

# **Analogue electronics for BPMs at GSI - Performance and limitations**

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Acknowledgments to  
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Hansi Rödl & Christian Schmidt (GSI)

# Our Team – GSI beam diagnostics RF lab



**You are invited to contact us, if there are any questions!**

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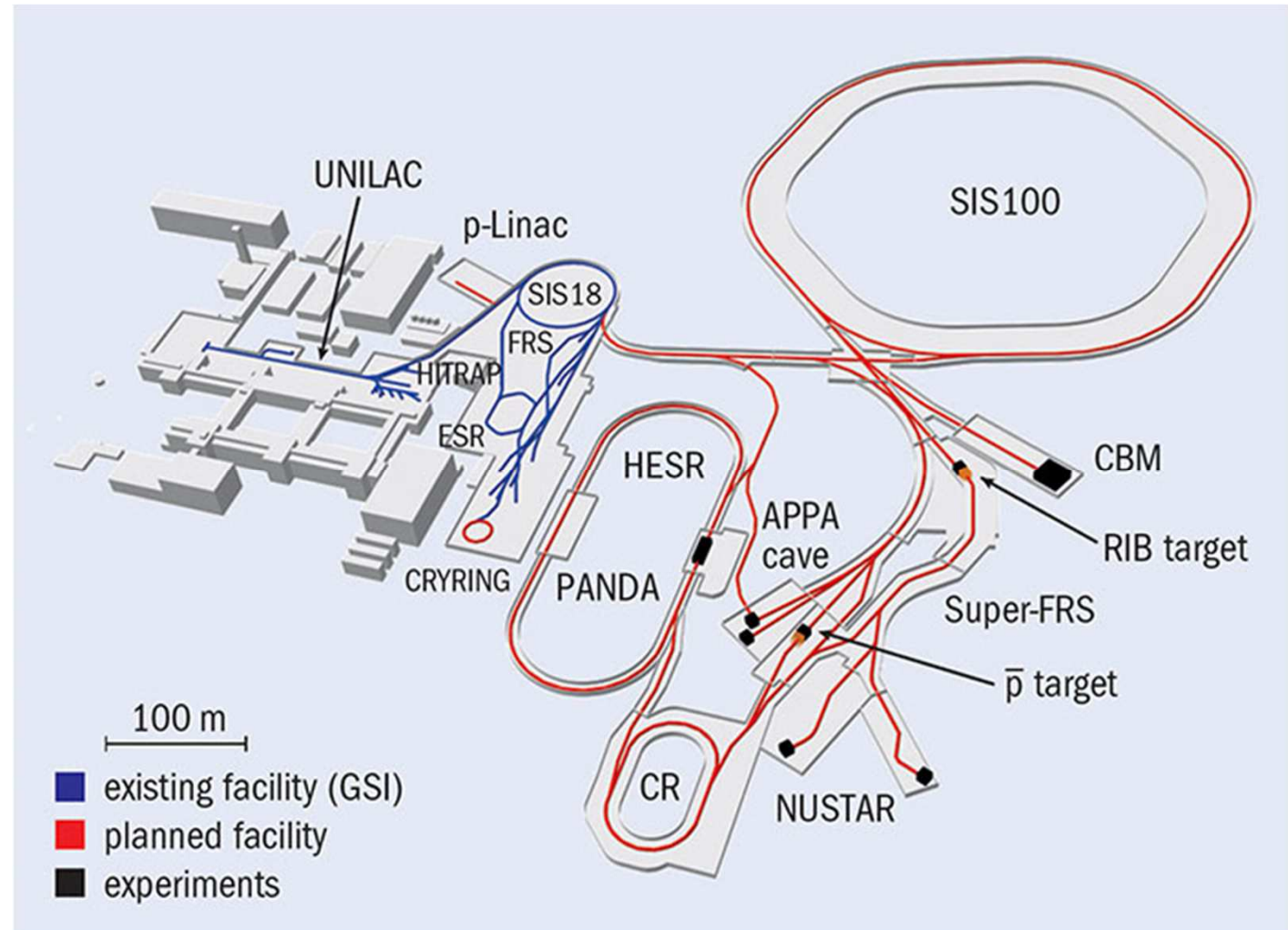


# Overview



## BPMs and amplifiers for Fair

- Cryring 9
- SIS100 84
- HESR 76
- HEBT 39
- SIS18 12
- ESR 12
- total 223
- p-Linac 15



## 3 different concepts for BPM pickup amplifiers

- Low energy synchrotron

**Machine:** Crying

**Bandwidth:** 10 kHz to 40 MHz

**input impedance:** 1 M $\Omega$

**Output impedance:** 50  $\Omega$

**Gain:** 40 dB / 60 dB +/- 0.05 dB

**Equivalent input noise:** 2.3 nV /  $\sqrt{\text{Hz}}$

- High energy synchrotron

**Machines:** SIS100, SIS18, HESR, ESR, High energy beam lines

**Bandwidth:** 40 kHz to 55 MHz

**Input / Output impedance:** 50  $\Omega$

**Gain:** -50 dB to +60 dB

**Equivalent input noise:** 1.6 nV /  $\sqrt{\text{Hz}}$

- Proton Linac

**Machine:** p-Linac

**Bandwidth:** 325 MHz to 3.25 GHz

**Input / Output impedance:** 50  $\Omega$

**Gain:** 0 dB to +40 dB

**Equivalent input noise:** 1.4 nV /  $\sqrt{\text{Hz}}$



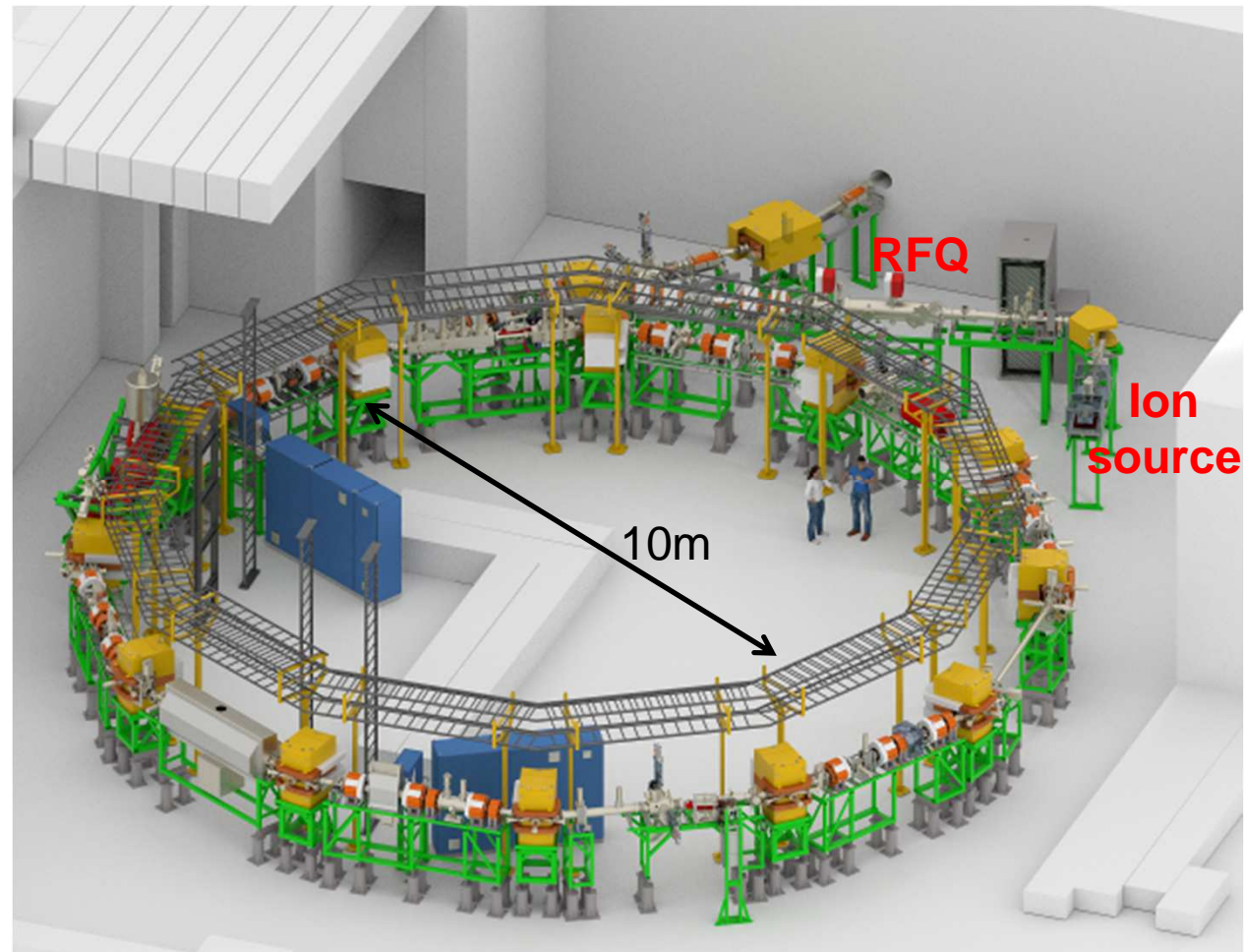
# Low energy synchrotron – Cryamp

## Cryring specs

- Dedicated 300 keV/u injector linac
- Injection possible with source potential 40 keV
- Beam current in the  $\mu\text{A}$  range
- RF frequency 100 kHz – 2 MHz

## Cryamp application

- Beam position measurement
- Integral bunch signals
- Relative beam current measurement
- Bunch shape monitoring



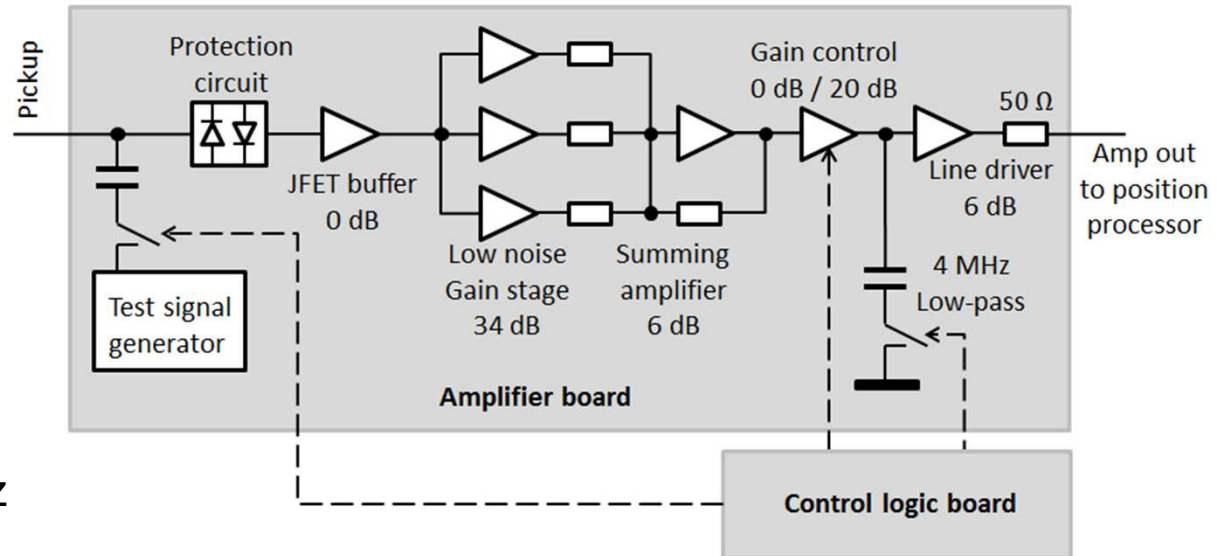
# Low energy synchrotron – Cryamp

## Design requirements

- Low noise architecture
- Fixed gain modes
- Gain accuracy through low tolerance components

## Cryamp data

- Bandwidth: 10 kHz to 40 MHz
- Selectable low pass: 4 MHz
- Input impedance: 1 M $\Omega$
- Output impedance: 50  $\Omega$
- Output level into 50  $\Omega$ : 6 dBV
- Equivalent input noise: 2.3 nV /  $\sqrt{\text{Hz}}$
- Gain: 40 dB / 60 dB +/- 0.05 dB
- Pickup electrode biasing up to 200 Vdc (variant for cooler bpm)
- Internal test signal generator 125 kHz square wave

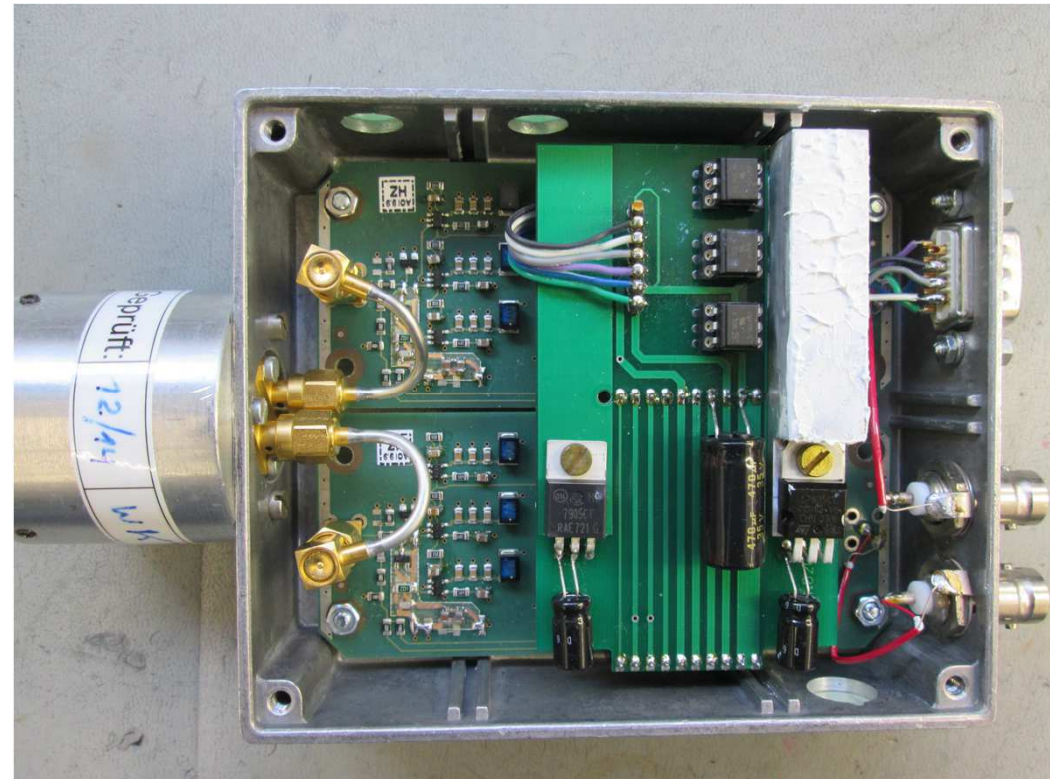


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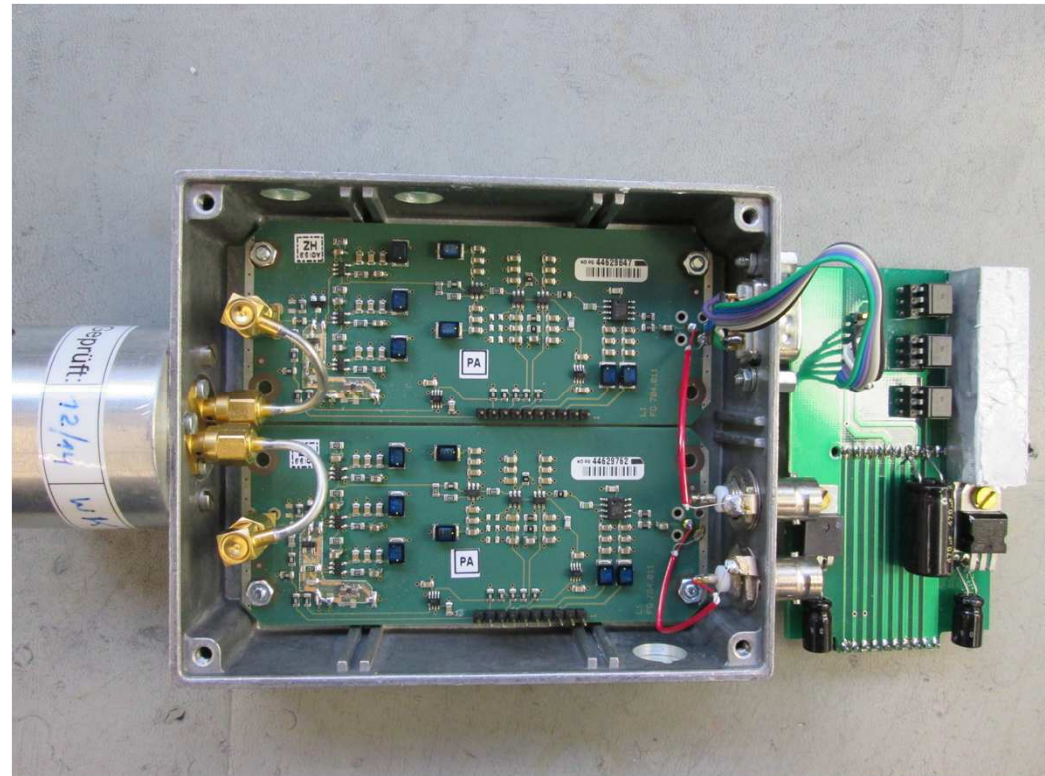


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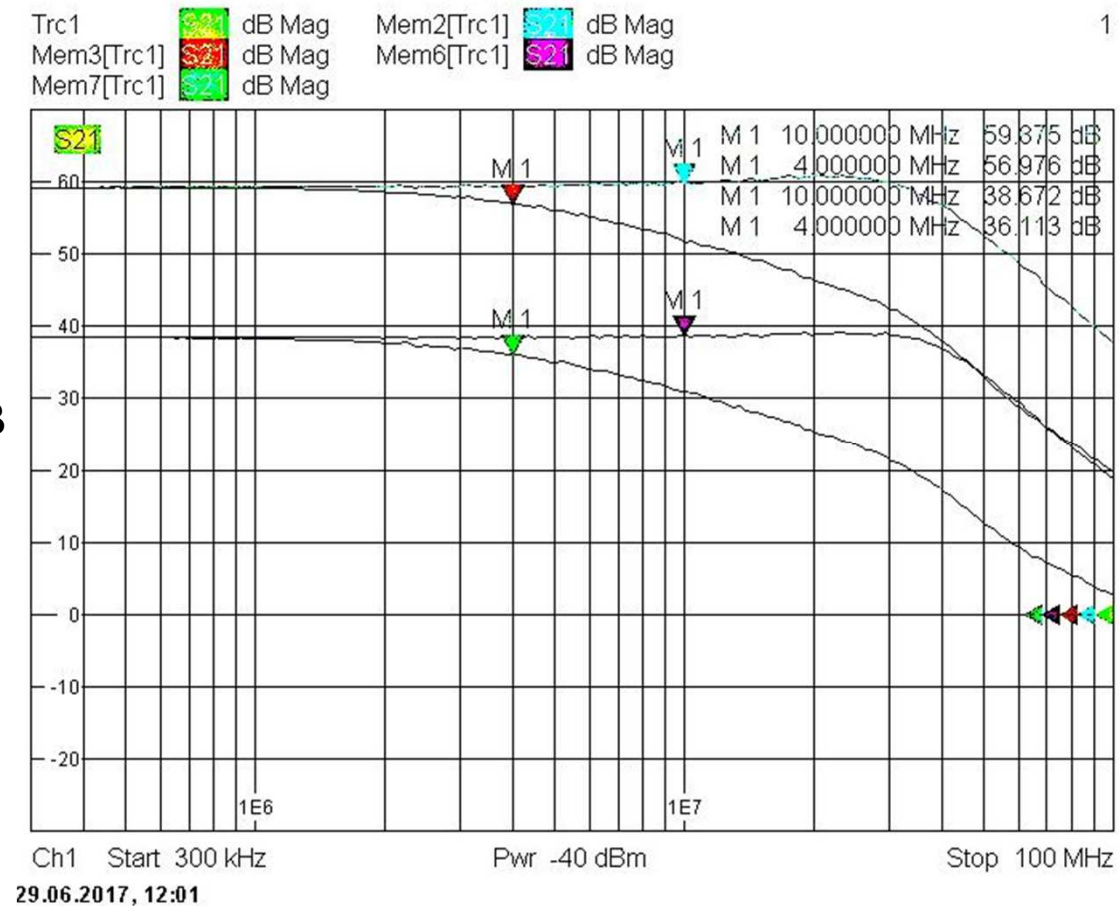






## Frequency Response

- Forward transmission:  $|S_{21}|$
- Measured with a R&S ZVB4 Network analyzer
- Shows operation in the 40 dB and 60 dB mode with internal low pass on and off



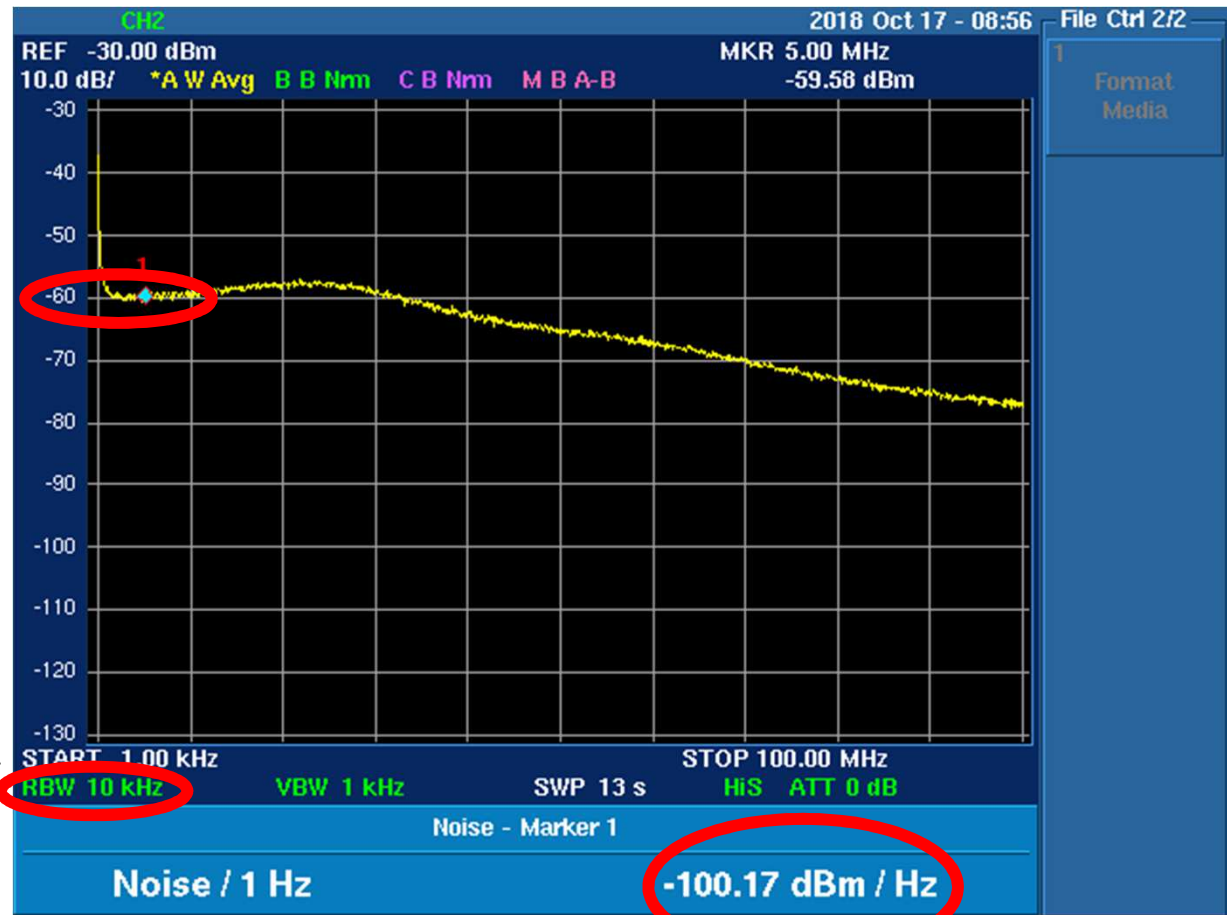
## Noise performance of the Cryamp at 60 dB

- Measured with an ADVANTEST U3841 FFT spectrum analyzer
- Resolution bandwidth must be taken into account to derive the spectral noise voltage density

$$-60 \frac{\text{dBm}}{10\text{kHz}} = -100 \frac{\text{dBm}}{\text{Hz}}$$

$$-100 \frac{\text{dBm}}{\text{Hz}} - 60\text{dB} = -160 \frac{\text{dBm}}{\text{Hz}}$$

$$2.3 \frac{\text{nV}}{\sqrt{\text{Hz}}} = \sqrt{50 \Omega \frac{1 \text{ mW}}{\text{Hz}} 10^{-16}}$$



# Low energy synchrotron – Cryamp



## Bunch signals in the time domain at 60 dB

- Measured with a Keysight DSOX2024A
- BPM sum signal as replacement for integrating current transformer for low currents ~10 nA
- BPM and ICT sense the bunches in this example



BPM  
sum signal 60 dB

Integrating current  
transformer 80 dB



# Low energy synchrotron – Cryamp



## Bunch signals in the time domain at 60 dB

- Measured with a Keysight DSOX2024A
- BPM sum signal as replacement for integrating current transformer for low currents ~10 nA
- Only the BPM detects the bunch in this example



BPM  
sum signal 60 dB

Integrating current  
transformer 80 dB

# High energy synchrotron – Amplifier 110

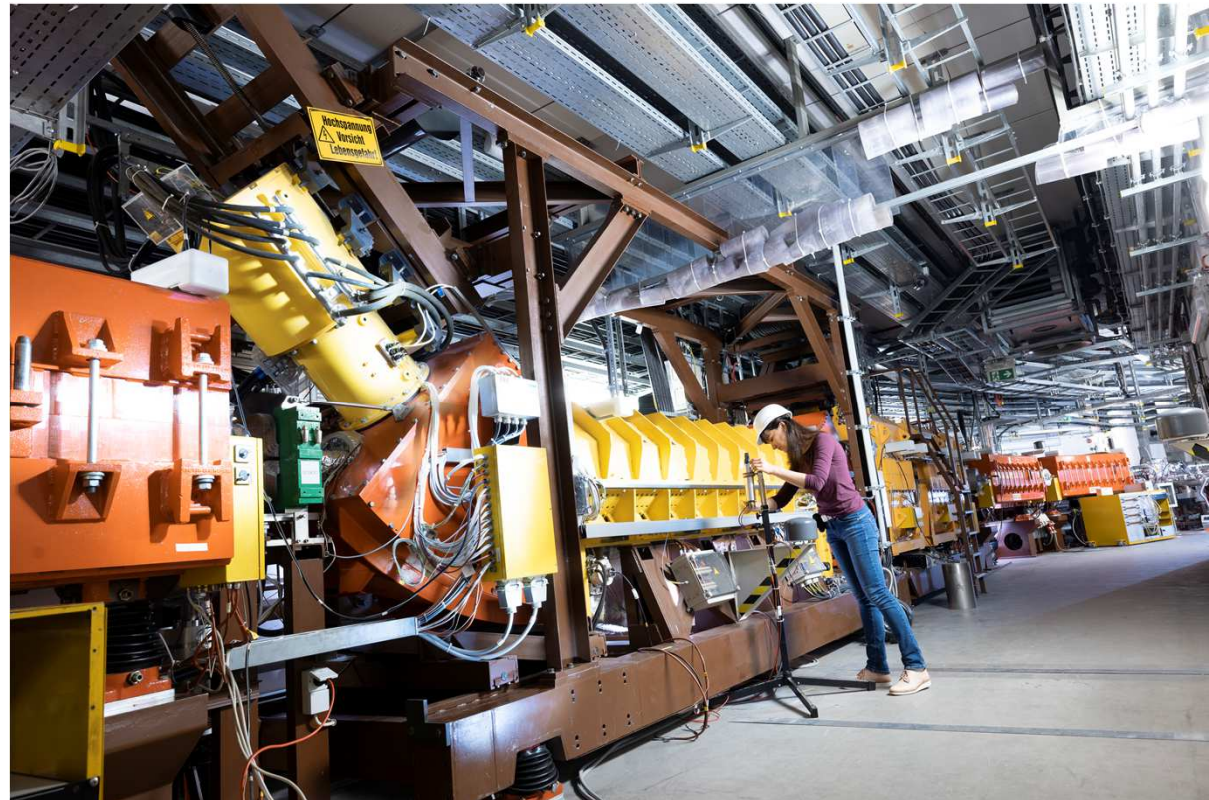


## SIS 18 specs

- 216 m circumference
- 18 Tm maximum rigidity
- 2 GeV/u maximum energy, depending on the element
- RF frequency  
800 kHz – 5.6 MHz

## Amplifier 110 application

- Beam position measurement
- Closed orbit feedback
- Integral bunch signals
- Pickup tap with matching transformer



# High energy synchrotron – Amplifier 110

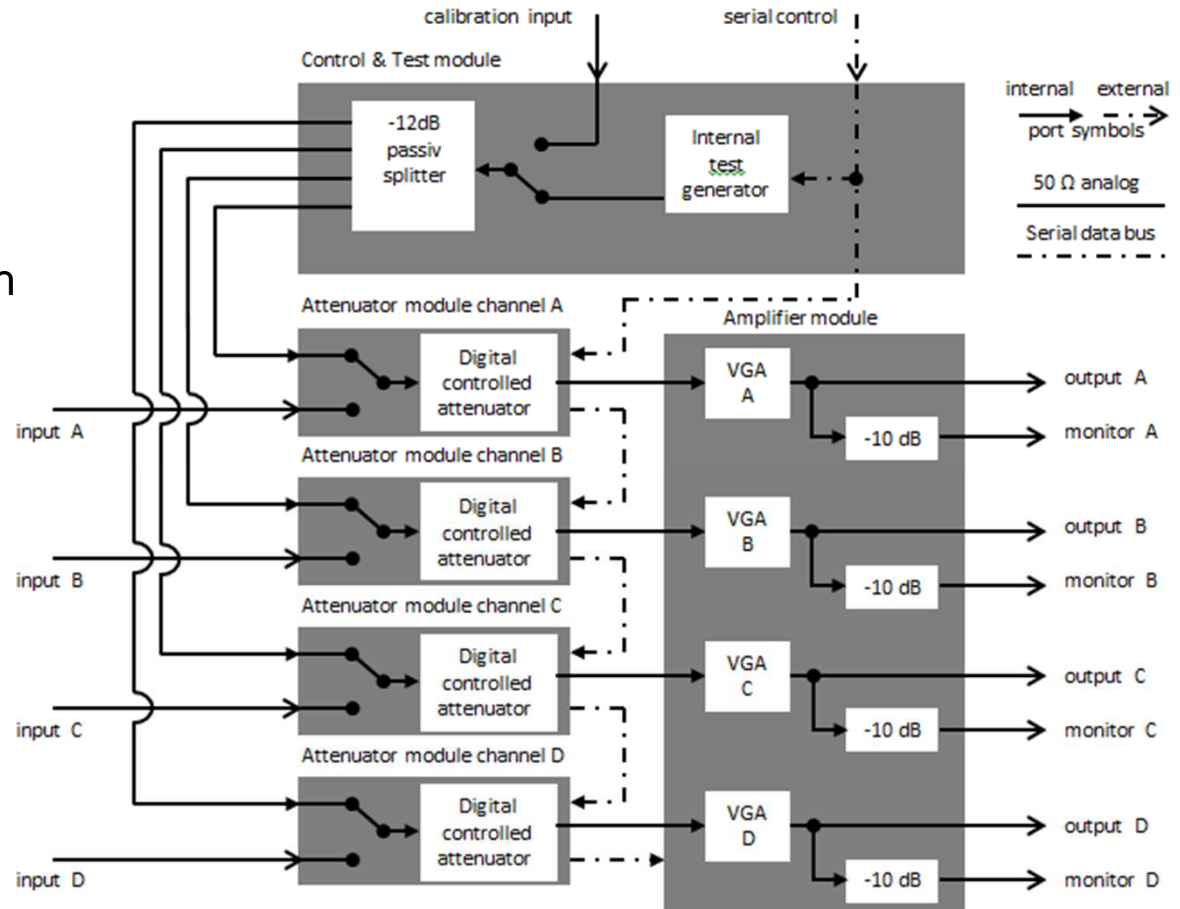


## Design requirements

- 110 dB Dynamic range
- Gain fine tuning through VGAs
- Automatic gain matching bench

## Amplifier 110 Data

- Bandwidth: 40 kHz to 55 MHz
- Selectable low pass: 7 MHz
- Input impedance: 50  $\Omega$
- Output impedance: 50  $\Omega$
- Max output level: 6 dBV
- Equivalent input noise:  $1.6 \text{ nV} / \sqrt{\text{Hz}}$
- Gain: -50 dB to 60 dB  $\pm 0.01 \text{ dB}$
- Internal test signal generator



# High energy synchrotron – Amplifier 110



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- Gain: -50 dB to 60 dB +/- 0.01 dB
- Internal test signal generator



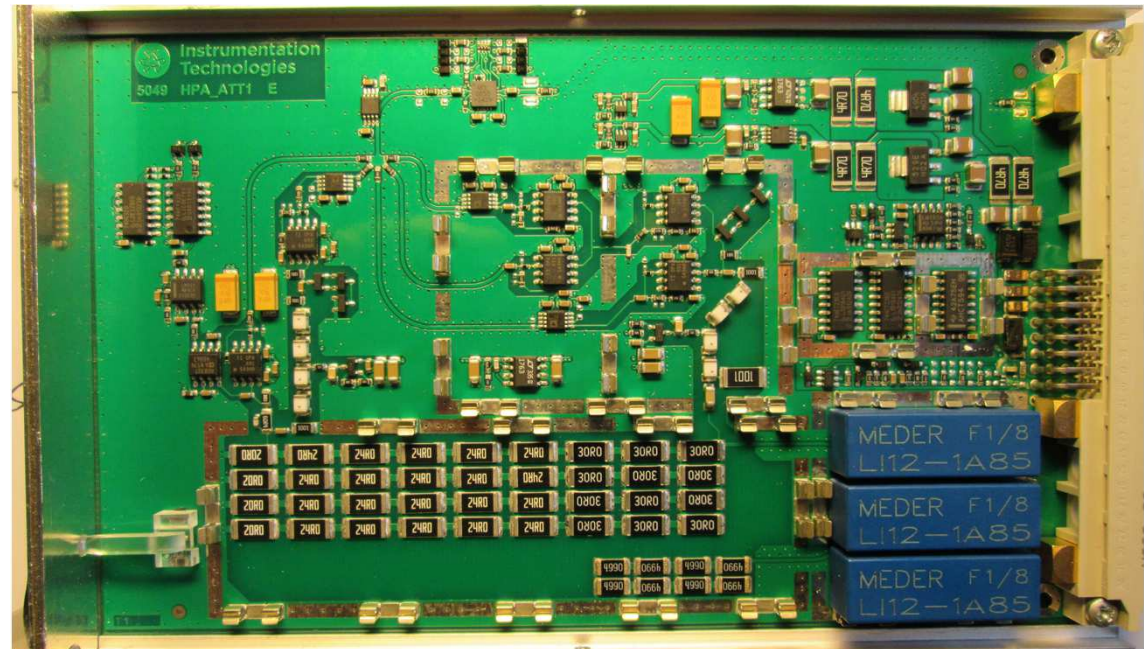


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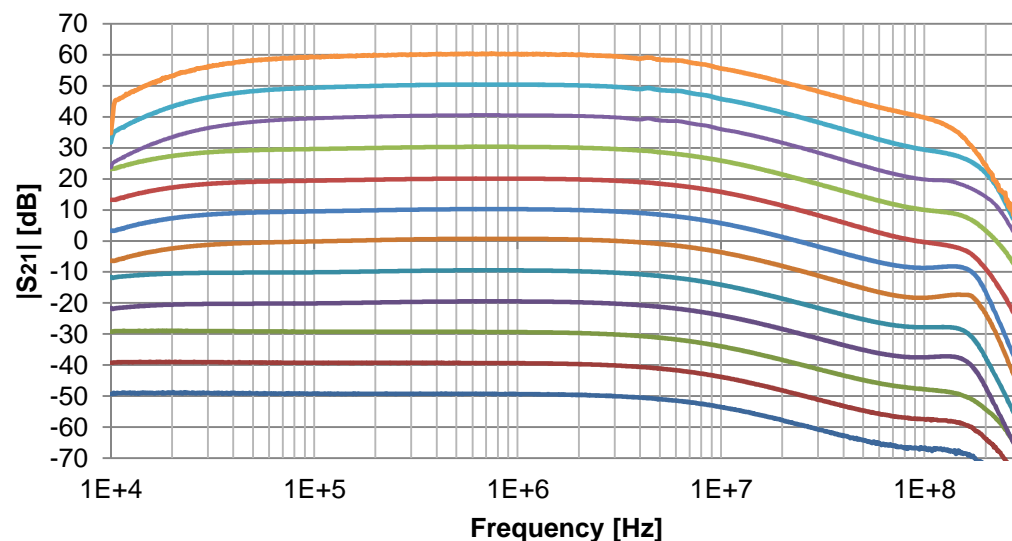
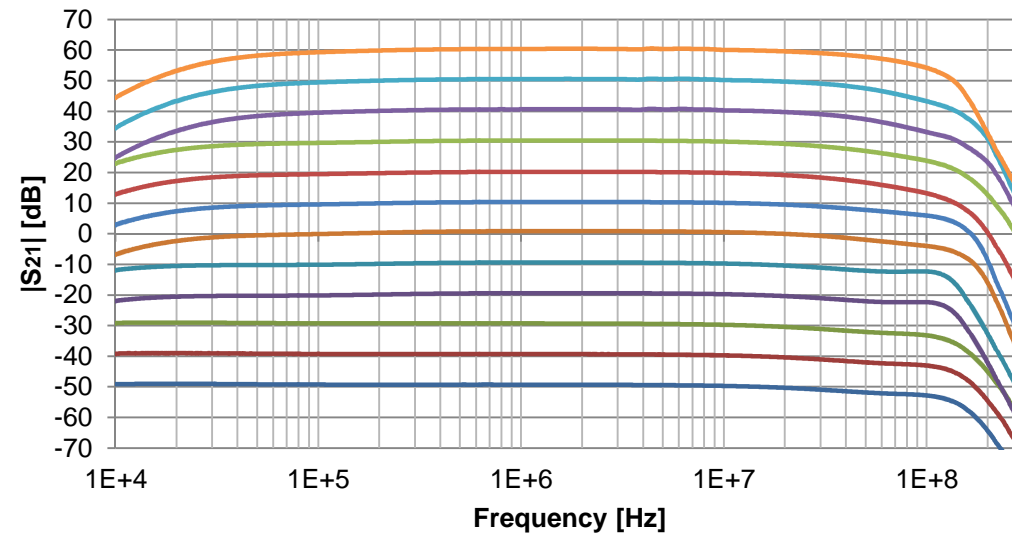
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## Frequency Response

- Forward transmission:  $|S_{21}|$
- Measured with an Agilent E5071C Network analyzer
- Shows operation in all modes from -50 dB to 60 dB

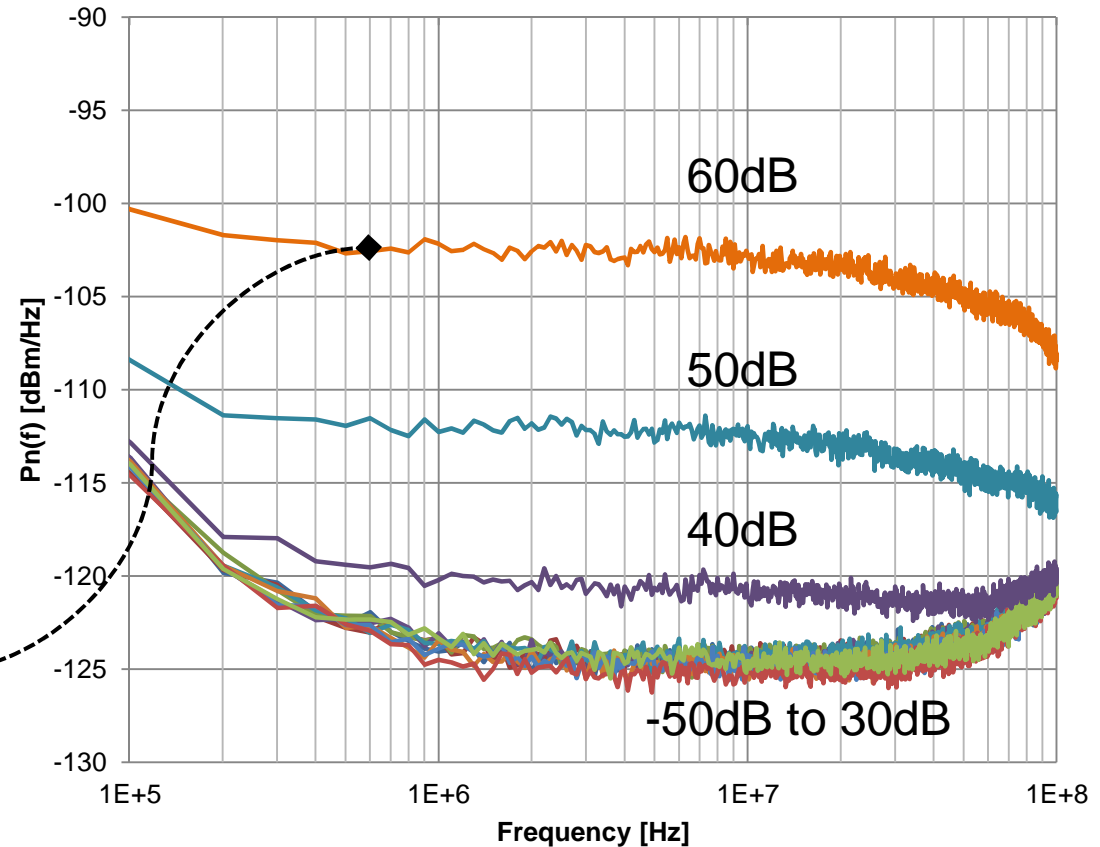


## Noise performance of the Amplifier 110

- Measured with an Agilent N9020A spectrum analyzer
- Shows output noise power density in all modes from -50 dB to 60 dB

$$-103 \frac{\text{dBm}}{\text{Hz}} - 60\text{dB} = -163 \frac{\text{dBm}}{\text{Hz}}$$

$$1.6 \frac{\text{nV}}{\sqrt{\text{Hz}}} = \sqrt{50 \Omega \frac{1 \text{ mW}}{\text{Hz}} 10^{-16.3}}$$

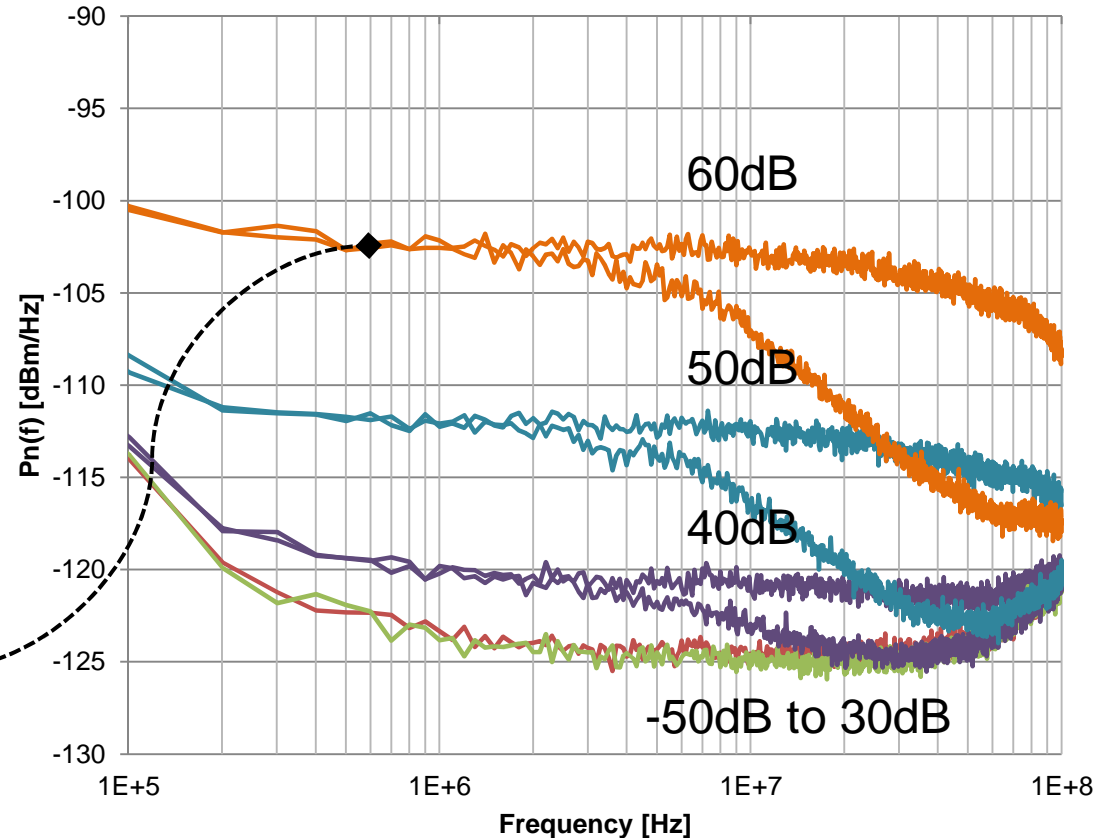


## Noise performance of the Amplifier 110

- Measured with an Agilent N9020A spectrum analyzer
- Shows output noise power density with internal low pass filter

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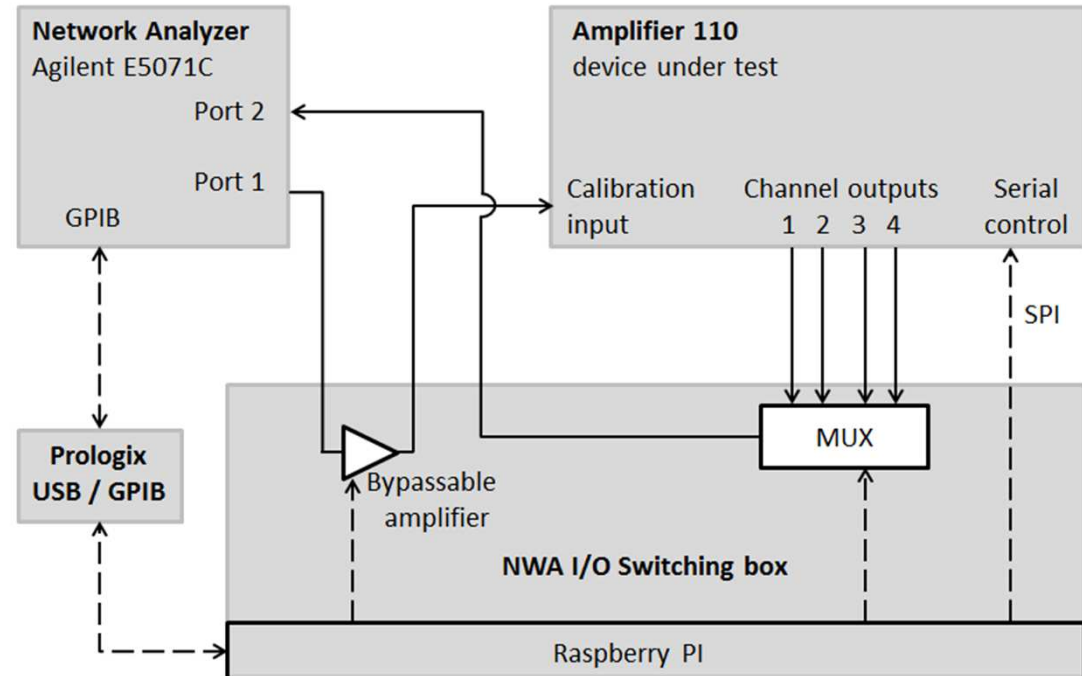
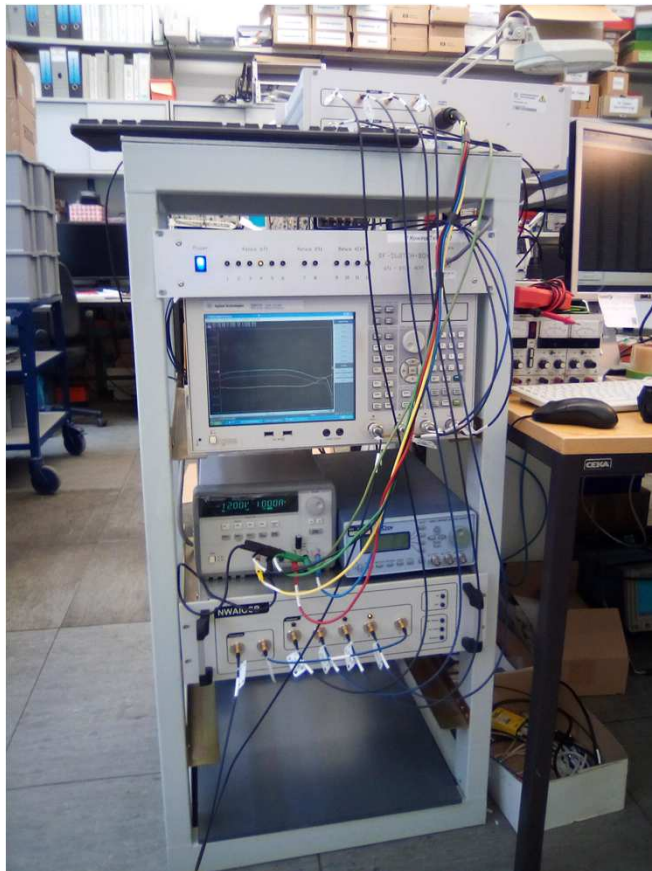




# High energy synchrotron – Amplifier 110



## Automatic gain matching bench for the Amplifier 110



## Automatic gain matching bench for the Amplifier 110

- Basic principle for finding optimal VGA settings based on S-parameter comparison
- In the case of perfect matching the integral equates zero
- Numeric approximation of the integral through a sum

$$\begin{aligned} I &= \int_{f_1}^{f_2} \left| |S_{21Ref}(f)|_{dB} - |S_{21Measure}(f)|_{dB} \right| df \\ &= \int_{f_1}^{f_2} |\Delta_{|S_{21}|dB}(f)| df \end{aligned}$$

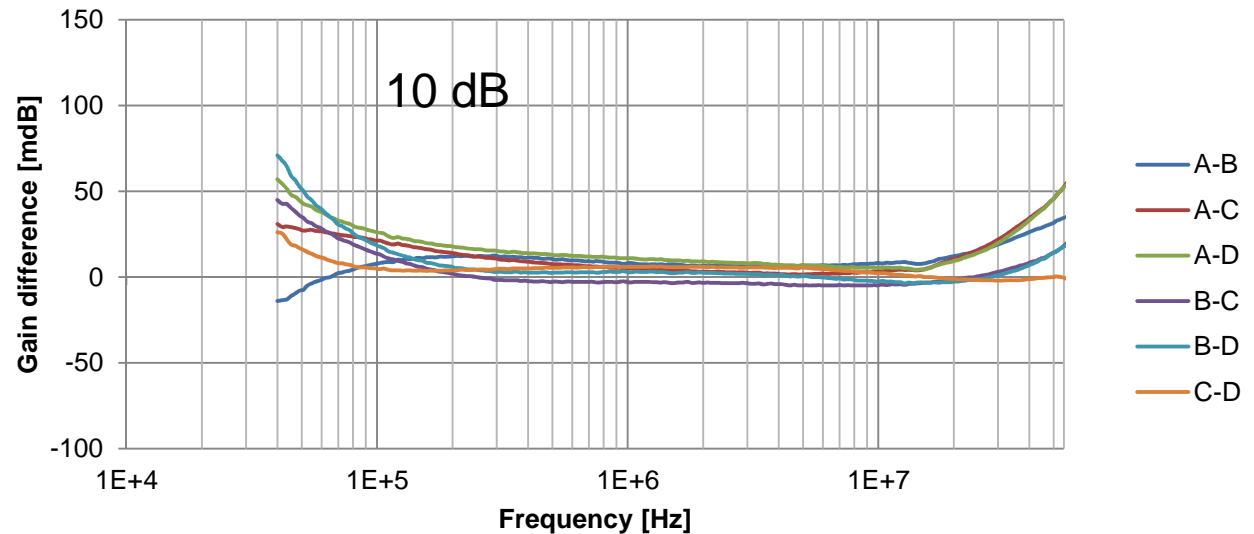
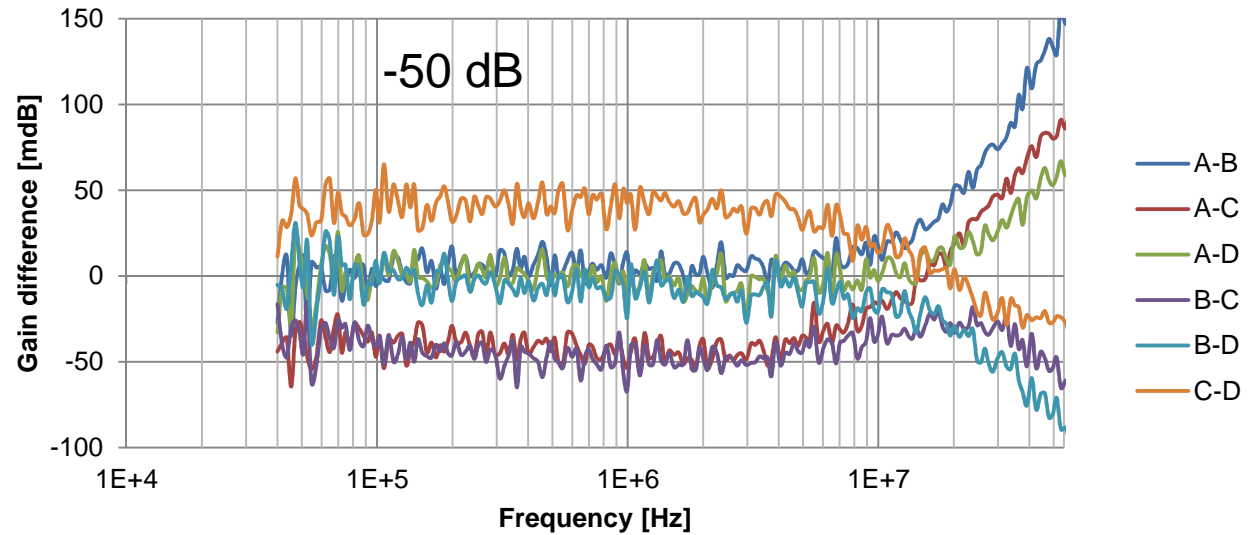
$$\begin{aligned} I &\approx \sum_{n=0}^{N-1} \left| |S_{21Ref}(n)|_{dB} - |S_{21Measure}(n)|_{dB} \right| \\ &= \sum_{n=0}^{N-1} |\Delta_{|S_{21}|dB}(n)| \end{aligned}$$

# High energy synchrotron – Amplifier 110



## Gain differences over frequency between channels, after calibration

- Mismatch is in the  $10^{-3} \text{ dB} = 1 \text{ mdB}$  range
- Noise is a challenge for the calibration routine

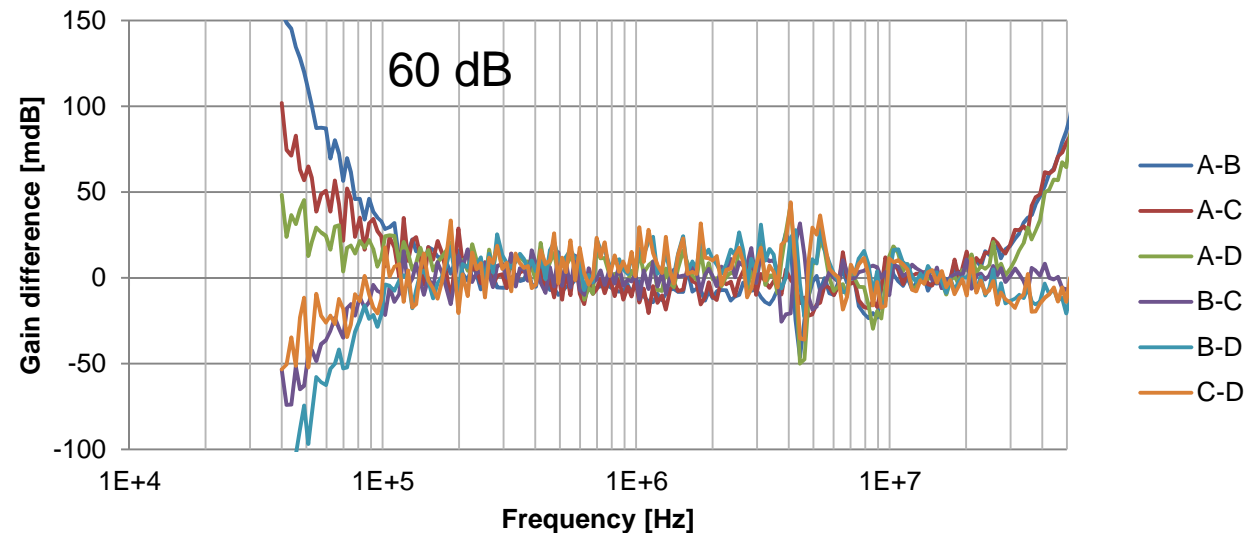
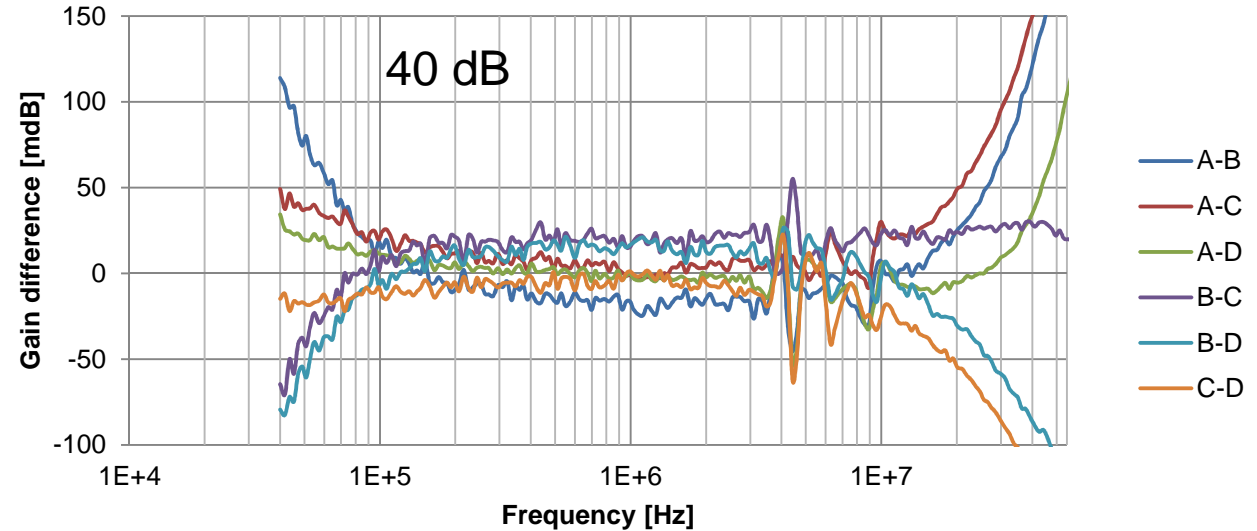


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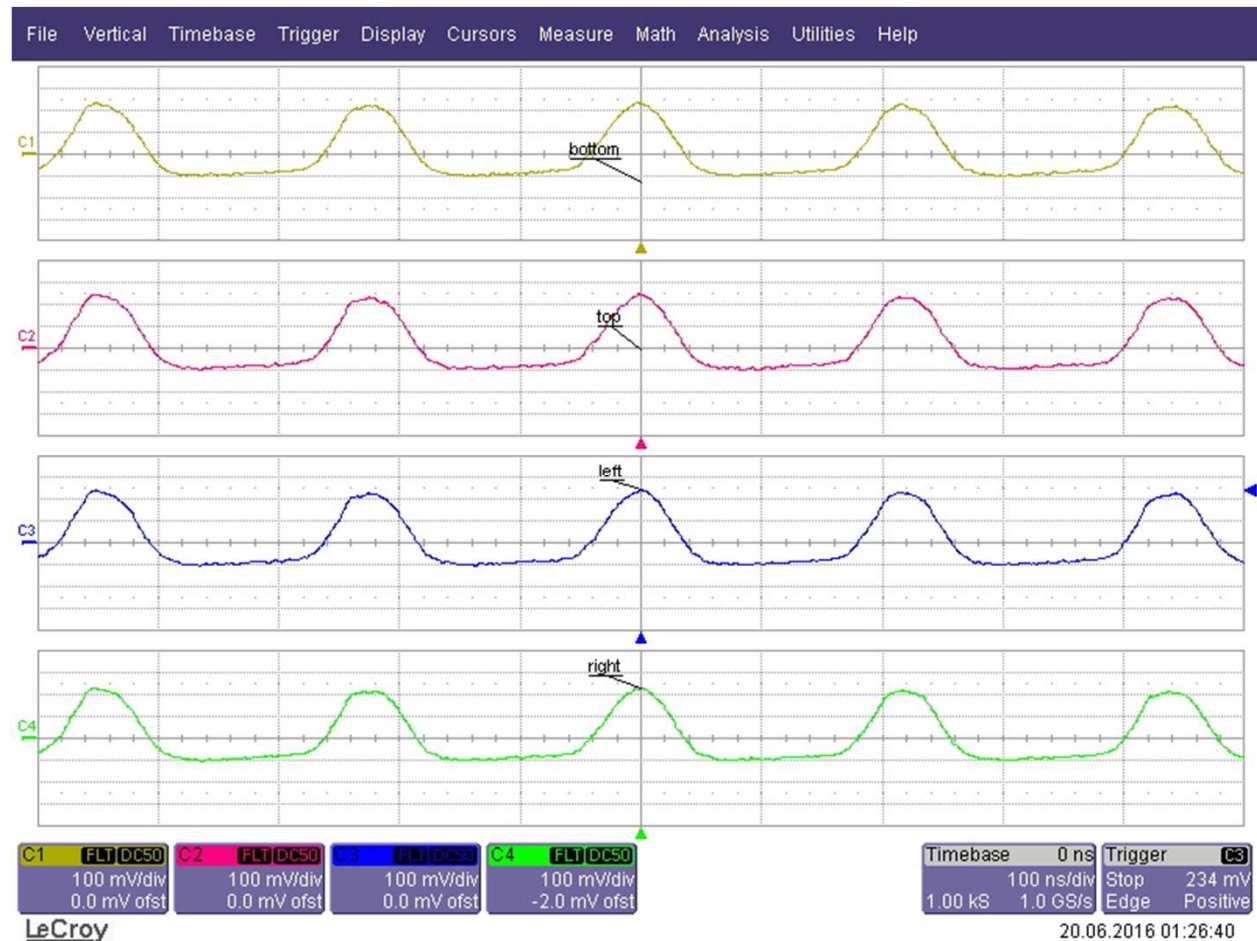


# High energy synchrotron – Amplifier 110



## SIS 18 Bunch signals In the time domain

- Measured with a LeCroy Wave Runner 6200A DSO
- 2.2 mA beam current
- Amplifier 110 set to 30 dB amplification



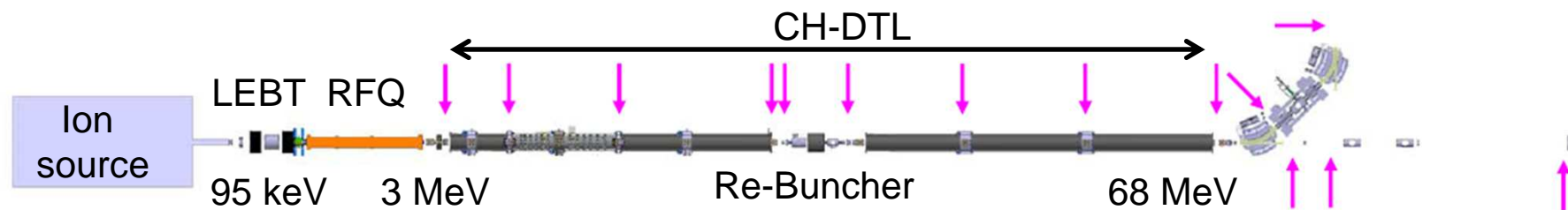
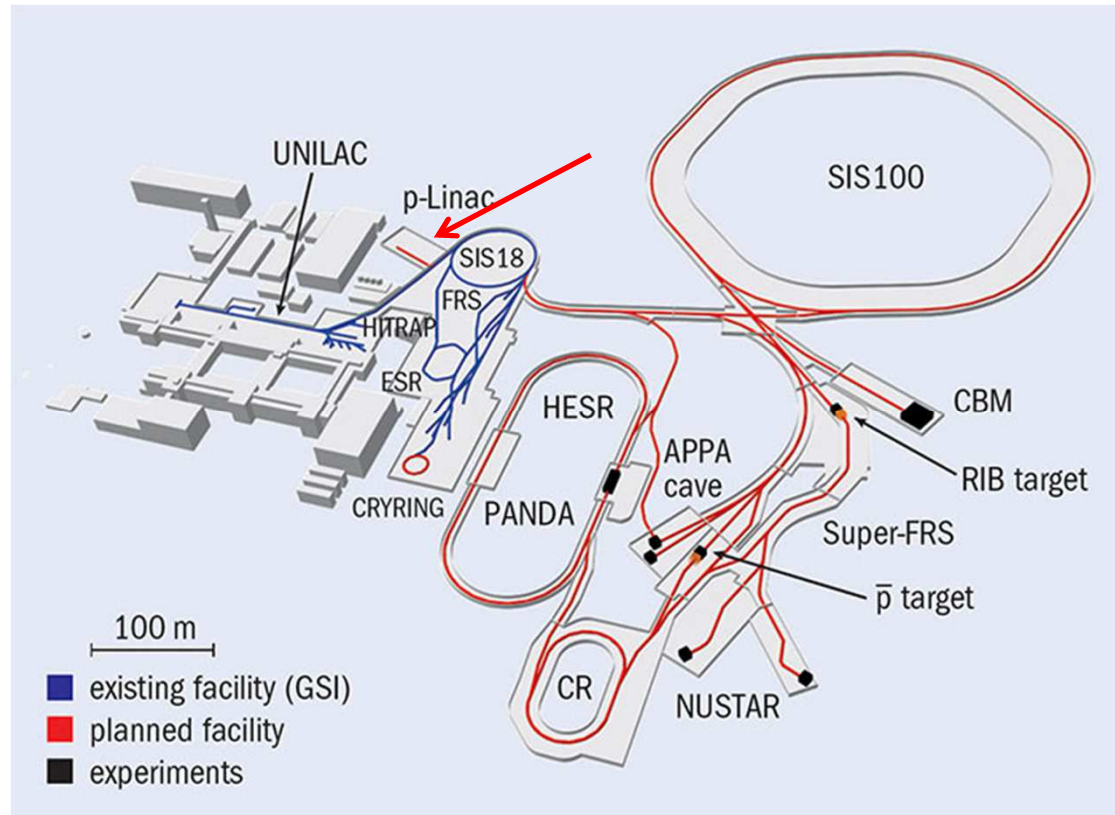
# Proton Linac – p-Linac Amplifier

## p-Linac specs

- 325 MHz RF frequency
- 68 MeV/u energy
- 70 mA nominal beam current
- 30  $\mu$ s makropulse time
- 250 ps bunch length

## p-Linac Amplifier application

- Beam position measurement
- Time of flight measurement
- differentiated bunch signals

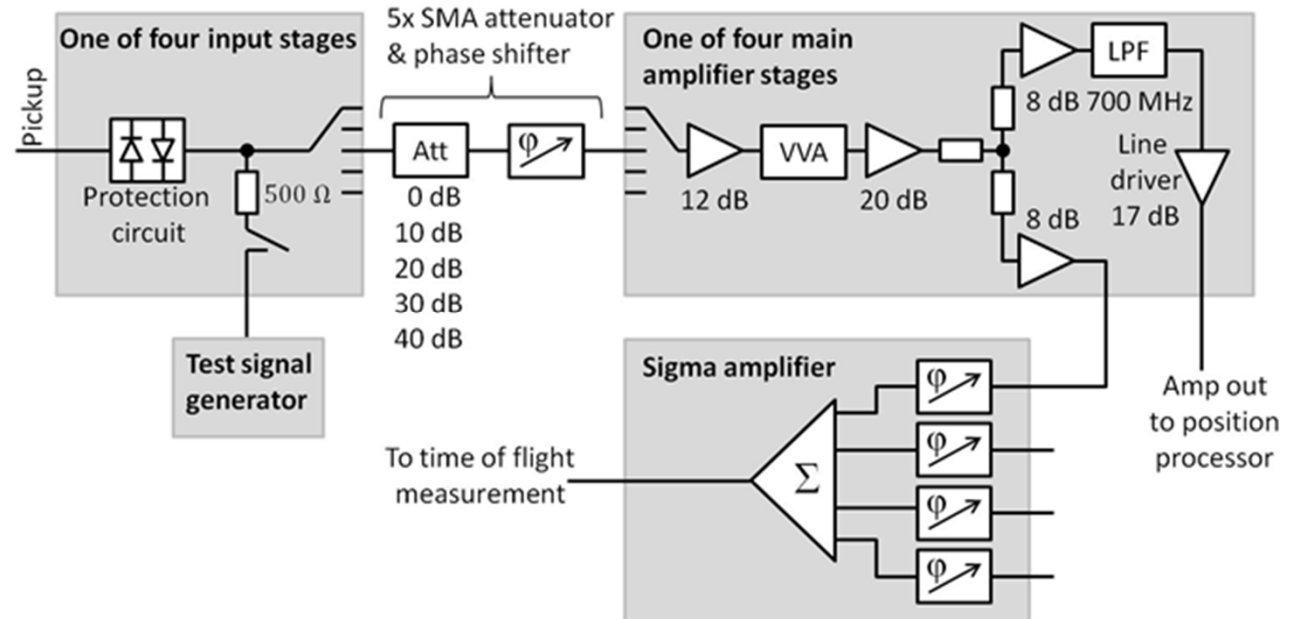


# Proton Linac – p-Linac Amplifier



## Design requirements

- Broad band architecture
- 0.1° phase accuracy @ 325 MHz
- Delay matching through phase shifters



## p-Linac Amplifier Data

- Bandwidth: 325 MHz to 3.25 GHz
- Input impedance: 50 Ω
- Output impedance: 50 Ω
- Equivalent input noise:  $1.3\text{nV} / \sqrt{\text{Hz}}$
- Gain: 0 dB to 40 dB +/- 0.1 dB
- internal test signal generator

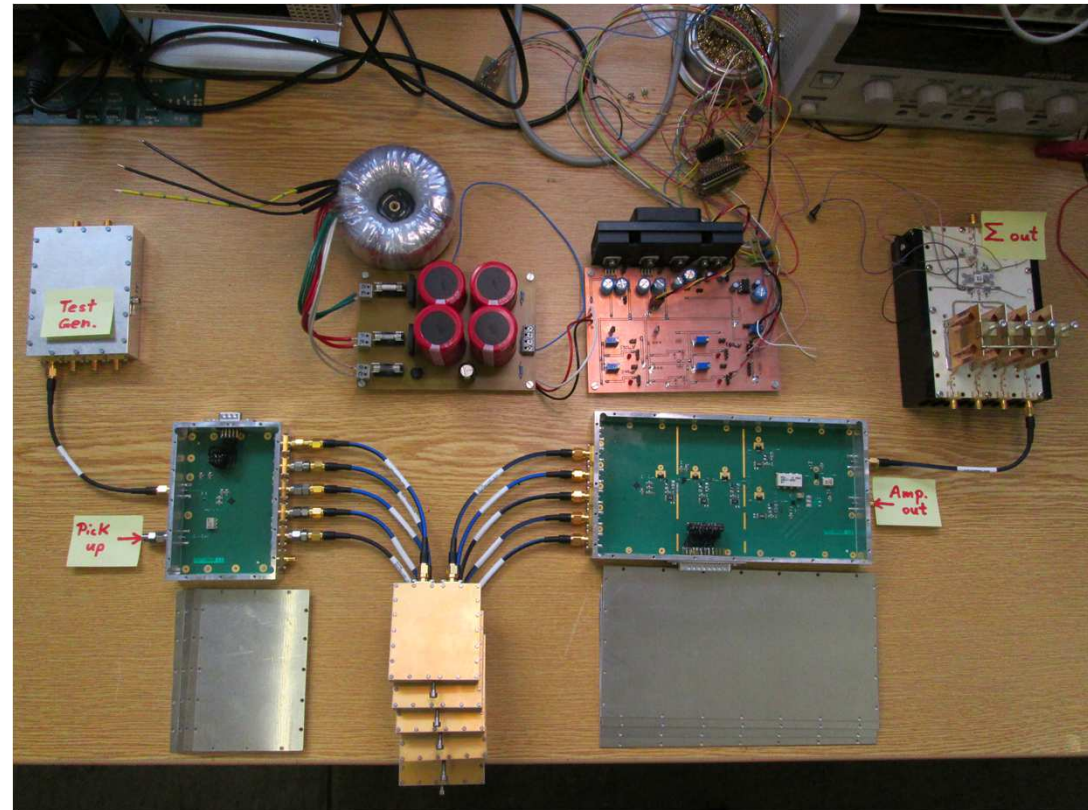
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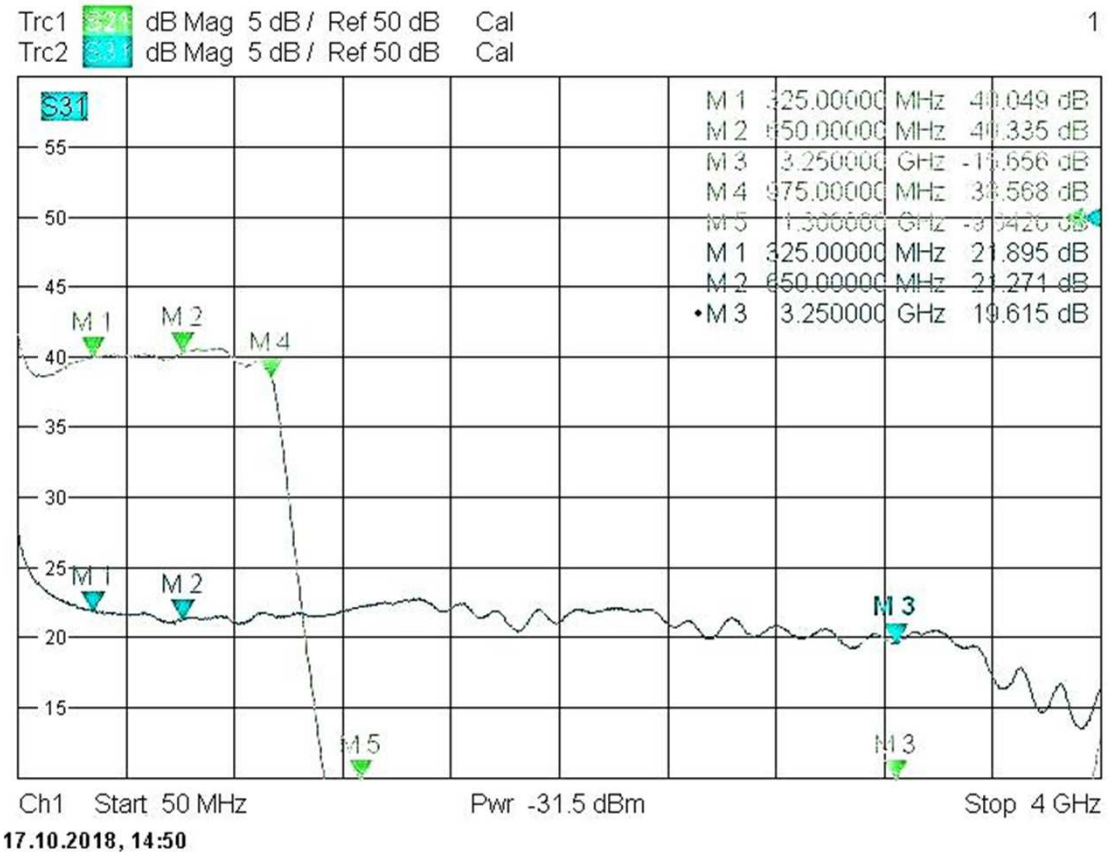


# Proton Linac – p-Linac Amplifier



## Frequency Response

- Forward transmission:  $|S_{21}|$
- Measured with a R&S ZVB4 Network analyzer
- Shows the transmission through the two different outputs



# Proton Linac – p-Linac Amplifier



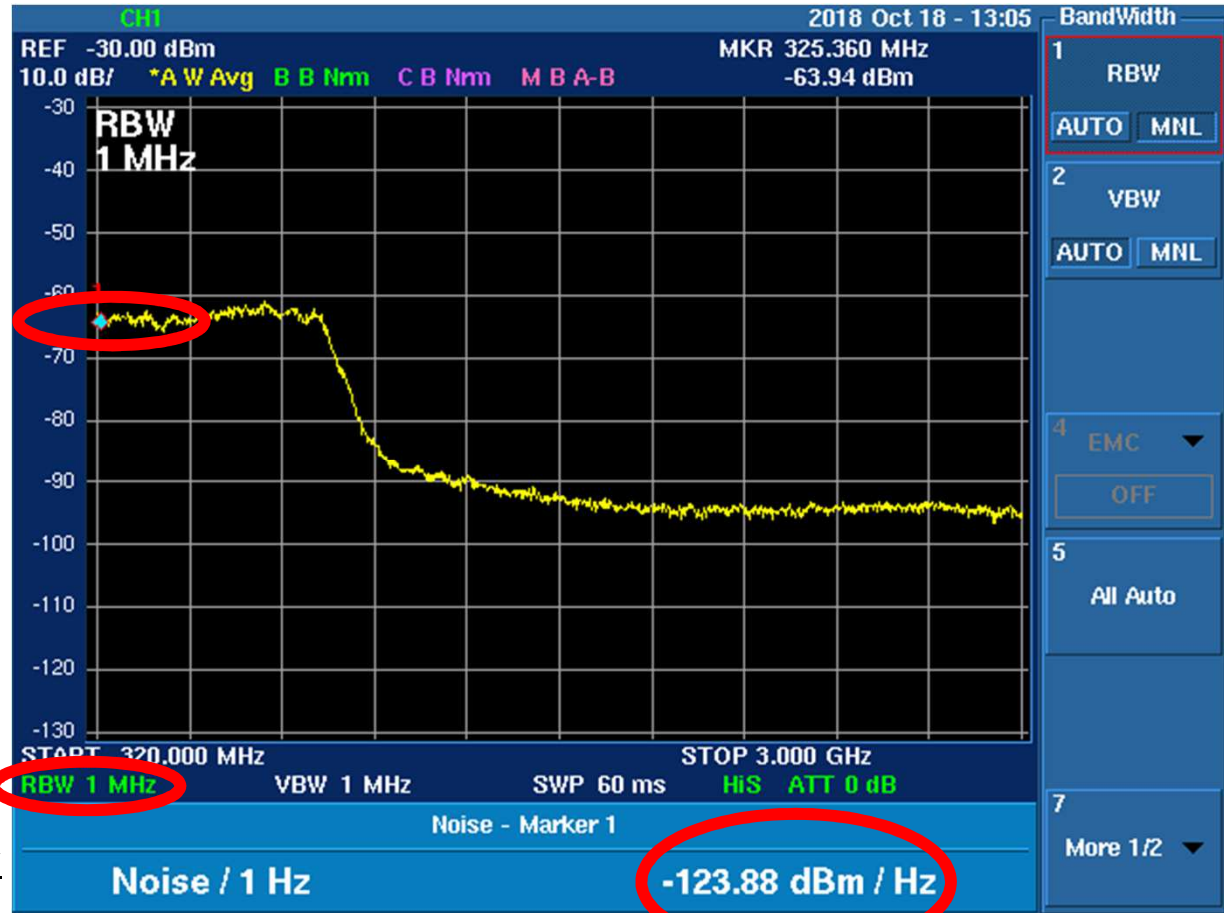
## Noise performance of the p-Linac amp at 40 dB

- Measured with an ADVANTEST U3841 FFT spectrum analyzer
- Resolution bandwidth must be taken into account to derive the spectral noise voltage density

$$-63.9 \frac{\text{dBm}}{1\text{MHz}} = -123.9 \frac{\text{dBm}}{\text{Hz}}$$

$$-123.9 \frac{\text{dBm}}{\text{Hz}} - 40\text{dB} = -163.9 \frac{\text{dBm}}{\text{Hz}}$$

$$1.4 \frac{\text{nV}}{\sqrt{\text{Hz}}} = \sqrt{50\Omega \frac{1 \text{ mW}}{\text{Hz}} 10^{-16.39}}$$



# Proton Linac – p-Linac Amplifier



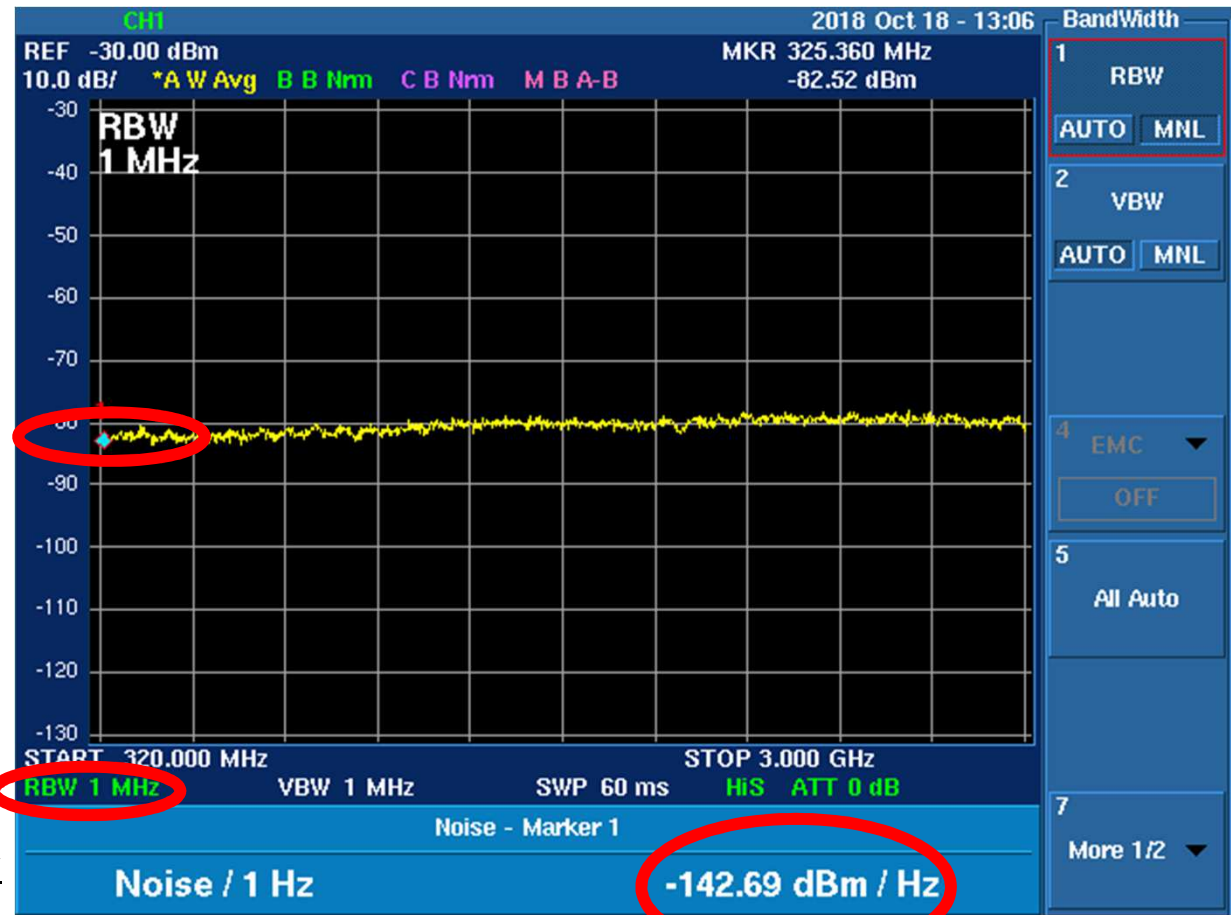
## Noise performance of the p-Linac amp at 40 dB

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- Resolution bandwidth must be taken into account to derive the spectral noise voltage density

$$-82.5 \frac{\text{dBm}}{1\text{MHz}} = -142.5 \frac{\text{dBm}}{\text{Hz}}$$

$$-142.5 \frac{\text{dBm}}{\text{Hz}} - 22\text{dB} = -164.5 \frac{\text{dBm}}{\text{Hz}}$$

$$1.3 \frac{\text{nV}}{\sqrt{\text{Hz}}} = \sqrt{50\Omega \frac{1 \text{ mW}}{\text{Hz}} 10^{-16.45}}$$



## Summary

- In house design was necessary due to special requirements
- High input impedance was needed for the Cryamp
- Large dynamic range for the high energy synchrotron amplifiers
- All three amplifiers are low noise designs
- Good gain flatness
- Fully remote controlled
- Internal test generators for gain drift check
- Recently brought in operation (Cryamp & Amplifier 110)

## Outlook

- Further testing during the beam time in February 2019
- Completion of the p-Linac amplifier
- Thank you!!!