

# Development of semiconductor solid-state detectors with sub-100ps time resolution.

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On behalf of the **TT-PET collaboration**

# The fast silicon pixel detector

- At the beginning of the LHC era, the RPC was the only detector capable of sub-100ps time resolution.
- Silicon pixel technology was focusing on tracking, power consumption and radiation hardness, with a time resolution in the range from few ns to some tens of ns.
- The roadmap for the development of **silicon pixel detectors with a 100ps (or better) time resolution** will be described **through the experience of the TT-PET collaboration**.

# The TT-PET project

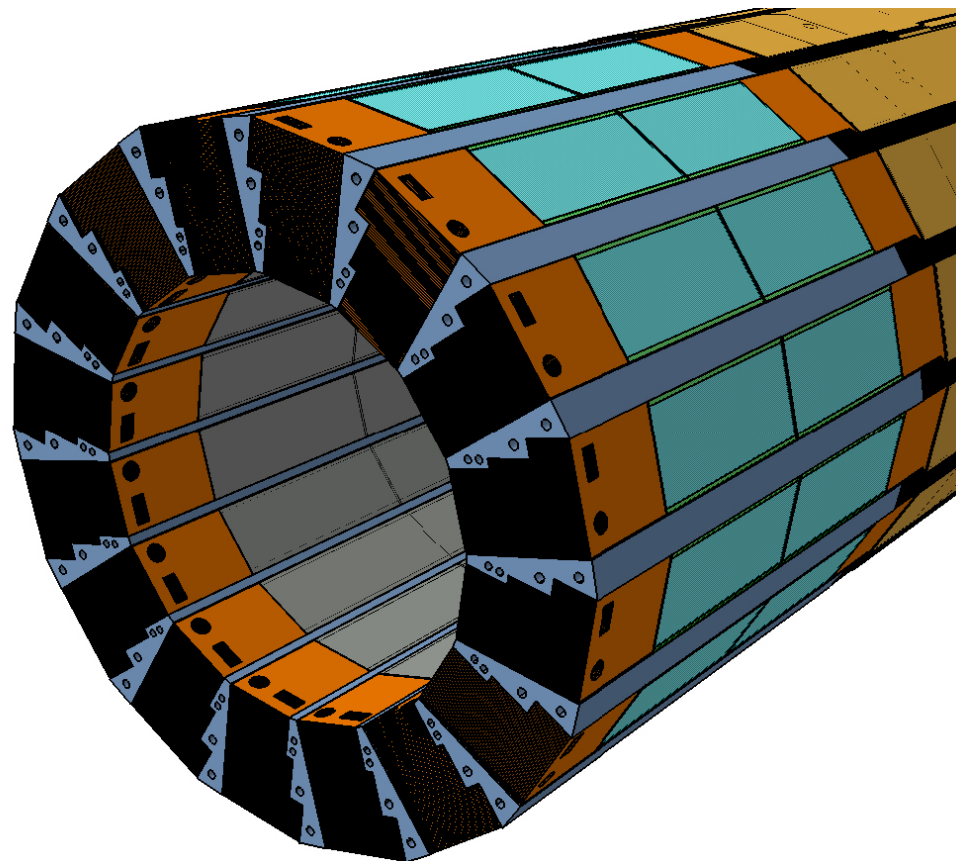
A 3-year project financed by SNSF to produce a PET Scanner for small animals **based on silicon detector technology**, insertable in an MRI machine and with 30ps RMS time resolution. **The project started in March 2016.**

## The TT-PET collaboration:

- University of Geneva
- University of Bern
- Hôpital cantonale de Genève
- INFN of Roma Tor Vergata
- CERN
- Stanford University

## Other collaborators:

- IHP Microelectronics



# Targets of the TT-PET project

➡ 1. Make a 30ps time resolution detector for 511 keV photons and...

What are the main parameters to improve for the time resolution of semiconductor detectors?

➡ Read out geometry.

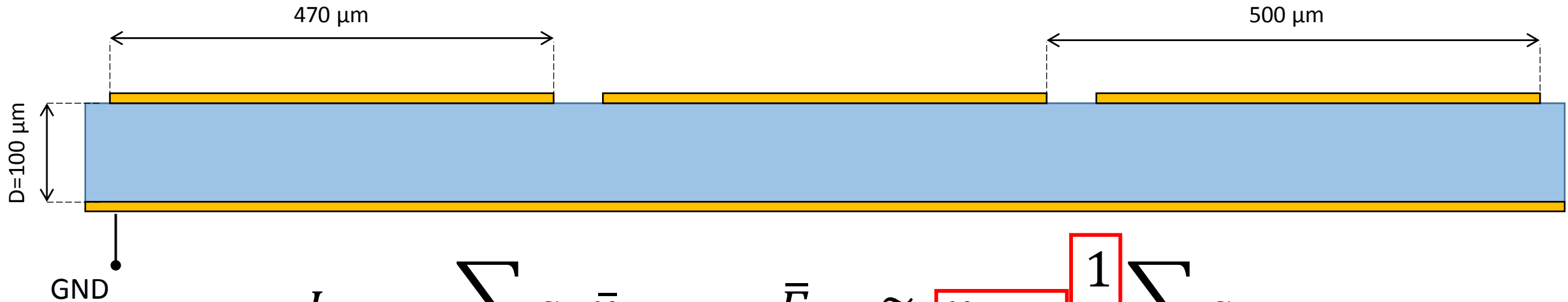
➡ Electronics noise.

➡ Charge collection noise.

$$I_{ind} = \sum_i q_i \bar{v}_{drift,i} \cdot \bar{E}_{w,i}$$

# Read out geometry

The “parallel plate” read out is fundamental to guarantee the uniformity of the weighting and the electric field.



$$I_{ind} = \sum_i q_i \bar{v}_{drift,i} \cdot \bar{E}_{w,i} \cong \boxed{v_{drift}} \boxed{\frac{1}{D}} \sum_i q_i$$

Scalar, possibly saturated

Scalar, uniform

## Drawback:

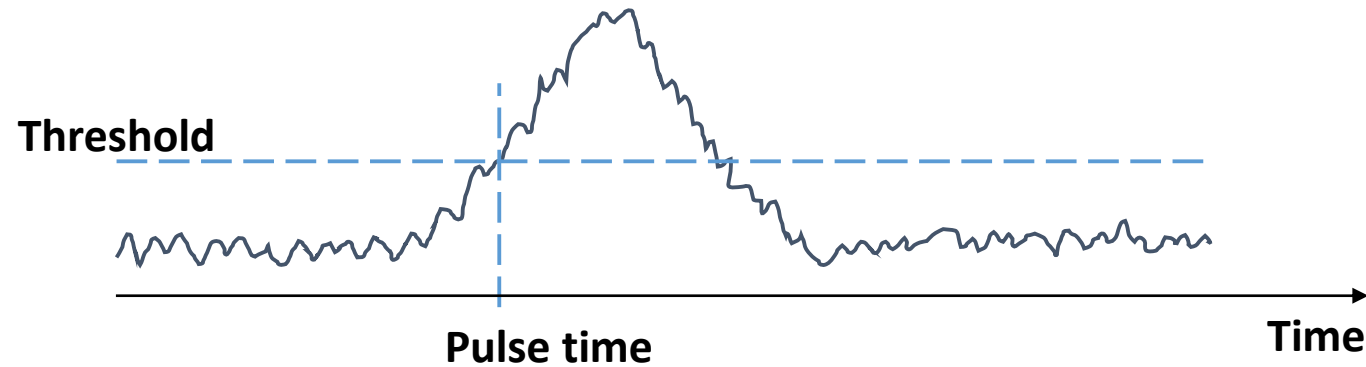
Increase of the pixel capacitance



Larger equivalent noise from the amplifier.

# Electronics noise

Detector time resolution depends mostly on the amplifier performance!



$$\sigma_t = \frac{\sigma_V}{\frac{dV}{dt}} \cong \frac{\text{Rise Time}}{Q/ENC} \quad \boxed{I_{ind}} = v_{drift} \frac{1}{D} \sum_i q_i$$

1. Fast, low noise electronics: 1 ns rise time,  $< 1000 e^- ENC$  on 1 pF capacitance.



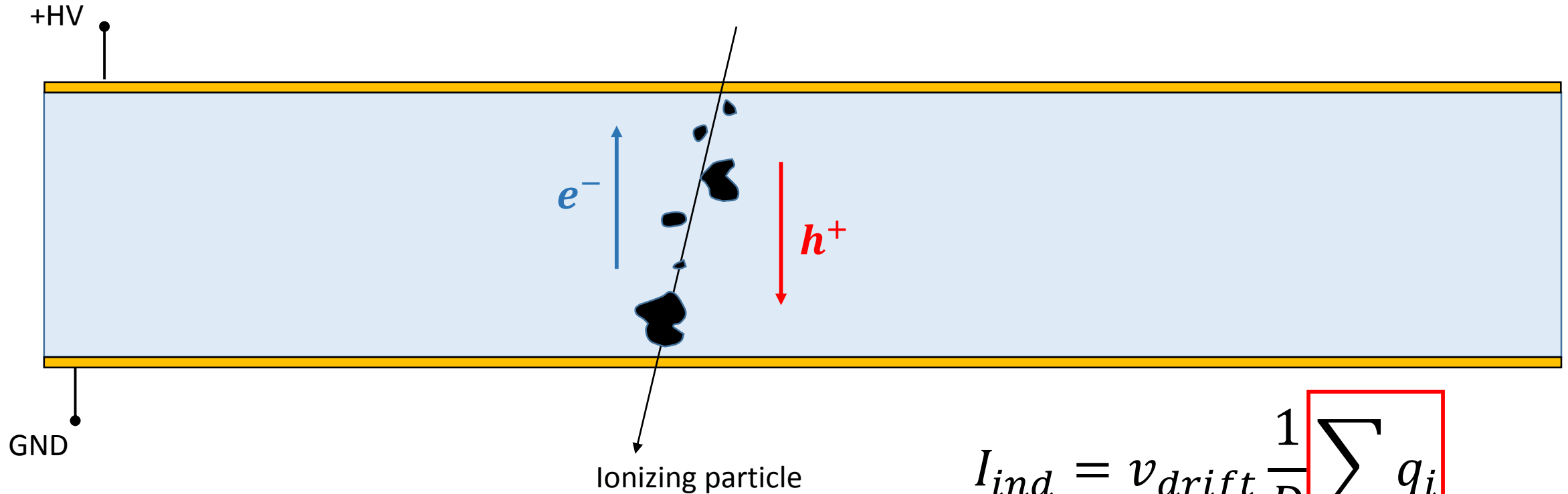
High  $f_t$ , single transistor preamplifier.



**SiGe HBT** technology.

2. Gain...?

# Charge collection noise



$$I_{ind} = v_{drift} \frac{1}{D} \sum_i q_i$$

- Intrinsic limit to the time resolution for a semiconductor detector.
- Can be reduced by reducing the sensor thickness.

# Targets of the TT-PET project

1. Make a 30ps time resolution detector for 511 keV photons and...

 2. ... make it monolithic.

Both sensor and electronics **integrated in the same chip**, in a commercial microelectronics process.

## Advantages:

- Simplified interconnections.
- Integrated front end, TDC, logic, serializer.
- Cost reduction



# The TT-PET ASIC

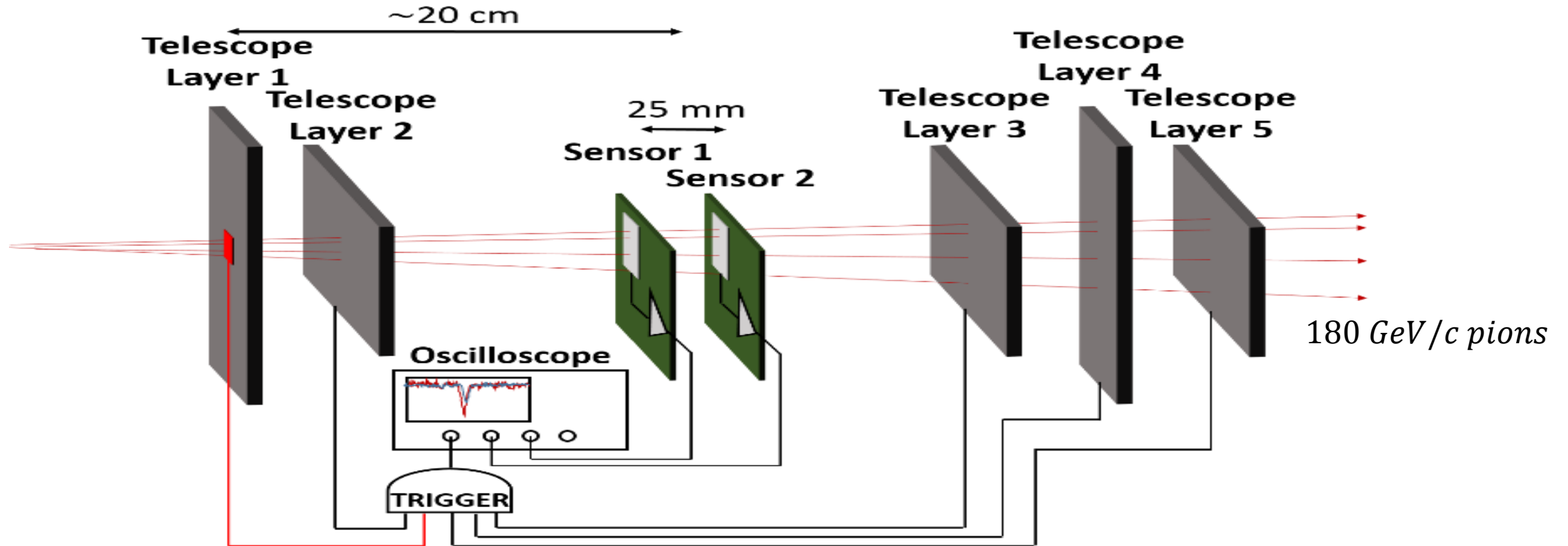
A monolithic silicon pixel detector:

ASIC length	24 <i>mm</i>
ASIC width	7, 9, 11 <i>mm</i>
Pixel Size	500 × 500 $\mu m^2$
Pixel Capacitance (comprised routing)	750 <i>fF</i>
Preamplifier power consumption	200 $\mu W / channel$
Preamplifier E.N.C.	600 $e^- RMS$
Preamplifier Rise time (10% - 90%)	800 <i>ps</i>
Time resolution for MIPs	100 <i>ps RMS</i>
TDC time binning	20 <i>ps</i>
TDC power consumption	< 1 <i>mW / channel</i>

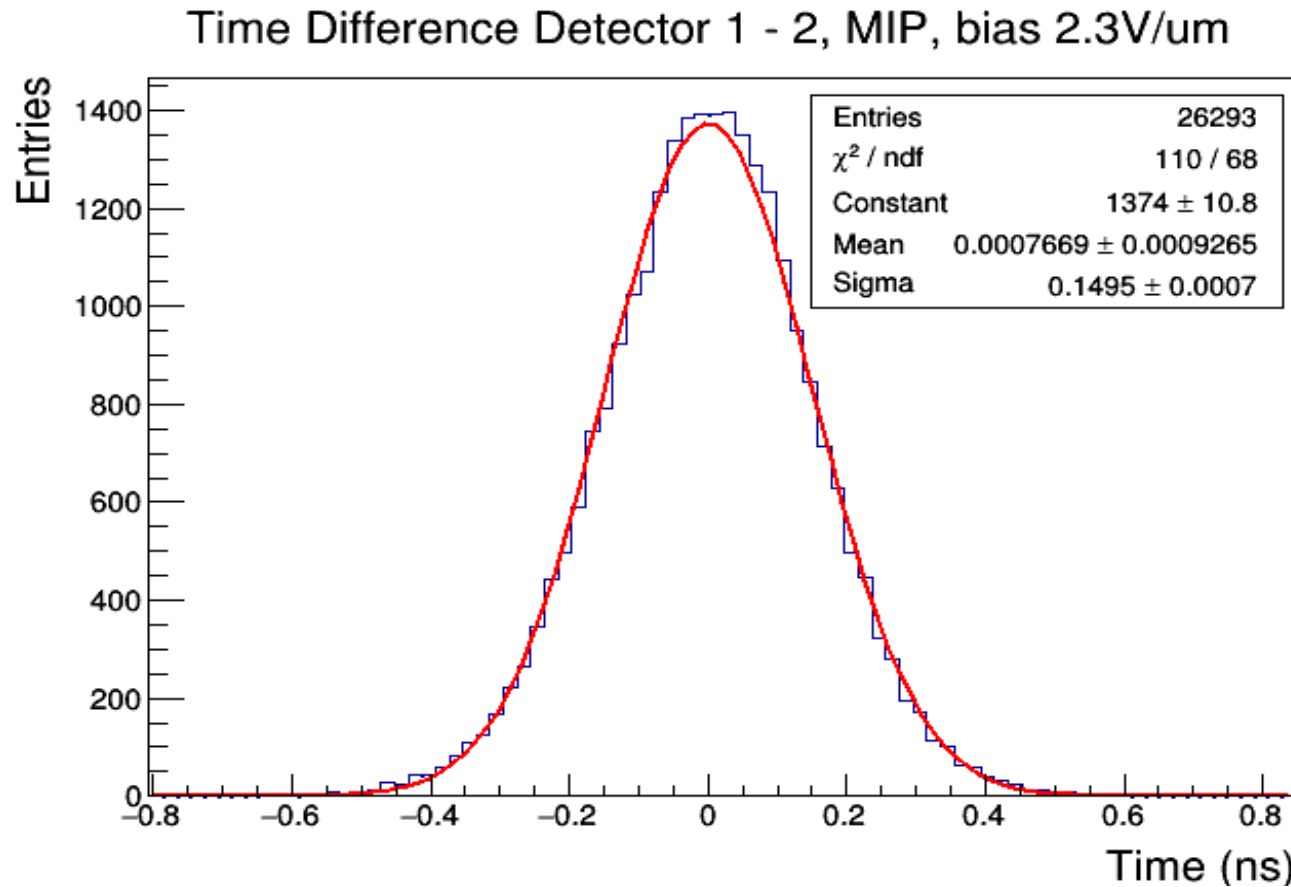
# Results from the R&D

# Proof of principle

Test of external sensor and custom SiGe HBT preamplifier at H8 beam line at SPS (CERN).



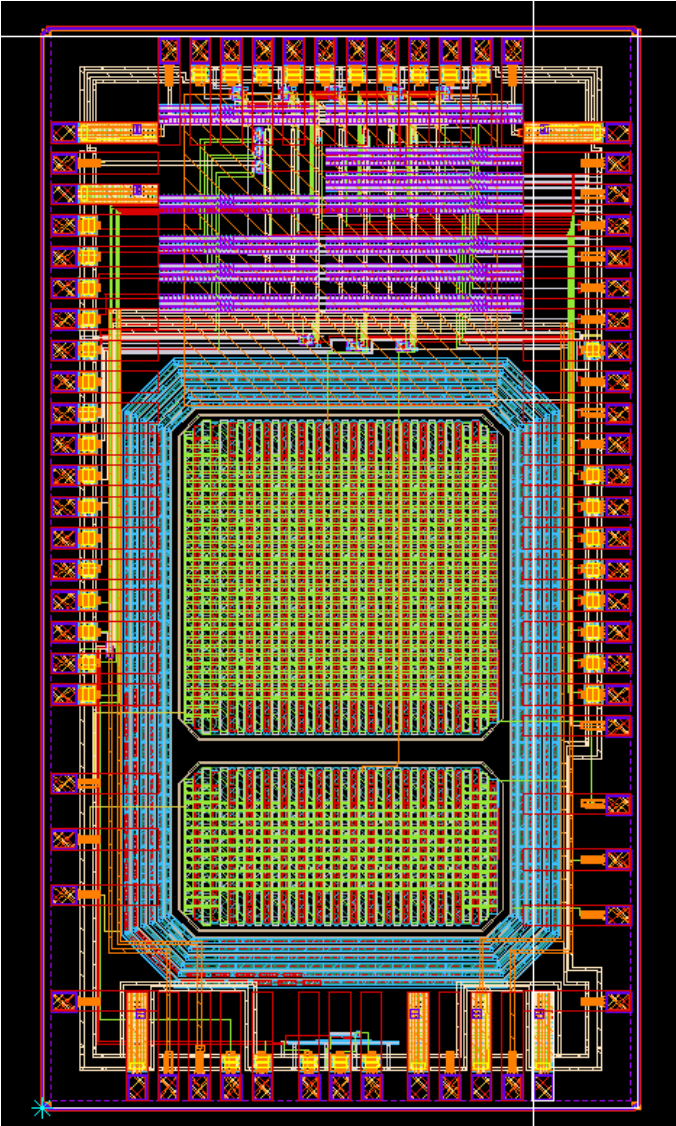
# Proof of principle



$$\sigma_t = \frac{(150 \pm 1)ps}{\sqrt{2}} = (106 \pm 1)ps.$$

**100ps time resolution  
measured with MIPs**

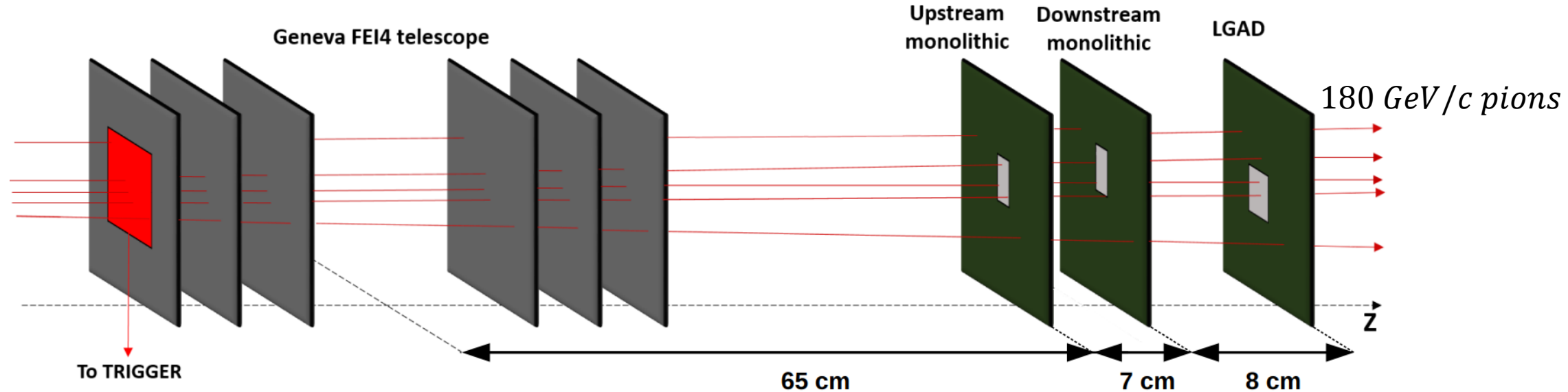
# 1<sup>st</sup> monolithic prototype



- Two pixels + amplifier + discriminator in standard IHP SG13S process.
- Pixel size:  $900 \times 900 \mu m^2$  and  $900 \times 450 \mu m^2$ .
- Higher wafer resistivity (1 kOhmcm)
- No thinning, no backplane metallization.

# 1<sup>st</sup> monolithic prototype

Test of the 1<sup>st</sup> monolithic prototype at H8 beam line at SPS (CERN).

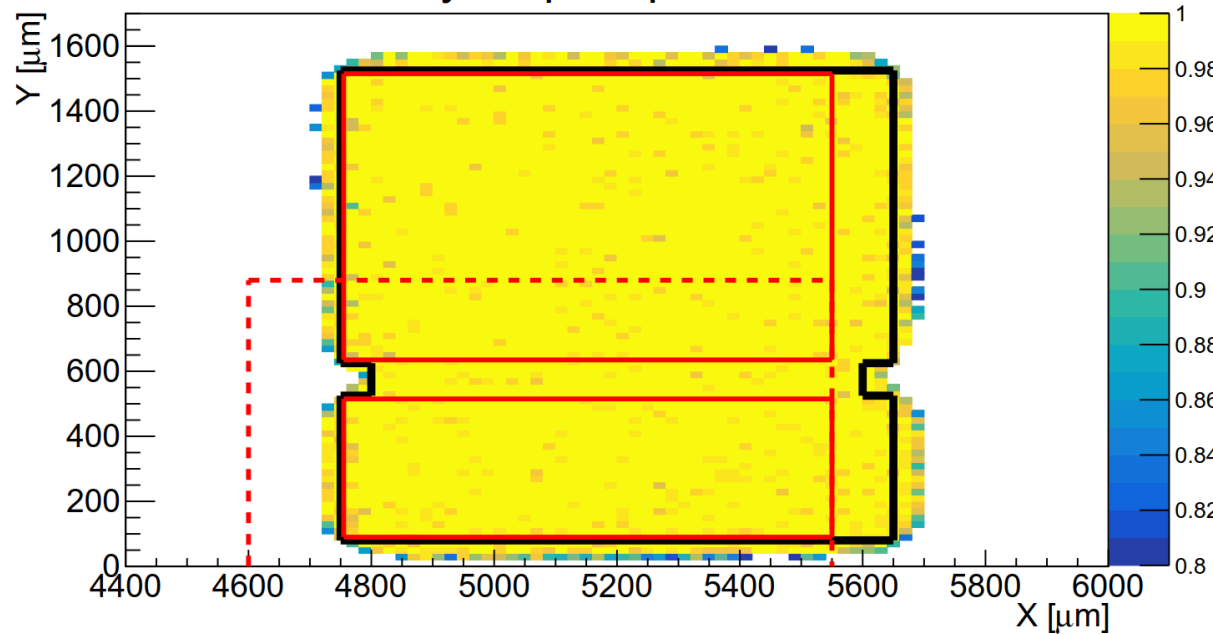


. Events triggered by the Geneva telescope.

. Typical signal charge: 1.6 fC.

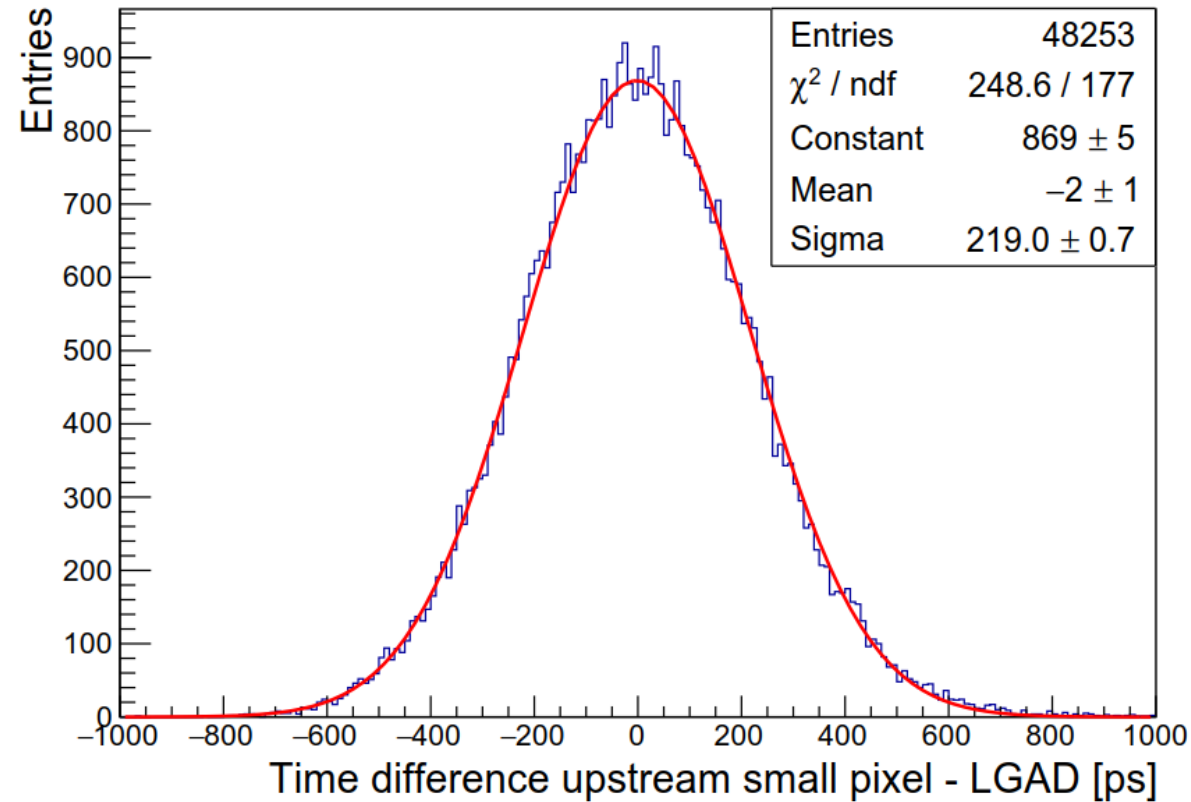
# 1<sup>st</sup> monolithic prototype

Efficiency Map - Upstream Sensor



- Efficiency 99.8%.
- Amplifier  $ENC < 600 \text{ electrons RMS}$ .
- Amplifier power consumption:  $< 350 \frac{\mu W}{channel}$

# 1<sup>st</sup> monolithic prototype

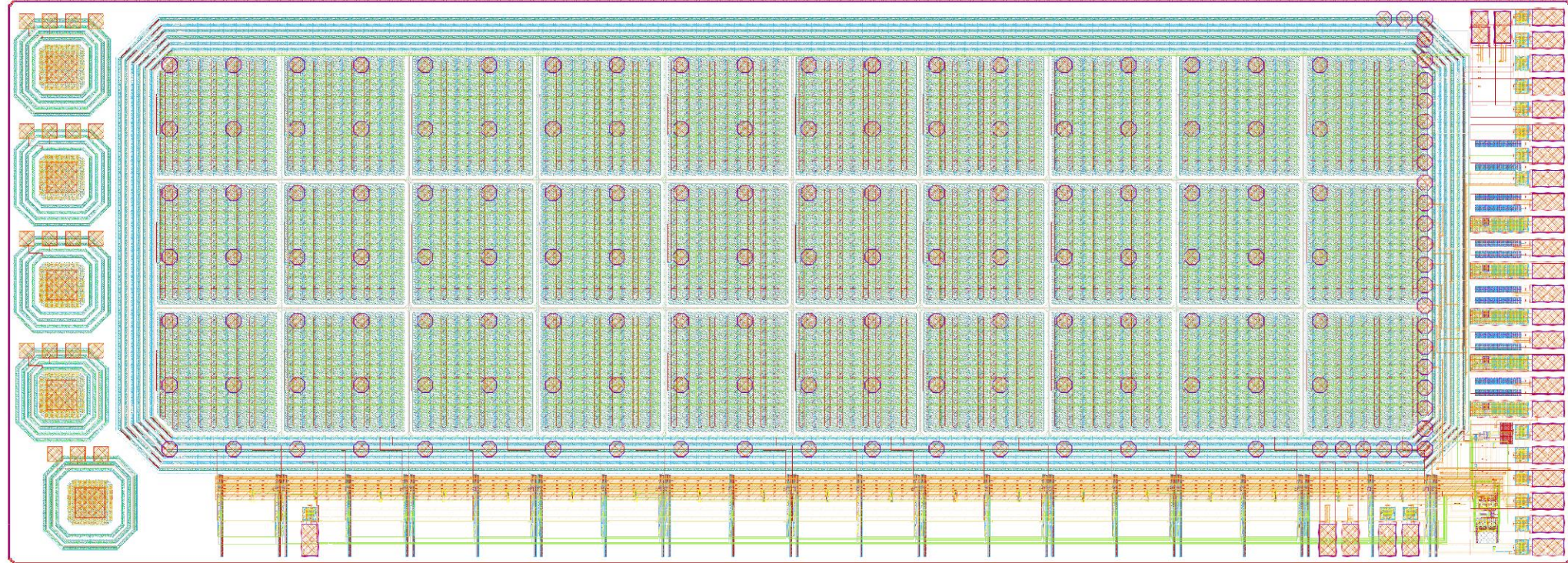


**Time resolution  $\sim 220$  ps RMS**

(Strongly affected by the absence of fundamental backside processing steps.)



## 2<sup>nd</sup> monolithic prototype



- $3 \times 10$  matrix,  $500 \times 500 \mu\text{m}^2$  pixels.
- Preamplifier, discriminator, 20 ps time resolution TDC, logic, serializer integrated in chip.
- Thinned to  $100 \mu\text{m}$ .
- Full backside processing.
- **Back from foundry next month.**

# Is it possible to go below 100 ps?

Possible approaches:

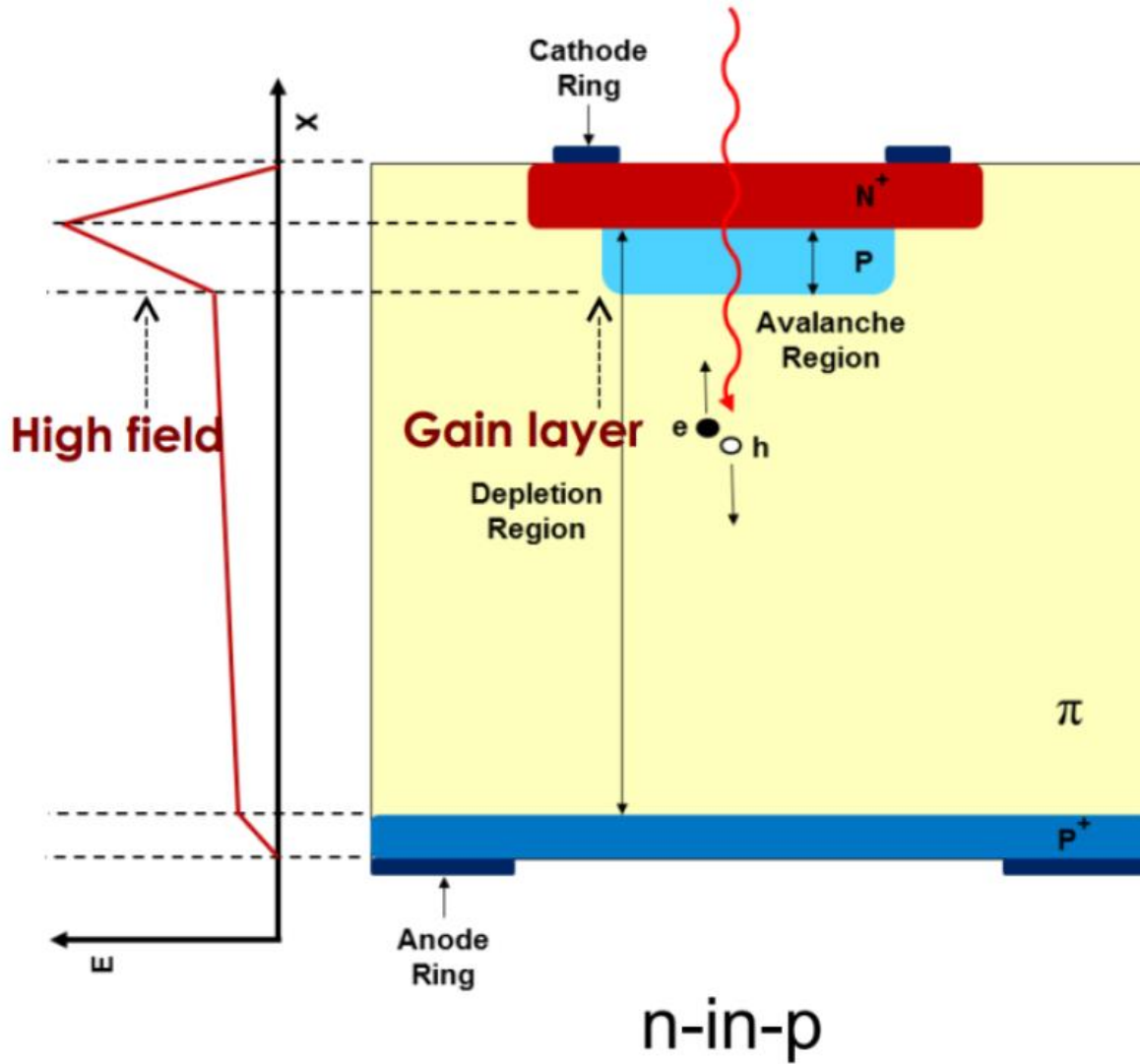
1. Reduce pixel size and sensor thickness.

 New monolithic design (under study)

2. The Low Gain Avalanche Diode (LGAD).

 See next slide

# The LGAD



- Proposed as timing detector for the new ATLAS HGTD.
- The is kept low:  $G \lesssim 20$



# Test beam with CNM LGAD sensors

Our discrete-component SiGe amplifier was coupled to a LGAD produced by CNM and kindly provided by Sebastian Grinstein (IFAE Barcelona)

Thickness:  $\sim 45 \mu m$

PAD size:  $1.3 \times 1.3 \text{ mm}^2$

Performance (bias 220V):

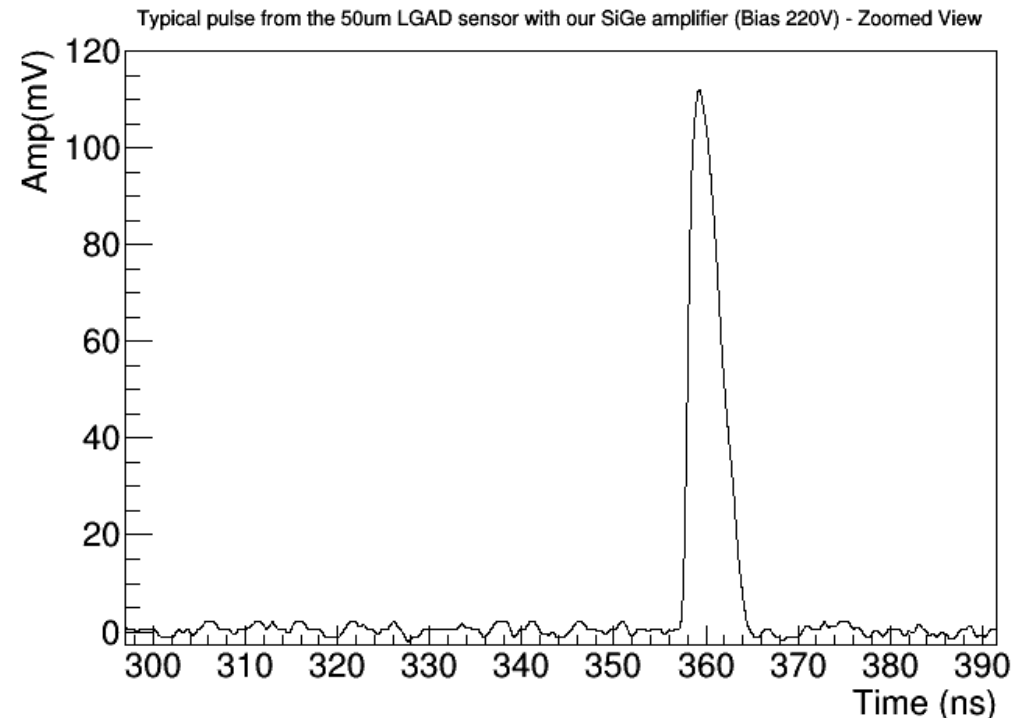
$$S/N = 124$$

$$t_{\text{rise},20\%-80\%} = 600 \text{ ps}$$

$$\frac{\sigma_V}{dV/dt} = 8 \text{ ps}$$

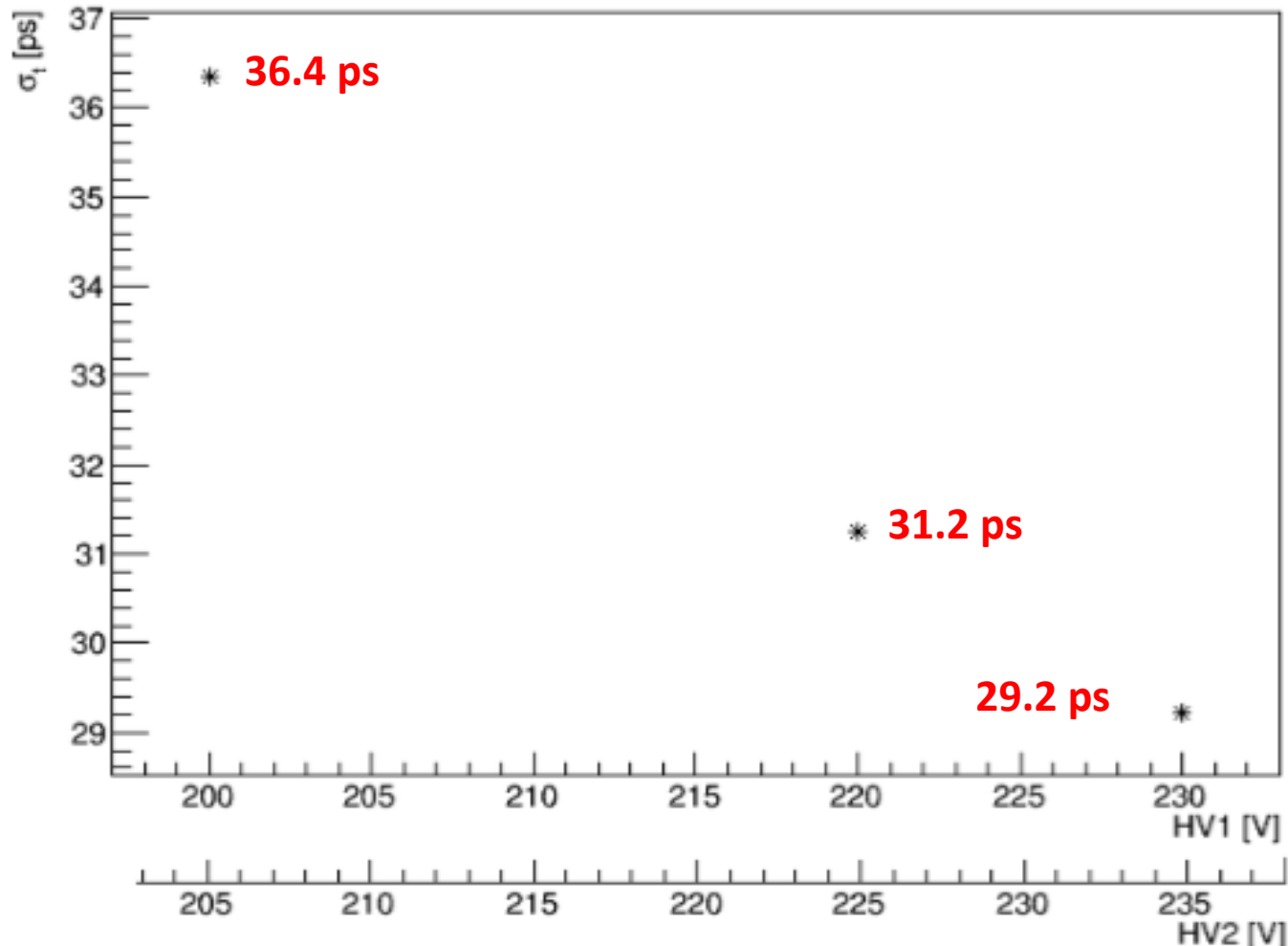


Expected electronics contribution to the time resolution.



# Test beam with CNM LGAD sensors

LGAD time resolution with MIPs measured at H8 beam line at SPS (CERN).



Compatible with expectations from charge collection noise.

# Conclusions

- Silicon pixel technology is now able to combine its sub-millimetre space resolution, very high counting rate capability and excellent radiation hardness with a 100 ps time resolution.
- The TT-PET collaboration is implementing this fast pixel detector in a monolithic structure in a SiGe BiCMOS process.
- Even better time resolution is possible introducing a gain layer in the sensitive volume.
- An important part of the expertise at the base of this development comes from the experience obtained on the development of the Resistive Plate Chambers and their front-end electronics.

# Extra Material

## Minimization of ENC for a fast integrator

$$ENC^2 \propto \left( 2q_e I_C + \frac{4kT}{R_P} + i_{na}^2 \right) \cdot \tau + \boxed{(4kTR_S + e_{na}^2) \cdot \frac{C_{in}^2}{\tau}} + 4A_f C_{in}^2$$

Dominating term

Excellent performance in terms of series noise for fast shaping are achievable with the BJT technology

$$ENC_{series\ noise} \propto \sqrt{2kT \langle SNI \rangle \left[ (C_{in})^2 \frac{h_{ie}}{\beta} + R_{bb} C_{in}^2 \right]}$$

Transistor ENC contribution depends on current gain and base spreading resistance



# SiGe technology for very low noise and fast amplifiers

Amplifier current gain can be expressed as (NPN BJT)

$$\beta = \frac{i_C}{i_B} = \frac{\tau_p}{\tau_t}$$

$\tau_p$  = hole recombination time in base

$\tau_t$  = **electron transit time (E to C)**

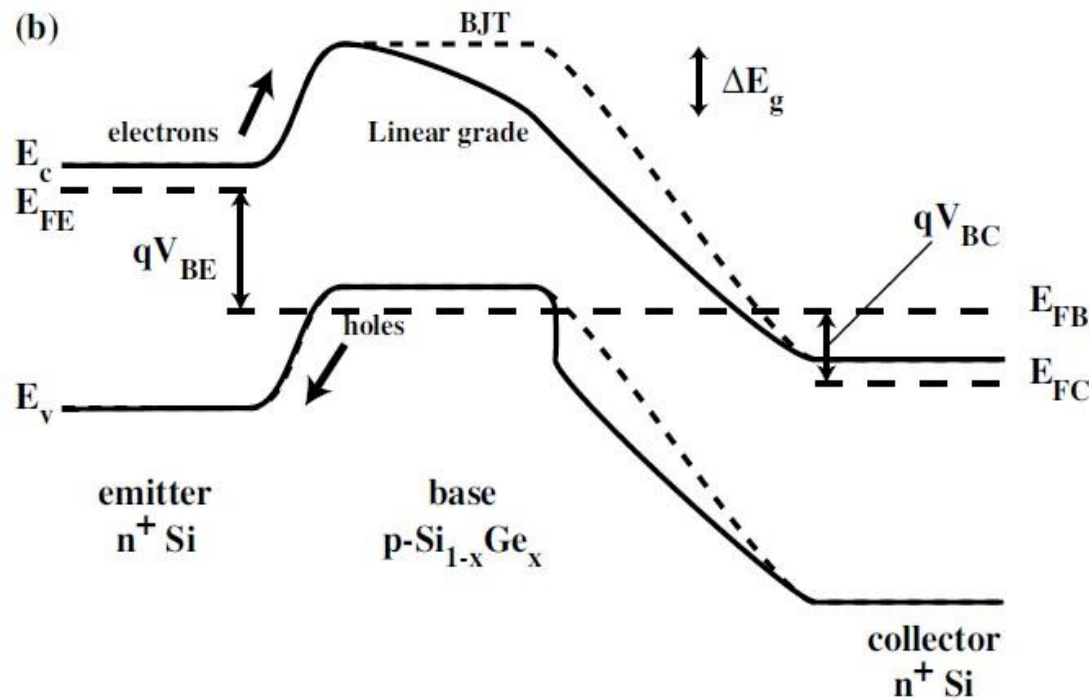
**Need to minimize electron transit time in the base**

Increase gain  Reduce base width  Reducing base doping

**Spreading resistance increases!**

# SiGe technology for very low noise and fast amplifiers

A possible approach: changing the charge transport mechanisms in the base from diffusion to drift.



**SiGe heterojunction bipolar transistor technology.**

The technology we have chosen is **SG13S from IHP:**

$$\beta = 900$$

$$f_t = 250 \text{ GHz}$$

Equivalent to introducing an electric field in the base.