

Electronics, Trigger and Data Acquisition part 1

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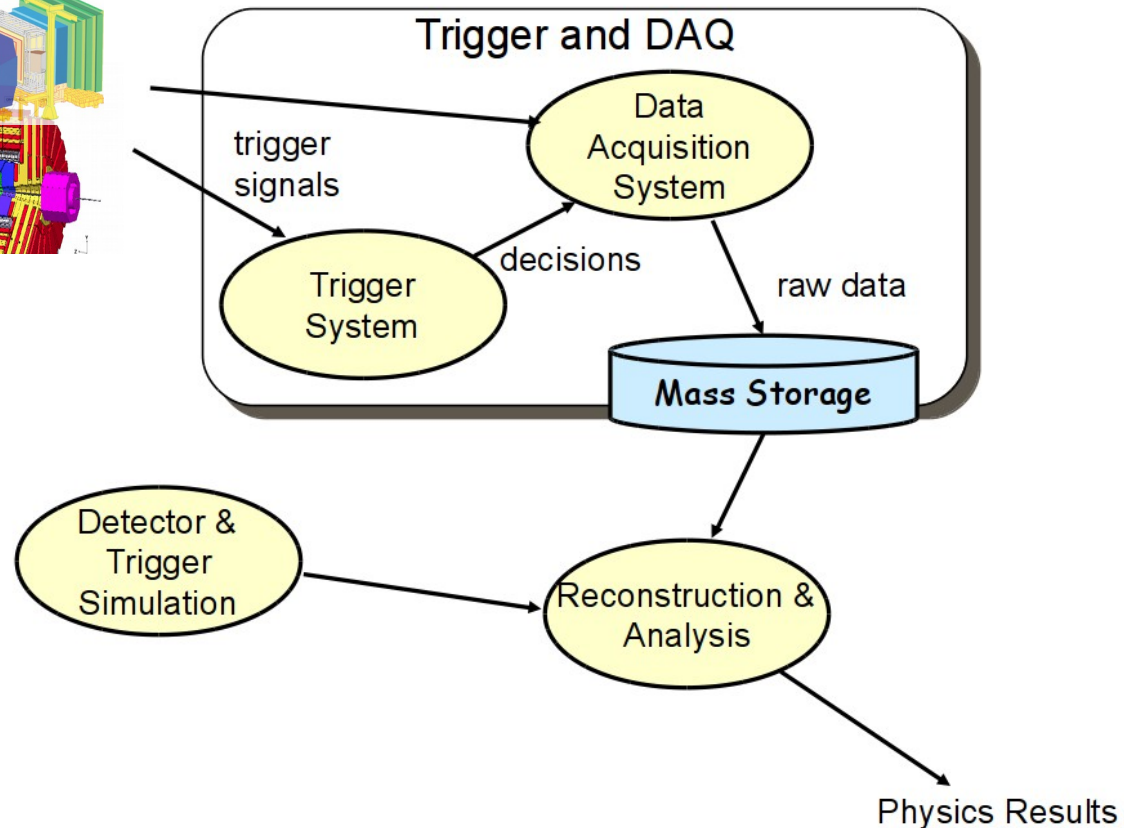
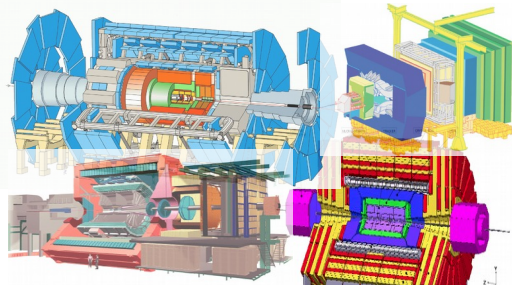
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Introduction

- Lectures will be concerned with Data Acquisition (DAQ) in High-Energy Physics ...
- Data acquisition is an **alchemy** of electronics, computer science, networking, physics
 - ..., resources and manpower matter as well, ...
- Topics are pretty much correlated → you will realise this in the lecture non-linearity

Credits: material and ideas come largely from my predecessor (Wainer Vandelli) and from the lectures of ISOTDAQ schools (<http://isotdaq.web.cern.ch/isotdaq/isotdaq/Home.html>)

General Overview



- main role of Electronics, Trigger & DAQ is to process signals generated in “a detector”, likely storing all useful information in some safe place

Signals ?

- Sometimes, somewhere in our detector, something (“one event”) happens, i.e. in some short time several particles interact within it.
- In High-Energy Physics, even a single event is composed by sequences of many different probabilistic (quantum-mechanics) processes → fluctuations are built-in
- At the end, “electrical” signals arrive at detector output terminals, different signals:
 - a) have different characteristics (size, arrival time, duration, ...)
 - b) carry different, (likely) independent, information
 - c) need quite some **electronics** in order to become “**profitable**”

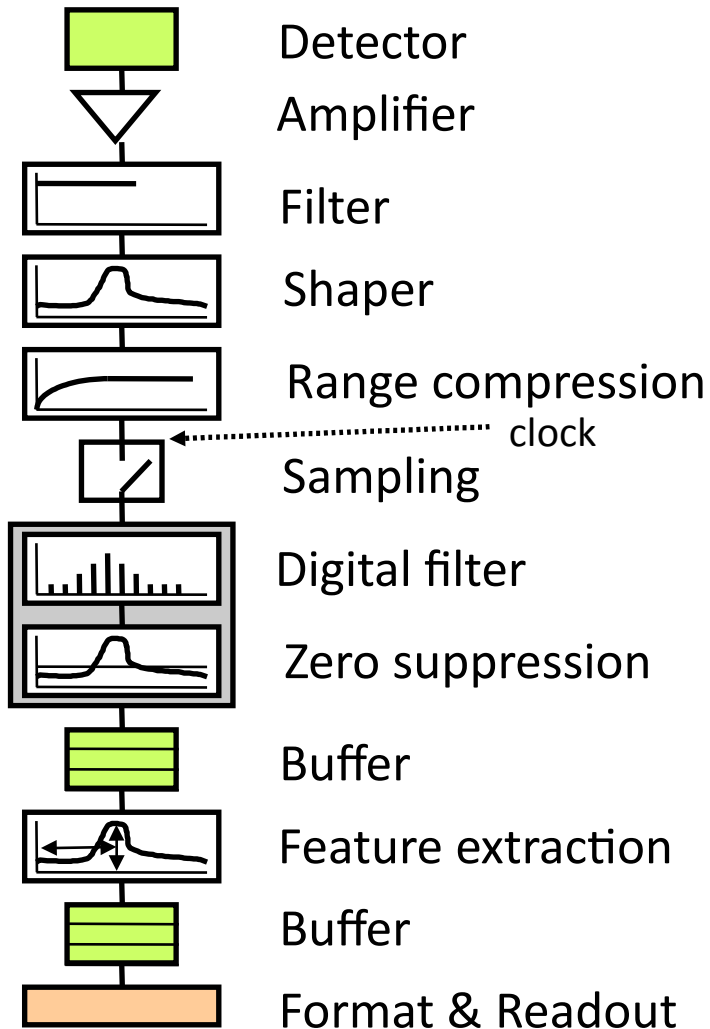
Electronics

[Horowitz & Hill, The Art of Electronics]

Role ?

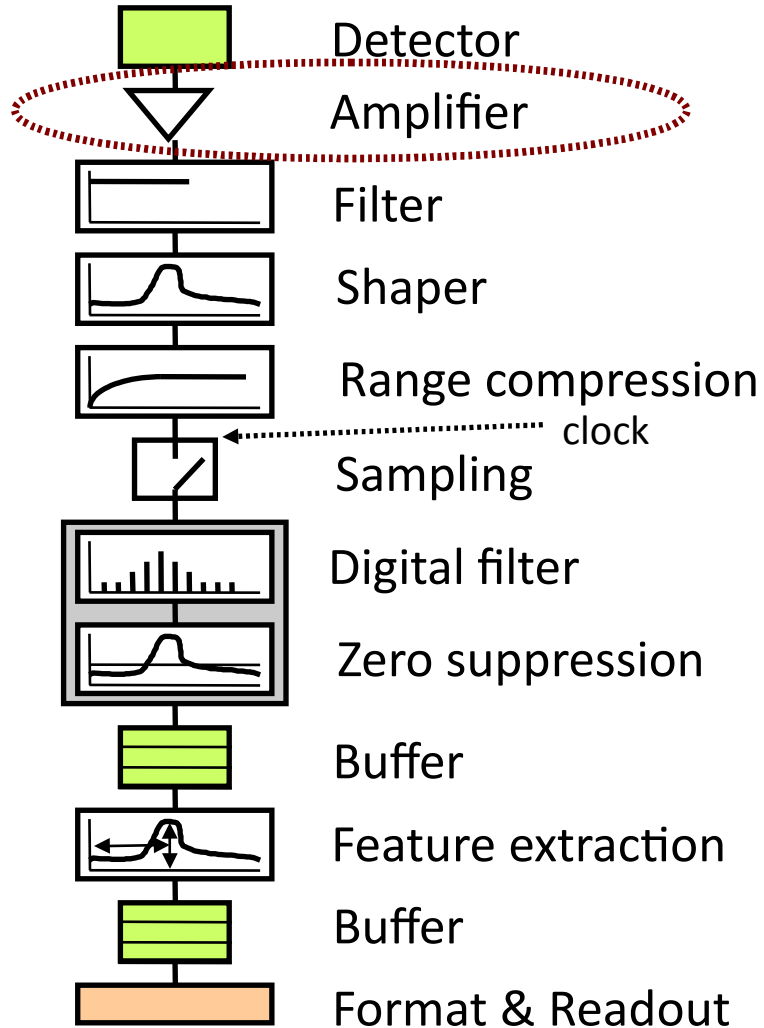
- Sense, transform and collect electrical signals from detector(s) – very often short current/light pulses (i.e. bunches of electrons or photons)
 - you may be interested in total charge or in signal time or in both
- Adapt signals to optimize different, **conflicting**, characteristics → balance:
 - minimum detectable signal (min. noise and max. signal-to-noise ratio)
 - maximum detectable signal (dynamic range)
 - speed (signal rate)
 - timing
 - pulse shape dependence
- Digitize and preserve information
 - allow for subsequent processing, transmission, storage using digital electronics
→ computers, networks, ...

Read-Out Chain



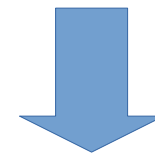
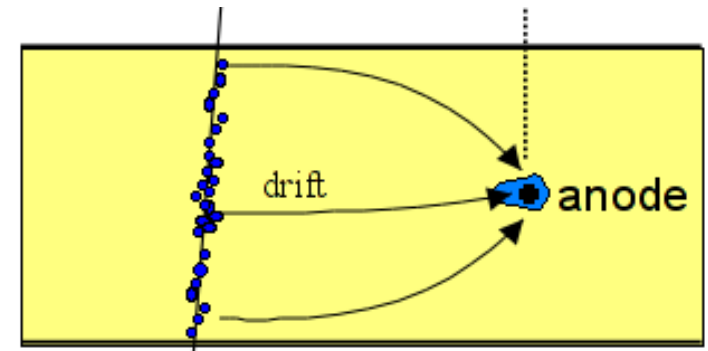
- Front-end electronics very specialized
 - custom-built to match detector characteristics
- Cannot discuss all design and architecture details
 - if you are into electronic design you already know more than me
- Find yourself dealing or choosing commercial electronics
 - provide you with base guidelines
- Selected functions and principles

Read-Out Chain

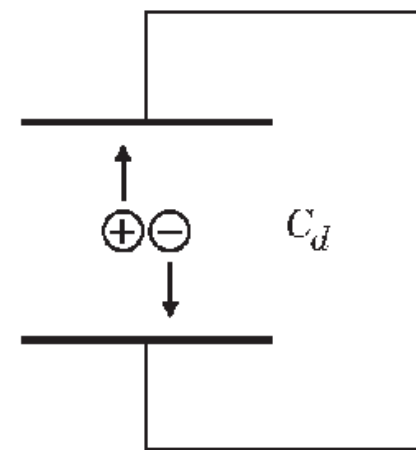


What can a Signal be ?

- Detectors may be electrically represented as a capacitor (C_d)
 - more realistic schemes will include other contributions
- Interactions of passing particle \rightarrow energy release $E \rightarrow$ short current pulse i_s



DETECTOR



$$E \propto Q_s = \int i_s(t) dt$$

- Pulse duration may range from $O(100 \text{ ps})$ up to $O(10 \mu\text{s})$

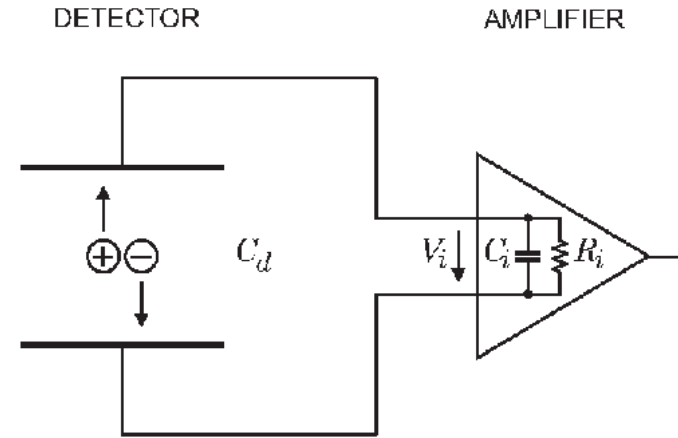
Voltage-Sensitive Amplifier

- Signals are possibly very small, amplify to:
 - improve signal resolution, adapt it to next stages
 - avoid SNR degradation ...

Using a simple voltage amplifier, the sensed **input voltage V_i** depends on the **detector capacitance**.

Detector capacitance could be a function of the operation point (e.g. high voltage) and/or detector dimension.

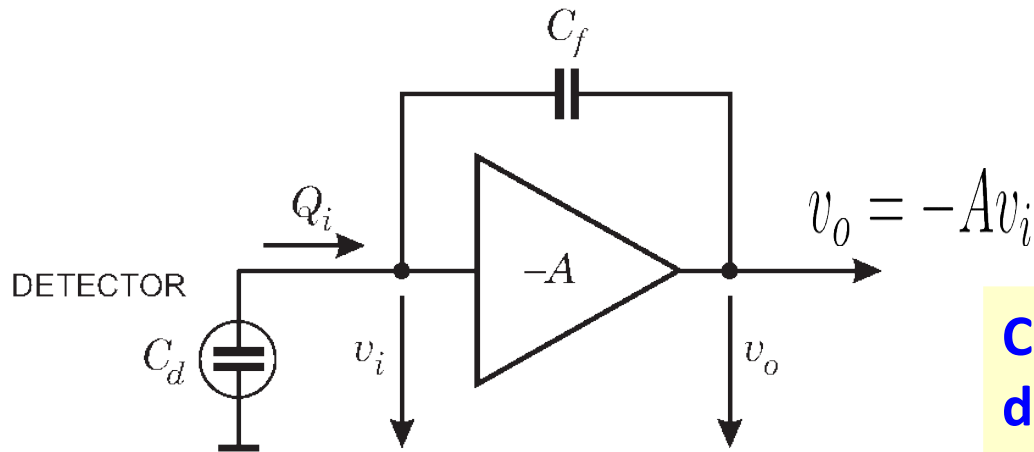
Additional **calibration** efforts



$$V_i = \frac{Q_s}{C_d + C_i}$$

A red arrow points from the text 'Additional calibration efforts' to the C_d term in the denominator of the equation.

Charge-Sensitive Amplifier



Charge amplification only depends on a well-controlled component

Effective (dynamic) input capacitance:

$$C_i = \frac{Q_i}{v_i} = C_f (A + 1)$$

Gain:

$$A_Q = \frac{v_o}{Q_i} = \frac{Av_i}{C_f(A+1)v_i} = \frac{A}{A+1} \frac{1}{C_f} \approx \frac{1}{C_f} \quad (A \gg 1)$$

Charge Transfer

$$C_i = \frac{Q_i}{v_i} = C_f (A + 1)$$

What total-charge fraction is measured ?

$$\frac{Q_i}{Q_s} = \frac{C_i v_i}{Q_{\text{det}} + Q_i} = \frac{C_i}{Q_s} \cdot \frac{Q_s}{C_i + C_{\text{det}}} = \frac{1}{1 + \frac{C_{\text{det}}}{C_i}} \approx 1 \quad (\text{if } C_i \gg C_{\text{det}})$$

Need **large input capacitance** to maximize charge transfer

Signal-to-Noise Ratio (SNR)

- Improving SNR improves sensitivity (minimum detectable signal)
- Electronic noise does not necessarily dominate each measurement

$$\Delta E = \sqrt{\Delta E^2_{fluc} + \Delta E^2_{noise}}$$

Fluctuations due to physical detection process (e.g. energy deposition) or detector readout (e.g. photomultiplier)

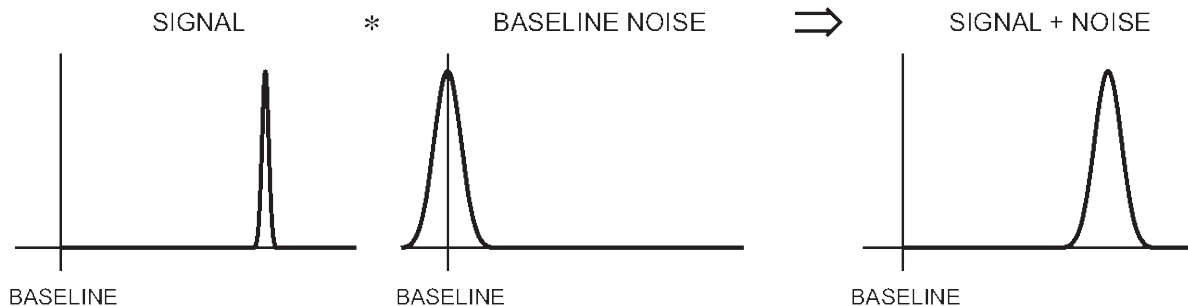
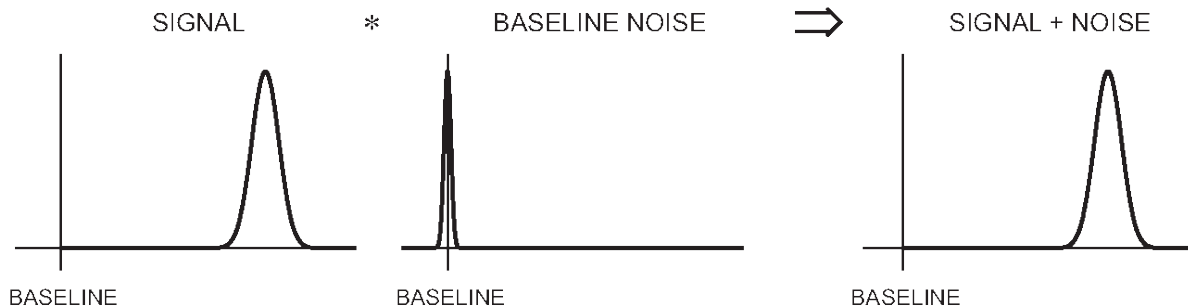
Electronic noise

- Thermal noise: velocity fluctuations in charge carriers
- Shot noise: fluctuations in carrier number (e.g. diode barrier crossing)

Signal-to-Noise Ratio (SNR)

- Improving SNR improves sensitivity (minimum detectable signal)
- Electronic noise does not necessarily dominate each measurement

$$\Delta E = \sqrt{\Delta E^2_{fluc} + \Delta E^2_{noise}}$$



S/N needs optimization

SNR .vs. Detector Capacitance

voltage-sensitive amplifier

→ given signal charge Q_s

$$V_s = \frac{Q_s}{C_d + C_i}$$

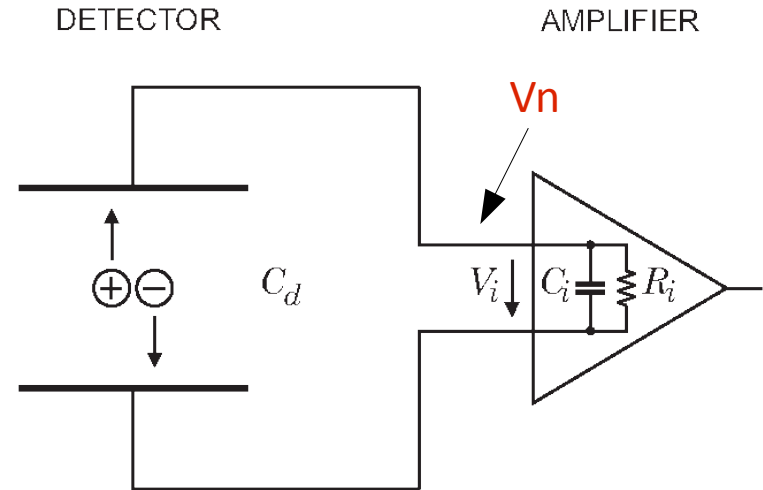
→ assuming input noise V_n

$$\frac{V_s}{V_n} = \frac{Q_s}{V_n \underbrace{(C_d + C_i)}}{}$$

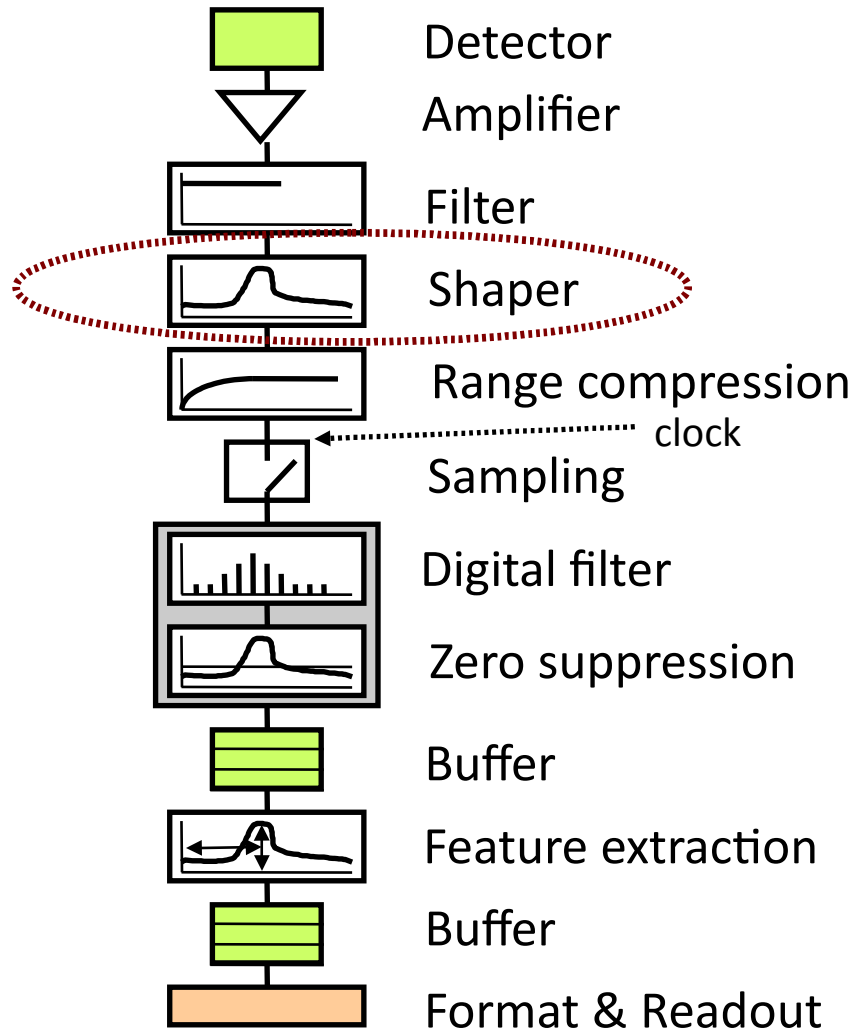
SNR inversely proportional to total input capacitance



Thick detectors normally provide larger signals **and noise**

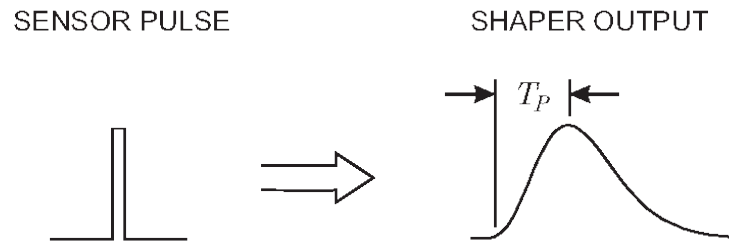


Read-Out Chain

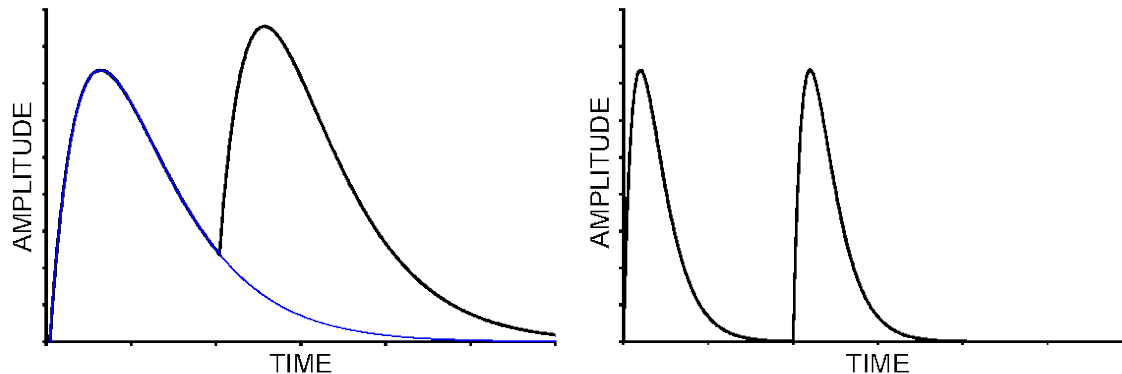


Pulse Shaping

- Reduce signal bandwidth (low-pass filter) → improve SNR
 - fast rising signals have large bandwidth
 - shaper broadens signals

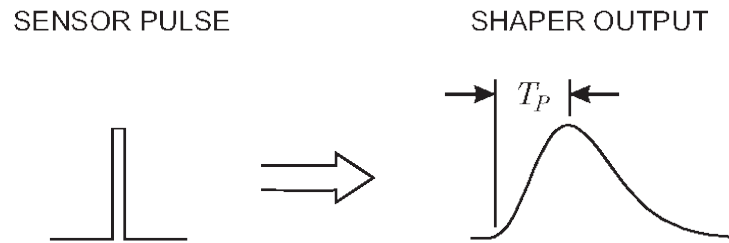


- Limit pulse width (high-pass filter) → avoid overlap of successive pulses
 - increase maximum signal rate at the cost of more noise

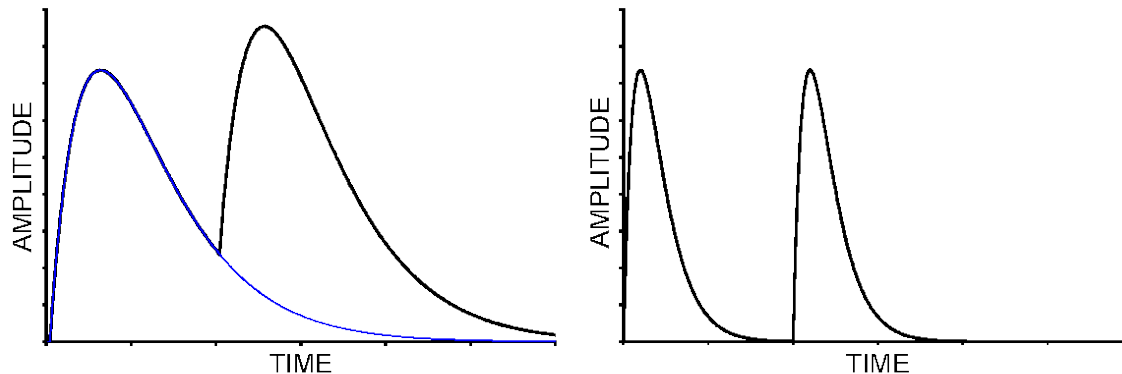


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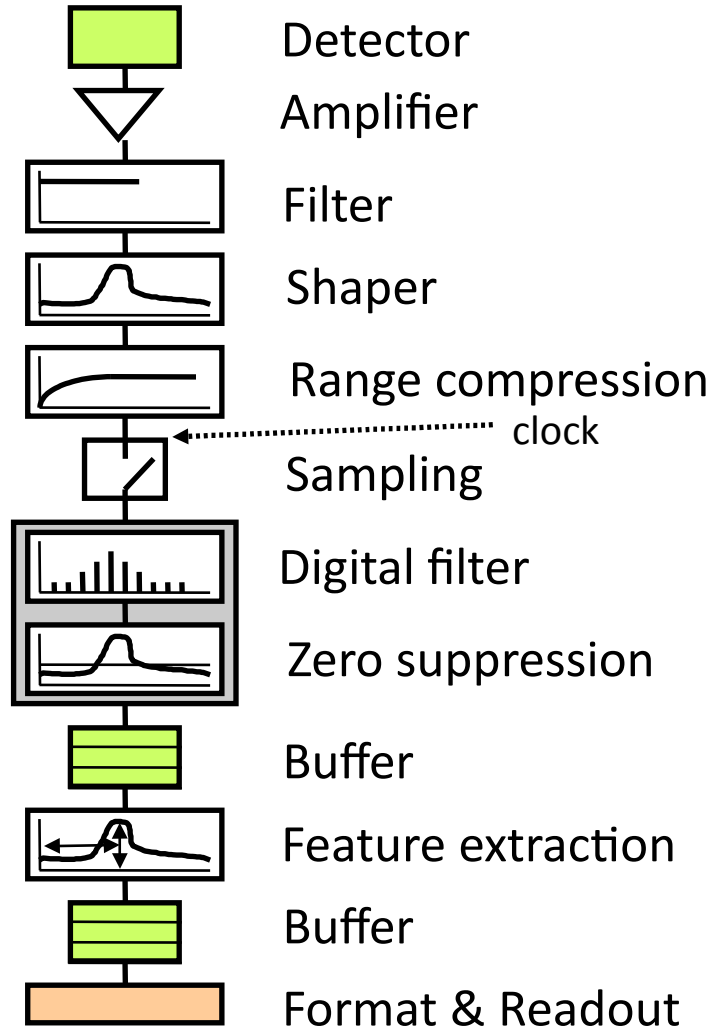
- Limit pulse width (high-pass filter) → avoid overlap of successive pulses
 - increase maximum signal rate at the cost of more noise



Shaping is often a compromise

$CR-(RC)^n$ shaping is often a good compromise

Read-Out Chain

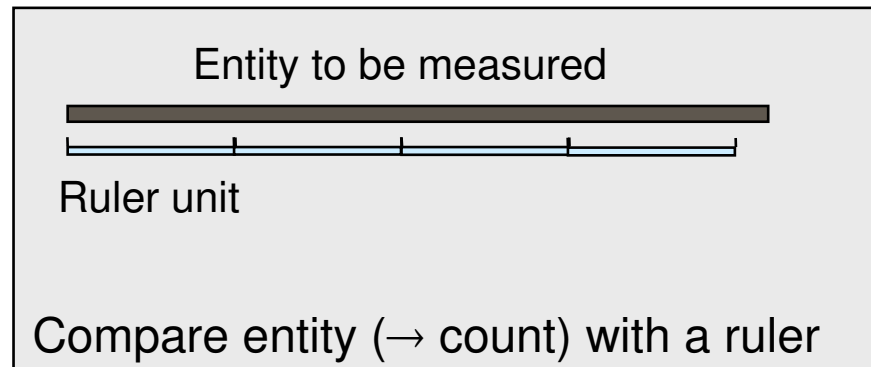


Analog-to-Digital Conversion Introduction

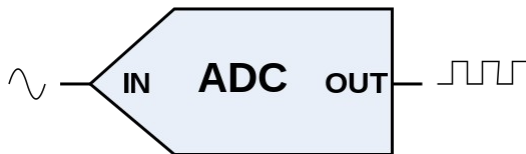
Digitization → encode analog information into a numerical (discrete) format

Why ? Much easier to handle and preserve

Which format ? Usually binary, sometimes BCD



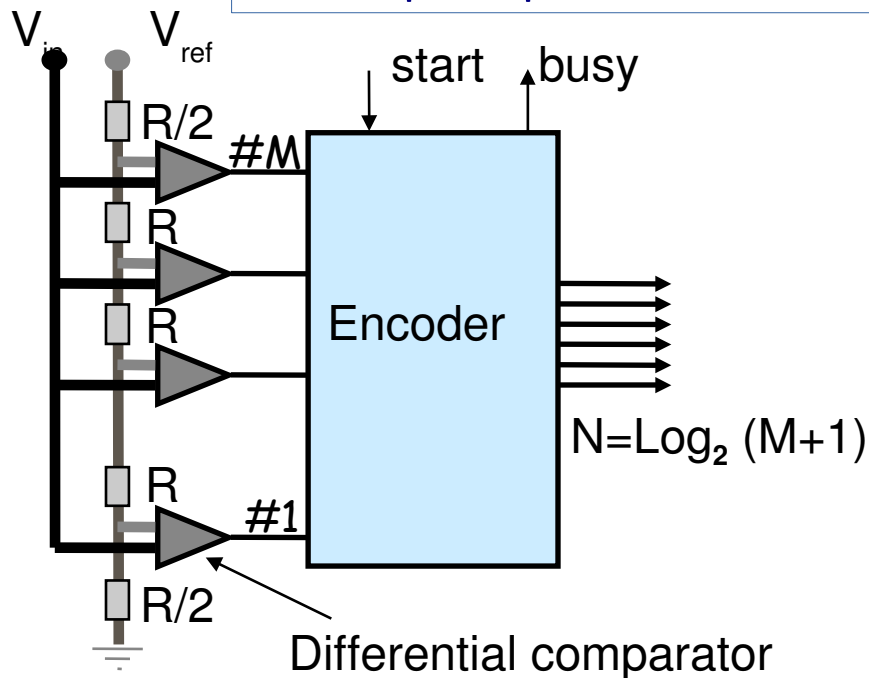
Electrical Symbol:



Allow further processing and storage in digital electronics
and **computers**

Flash ADC

- Simplest and fastest implementation
- Performs M comparisons in parallel
 - input V compared with M fractions of a reference voltage:
 $0.5 * V_{ref}/M, 1.5 * V_{ref}/M, \dots (M-0.5) * V_{ref}/M$
in step of V_{ref}/M
- Result encoded into compact N -bit binary value
- Sample input at fixed rate and fixed sample depth

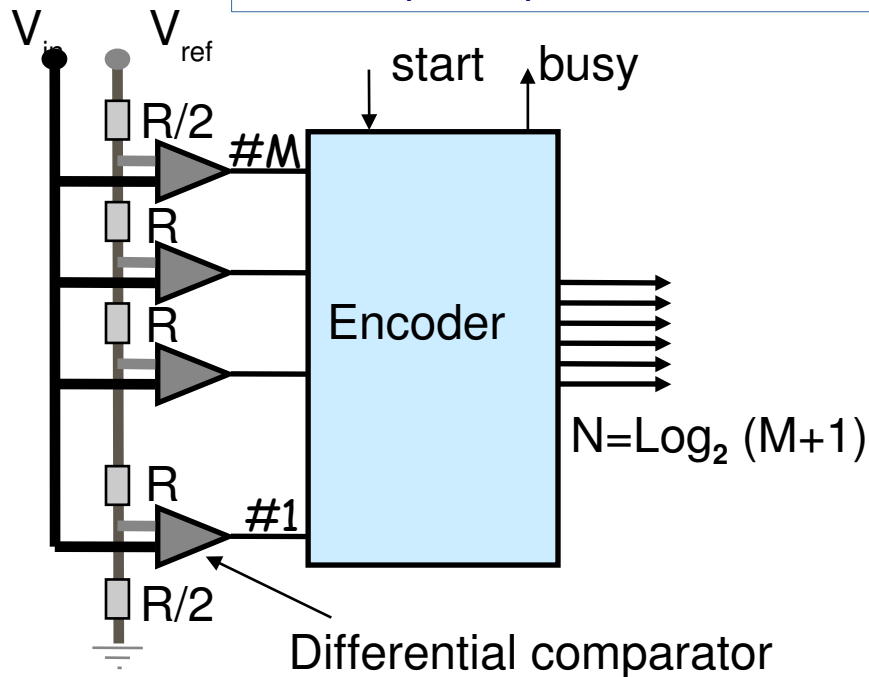


- Simple ? Yes
- Fast ? Yes
- Could be non-linear ? Yes
- Cheap ? Oopss ...

Flash ADC

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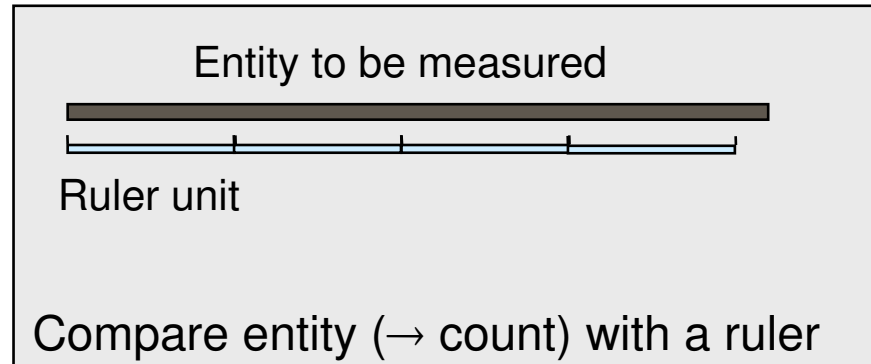


Example $M=3 \rightarrow N=2$

V_{in}/V_{ref}	Comparison results	Encoded form
$<1/6$	000	00
$1/6 \leq <3/6$	001	01
$3/6 \leq <5/6$	011	10
$5/6 \leq$	111	11

ADC Characteristics

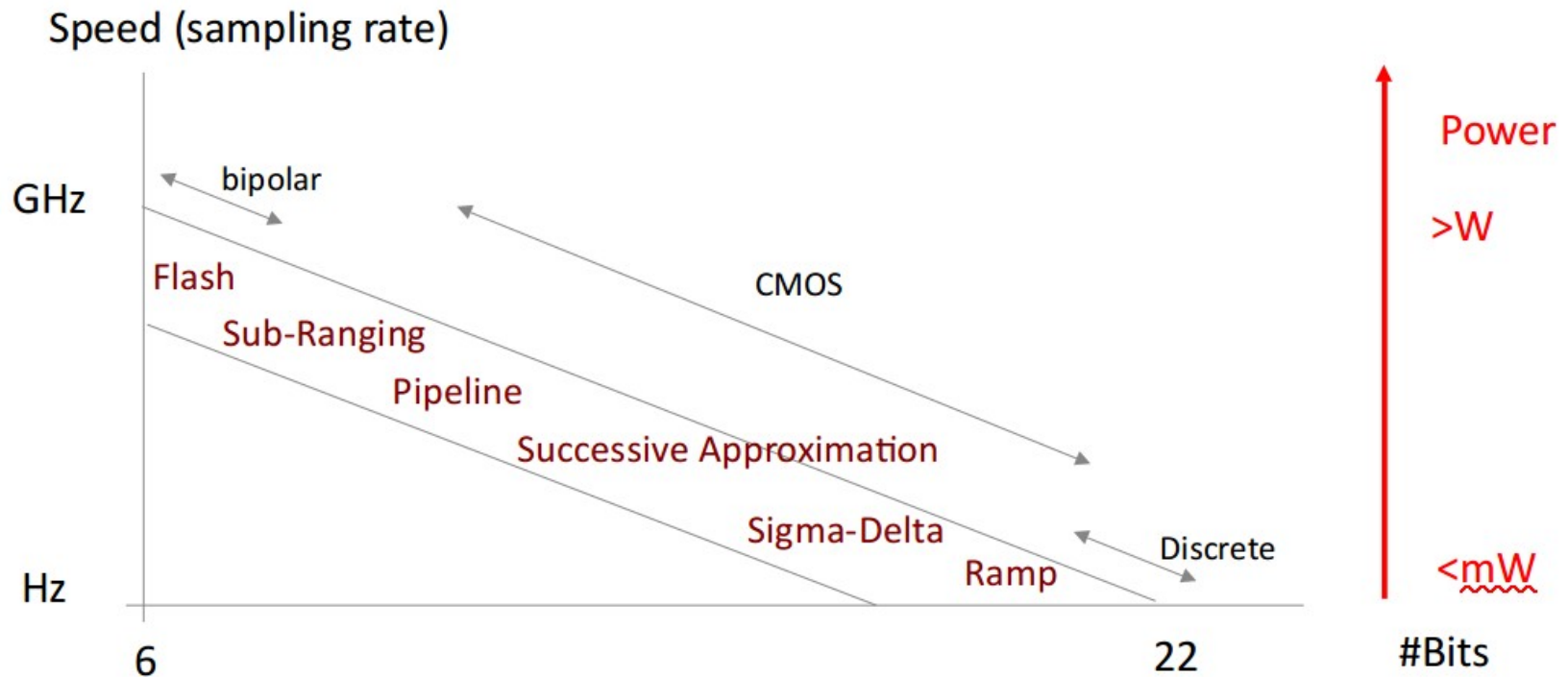
Digitization → encode analog information into a numerical (discrete) format



- Resolution (LSB), the ruler unit: $V_{\max}/2^N$
 - 8bit, 1V → LSB=3.9mV
- Quantization error: $\pm\text{LSB}/2$
- Dynamic range: V_{\max}/LSB
 - N for linear ADC
 - >N for non-linear ADC; if logarithmic:
constant relative resolution on the valid input range

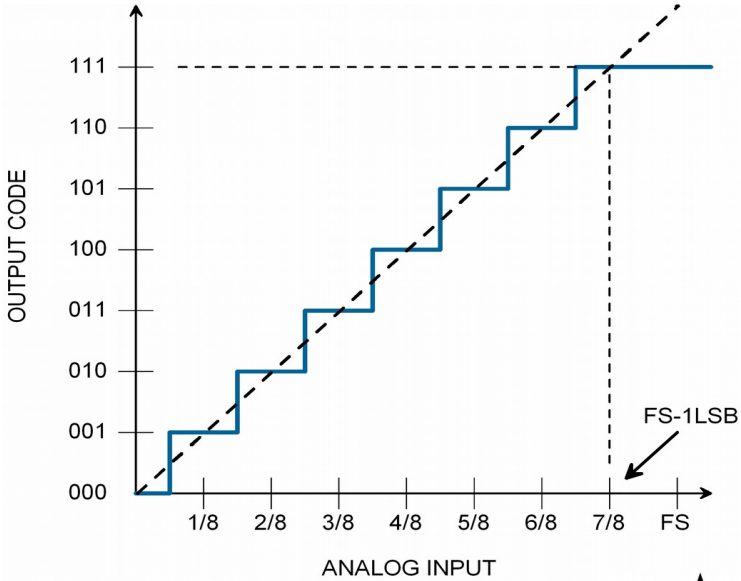
ADC Phase-Space

Many different ADC techniques exist, with different trade-offs between speed, resolution, power consumption, cost, ...

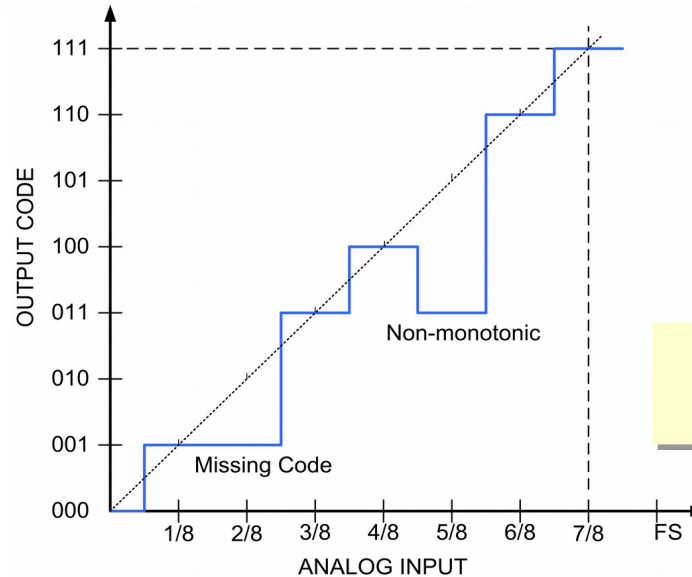
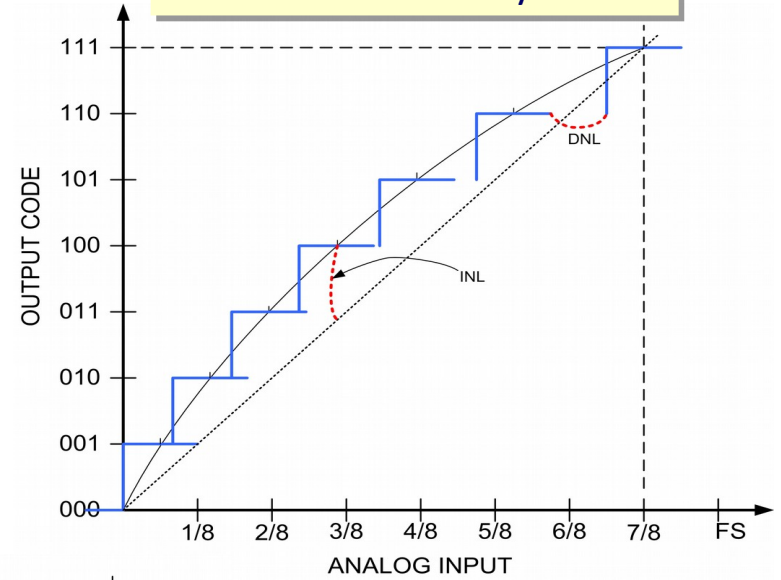


ADC (In)Accuracies

ADC Transfer function



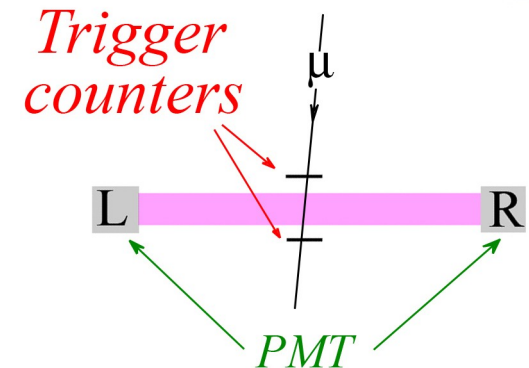
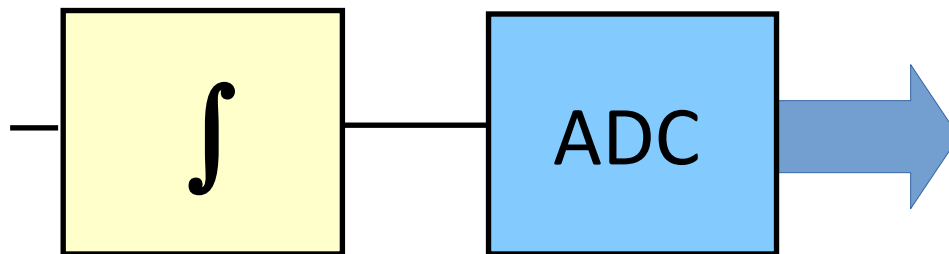
Integral & Differential Non-Linearity



Extreme Differential Non-Linearity

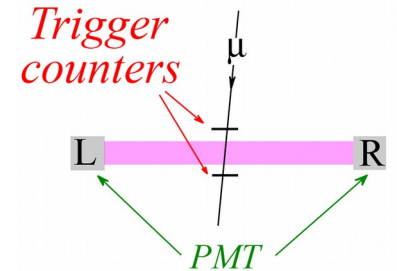
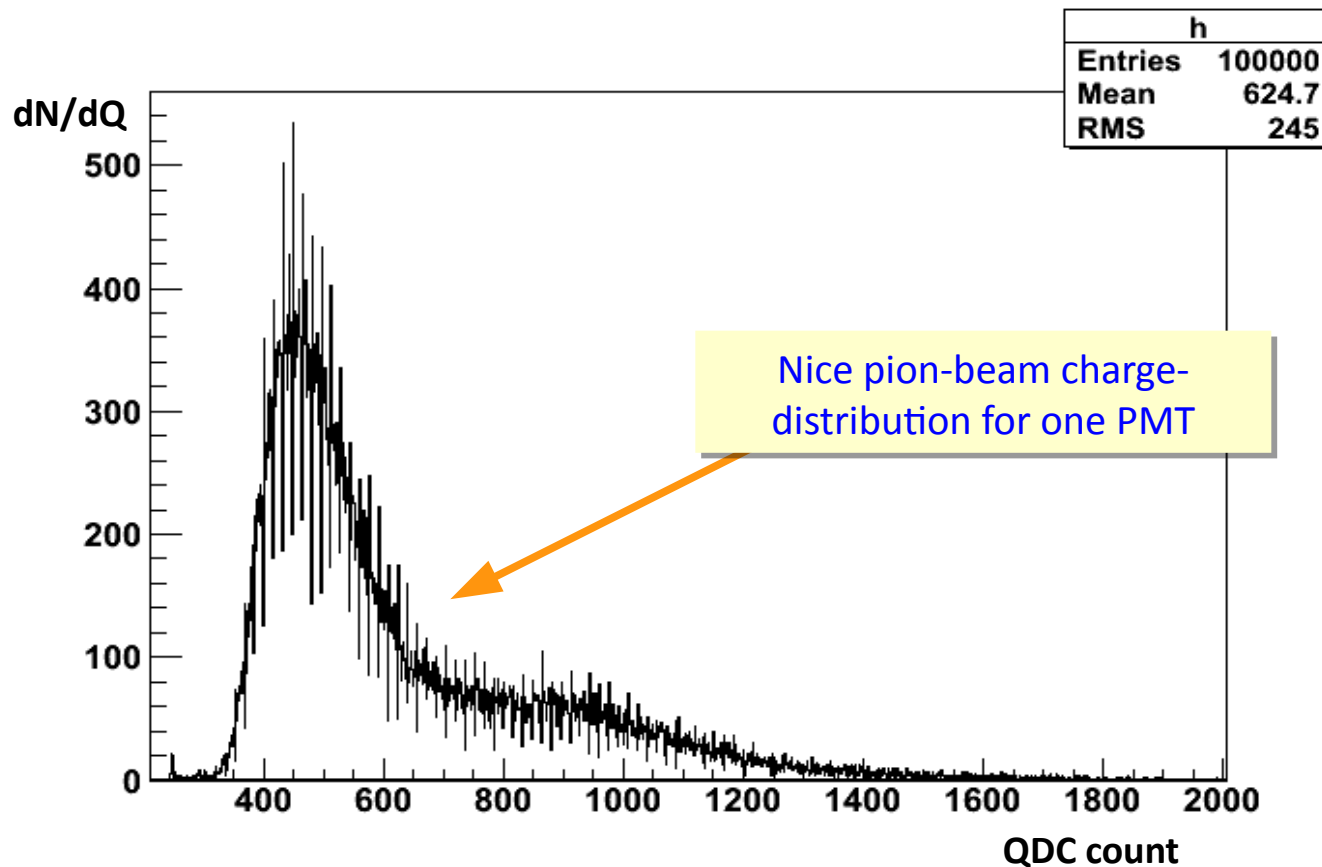
Real ADC.s at Work

- Real data from a beam test @CERN
- PbWO_4 (scintillating) crystal equipped with two PMTs and exposed to e , μ and π beams
- QDC \rightarrow (gated) charge integrator followed by ADC
 - \rightarrow (fixed-duration) gate opened after trigger
 - \rightarrow digital conversion started after gate closing
 - \rightarrow integrator discharged after conversion



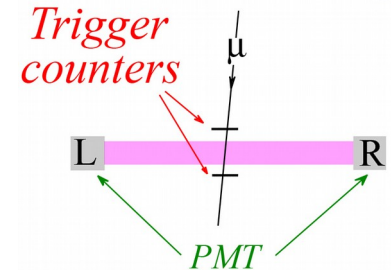
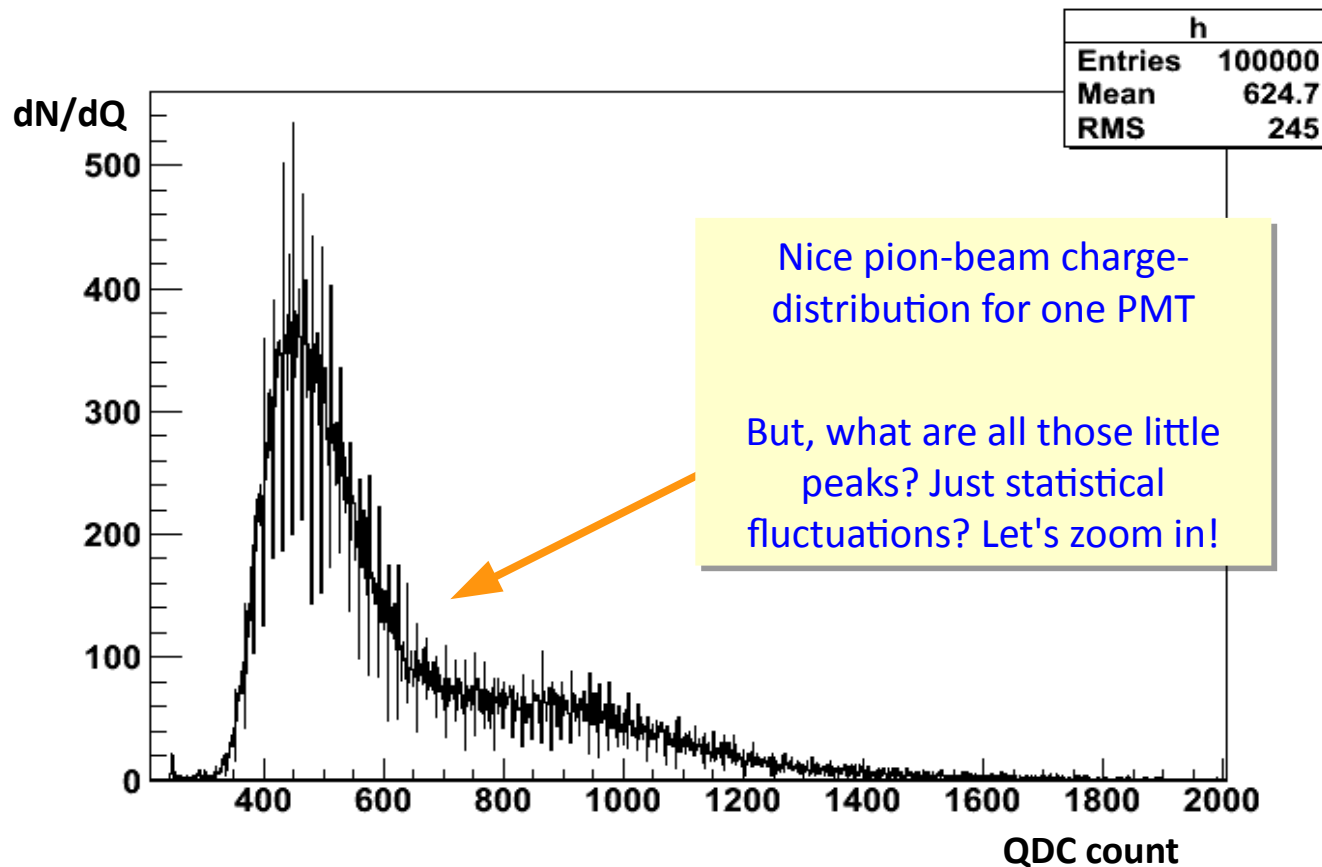
Real QDC.s at Work

PbWO₄ crystal equipped with two PMTs and exposed to e, μ and π beams

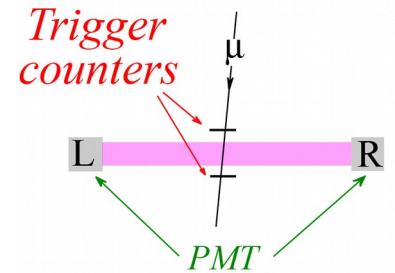


Real QDC.s at Work

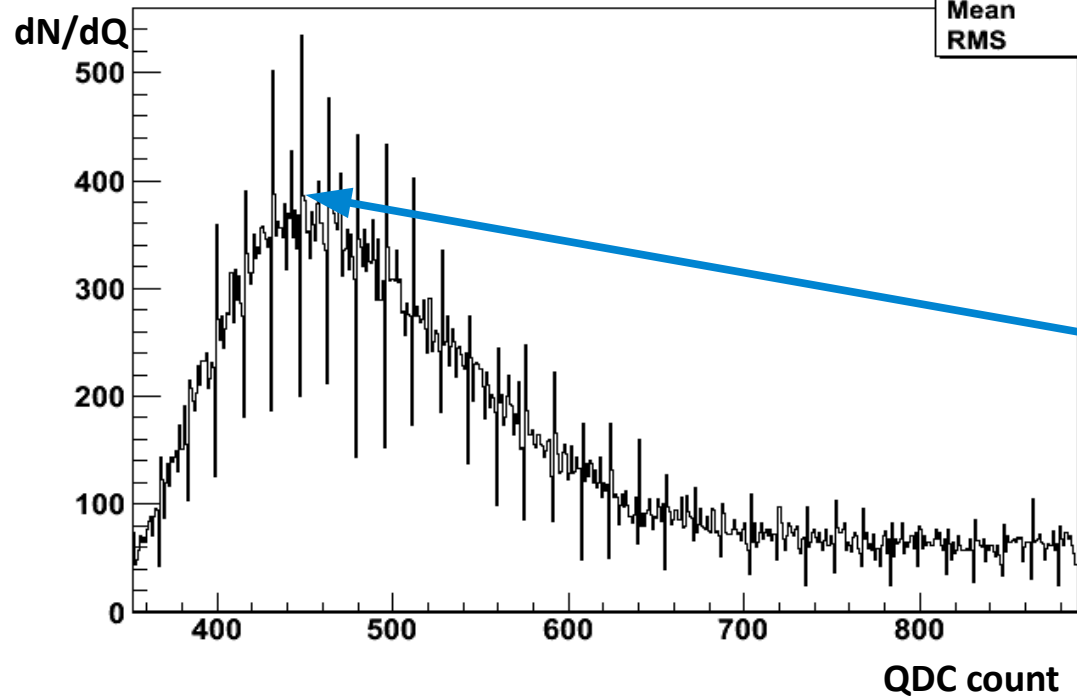
PbWO₄ crystal equipped with two PMTs and exposed to e, μ and π beams



Real QDC.s at Work



h	
Entries	100000
Mean	541.3
RMS	130



Bin with N entries shall fluctuate with:

- $\sigma = \sqrt{N}$

$$\sqrt{360} \sim 19 \rightarrow (540 - 360) / 19 \sim 10$$

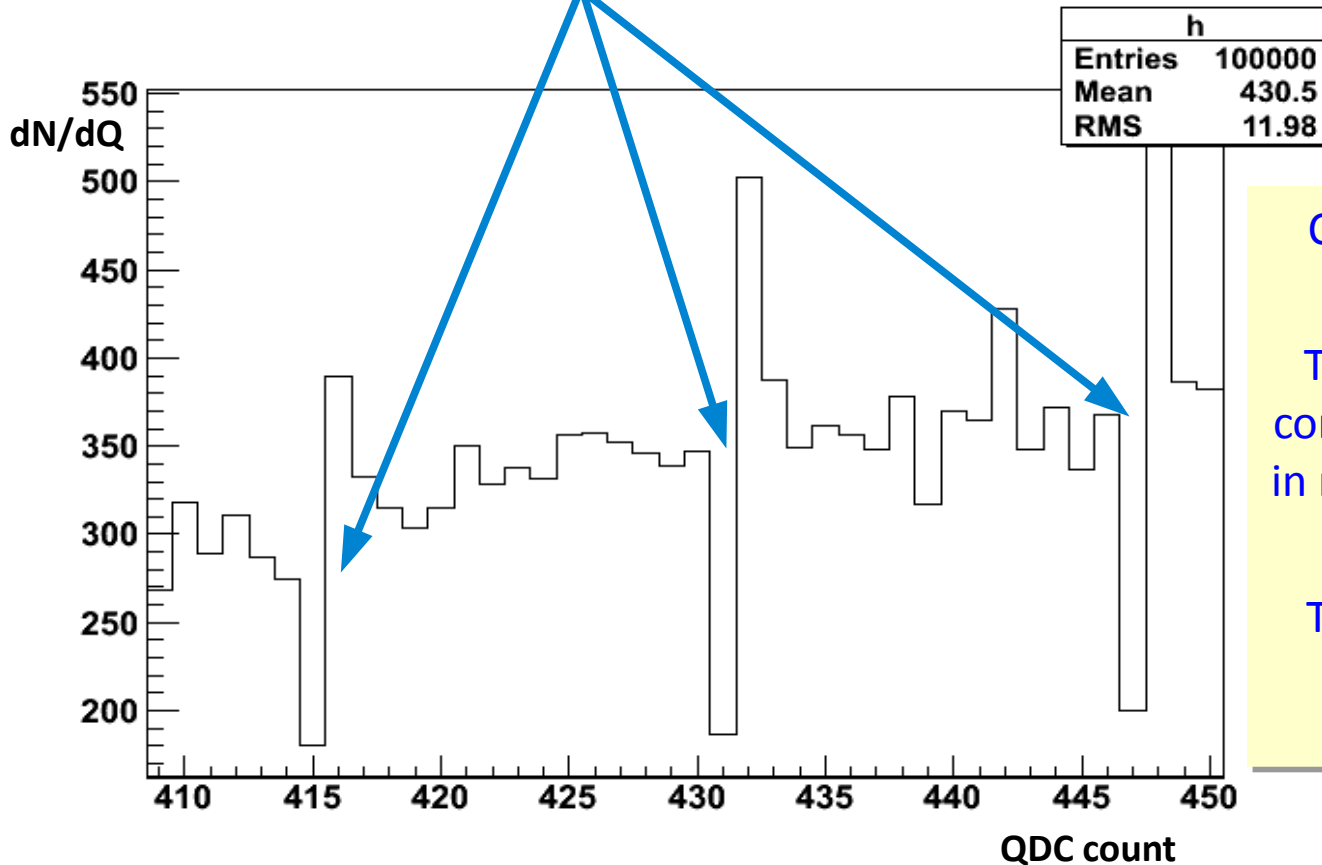
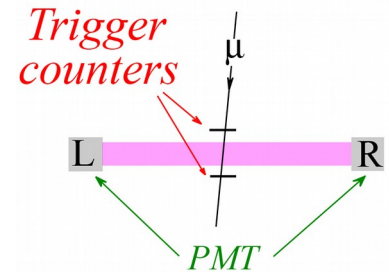
Spikes are regularly distributed

- some systematic effect is taking place

Let zoom in a bit more!

Real QDC.s at Work

415 & 416 → 0x19f & 0x1a0
 431 & 432 → 0x1af & 0x1b0
 447 & 448 → 0x1bf & 0x1c0



Can you see the effect?

The QDC prefers output configurations of type xxx0 in respect of those like xxxf

Typical differential non-linearity of successive approximation ADCs

Homework: can you find a simple way to fix this problem in the data? What the cost?

Some more ... Peak-Sensing ADC

- QDC signals need to be fully integrated, gate need not to be too long
→ need precise and stable timing .vs. gate:
- high-rate unipolar pulses may suffer of baseline drift

→ to overcome such problems:

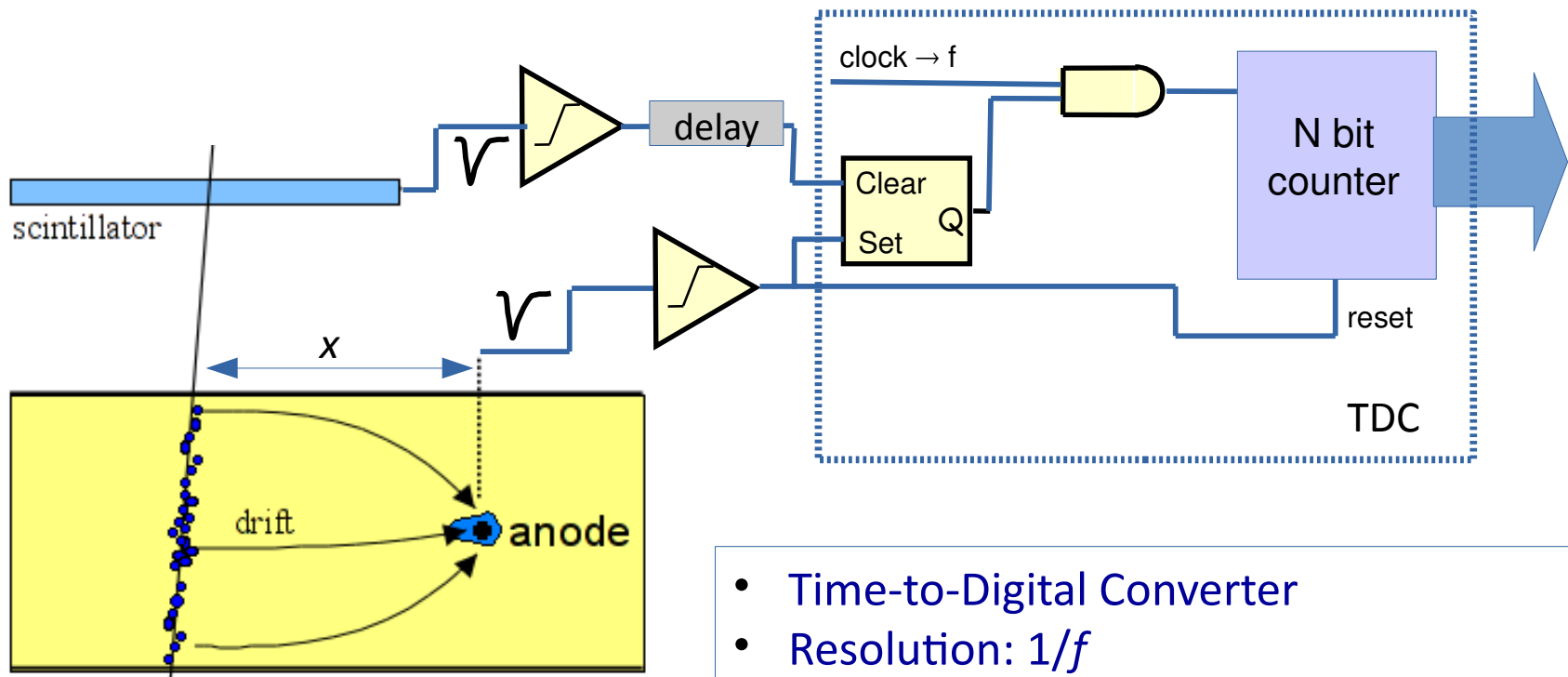
1) charge amplifier + shaper:

- signal shape stable
- peak value proportional to Q

2) peak-sensing ADC:

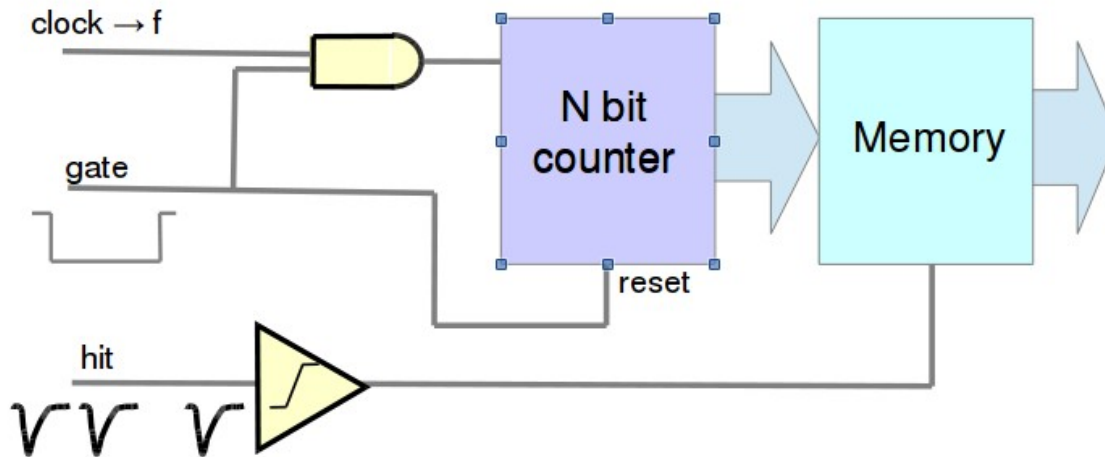
- find V_{max} within a gate (rise-time shouldn't be too fast)
- at end-of-gate, convert

Time Measurements → TDC



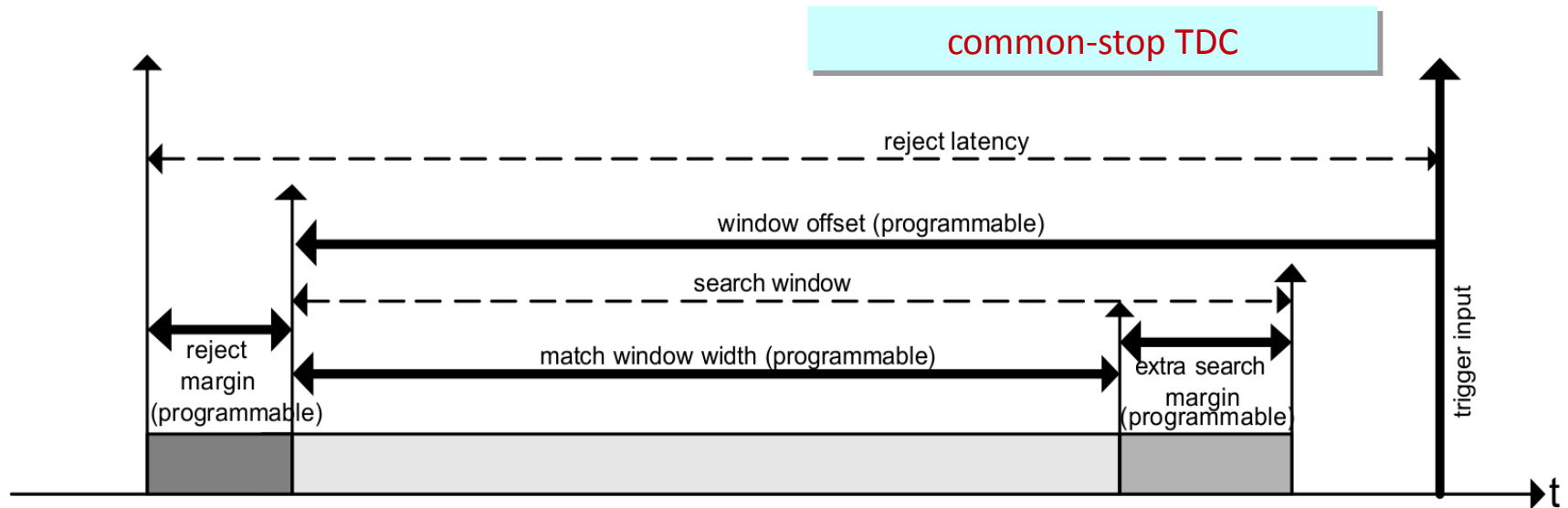
- Time-to-Digital Converter
- Resolution: $1/f$
- Dynamic range: N
- Single-hit TDC:
 - e.g. noise spike arriving just before the signal → measure lost

Multi-Hit TDC



- Gate resets and starts counter. Also provides measurement period
- Gate duration $< \sim 2^N/f$
- Each “hit” (i.e. signal) forces memory (FIFO) to load current counter value, that's delay wrt. gate opening
 - in order to distinguish hits belonging to different gates, some additional logic needed to tag data (end/start-of-gate marker)
- Common-start configuration

Real TDC.s



- Real TDCs provide advanced functionalities for fine-tuning hit-trigger matching
 - internal programmable delays
 - internal generation of programmable gates
 - programmable rejection frames

Digital Converters - Summary

Speaking about measuring fast (short) pulses ...

- Flash ADC: high-speed sampling, measure V
- QDC: pulse integration, measure Q , pedestal* proportional to gate length
- Peak Sensing ADC: measure $V(\text{peak})$
- TDC: measure time intervals
- **conversion (dead) time of commercial VME boards \sim few μs**

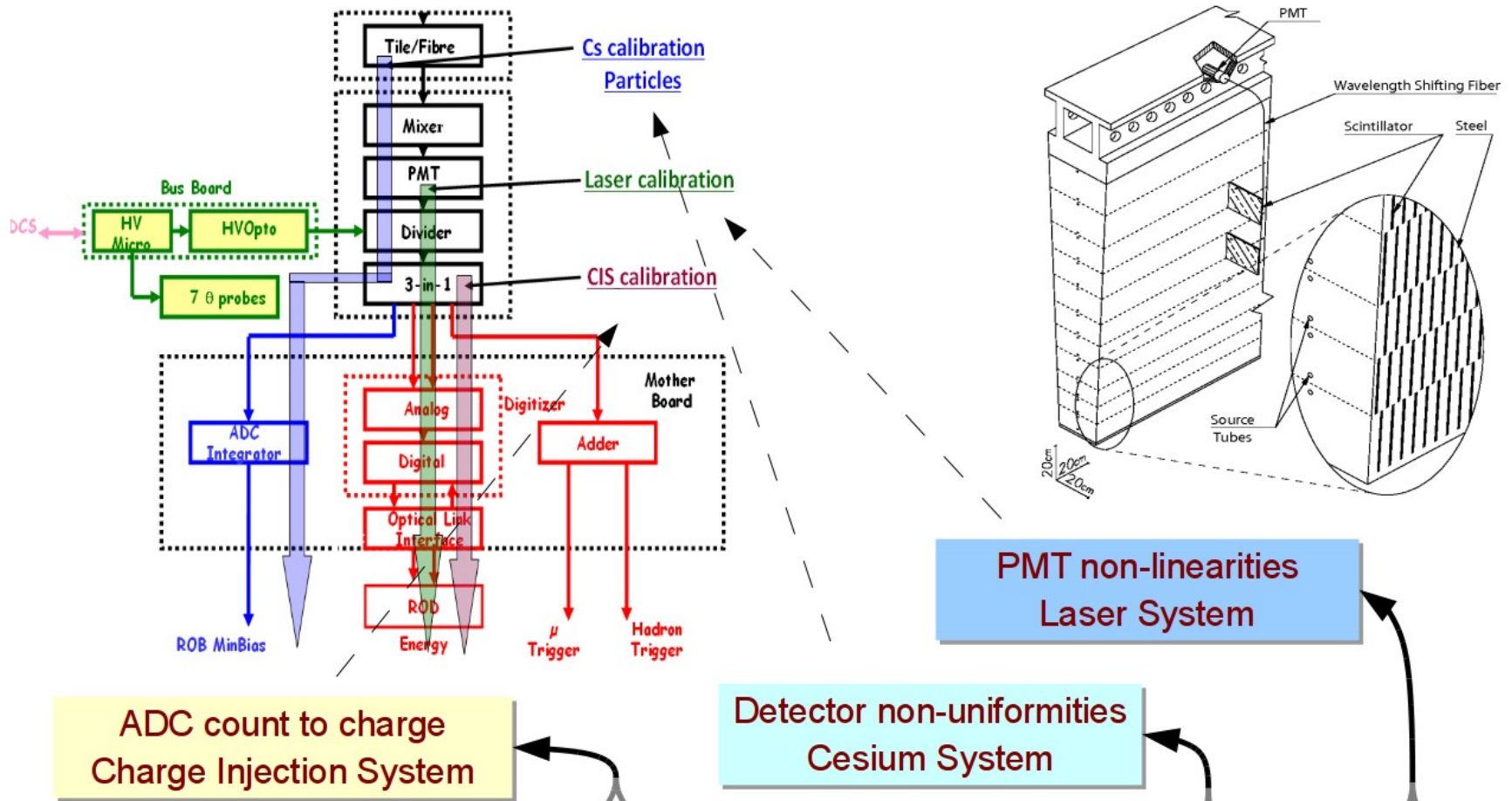
*pedestal = measurement of input noise (absence of signal)

Calibration

Calibration

- Our elx chain provides us with numbers (raw data) but ... what do they really mean ?
- Likely related with physics quantities but with relations (transfer functions) affected by several **uncertainties**:
 - due to physical detection mechanisms
 - due to signal processing
- Transfer functions usually parametrized, sometimes based on (look-up) numeric tables
- All system elements need to be **calibrated** to keep optimal knowledge of all parameters → **calibration procedures** → **calibration constants**
- Calibration constants change with ageing (mainly due to radiation), beam conditions (electronics may have baseline drifting with pile-up), time ... HV, LV, ...
- The design of our detector and DAQ has to foresee calibration mechanisms/procedures
 - injection of known signals
 - dedicated calibration *triggers* and data streams

ATLAS Tile Calorimeter Calibration



$$E_{channel} = A \cdot C_{ADC \rightarrow pC} \cdot C_{pC \rightarrow GeV} \cdot C_{Cs} \cdot C_{laser}$$

to be continued...