

Max-Planck-Institut für Physik (Werner-Heisenberg-Institut)

# Detector/Receiver "Cold" Measurements

MadMax-Workshop MPP Munich

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### Outline

- Microwave radiometer (short reminder)
  - Comparison between photon- and heterodyne detection
- Current lab system
  - Schematic
  - First cold tests
- Conclusion

## Photon Detection Setups

#### • Two principle ways:

- Photon counting
- Measurement of mean photon flux

#### Photon counting

- Limited by photon energy (Needs "high energy" photons)
- Energy (frequency) resolution is limited

#### Photon flux measurement

- Not limited by low energy photons
- Excellent frequency (energy) resolution (easily it can be better than 10<sup>-9</sup>), because of usually used "coherent" detection (normally heterodyne detection)

#### Photon Detection Setups

Photon counting:



Photon flux measurement:



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### Choosing the Right Bandwidth

• What is the detectable noise temperature for a given system noise temperature (Dicke-formula):

$$\Delta T = T_{Sys} \sqrt{\frac{1}{\Delta f_F \tau} + \left(\frac{\Delta G}{G}\right)^2}$$

- $\Delta f_F$ : Filter bandwidth
- $\tau$ : Averaging time
- T<sub>Sys</sub>: Total system noise temp.
- Detectable noise power (assuming no gain fluctuation)

$$P_N = k_B T_{Sys} \sqrt{\frac{\Delta f_F}{\tau}}$$
 with  $P_N = k_B \Delta T \Delta f_F$  and  $\Delta G = 0$ 

• Averaging time for a given signal/noise ratio:

$$\tau = \left(s \ \frac{k_B \ T_{Sys}}{P_S}\right)^2 \Delta f_F \qquad \text{with} \ s = P_S / P_N$$

## Choosing the Right Bandwidth

- What is the best bandwidth for line detection
  - Detectable background noise power increases with frequency (Square root)
     $k_B T_{Sys} \sqrt{\frac{\Delta f_F}{\tau}}$
  - Signal noise increases with frequency  $k_B \Delta T \Delta f_F$ (Linear, if rect. distribution)
- Integration time: 1.6x10<sup>-23</sup> 50h Bandwidth should not be larger than line-width for best signal-noise ratio
   Mod solution 100h 1.4x10<sup>-23</sup> 200h 400h 1.2x10<sup>-23</sup> Signal 1.0x10<sup>-23</sup> 8.0x10<sup>-24</sup> 6.0x10<sup>-24</sup> 4.0x10<sup>-24</sup> Example Receiver: T<sub>Svs</sub>=5K Signal linewidth limit 2.0x10<sup>-24</sup> Signal: 10<sup>-23</sup>W (1photon/s @ 15GHz), 0.0 linewidth 10kHz, equal distributed 2.0k 10.0k 12.0k 14.0k 16.0k 18.0k 20.0k 0.0 4.0k 6.0k 8.0k Filter Bandwidth (Hz)

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#### Receiver

- Axion mass range: 40  $\mu$ eV ... 400  $\mu$ eV Frequency range: 10 GHz ... 100 GHz ( $\lambda$  = 3 cm ... 3 mm)
- Detection of signal line in frequency domain with  $\Delta \nu_{\text{A}}$  = 10^{-6}  $\nu_{\text{A}}$

• • •

### Low-Noise Amplifiers

- 2 different devices (Low Noise Factory, Chalmer University)
- Same characteristics @ RT but 1 is for cryo temperatures



6-20 GHz Cryogenic Low Noise Amplifier, 5K @ 8-10K 1-15 GHz Low Noise Amplifier, 75K @ RT





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- Inject fake axion signal with 1.2.10<sup>-22</sup> W at LHe temp.
  - Frequency: 18.85 GHz
  - Frequency modulated with gaussian noise
  - Signal bandwidth: 8 kHz, Lorentz-shaped

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  - Signal bandwidth: 8 kHz, Lorentz-shaped
  - Received signal after
    28h measurement
    (averaged signal):



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  - Frequency: 18.85 GHz
  - Frequency modulated with gaussian noise
  - Signal bandwidth: 8 kHz, Lorentz-shaped
  - Received signal after
    baseline subtraction
    and gain correction:



Inject fake axion signal with 1.2.10<sup>-22</sup> W at LHe temp.

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- Frequency: 18.85 GHz
- Frequency modulated with gaussian noise
- Signal bandwidth: 8 kHz, Lorentz-shaped
- X-Correlation signal with 8kHz width:

$$X(\tau) = \int s(t) T(t+\tau) d\tau$$

s: Signal T: Testfunction (Lorentz, Gauss, ...)



Highest peak

- Inject fake axion signal with 1.2.10<sup>-22</sup> W at LHe temp.
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  - Why 8kHz Bandwidth?
    Algorithm is searching for best S/N-ratio:



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## Additional Facts:

- Comparison with Allen's run statistic algorithm showed good agreement
- Cold tests are ongoing
  - 5.10<sup>-23</sup> W in 10kHz linewidth already reached within one week in 10K T<sub>sys</sub> (Physical limit)
  - Different tests runs should give a clearer insight to possible problems (quantization noise, ...)

### Conclusion

#### Receiver concept is OK

- Dead time 1,4%
- Sensitivity in warm and cold is OK

#### • Next Tests:

- Systematic cold measurements
- Better Antenna measurements
- Cold background measurements in cryostat



#### Spectral Power Density of (BB)-Noise

 Contribution of a detector: (no phase preservation)

$$E_N = \frac{h\nu}{e^{\left(\frac{h\nu}{k_B T}\right)} - 1} \qquad [E_N] = \frac{W}{Hz}$$

 Contribution of an amplifier or mixer: (phase preservation)

$$E_N = h\nu \left(\frac{1}{e^{\left(\frac{h\nu}{k_BT}\right)} - 1} + 1\right)$$

• Limit for low frequencies and/or high temperatures:

$$E_N = k_B T_{\text{Noise temperature}}$$
  
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#### Spectral Power Density of (BB)-Noise

- Example:
  - Spectral power density for different temperatures



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## Noise Equivalent Power

- System noise temperature  $T_{\text{Sys}}$  and bandwidth  $\Delta f_{\text{F}}$  are difficult to measure for broadband detectors
  - Johnson noise
  - Phonon-electron coupling
  - Generation-recombination noise
  - Background noise
- $\rightarrow$  Using noise equivalent power (NEP):

$$NEP_{\tau} = k_B T_{Sys} \sqrt{\Delta f_F}$$
  $[NEP_{\tau}] = \frac{W}{\sqrt{Hz}}$ 

Sometimes a little bit different NEP definitions are used, most of them have factor 2 or 2<sup>1/2</sup> included
 (Because of 2 polarizations or time to bandwidth conversion)

$$P_N = k_B T_{Sys} \sqrt{\frac{\Delta f_F}{\tau}}$$

...

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## Broadband detectors

- Types of broadband detectors
  - Bolometers
  - Microwave kinetic inductance detector (MKID)
  - Double quantum well detectors
  - Transition edge sensors (TES)
- Usually they work good only at higher frequencies (> 50 ... 100 GHz)
- Often the devices are background limited
  - Example: Background temperature 300 K, bandwidth 50 GHz  $\rightarrow$  NEP = 9.2 10<sup>-16</sup> W Hz<sup>-1/2</sup>
- Temperature and bandwidth can be reduced, but then again the other noise sources start to dominate (see later)!

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#### Comparison: Heterodyne $\leftrightarrow$ Direct Det.

Noise equivalent power of a heterodyne system:



[1] K.C. Fong and K.C. Schwab, "Ultra-sensitive and Wide Bandwidth Thermal Measurements of Graphene at Low Temperatures", 2012

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#### Heterodyne Detection: Real Devices

#### • Noise temperature limit for InP devices:

• Mainly phonon self heating  $\rightarrow$  Inner bulk black body radiator

InP-HFET, Bryerton et. al. "Ultra Low Noise Cryogenic Amplifiers for Radio Astronomy", 2013

Shi, et. al.

A 100-GHz Fixed-Tuned Waveguide SIS Mixer Exhibiting Broad Bandwidth and Very Low Noise Temperature, 1997

InP-HEMT Our amplifier, LNF



#### First Cold Measurement

- First quick and dirty test:
  - Very simple test in LHe-dewar
  - Amplifier at LHe-temperature (4.1K)



#### Room for improvement!

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#### **Run Optimization**

- Measurement time vs. analysis threshold level and power boost factor:
  - 80 disks, LaAlO<sub>3</sub>, T<sub>sys</sub>=8K, effectivity: 75%, 1day adjustment time



#### Sensitivity in terms of Axions



Axion mass (eV)

### Photon Noise Equivalent Power (NEP $_{\gamma}$ )

Photon energy: Noise equivalent power "of a photon":

$$E_{\gamma} = hv$$

 $NEP_{\gamma} = h v \sqrt{2n}$ 

**n:** 

Mean photon flux (background + signal) in 1/s

Frequency v	Photon Energy $E_{\gamma}$	1 γ/s =	NEP <sub><math>\gamma</math></sub> for 1 $\gamma$ /s
10 GHz	6.62 10 <sup>-24</sup> J (41.36 µeV)	6.62 10 <sup>-24</sup> W	9.4 10 <sup>-24</sup> W Hz <sup>-1/2</sup>
20 GHz	1.33 10 <sup>-23</sup> J (82.71 µeV)	1.33 10 <sup>-23</sup> W	1.87 10 <sup>-23</sup> W Hz <sup>-1/2</sup>
50 GHz	3.31 10 <sup>-23</sup> J (206.8 µeV)	3.31 10 <sup>-23</sup> W	4.69 10 <sup>-23</sup> W Hz <sup>-1/2</sup>
100 GHz	6.62 10 <sup>-23</sup> J (413.6 µeV)	6.62 10 <sup>-23</sup> W	9.4 10 <sup>-23</sup> W Hz <sup>-1/2</sup>

#### Double quantum dot

- Function principle
  - Absorption of photon with energy hv
  - Electron in QD1 is excited to QD2 (tunneling)
  - Electron can leave to drain lead and new electron enters from the source
  - Then cycle can be repeated
  - Current flow through the system

•  $\delta$  can be changed by electric field



#### Microwave Kinetic Inductance Detector

#### • Function principle:

- Breaking cooper pairs in a superconductor (inductor) by photons
- Stored energy (inner inductance) is changed
- Resonance frequency of the resonator shifts



#### Superconducting Gap Energy

<u>Aluminum</u>	3.4×10 <sup>-4</sup> eV
Cadmium	1.5×10 <sup>-4</sup> eV
<u>Gallium</u>	3.3×10 <sup>-4</sup> eV
Indium	10.5×10 <sup>-4</sup> eV
Lanthanum	
<u>β-lanthanum</u>	19×10 <sup>-4</sup> eV
Lead	27.3×10 <sup>-4</sup> eV
Mercury	
<u>α-mercury</u>	16.5×10 <sup>-4</sup> eV
Molybdenum	2.7×10 <sup>-4</sup> eV
<u>Niobium</u>	30.5×10 <sup>-4</sup> eV
<u>Tantalum</u>	14×10 <sup>-4</sup> eV
<u>Thallium</u>	7.35×10 <sup>-4</sup> eV
Tin	
white tin	11.5×10 <sup>-4</sup> eV
Vanadium	16×10 <sup>-4</sup> eV
Zinc	2.4×10 <sup>-4</sup> eV

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# SIS-Mixer (Principle)

- Cooper pairs break into quasi-particles and tunnel over the barrier
- Using photon assistant tunneling for mixing
- Slope of I-V curve has sharp discontinuity: efficient



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# SIS-Mixer (Principle)

- Mixer loss -> higher noise temperature
- Double-sideband feature -> looking at two frequencies at the same time



J. Zmuidzinas, "COHERENT DETECTION AND SIS MIXERS", 2002

#### Detection of a broadband noise signal

- Frequency: 15 GHz
- Linewidth: 200 kHz
- Detection bandwidth: 10 kHz



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- Detection of a line signal (Examples)
  - Frequency: 15 GHz
  - Detection bandwidth: 10 kHz



Real signal:

@ Signal Analyzer (10<sup>-13</sup> W)

Power ( 5,17

5,21

5,20 5,19

5,18

5.16

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