

# HOM power in FCC-ee cavities

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# FCC-ee options

	Z	W	H	$t\bar{t}$
Bunches / beam, $M$	71200	6000	740	62
Bunch spacing, $t_{bb}$ [ns]	2.5	50	400	4000
Bunch population, $N_b$	$0.4 \times 10^{11}$	$0.5 \times 10^{11}$	$0.8 \times 10^{11}$	$2.1 \times 10^{11}$
Bunch length, $\sigma_t$ [ps]	12	8.3	7.7	9.2
Beam current, $J_A$ [mA]	<b>1399</b>	147	29	6.4

Harmonic number,  $h = 130680$

Ring circumference,  $C = 97.75$  km

# HOM power loss in FCC cavities

Simulated cavity  
impedance

Normalized  
Fourier harmonics  
of beam current

$$P = J_A^2 \sum_{k=-\infty}^{\infty} \operatorname{Re}[Z_{||}(k f_0)] |J_k|^2$$

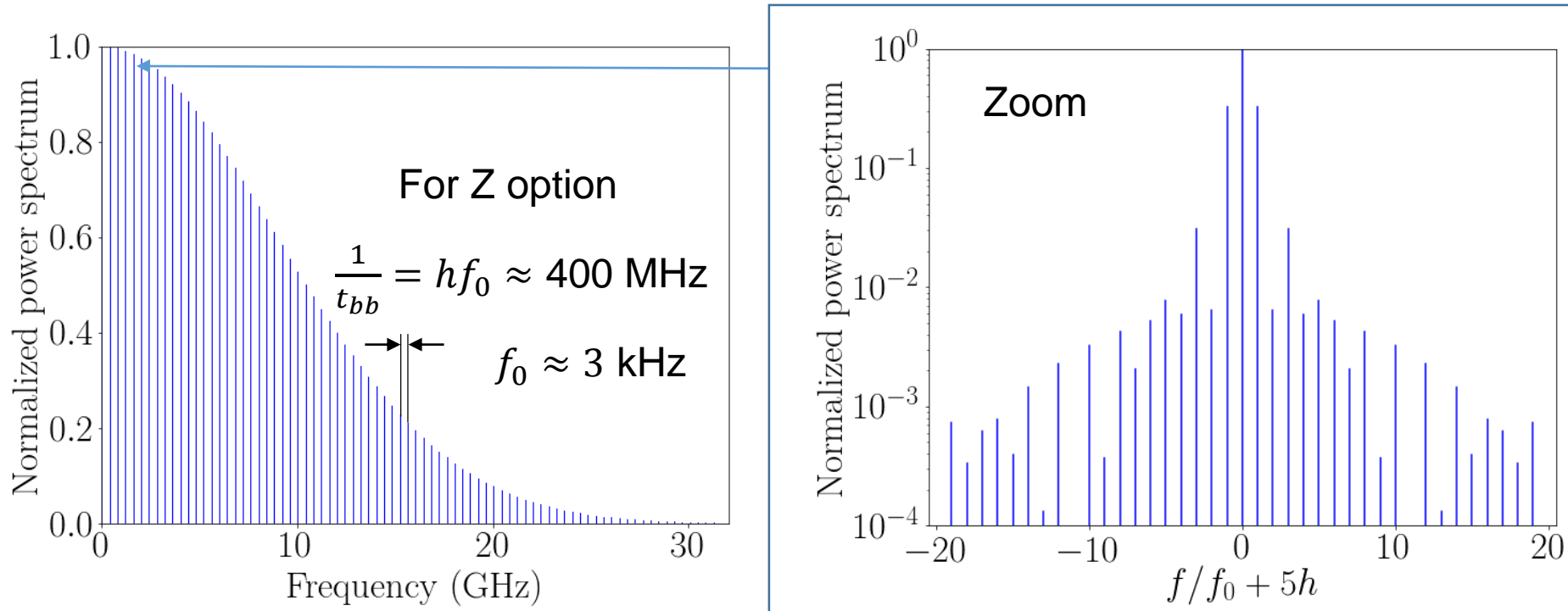
$f_0$  - revolution frequency

$J_A$  - average beam current

→ HOM power should be extracted (max 1 kW per coupler)

# Beam power spectrum

Spectrum contains multiples of  $1/t_{bb}$  and  $f_0$  harmonics



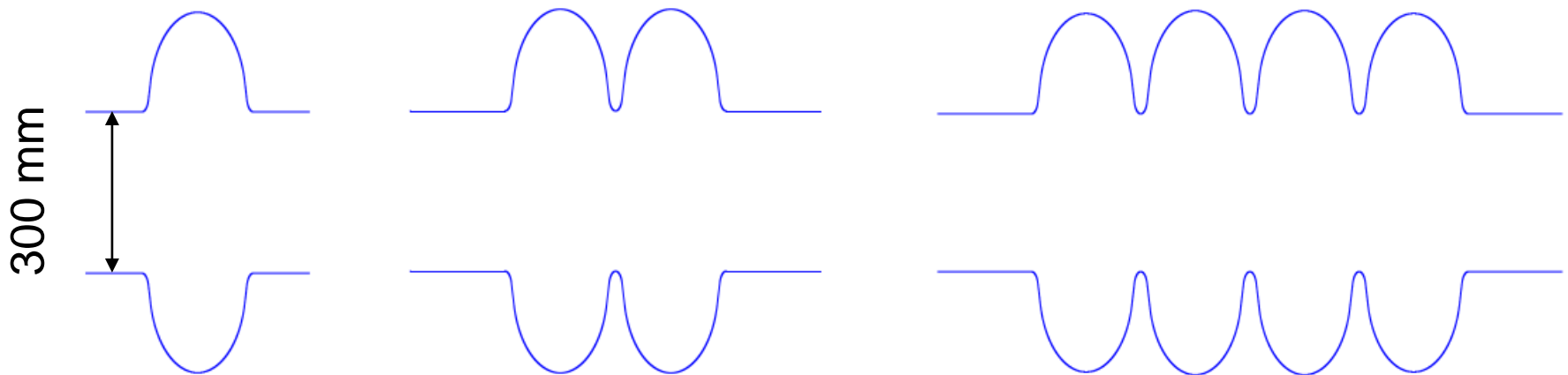
For Gaussian bunches

$$|J_k|^2 = e^{-(2\pi k f_0 \sigma_t)^2} \frac{\sin^2(M\pi k f_0 t_{bb})}{M^2 \sin^2(\pi k f_0 t_{bb})}$$

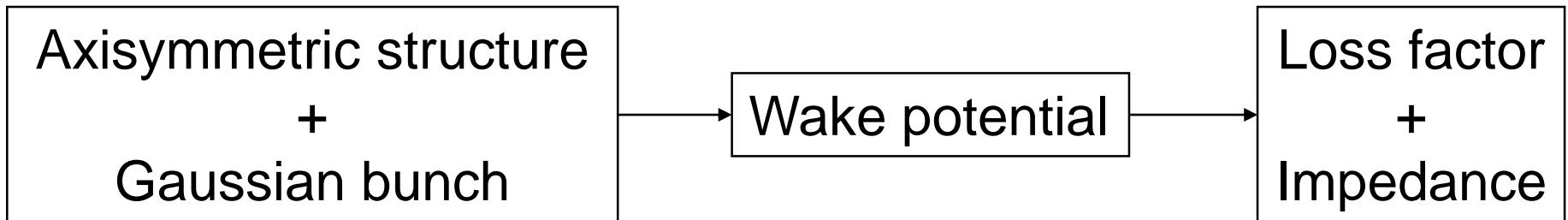
$M$  is number of bunches

# Example of 400 MHz cavities

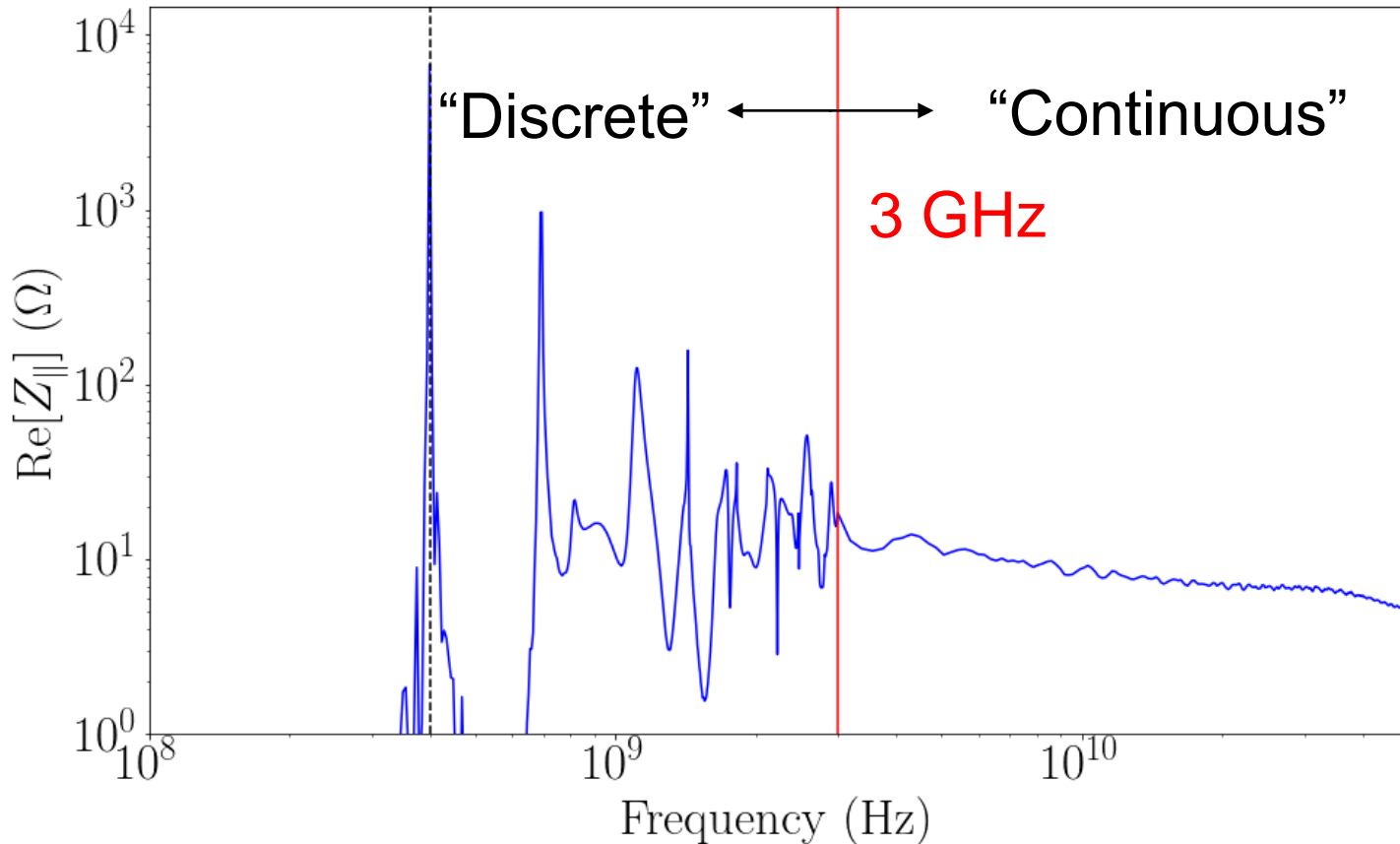
Cavity options for FCC-ee (Input from R. Calaga)



Impedance calculation using ABCI

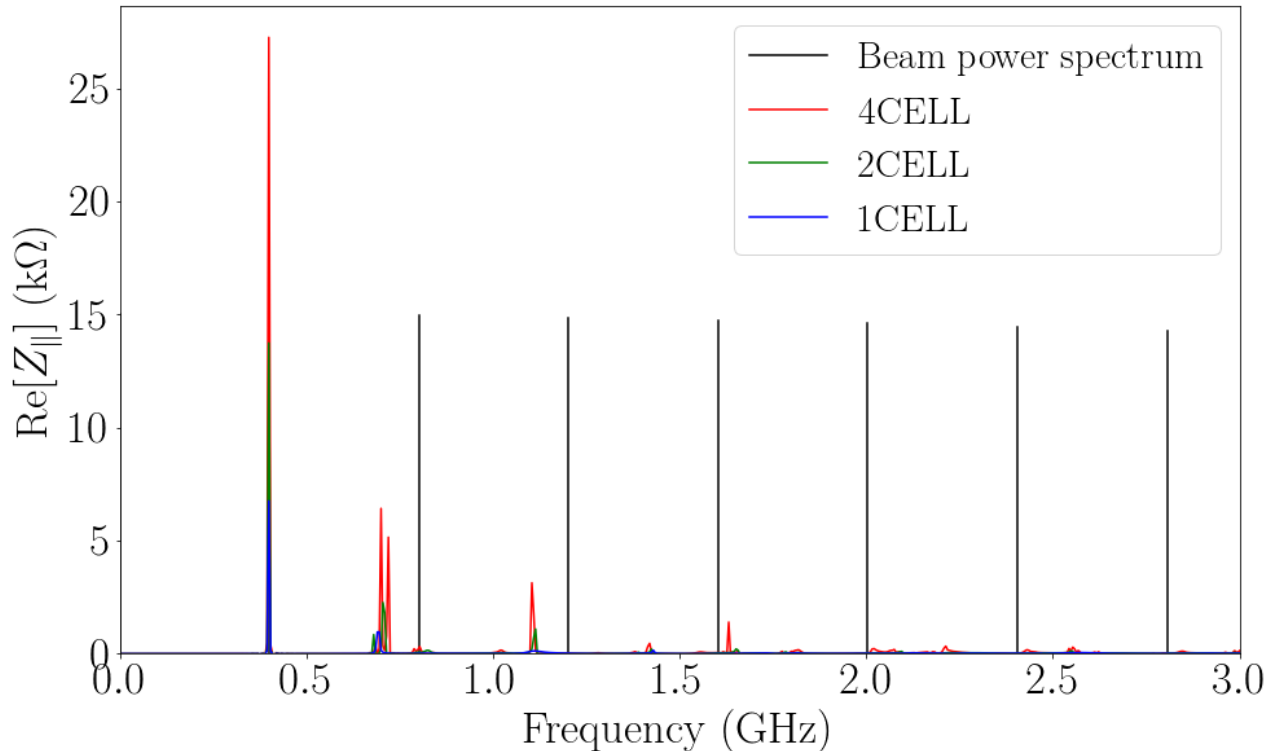


# Example of 400 MHz single-cell cavity impedance



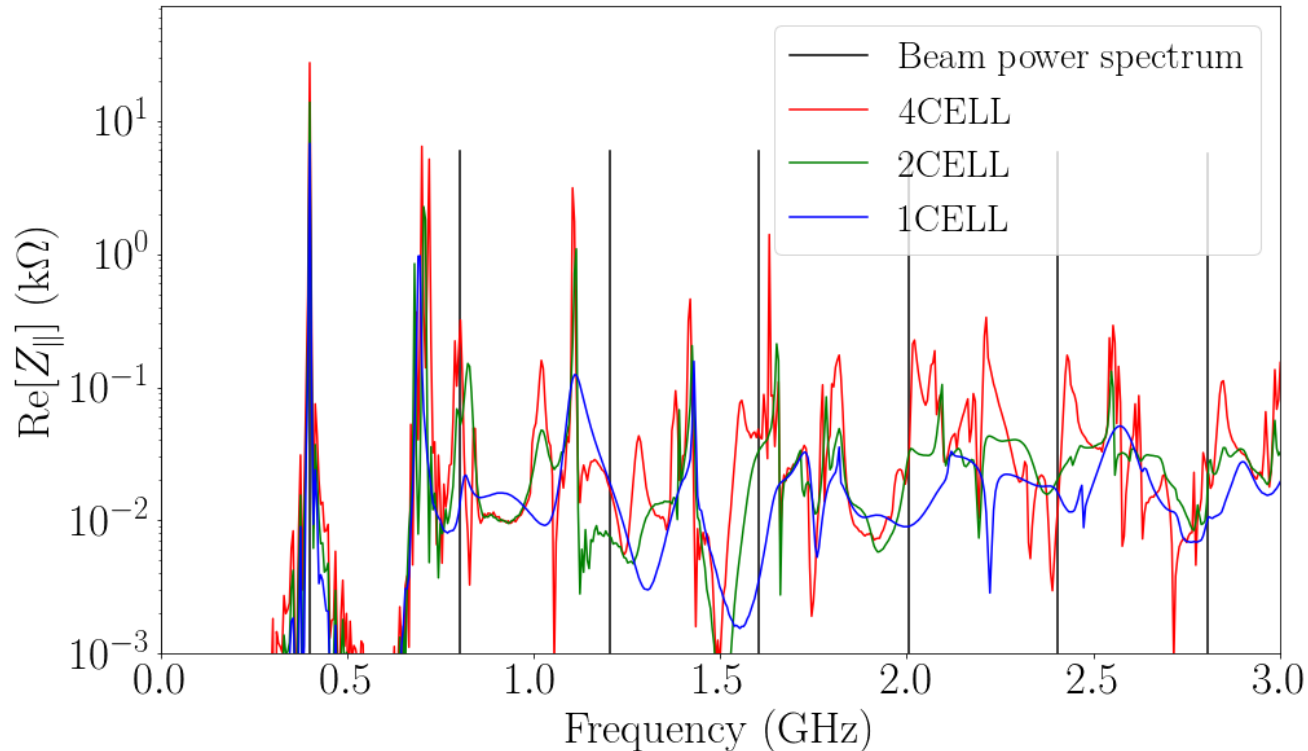
Cutoff for all trapped modes  $\leq 3$  GHz

# Example of 400 MHz cavities. Impedance **below** 3 GHz



Resonant lines are far from the beam spectrum harmonics  
 $k \times 400$  MHz (care should also be taken in any future design)  
→ This impedance can be excluded from power loss calculations

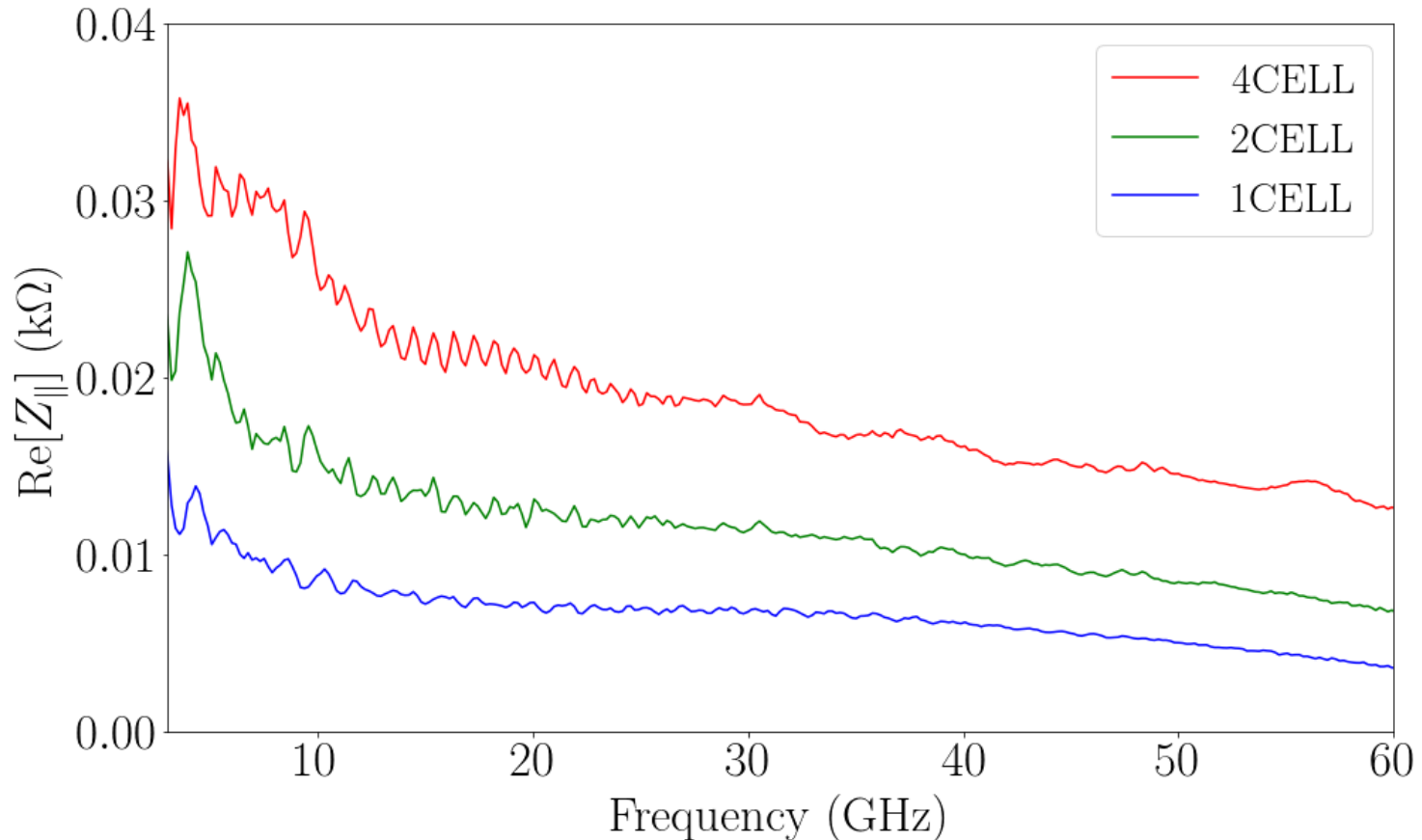
# Example of 400 MHz cavities. Impedance **below** 3 GHz



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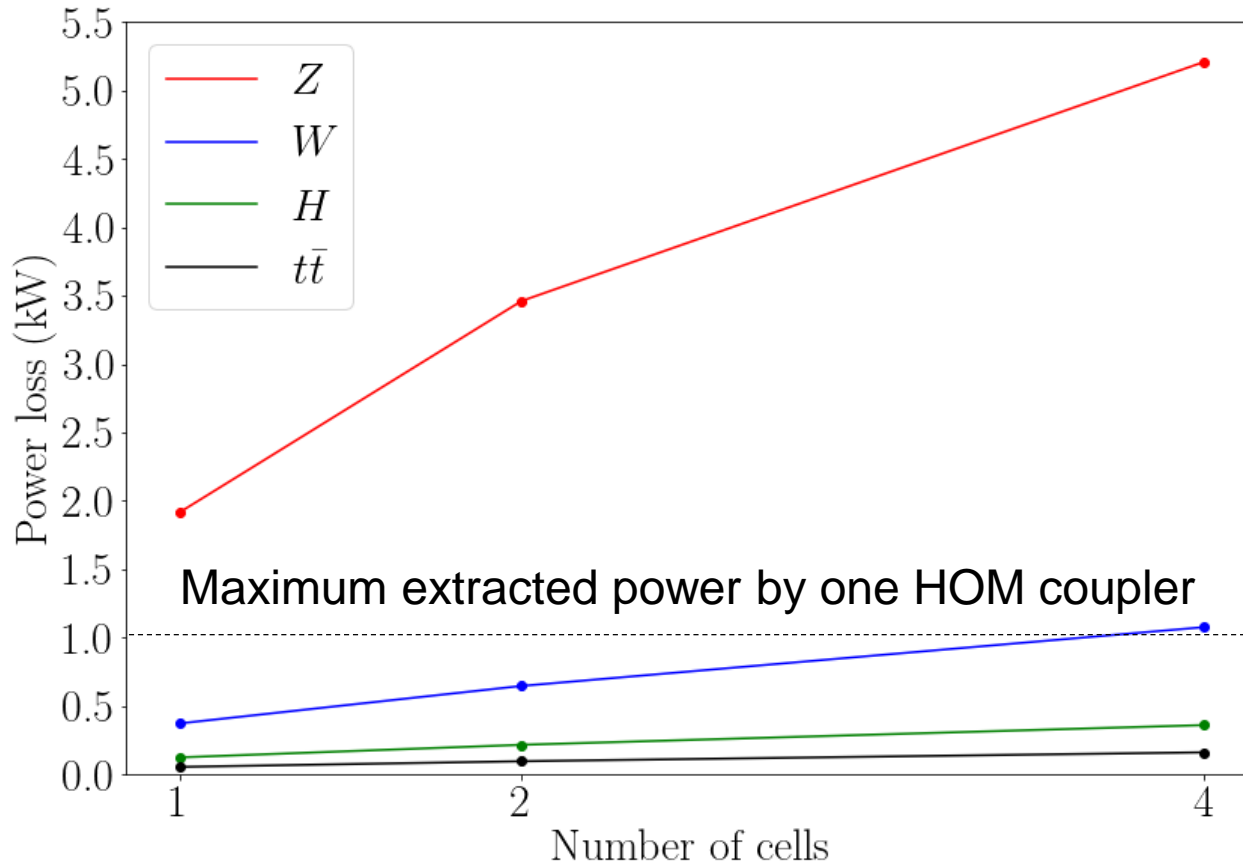


# Impedance **above** 3 GHz



→ Larger cavity impedance for larger number of cells  
(opposite per cell)

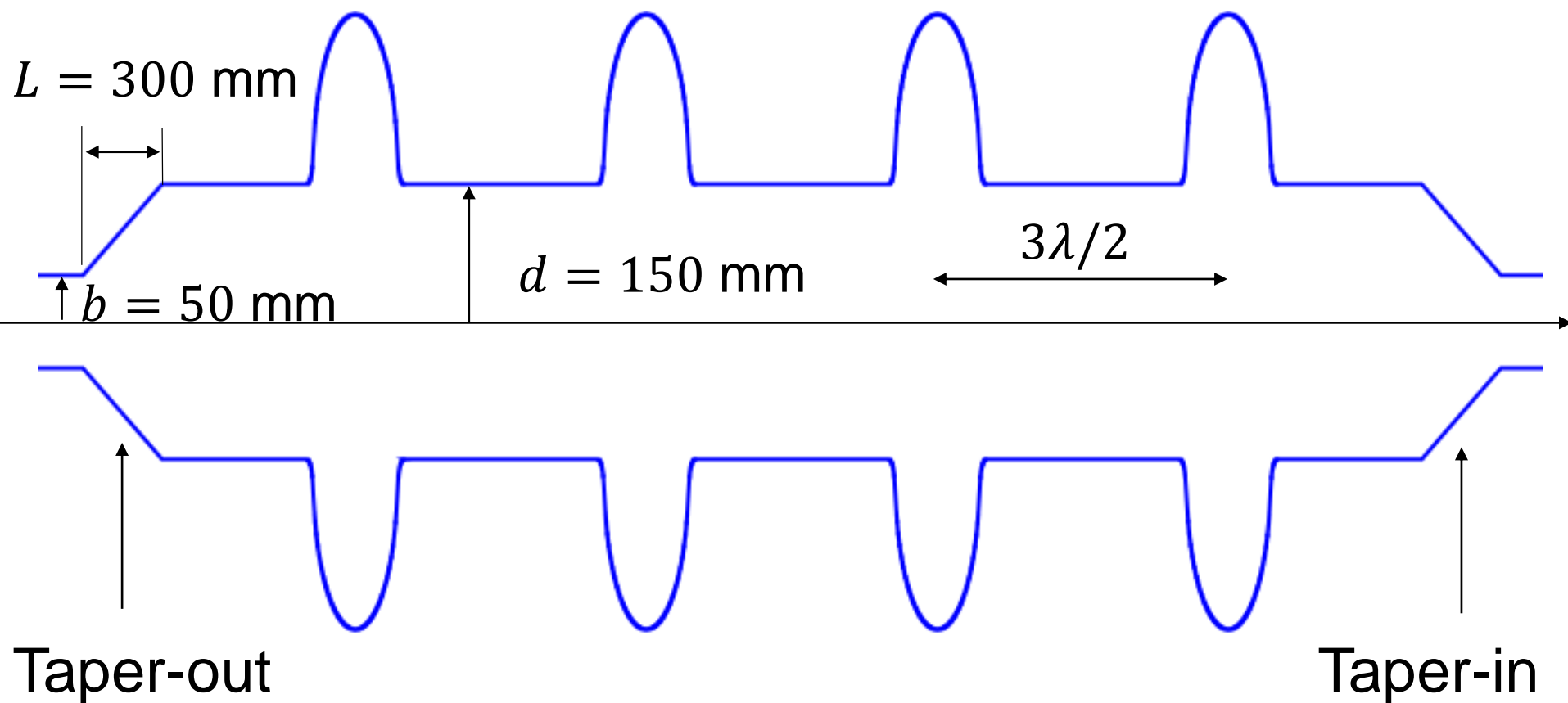
# Power loss for different number of cells in FCC-ee machines



Discrete impedance lines are excluded.

Single-cell cavity design is feasible for Z machine

# LHC-like layout of cryomodule



# Loss factor of taper-out

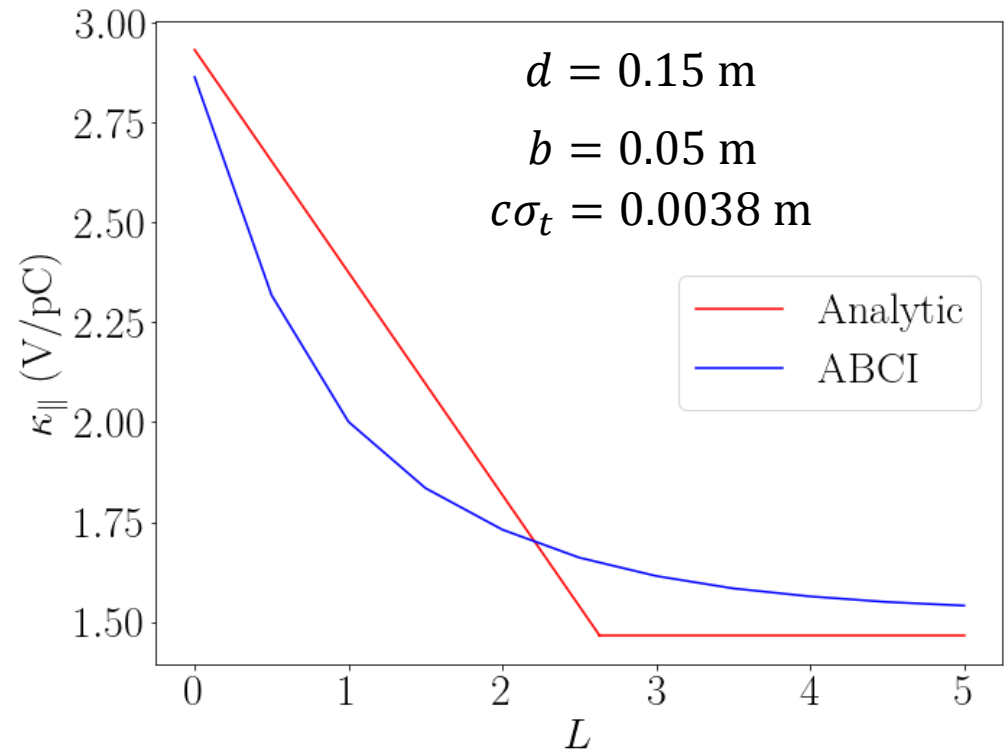
Analytic estimates\*

$$c\sigma_t \ll b < d$$

$$\kappa_{\parallel}(\sigma) = \frac{1}{2\pi\epsilon_0 c\sigma_t\sqrt{\pi}} \ln\left(\frac{d}{b}\right) \left[1 - \frac{\tilde{\eta}}{2}\right]$$

$$\tilde{\eta} = \min(1, \eta)$$

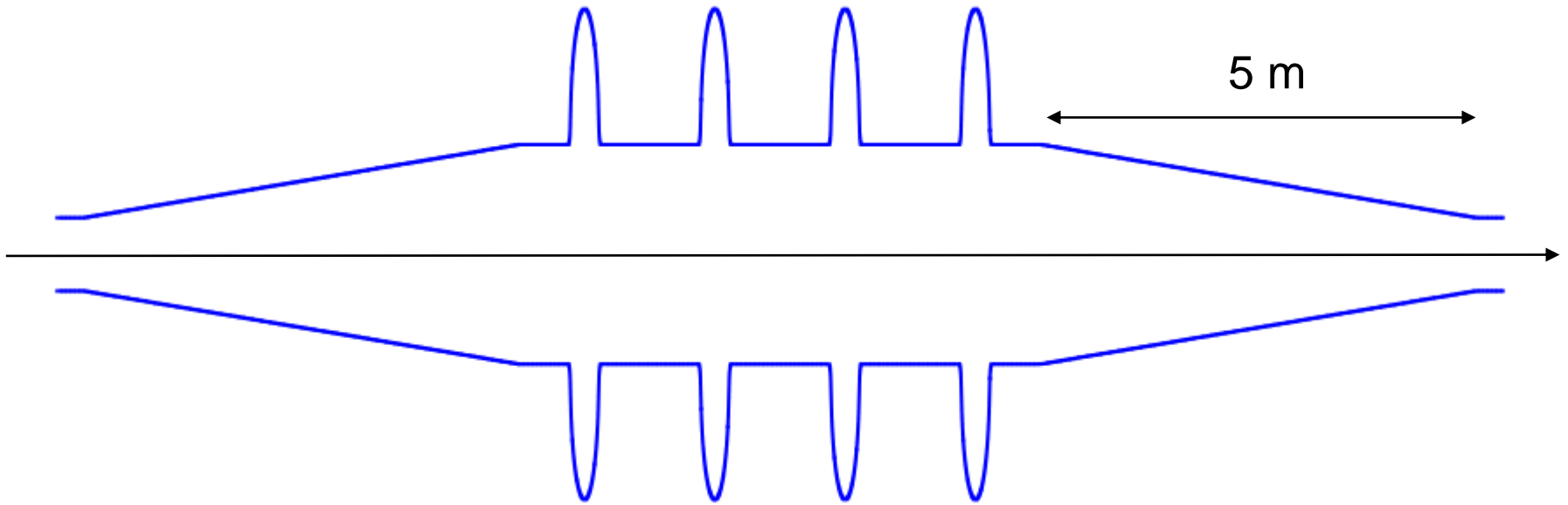
$$\eta = \frac{Lc\sigma_t}{(d-b)^2}$$



\*S. A. Heifets and S. A. Kheifets  
Rev. Mod. Phys. **63**, 631 (1991)

Short wake potentials  
are used in simulations

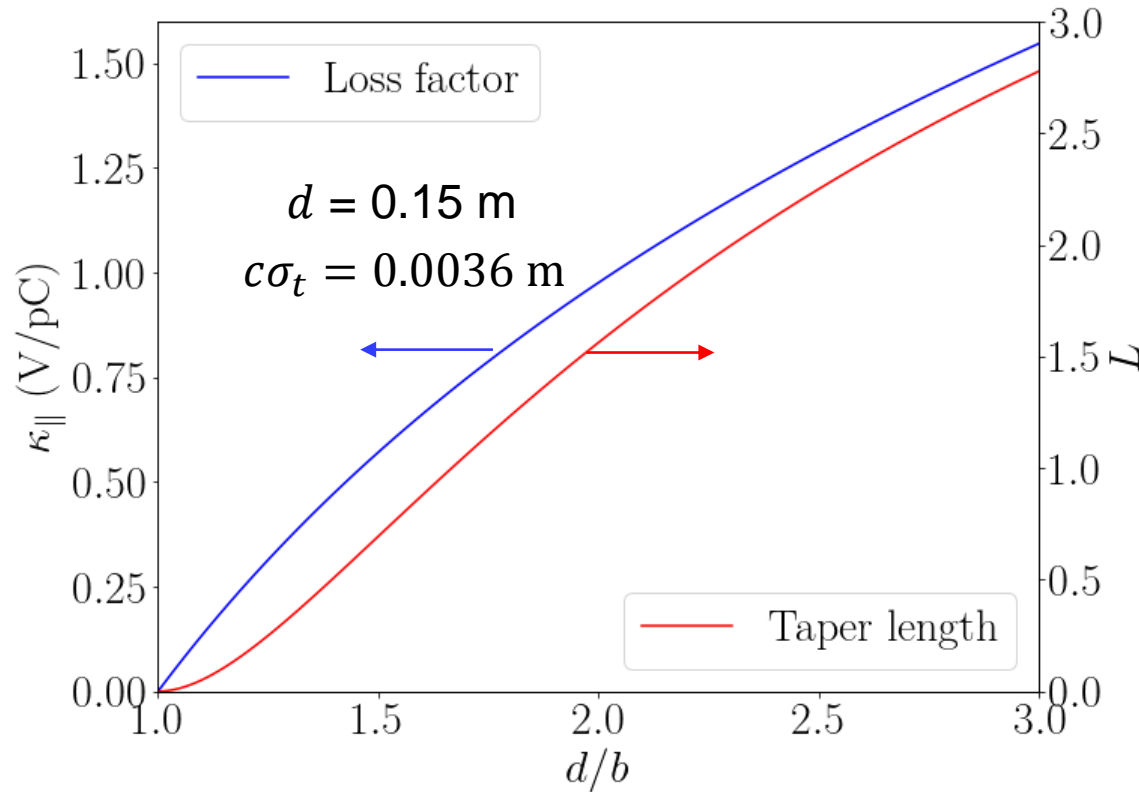
# Loss factors of 4-cavity structure



Structure	Loss factor $\kappa_{\parallel}$ (V/pC)		
	Steps	Taper-out	Tapers
With cavities	3.82	2.78	1.41
Without cavities	3.23	1.54	0.13

Loss factor of 4 cavities is 1.21 V/pC

# Optimization of transitions



Asymptotic value of loss factor for taper-out

$$\kappa_{||}(\sigma) = \frac{1}{4\pi\epsilon_0 c\sigma_t \sqrt{\pi}} \ln\left(\frac{d}{b}\right)$$

Minimum required length

$$L = \frac{(d - b)^2}{c\sigma_t}$$

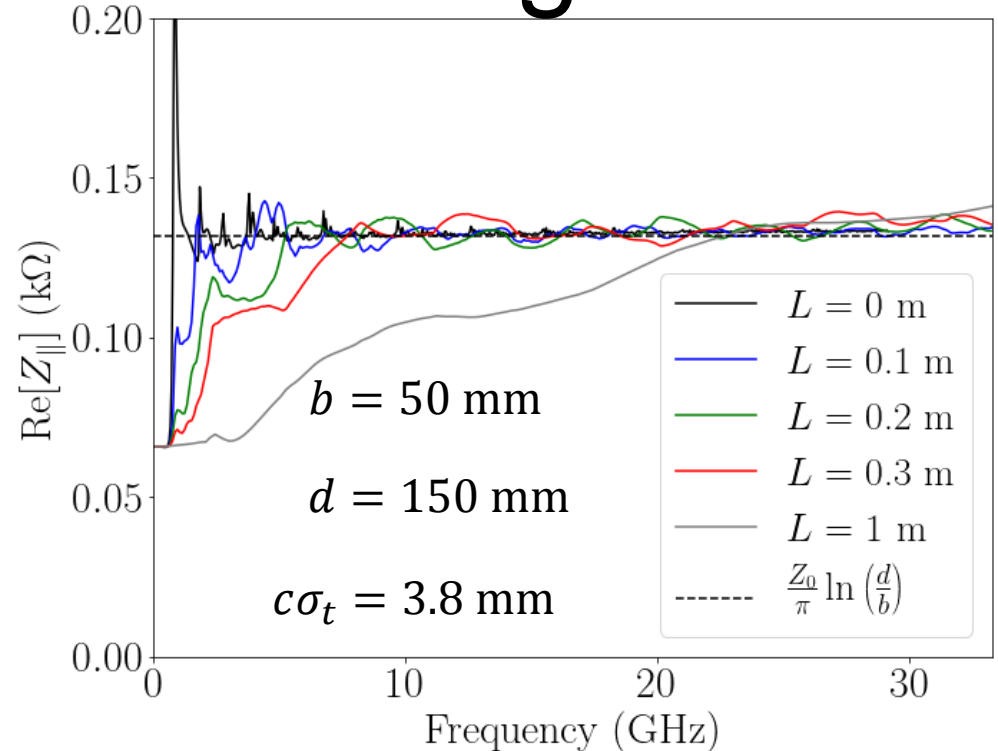
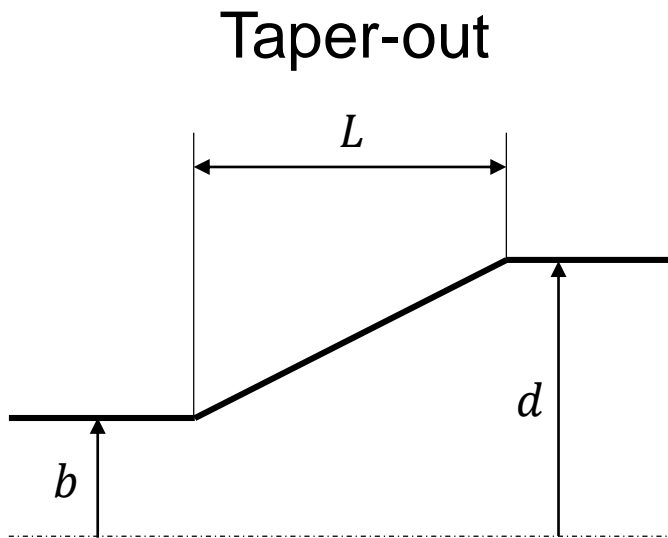
# Conclusions

- Estimations of power loss for all FCC-ee machines (400 MHz cavities)
  - Maximum power losses are for Z machine: ~ 2 kW for single cell cavity, main contribution is given by impedance above 3 GHz
  - For higher energy machines power losses are below 1 kW
- Significant contribution to the total power loss from tapers for FCC-ee bunch length.
  - For transition from 150 mm to 50 mm loss factor of taper-out ~ 1.5 V/pC is achieved for 5 m length
  - Optimization of cold-warm transitions is ongoing

**Thank you for your attention!**



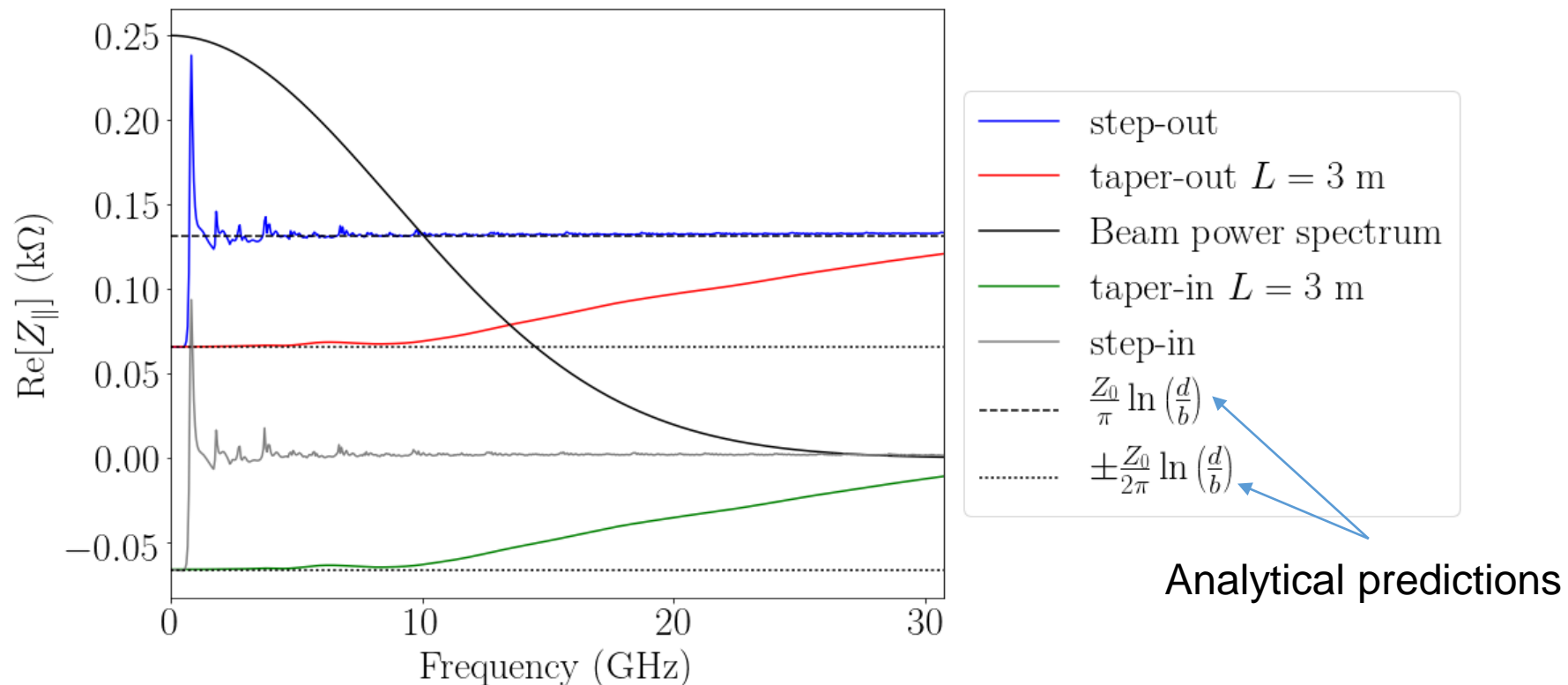
# Impedance of tapers: dependence on length



Asymptotic behavior at high frequencies

$$f \gg \frac{c}{2\pi b} \rightarrow Z_0^{\parallel} = \frac{Z_0}{\pi} \ln \frac{d}{b}$$

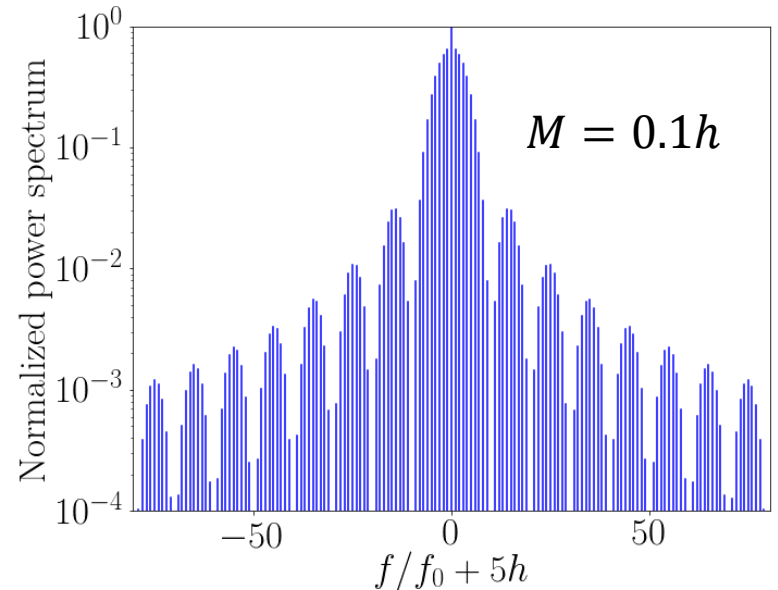
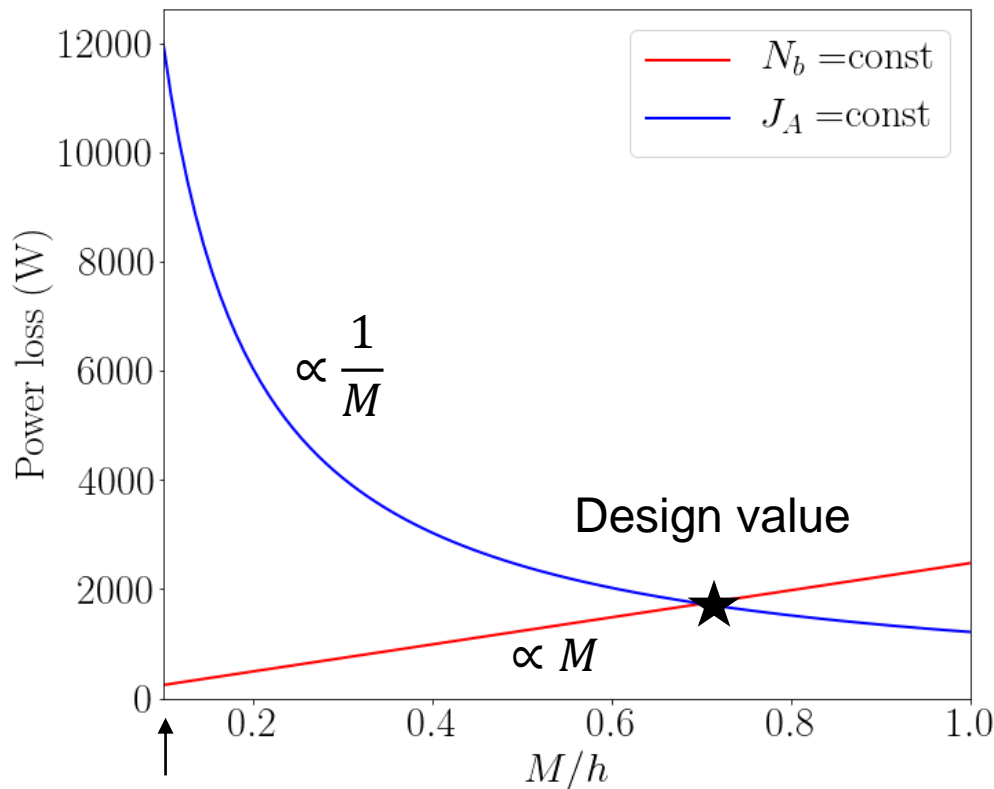
# Comparison of impedances of steps and tapers



-> Reduction of loss factor due smaller impedance at frequencies below 20 GHz

# Role of gap

Example for Z1 option (2.5 ns bunch spacing) and single-cell cavity



- Scaling corresponds to the case of broadband impedance
- Only for very small  $M$  one can expect that resonances can hit revolution harmonics