

Port-Hamiltonian Description of Electro-Thermal Field-Circuit models

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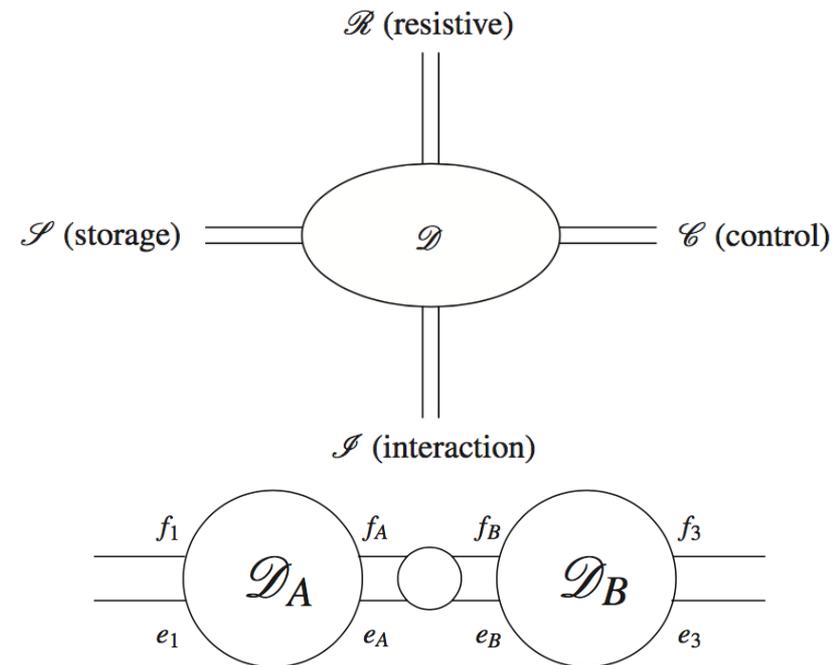


Outline

- Motivation
- Introduction
- Electromagnetic Field
- Irreversible Thermodynamics
- Conclusions

Motivation

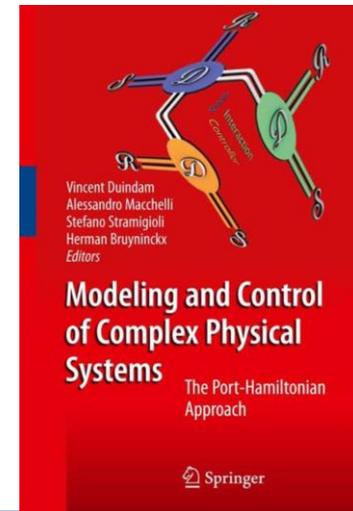
- General framework for the multi-physics modelling
- Computational causality
- Analysis of the coupling strength between physical systems
- Distributed and lumped element phenomena



1.12 Future Trends

The following general future trends in bond graphs and port-based modeling can be distinguished:

- ...
- use of the port-based approach for *co-simulation*.



V. Duindam, et al. Modelling and Control of Complex Physical Systems
The Port-Hamiltonian Approach, Springer, 2009.

Modelling with bond graphs

Effort and Flow



$$\text{effort} \cdot \text{flow} = \text{Power}$$

Causality assignment

1. Fixed
2. Arbitrary
3. Preferred
4. Restricted

Flow and effort



Resistor



Inductor and Capacitor



$$f(t) = \frac{1}{I} \int e(t) dt$$

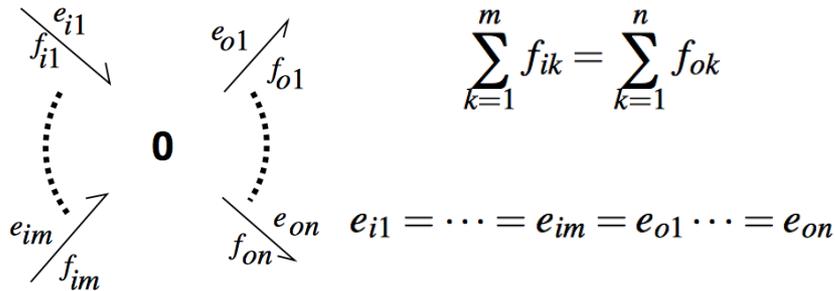
$$e(t) = \frac{1}{C} \int f(t) dt$$

	f flow	e effort	$q = \int f dt$ generalized displacement	$p = \int e dt$ generalized momentum
electro-magnetic	i current	u voltage	$q = \int i dt$ charge	$\lambda = \int u dt$ magnetic flux linkage
mechanical translation	v velocity	F force	$x = \int v dt$ displacement	$p = \int F dt$ momentum
mechanical rotation	ω angular velocity	T torque	$\theta = \int \omega dt$ angular displacement	$b = \int T dt$ angular momentum
hydraulic pneumatic	ϕ volume flow	p pressure	$V = \int \phi dt$ volume	$\Gamma = \int p dt$ momentum of a flow tube
thermal	T temperature	f_S entropy flow	$S = \int f_S dt$ entropy	

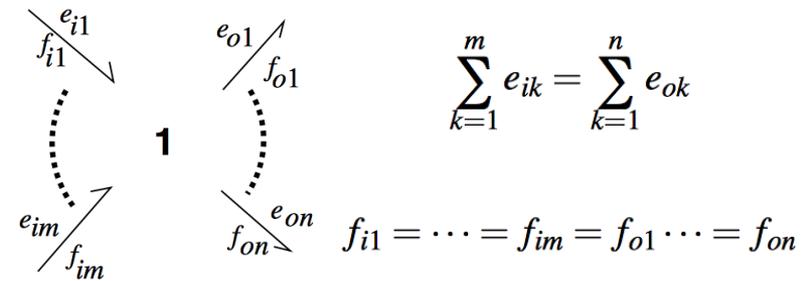


Modelling with bond graphs

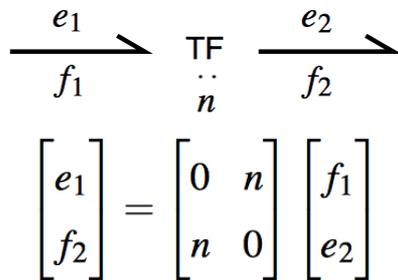
0-junction



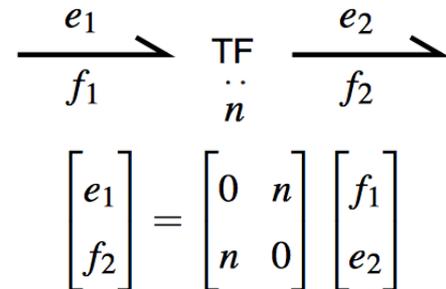
1-junction



Transformer



Gyrator



Input-state-output Port Hamiltonian Model

$$\begin{aligned}\partial_t x &= [J(x) - R(x)]\delta_x \mathbb{H}(x) + g(x)u \\ y &= g^\top(x)\delta_x \mathbb{H}(x) + S(x)u,\end{aligned}$$

where $x \in \mathbb{R}^n$ is the state vector,

$J(x) = -J(x)^\top \in \mathbb{R}^n \times \mathbb{R}^n$ is a skew-symmetric interconnection matrix

$R(x) = R(x)^\top \in \mathbb{R}^n \times \mathbb{R}^n$ is a symmetric matrix corresponding to the resistive port.

$g(x) \in \mathbb{R}^n \times \mathbb{R}^m$ is the input vector field

$S(x) \in \mathbb{R}^m \times \mathbb{R}^m$ is a direct feed-through matrix

B. Maschke and A. van der Schaft, "From conservation laws to port-hamiltonian representations of distributed-parameter systems," in *16th IFAC World Congress*, Elsevier, 2005.

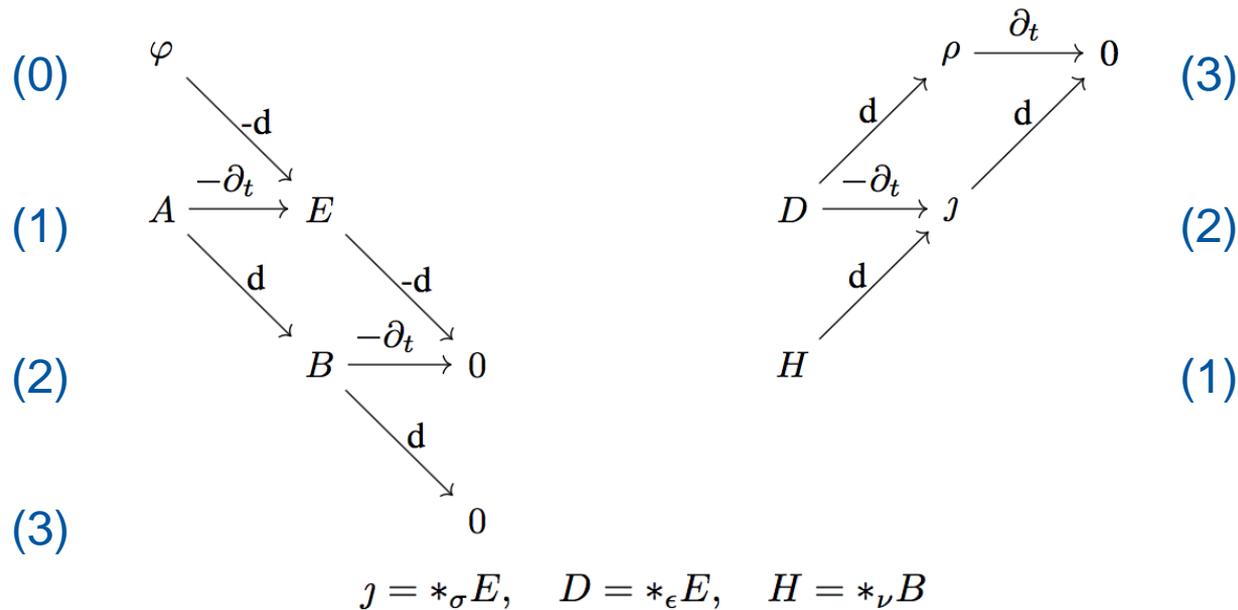
Modelling of Magnetoquasistatic Domain

For a superconducting magnet model we consider a magnetoquasistatic model including

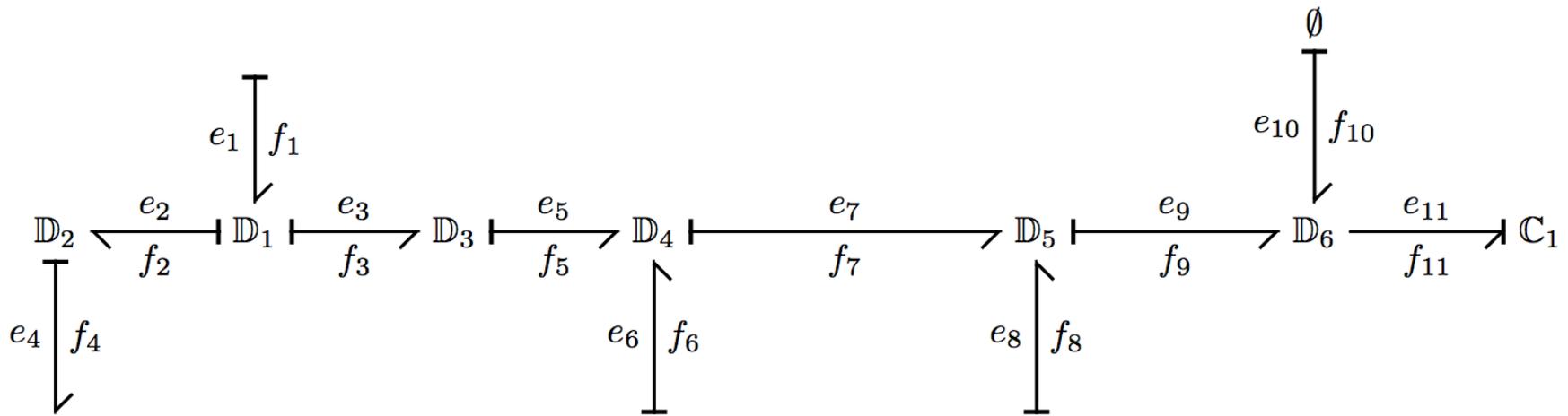
1. energy stored in the magnetic field,
2. resistive voltage due to a quench,
3. induced voltage influenced by the eddy currents and cable coupling currents.

Electromagnetic Field – Topological Diagram

Symbol	Form order	Unit	Description
φ	$\Lambda^0(\Omega)$	V	Electric scalar potential
Ψ	$\Lambda^0(\Omega)$	A	Magnetic scalar potential
A	$\Lambda^1(\Omega)$	Wb	Magnetic vector potential
E	$\Lambda^1(\Omega)$	V	Electric field strength
H	$\Lambda^1(\Omega)$	A	Magnetic field strength
D	$\Lambda^2(\Omega)$	C	Electric flux density
B	$\Lambda^2(\Omega)$	Wb	Magnetic flux density
j	$\Lambda^2(\Omega)$	A	Electric current density
ρ	$\Lambda^3(\Omega)$	C	Electric charge density



Electromagnetic Field – Bond Graph Model (1/x)



For a superconducting magnet model we consider a magnetoquasistatic model including

1. energy stored in the magnetic field,
2. resistive voltage due to a quench,
3. induced voltage influenced by the eddy currents and cable coupling currents.

Power-continuity is the key.

Electromagnetic Field – Bond Graph Model (1/x)

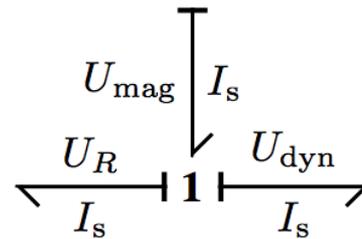
The voltage across a magnet is composed of a resistive and dynamic component

$$U_{\text{mag}} = U_R + U_{\text{dyn}}$$

which forms a 0-junction

$$\mathcal{D}_1 = \{(f_1, f_2, f_3, e_1, e_2, e_3) \in \mathcal{F}_1 \times \mathcal{E}_1 : f_1 = f_2 = f_3, e_1 + e_2 + e_3 = 0\}$$

$$\text{with } (f_1, f_2, f_3, e_1, e_2, e_3) = (I_s, I_s, I_s, U_{\text{mag}}, -U_R, -U_{\text{dyn}})$$



Electromagnetic Field – Bond Graph Model (1/4)

The winding density function distributes current in the domain and forms a transformer

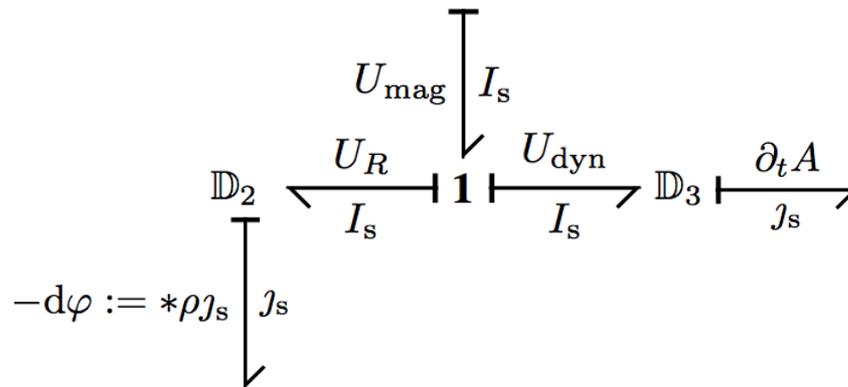
$$\begin{bmatrix} e_a \\ f_b \end{bmatrix} = \begin{bmatrix} 0 & -\chi^\top \\ \chi & 0 \end{bmatrix} \begin{bmatrix} f_a \\ e_b \end{bmatrix} \quad f^\top e = 0$$

For the resistive voltage, the conjugate power pair takes form

$$\begin{bmatrix} e_a \\ f_a \end{bmatrix} = \begin{bmatrix} U_R \\ I_s \end{bmatrix}, \quad \begin{bmatrix} e_b \\ f_b \end{bmatrix} = \begin{bmatrix} d\varphi \\ j_s \end{bmatrix}$$

and for the inductive voltage

$$\begin{bmatrix} e_a \\ f_a \end{bmatrix} = \begin{bmatrix} U_{\text{dyn}} \\ I_s \end{bmatrix}, \quad \begin{bmatrix} e_b \\ f_b \end{bmatrix} = \begin{bmatrix} \partial_t A \\ j_s \end{bmatrix}$$



Electromagnetic Field – Bond Graph Model (2/4)

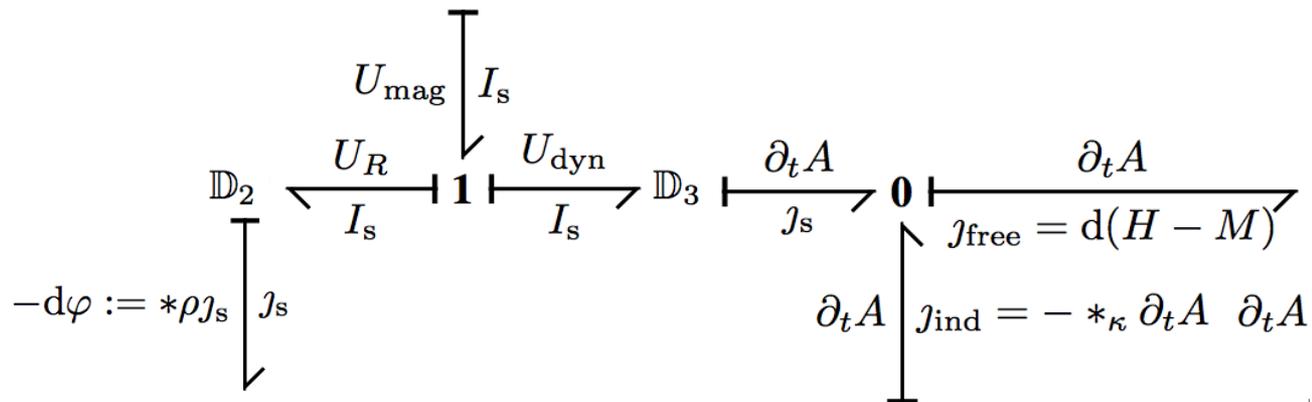
We consider eddy currents in copper wedges

$$j_s = j_{\text{free}} - j_{\text{ind}}.$$

forming a 0-junction

$$\mathcal{D}_4 = \{(f_5, f_6, f_7, e_5, e_6, e_7) \in \mathcal{F}_2 \times \mathcal{E}_2 : f_5 + f_6 + f_7 = 0, e_5 = e_6 = e_7\}$$

$$\text{with } (f_5, f_6, f_7, e_5, e_6, e_7) = (j_s, - *_{\kappa} \partial_t A, -d(H - M), \partial_t A, \partial_t A, \partial_t A).$$



Electromagnetic Field – Bond Graph Model (3/4)

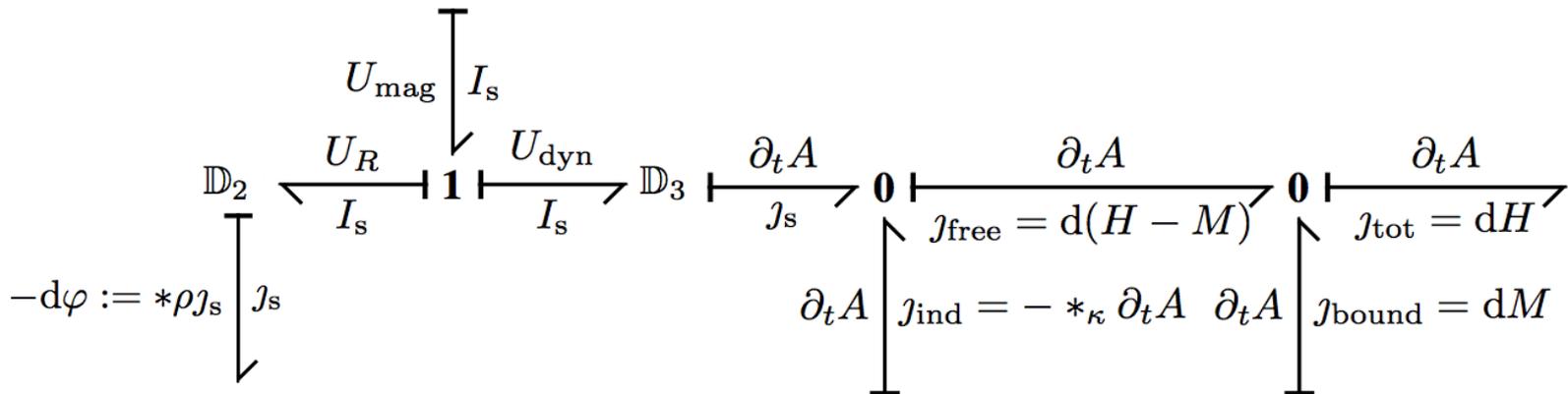
Coupling currents in the fully-transposed cables are accounted for as

$$j_{\text{tot}} = j_{\text{free}} + j_{\text{bound}} = dH$$

forming a 0-junction

$$\mathcal{D}_5 = \{(f_7, f_8, f_9, e_7, e_8, e_9) \in \mathcal{F}_3 \times \mathcal{E}_3 : f_7 + f_8 + f_9 = 0, e_7 = e_8 = e_9\}$$

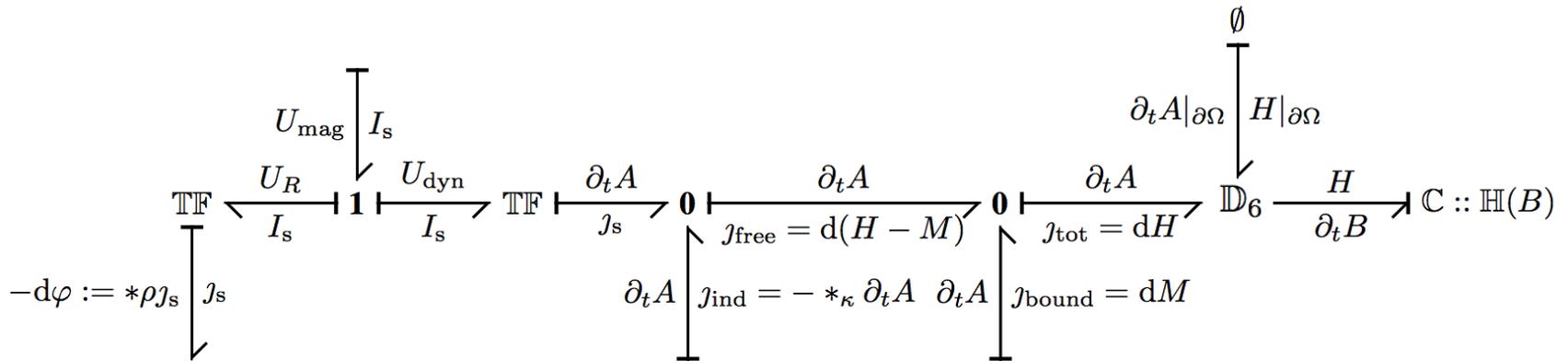
$$(f_7, f_8, f_9, e_7, e_8, e_9) = (-d(H - M), -dM, dH, \partial_t A, \partial_t A, \partial_t A)$$



Electromagnetic Field – Bond Graph Model (4/4)

$$\mathcal{D}_6 = \left\{ \begin{array}{l} (f_9, f_{10}, f_{11}, e_9, e_{10}, e_{11} \in \mathcal{F} \times \mathcal{E}) \\ \begin{bmatrix} f_9 \\ f_{10} \end{bmatrix} = \begin{bmatrix} 0 & -d \\ d & 0 \end{bmatrix} \begin{bmatrix} e_9 \\ e_{10} \end{bmatrix} \\ \begin{bmatrix} f_{11} \\ e_{11} \end{bmatrix} = \begin{bmatrix} e_9|_{\partial\Omega} \\ f_9|_{\partial\Omega} \end{bmatrix} \end{array} \right\}.$$

$$(-\partial_t B, -\partial_t D, f_{11}, H, \partial_t A, e_{11} \in \mathcal{F} \times \mathcal{E}) \in \mathcal{D}_6$$



$$\mathbb{H}(B, D) = \frac{1}{2} \int_{\Omega} *_{\nu} B \wedge B + *_{\frac{1}{\epsilon}} D \wedge D$$

$$\delta_B \mathcal{H} = H$$

$$\delta_D \mathcal{H} = \partial_t A$$

Electromagnetic Field – port-Hamiltonian Model

Combining the Dirac structures together, the state and output equations are given as

$$-\partial_t \begin{bmatrix} B \\ 0 \end{bmatrix} = \left(\begin{bmatrix} 0 & -d \\ d & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 \\ 0 & *_{\kappa} + d *_{\nu\tau} d \end{bmatrix} \right) \begin{bmatrix} H \\ \partial_t A \end{bmatrix} + \begin{bmatrix} 0 \\ \chi \end{bmatrix} I_s$$

$$U_{\text{mag}} = [0 \quad \chi^\top] \begin{bmatrix} H \\ \partial_t A \end{bmatrix} + [0 \quad \chi^\top] \begin{bmatrix} 0 & 0 \\ 0 & *_{\rho} \end{bmatrix} \begin{bmatrix} 0 \\ \chi \end{bmatrix} I_s,$$

which forms an input-state-output port-Hamiltonian system

$$\begin{aligned} \partial_t x &= (J(x) - R(x)) \delta_x \mathbb{H}(x) + gu \\ y &= g^\top \delta_x \mathbb{H}(x) + Su, \end{aligned}$$

The power of the system is bounded by the product of input and output

$$\begin{aligned} \partial_t \mathbb{H}(x) &= \delta_x \mathbb{H}(x)^\top \partial_t x = \delta_x \mathbb{H}(x)^\top ((J(x) - R(x)) \delta_x \mathbb{H}(x) + gu) \\ \partial_t \mathbb{H}(x) &= -\delta_x \mathbb{H}(x)^\top R(x) \delta_x \mathbb{H}(x) + y^\top u - u^\top S^\top u \leq y^\top u. \end{aligned}$$

Electromagnetic Field – port-Hamiltonian Model

Considering the input-state-output equations representing the magnetoquasistatic mode, the variation of the Hamiltonian function reads

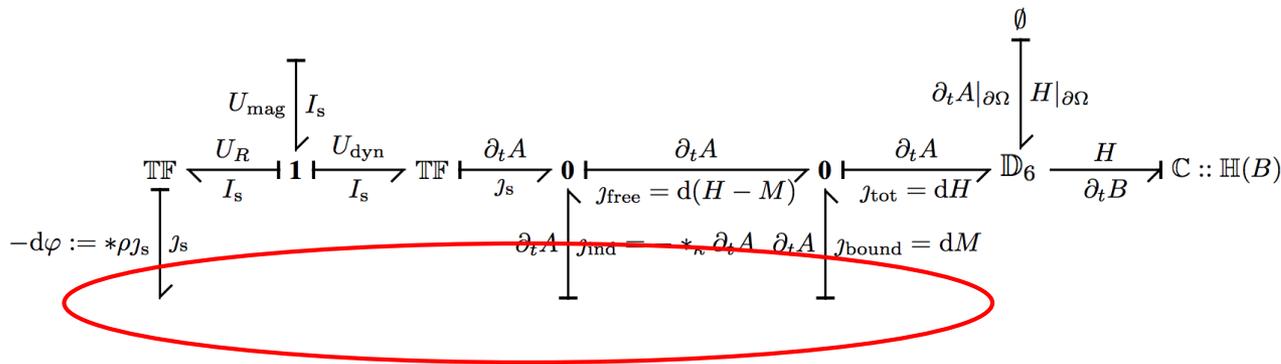
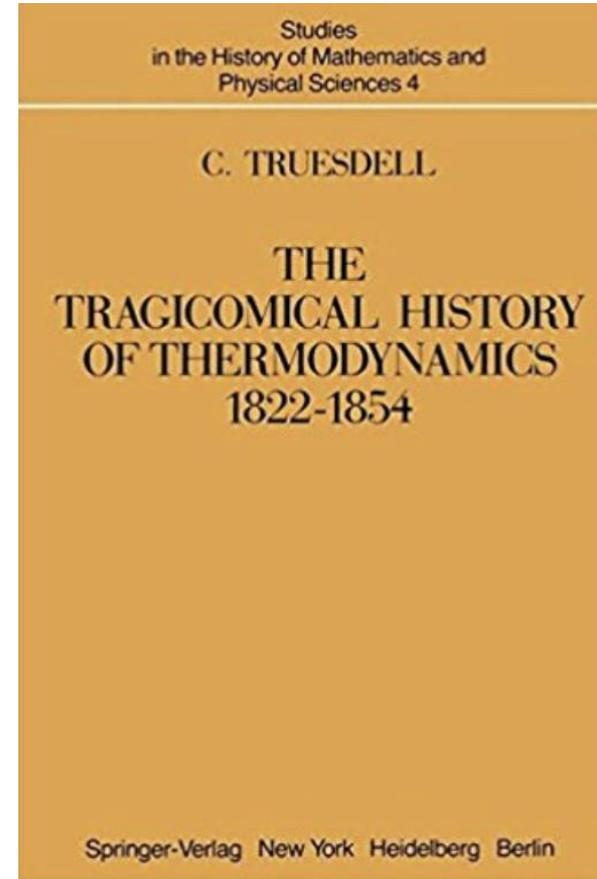
$$\begin{aligned}\partial_t \mathbb{H}(x) &= -\delta_x \mathbb{H}(x)^\top R(x) \delta_x \mathbb{H}(x) + y^\top u - u^\top S^\top u \\ &= - \begin{bmatrix} H & \partial_t A \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 0 & *_\kappa + d *_{\nu\tau} d \end{bmatrix} \begin{bmatrix} H \\ \partial_t A \end{bmatrix} + (\chi^\top \partial_t A + \chi *_\rho \chi^\top I_s) I_s - I_s \chi *_\rho \chi^\top I_s \\ &= \partial_t A \chi I_s - \partial_t A (*_\kappa + d *_{\nu\tau} d) \partial_t A\end{aligned}$$

The magnetic energy **decreases only due** to the eddy and coupling current losses and **is not directly** influenced by the Joule losses after a quench.

Modelling of Irreversible Thermodynamics

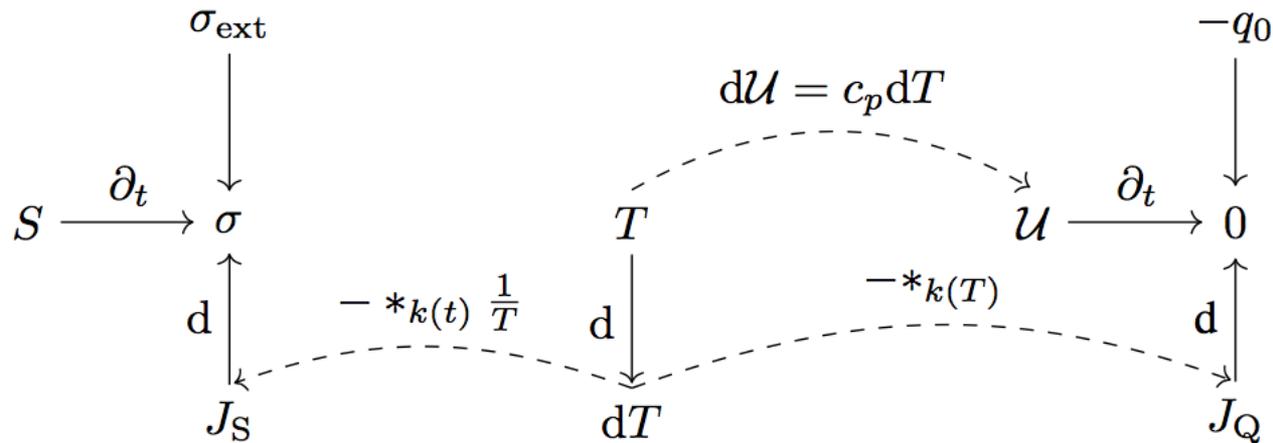
Irreversible thermodynamics modelling considers

1. storage of the thermal energy
2. heat conduction
3. heat generation in the EM domain



Irreversible Thermodynamics – Topological Diagram

Symbol	Form order	Unit	Description
T	$\Lambda^0(\Omega)$	K	Temperature
J_S	$\Lambda^2(\Omega)$	W/K	Entropy flux
J_Q	$\Lambda^2(\Omega)$	W	Heat flux
\mathcal{U}	$\Lambda^3(\Omega)$	J	Internal energy density
S	$\Lambda^3(\Omega)$	J/K	Entropy density
σ	$\Lambda^3(\Omega)$	W/K	Irreversible entropy flow
f_s	$\Lambda^3(\Omega)$	W/K	Reversible entropy flow



D. Eberard and B. Maschke, "Port hamiltonian systems extended to irreversible systems: The example of heat conduction," in *IFAC Nonlinear Control Congress*, Elsevier, 2004.

Irreversible Thermodynamics – Bond Graph

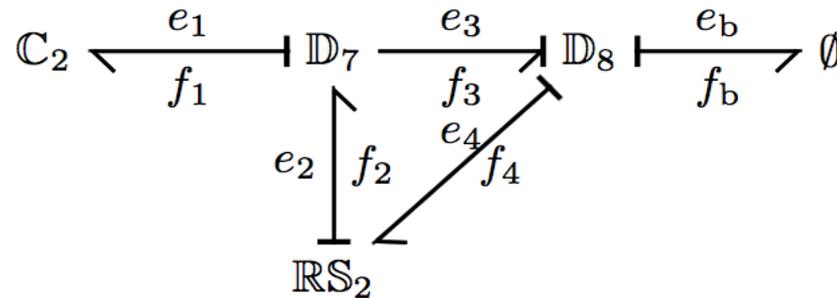
Jaumann's entropy balance forms a 0-junction

$$\mathcal{D}_7 = \{(f_{13}, f_{14}, f_{15}, e_{13}, e_{14}, e_{15}) \in \mathcal{F}_s \times \mathcal{E}_s : f_{13} + f_{14} + f_{15} = 0, e_{13} = e_{14} = e_{15}\}$$

with $((f_{13}, f_{14}, f_{15}, e_{13}, e_{14}, e_{15}) = (\partial_t s, -\sigma_s, dJ_s, T, T, T).$

Stokes-Dirac structure (symplectic gurator) accounts for the interaction through a boundary

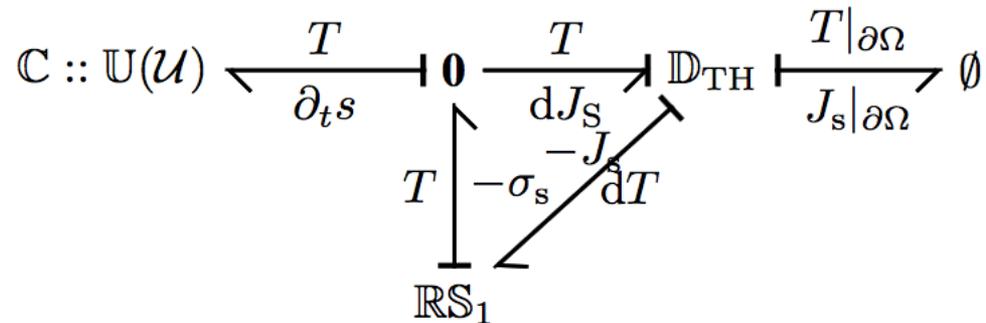
$$\mathcal{D}_2 = \left\{ \begin{array}{l} (f_3, f_4, f_b, e_3, e_4, e_b \in \mathcal{F} \times \mathcal{E}) \\ \begin{bmatrix} f_3 \\ f_4 \end{bmatrix} = \begin{bmatrix} 0 & -d \\ d & 0 \end{bmatrix} \begin{bmatrix} e_3 \\ e_4 \end{bmatrix} \\ \begin{bmatrix} f_b \\ e_b \end{bmatrix} = \begin{bmatrix} e_{th}|\partial\Omega \\ f_{th}|\partial\Omega \end{bmatrix} \end{array} \right\} \quad (dJ_s, dT, J_s|_{\partial\Omega}, T, -J_s, T|_{\partial\Omega} \in \mathcal{F} \times \mathcal{E})$$



Irreversible Thermodynamics – Bond Graph

Heat conduction (the Fourier’s law) in the thermal domain is represented as an RS element

$$\mathcal{I}_3 = \left\{ \begin{array}{l} (f_2, f_4, e_2, e_4 \in \mathcal{F} \times \mathcal{E}) \\ \begin{bmatrix} e_4 \\ f_2 \end{bmatrix} = \begin{bmatrix} 0 & e_R \\ -e_R & 0 \end{bmatrix} \begin{bmatrix} f_4 \\ e_2 \end{bmatrix} \end{array} \right\} \quad e_R = \frac{*\lambda}{T^2} dT \wedge$$



H. Ramirez, B. Maschke, and D. Sbarbaro, “Irreversible port-hamiltonian systems: A general formulation of irreversible processes with application to the cstr,” *Chemical Engineering Science*, vol. 89, 2020, pp. 1–12.



Irreversible Port-Hamiltonian Model

Hamiltonian function is obtained with the internal energy

$$\mathbb{U} = \int_{\Omega} \mathcal{U}.$$

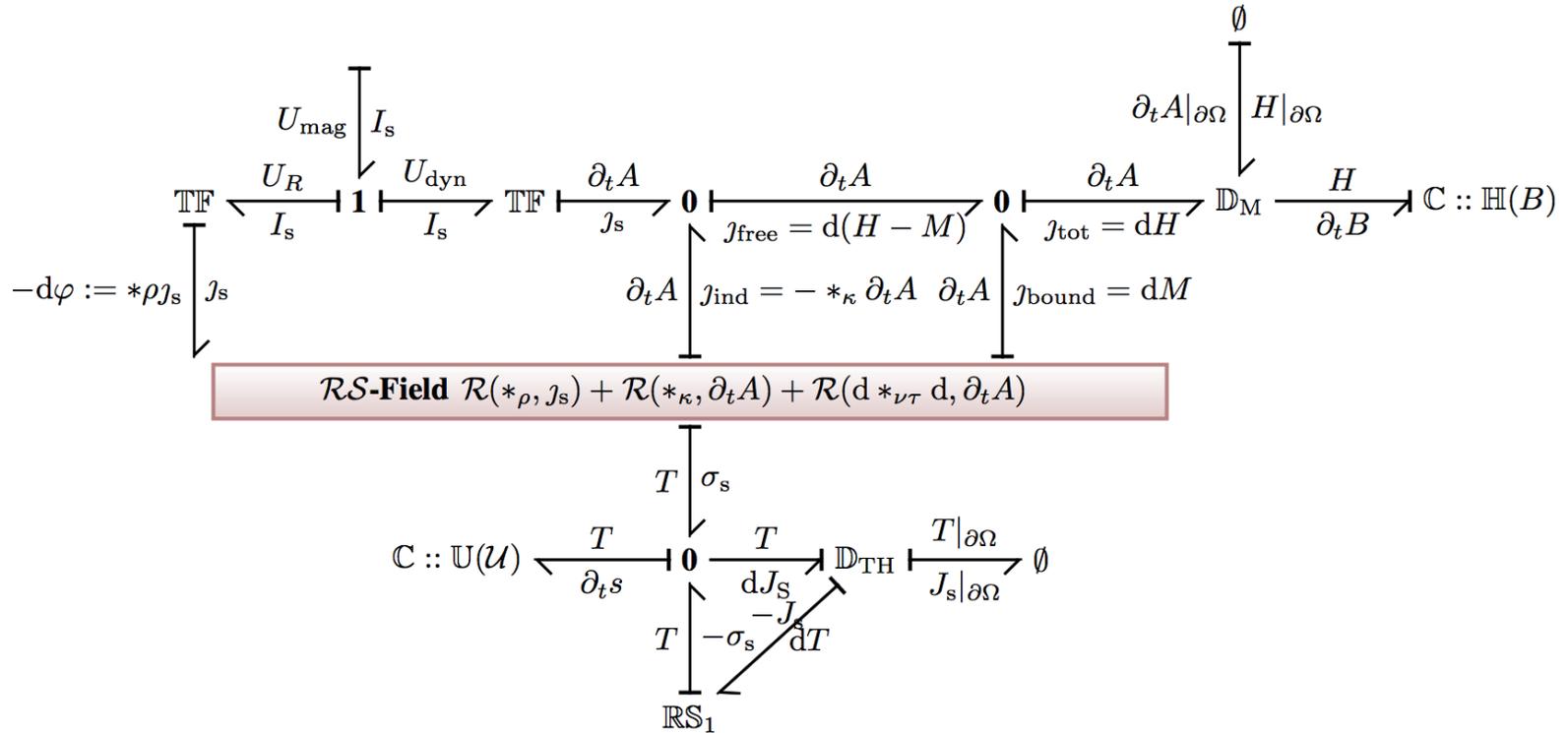
Dynamics of the thermal model can be represented as

$$\begin{aligned}\partial_t x &= (de_R + e_R d)\delta_x \mathcal{U} + u \\ y &= \delta_x \mathcal{U},\end{aligned}$$

The variation of the internal energy reads

$$\begin{aligned}\partial_t U &= \delta_x U \partial_t x = T(dR \wedge T - R \wedge dT) + Tu \\ &= T(dR \wedge T - R \wedge dT) + *_{\rho} j_s \wedge j_s + \partial_t A \wedge *_{\kappa} \partial_t A - \partial_t A \wedge d *_{\nu\tau} d\partial_t A\end{aligned}$$

Magneto-Thermal Model of an SC Magnet



Irreversible entropy creation

$$\sigma_{\text{em}} = \frac{1}{T} (*\rho j_s \wedge j_s + \partial_t A \wedge * \kappa \partial_t A - \partial_t A \wedge d * \nu_\tau d \partial_t A)$$

$$\mathcal{R} = \left\{ \begin{array}{l} (f_1, e_1, f_2, e_2, f_3, e_3, f_4, e_4) \in \Lambda^3(\Omega) \times \Lambda^0(\Omega) \\ \times \Lambda^2(\Omega) \times \Lambda^1(\Omega) \times \Lambda^2(\Omega) \times \Lambda^1(\Omega) \times \Lambda^2(\Omega) \times \Lambda^1(\Omega) \\ e_2 = *\rho f_2, f_3 = *\kappa e_3, f_4 = d * \nu_\tau d e_4, \sum_{i=1}^4 e_i \wedge f_i = 0 \end{array} \right\}$$

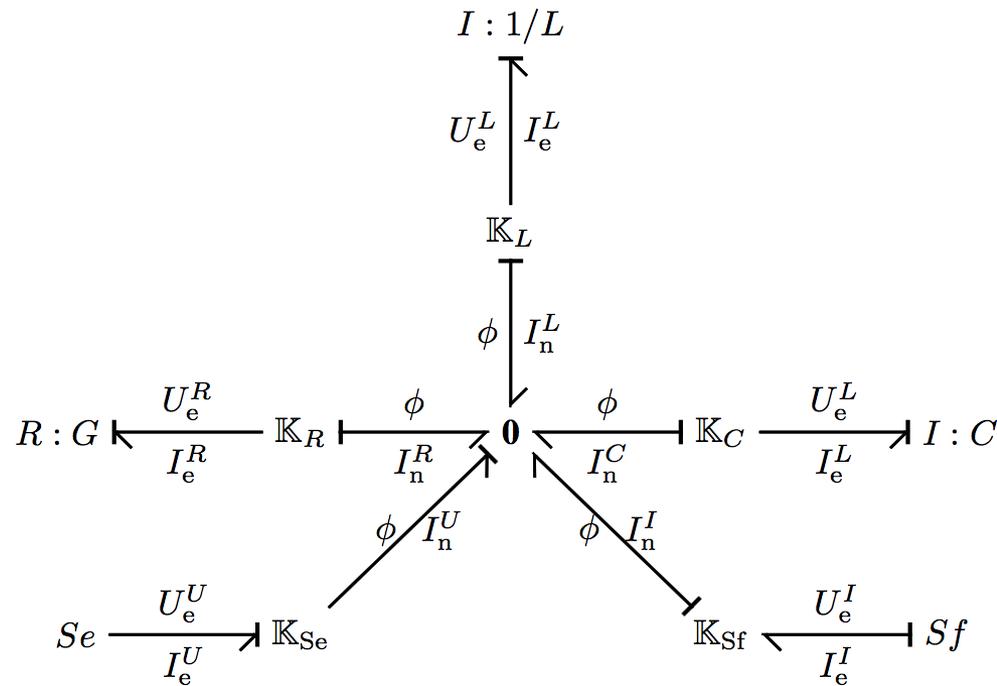
Port-based Model of an SC Circuit

Starting with the Modified Nodal Analysis

$$A_C \partial_t Q + A_L I_L + A_R g_R(A_R^\top \phi, t) + A_U I_U + A_I I(t) = 0,$$

$$L \partial_t I_L - A_L^\top \phi = 0$$

$$A_S^\top \phi = U_S$$



Summary

1. Port-based modelling allows for a generic representation of multi-physical systems (both, SC magnets, and circuits)
2. Variation of the magnetic and thermal energy is studied
3. Computational causality can be derived from a bond-graph model

- use of the port-based approach for *co-simulation*.



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