

Tests of Ultra Fast Silicon Detectors

Nicola Minafra

University of Kansas

R. Arcidiacono, N. Cartiglia, M. Ferrero, V. Sola C. Royon, H. Al Ghoul, P. Pare INFN di Torino KU

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Multi purpose board for a silicon/diamond detector



A two channels board that can be use for the characterization of different solid state detectors.



Guard ring pad (if present)

< 200 USD per board (10 produced) Sensors up to 16x13 mm² can be glued and bonded. The components can be easily adapted to accommodate:

- Diamond sensors: ~1 nA bias current, both polarities, small signal
- Silicon detectors: ~100 nA bias current, small signal
- UfSi: ~100 nA bias current, ~ larger signal
- SiPM: ~ 5 uA bias current, large signal



Amplifier with high input impedance



An amplifier with High Input impedance has some advantage of a Broadband Amplifier (50 Ω) and some of the Charge Sensitive Amplifier



The input impedance has to be selected according to the characteristics of the sensor.

The main advantage/disadvantage is that there are no general purpose commercial solutions!



Simulated SNR and time resolution for a diamond detector.

Amplifier with high input impedance

One implementation of a HImp amplifier is using a common emitter with a feedback resistor



The best value of R for timing has to be optimized according to the sensor:





TOTEM @ CERN See talks tomorrow

Optimization for UfSD

The signal generated at the passage of a MIP by a 50 µm UfSD can be simulated using Weightfield2*

Using Weightfield2 it is possible to simulate different

detectors, in different configurations.

Timing capabilities of Ultra-Fast Silicon Detector

The reliability of the simulations have been proved in several

occasions.

arXiv:1608.08681

Weightfield2 Build 3.3 Drift Potential Weighting Potential Currents and Oscilloscope Electronics G Si C Diamond C Free Set Potentials Currents Doping type Precision Strips © n C p Bulk Cn @p eh pairs followed (1=best, 100=fastest 1 100 🖨 Sampling [GigaSample]: Dimensions -Output File # of strips (1,3,5..): 1 Thickness[un] 200 100 \$ 000 \$ Batch Mode Width[um), Pitch[um)]: 1 ON Number of events Gain (1 = no G): 13 🛫 Select Particle MIP: NON uniform Q, Qtot = Landau Force Fixed Gain (Irr. OFF) 0 -0 h/e Gain ratio: Gain recess (um): 3000 100 X[un], Angle [deg]: 150- 승 9 승 Number of Particles Bias[V], Depletion[V]: 200 1 50 1 Plotting at: On Strips Between Strips 1505 🌩 Draw Field: |Ey| |Ex| Irradiatio Weighting Potentia Weighting Field Ew (1/m) CCE Effects 0 100]: 1 Read-Out Top Strip C Backplane 0.5 3000 Electron 1 I ON Plot Settings 0.8 Detector Cap[pF]: 1000 🔲 Draw Current Absolute Value Scope (50 [Ohm]) BW[GHz]: 0.6 ☑ Draw e/h motion CSA: Imp[Ohm]. Tr. Imp[mV/f0]: 1000 0.4 CSA:T_r,T_f[ns]: Current Settings 2000 □ B-Field on at[T]: CSA:Noise.Vth[mV, CFD if <1] Diffusion 0.2 00 BB:Imp[Ohm],BW.Gaint 50 \$2.5 \$100 \$ 🗖 Charge Cloud Dispersion (no Alpha 300 🜩 BB:Noise,Vth[mV, CFD if <1]: Temperature[K]: 200 150 200

*: <u>http://personalpages.to.infn.it/~cartigli/Weightfield2/Main.html</u>

A simplified signal can be used to simulate the behavior of several types of amplifiers.



2

t (ns)

2.5

3

-10

-12

1.5

 $50\,\mu m$ UfSD at 200V with a gain 15



The timing detector for CT-PPS

CMS TOTEM Proton Precision Spectrometer (CT-PPS) adds precision proton tracking and timing detectors in the very forward region on both sides of CMS to study central exclusive production (CEP) in proton-proton collisions.

Requirement of the timing detector:

- Small active area ($\sim 4 \text{ cm}^2$)
- Small dead region at the edge and between channels
- Low power consumption and low material budget
- Radiation hard (proton flux of 5×10^{15} cm⁻² per 100 fb⁻¹)
- Time resolution of 10-30 ps.

Data included from 2016-04-22 22:48 to 2016-09-06 07:09 UTC LHC Delivered: 28.42 fb⁻ ≞,, MS Recorded: 26.24 fb Luminosity 5 CMS Offline Luminosity (Preliminary) 15

Date (UTC)



CT-PPS proposed sensor geometry

1.0 mm

0.5 mm

50 micron inactive gap



0.45 mm





The amplifier was first test using a laser pulse to measure the behaviour with sensors of different capacitance



1080 nm picosecond laser, 50 ps wide pulses with peak power > 100 mW set at 10 cm away from the sensor board The support can be moved XY with micrometric accuracy



The amplifier was first test using a laser pulse to measure the behaviour with sensors of different capacitance



To test the amplifier with sensors of increasing capacitance, more than one pad were connected to the same amplifier

10.00-9.00-8.00-7.00-

ce (m 6.00-

Vertical Distar 5.00-4.00-3.00-2.00-1.00-

0.00-

4.00

6.00

8.00 Horizontal Distance (mm)

The amplifier was first test using a laser pulse to measure the behaviour with sensors of different capacitance



12.00

10.00

14.00 15.00



However, the fall-time shows a strong dependency

The amplitude of the signal has a week dependence on the position of the laser spot







The amplifier was first test using a laser pulse to measure the behaviour with sensors of different capacitance



Time difference between two neighboring pixels with the laser, measured using the SAMPIC evaluation board

Test of the amplifier in Turin

The amplifier was first test using a laser pulse, calibrated using cosmic rays and a radioactive source (Sr⁹⁰)



The laser pulse was focused on \sim 50 μm spot; the power was calibrated using a Sr⁹⁰ and

UFSD for CT-PPS



The CT-PPS sensor was used to test the amplifier with a pad of ~3 pF and one of ~12 pF



UFSD produced by Hamamatsu

Hamamatsu produced some 50 µm thick LGADs, a sample with gain ~10 at 600V was tested







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Performance with different sensor capacitances



The behaviour of the different approaches using 50 μm UfSD can be simulated for several values of the sensor capacitance.



Below 15 pF the CSA is the amplifier with the best time resolution

Performance with damaged sensors

Supposing that the gain become 50% lower because of radiation damage, the CSA is still the best approach.



Below 10 pF the CSA is the amplifier with the best time resolution.



Front-end electronics: amplifier

For $R < \sim 100 \Omega$ the signal is not separated from the

 $\sigma_t \sim \frac{\sigma_v}{MAX[\frac{dV}{dt}]}$

It is useful to analyse the simplest possible case: a diamond detector read using a simple resistor.



First stage with input resistor ~ $k\Omega$

Strategy suggested by HADES @ GSI

10.1016/j.nima.2010.02.113

SNR

And higher R means slower signal:

value of the read-out resistor.





10

10^t



 $C = 0.1 \, pF$

C = 1 pF

 $C = 10 \, pF$

SNR = 1

10⁶ R (Ω)

The TOTEM timing detector: timing performance

 V_{th}

 $t_0 t_1$

To measure the time resolution of two identical detectors it is possible to measure the arrival time of a particle crossing both sensors.

However, the time resolution depends on the capacitance of the detector!

A series of tests were done using a sensor with pads of different surface, i.e. capacitance.



The measured time difference will be distributed around the true value because of the limited resolution of the detectors:

$$\sigma_{TOT}^2 \sim \sigma_{det1}^2 + \sigma_{det2}^2 \sim 2\sigma_{det1}^2 \implies \sigma_{meas} \sim \sqrt{2}\sigma_{det}$$

Time difference between a sensor of 17.6 mm² (~1.7 pF) and sensors of different size



