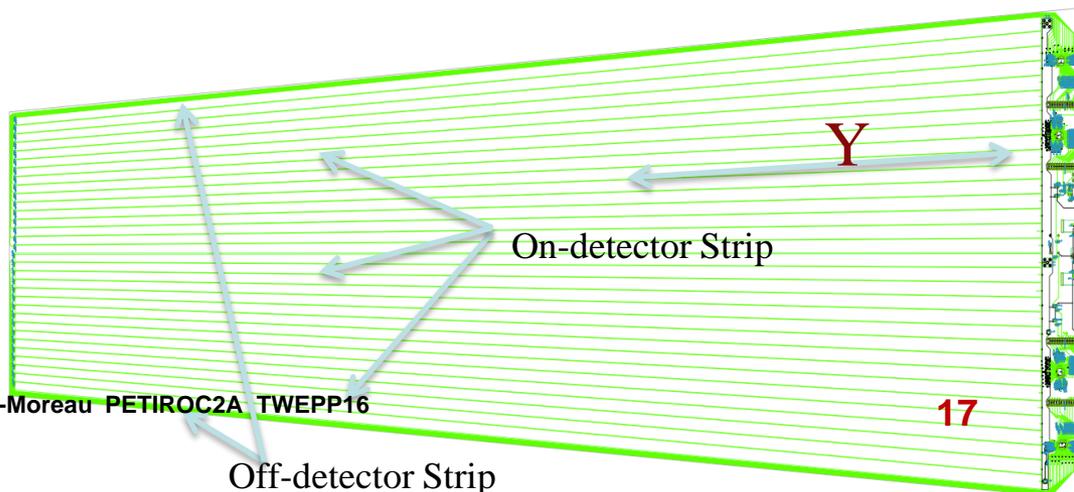
- PETIROC ASIC : 32-channel,
- high bandwidth preamp (GBWP > 10 GHz)
- < 3 mW/ch, dual time and charge measurement (160 fC-400 pC)
- very fast and low-jitter < 25 ps rms



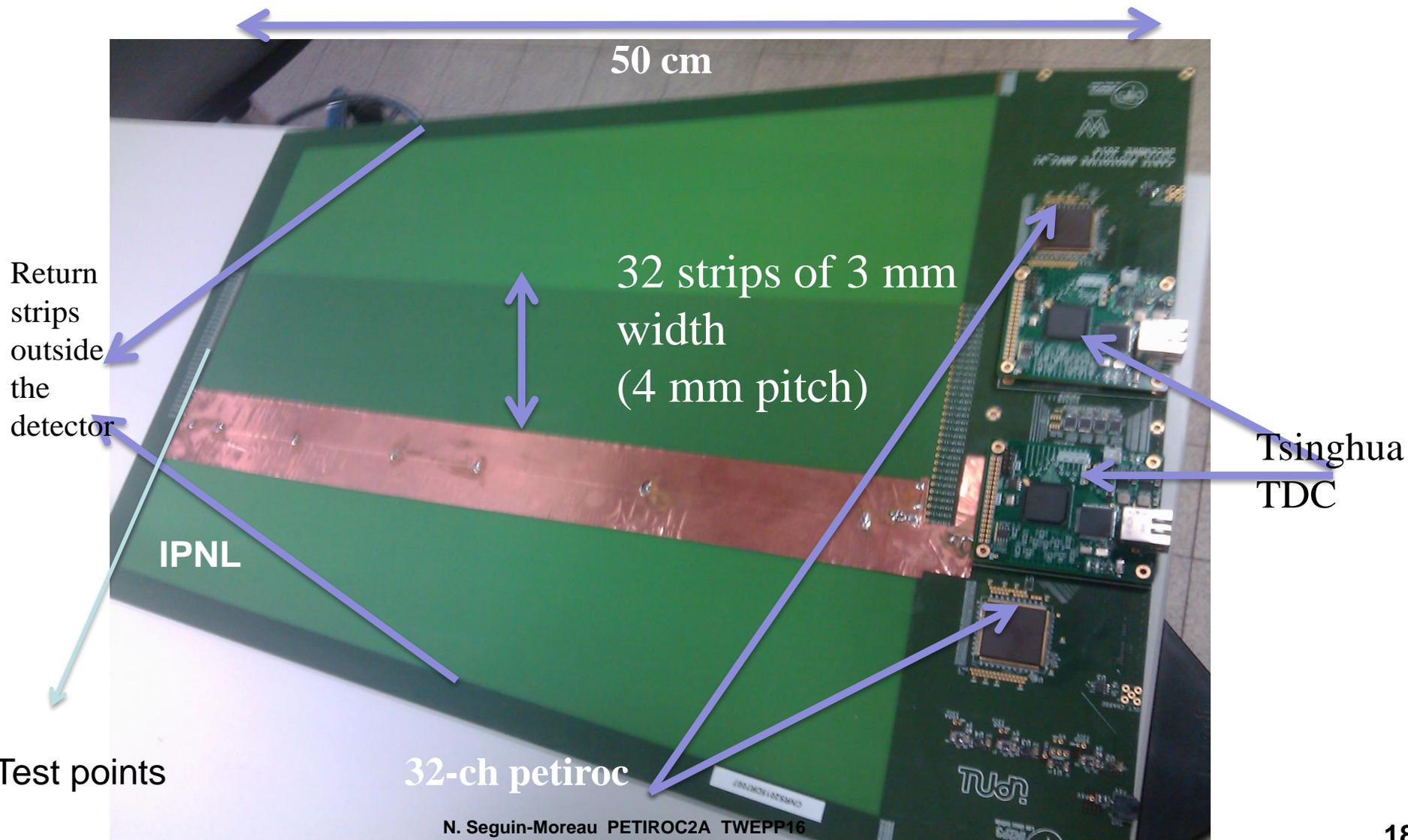
- 24-ch, TDC of 25 ps time resolution developed by the **Tsinghua university** will be used to demonstrate the RPC/MRPC time

- New PCB with pick-up strips read from both sides is being designed information with the aim to achieve Y-position determination  $Y = L/2 - v \cdot (t_2 - t_1)/2$ .

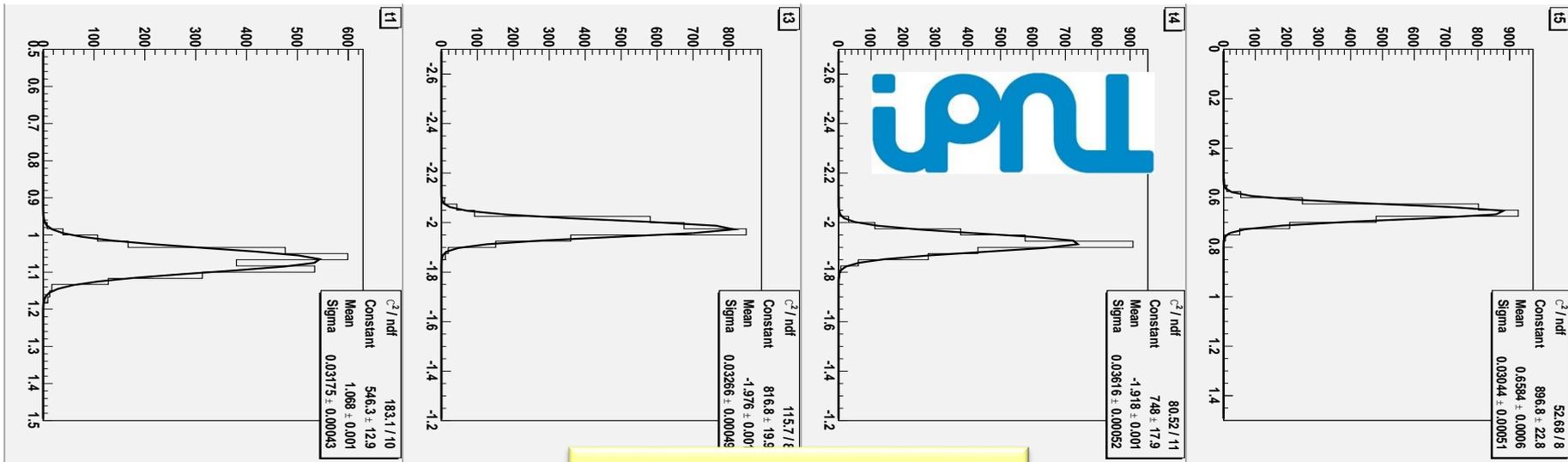
Time resolution can be measured:  
 $(t_1 + t_2) - L/v$



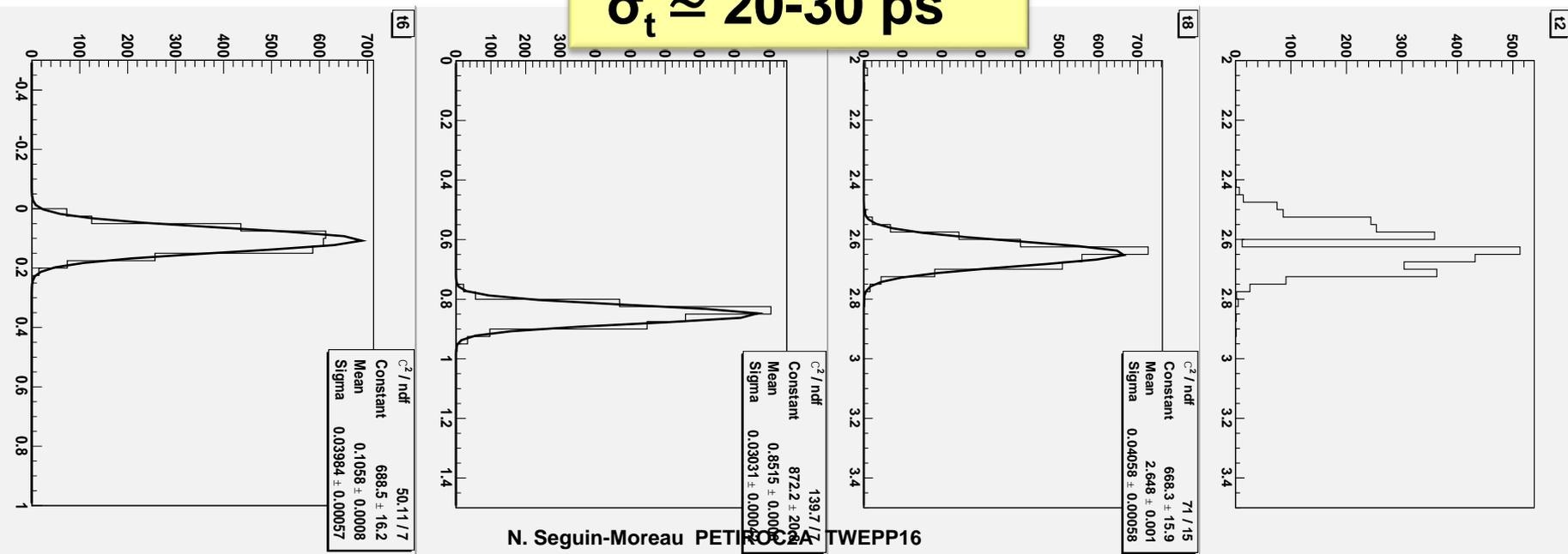
A PCB was conceived to host : Pickup strips, 2 PETIROC, 2 TDC  
A DAQ system was developed. The PCB is intended to equip large chambers



Injecting charge on test points and then recording time difference  $\Delta T = T_2 - T_1$



$\sigma_t \approx 20-30$  ps



- PETIROC2A features high speed time and charge measurement
  - 32 channels broadband amplifiers+discriminators
  - $G = 25$   $BW = 0.9$  GHz  $GBW = 20$  GHz
  - Minimum Threshold  $<1$  mV
  - Low power : 6 mW/ch
  - 10 ps jitter in analog operation (external TDC)
  - 50 – 90 ps with internal TDC (synchronous/asynchronous)



# CE and TZ frequency response for $C_d \gg 1\text{pF}$

## • CE using an OTA with gain $G_m$

- $V_{OUT} = -G_m (Z_F // Z_L) V_{IN}$

With  $Z_L$  neglected and  $Z_S = C_d$  and  $Z_F = R_F$

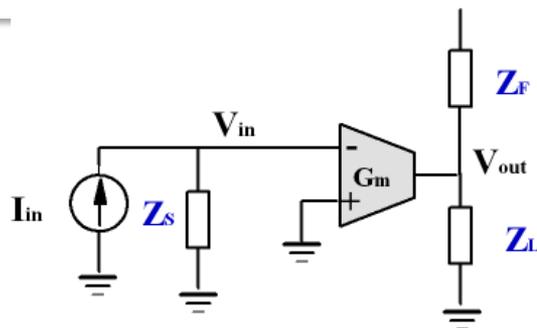
- $V_{IN} = Z_S I_{IN} = \frac{1}{j\omega C_d} I_{IN}$

- $V_{OUT} = -G_m Z_F V_{IN} = -\frac{G_m Z_F}{j\omega C_d} I_{IN}$

- $V_{OUT} = -G_m R_F \frac{Q_{IN}}{C_d}$

- $Z_{IN} = Z_S \Rightarrow C_d$

- $Z_{OUT} = Z_F = R_F$



## TZ using an OTA with gain $G_m$

- $I_{OUT} = V_{OUT} / Z_L + (V_{OUT} - V_{IN}) / Z_F$

- $I_{IN} = V_{IN} / Z_S - (V_{OUT} - V_{IN}) / Z_F$

- $I_{OUT} = -G_m V_{IN}$

With  $Z_L$  neglected and  $Z_S = C_d$  and  $Z_F = R_F$

- $I_{IN} = \frac{V_{in}}{Z_S} - I_{OUT}$

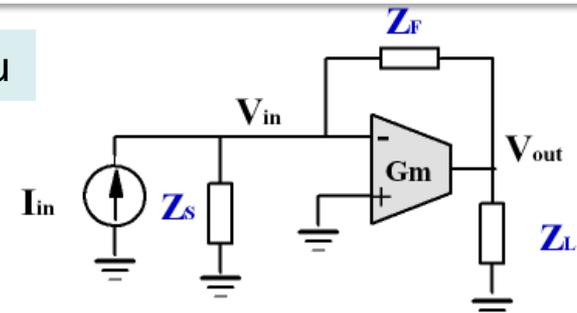
- $V_{IN} = \frac{I_{IN}}{\left(\frac{1}{Z_S} + G_m\right)} = \frac{\frac{1}{G_m}}{1 + j\omega \frac{C_d}{G_m}} I_{IN}$

- $V_{OUT} = (1 - G_m Z_F) V_{IN}$

- $V_{OUT} = \frac{\frac{1}{G_m} - R_F}{1 + j\omega \frac{C_d}{G_m}} I_{IN} \approx -G_m R_F \frac{I_{IN}}{j\omega C_d} = -G_m R_F \frac{Q_{IN}}{C_d}$

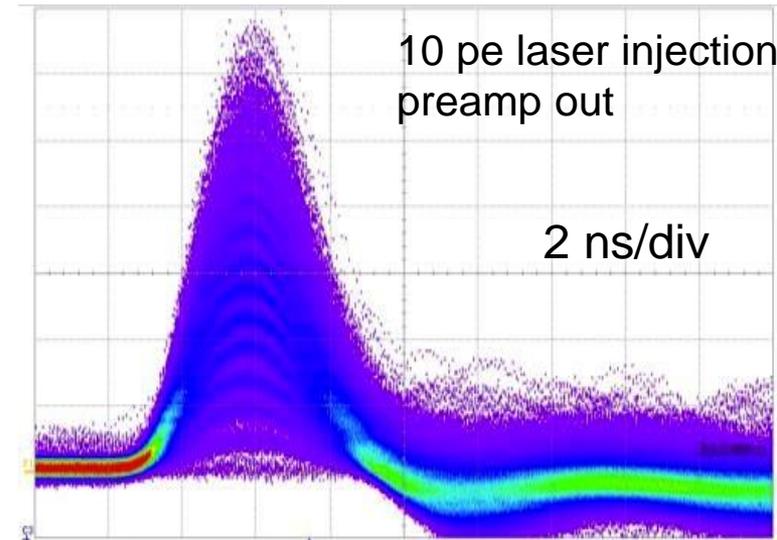
- $Z_{IN} = \frac{\frac{1}{G_m}}{1 + j\omega \frac{C_d}{G_m}} \Rightarrow 1 / G_m \text{ in } // \text{ with } C_d$

- $Z_{OUT} = \frac{R_F}{1 + G_m R_F} \sim \frac{1}{G_m}$



@Nathalie Seguin-Moreau

- 1x1mm SiPM Hamamatsu
- Laser for low light injection
  - 405nm
  - Jitter : 28 ps FWHM
- Low trigger mandatory for good timing resolution
- Petiroc can trigger on first photoelectron
- Petiroc is low noise : single photon identification



*Thanks to David Brasse and Jacques Wurtz- IPHC – IN2P3  
For their kind support for the SiPM-based measurements*

