



HOM couplers for crab cavities and challenges

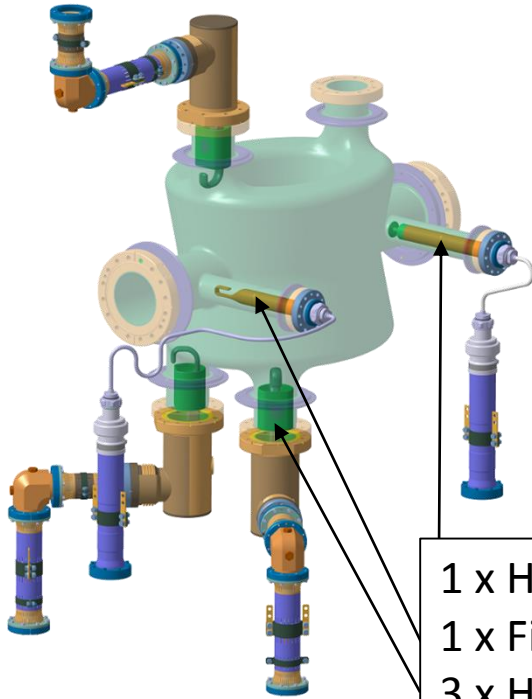
James Mitchell
CERN, BE-RF-PM

Damping requirements:

$$Z_{\parallel} < 200 \text{ k}\Omega$$

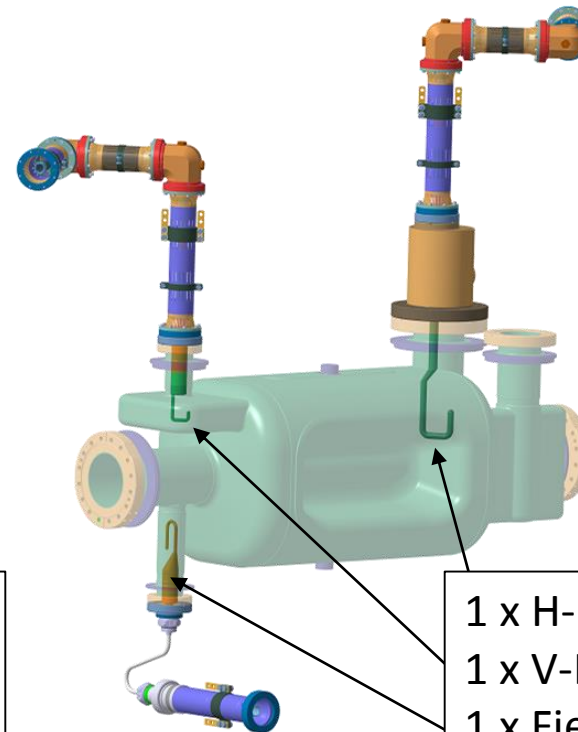
$$Z_{\perp}(x,y) < 1 \text{ M}\Omega/\text{m}$$

Crab cavity damping



- 1 x HF Damper
- 1 x Field Antenna
- 3 x HOM Coupler

**Double Quarter Wave
(DQW)**



- 1 x H-HOMC
- 1 x V-HOMC
- 1 x Field Antenna

**Radio Frequency Dipole
(RFD)**

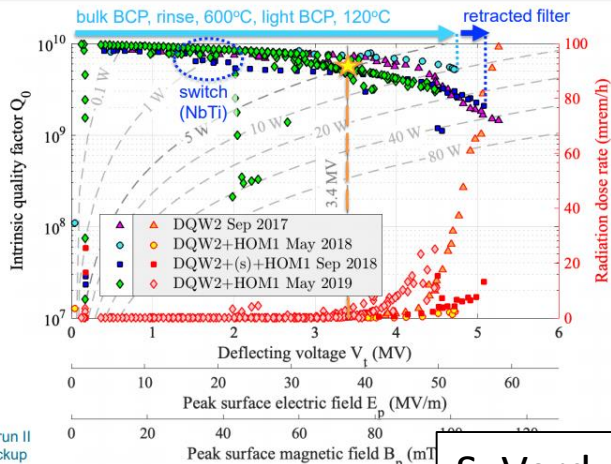
Crab cavity damping

- Dressed cavities tested without beam.
- Dressed DQW tested with beam!

Cryogenic (2K) RF perf. of DQW + HOM coupler

STUDY II – DISCRIMINATE QUENCH FROM CAVITY OR COUPLER

- Retracted filter using 20 mm spacer reduces B_p in hook by 50%, allows reaching $V_t \sim 5.1$ MV. Assume May18 test was limited by $B_p(\text{filter}) \sim 120$ mT. With spacer, the field in hook is only 60 mT, so the field in the cavity will be now the limiting factor. That is, we will expect voltages around 5.3 MV.
- Q-switch due to NbTi spacer becoming normal conductor: Q-switch $\sim 1.7e10 \Leftrightarrow \sigma = 1.3e6$ S/m.

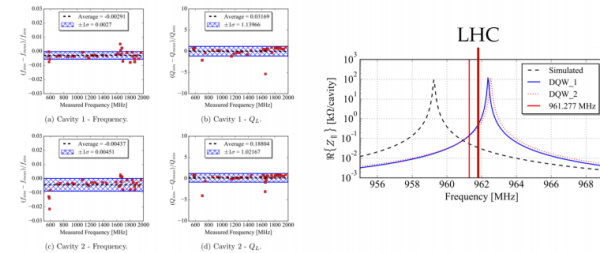


Jul'18 and Sep'18 run II not shown: see backup

S. Verdu-Andres

SPS Measurements: Pre-Installation

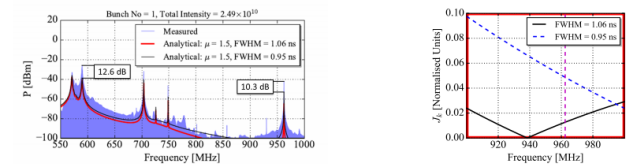
- Measured mode parameter deviation from simulations.



- f-range: $-0.9\% \rightarrow +1.0\%$, Q-range: $-50\% \rightarrow +100\%$
- 959 MHz mode
 - ▶ Frequency: $+ 3.31$ MHz and $+ 3.47$ MHz
 - ▶ Q-factors: $- 15\%$ and $- 30\%$

SPS test: Single Bunch

- Single bunch coast (one bunch for many hours).
- Measurements from each coupler compared to analytical calculations (impedance spectra altered with measured frequencies and Q-factors).



(a) Analytical and measured (average from 3 couplers) HOM power for single bunch. (b) Normalised bunch profile (form of current source).

- General form matches well (HOMs seen where predicted).
- Analytical power under-estimated.
 - ▶ Misrepresentation of proton bunch distribution.
 - ▶ Underestimation of impedance spectra.
 - ▶ Error in the measurement signal.

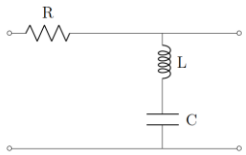
Topics

1. Dynamic heat loads (gasket heating)
2. Change of characteristic impedance (Z_0)

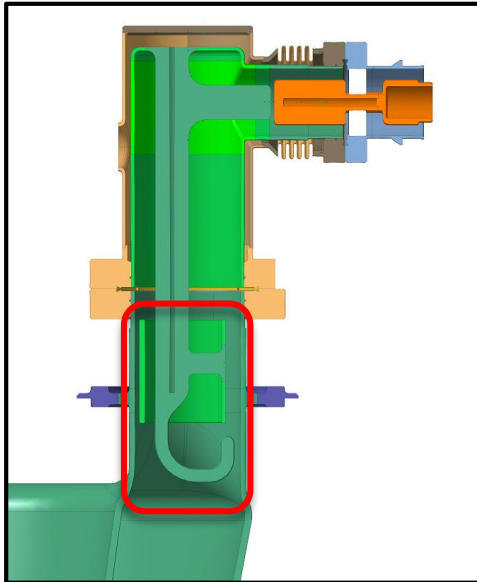
$$P = \int H^2 R_s dA$$

Dynamic Heat Loads

- Dynamic heat load on gaskets reduced by:

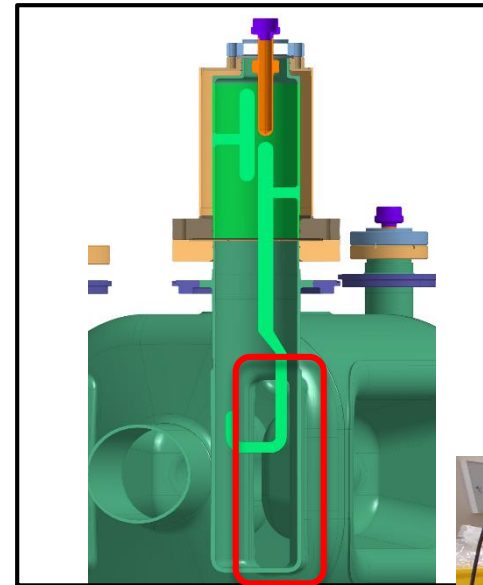


$$\omega_c = \frac{1}{\sqrt{LC}}$$



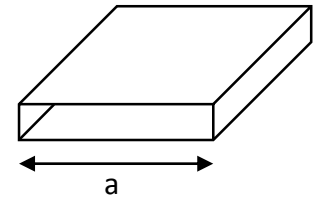
Rejection filter before the gasket.

- complicated geometry
- high fields on hook
- broad notch (mW level heat-load)

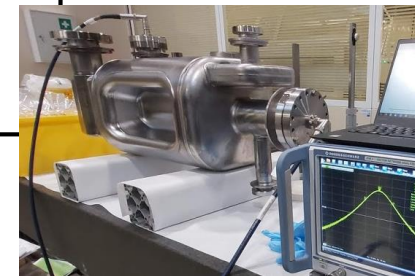


Waveguide with $f_c > 400$ MHz

- hard to machine and weld
- higher nominal gasket heat-load
- less sensitive to tolerances



$$\frac{c}{2f_{HOM1}} < a < \frac{c}{2f_0}$$



Could complex couplers and cavity shapes could be avoided with SC seals?

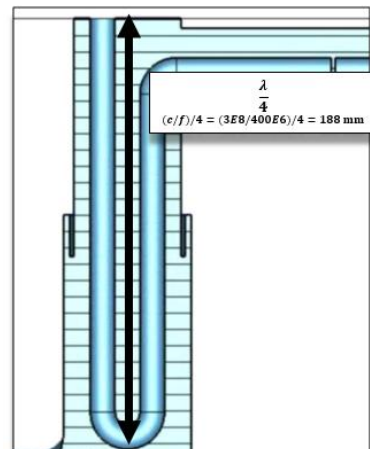
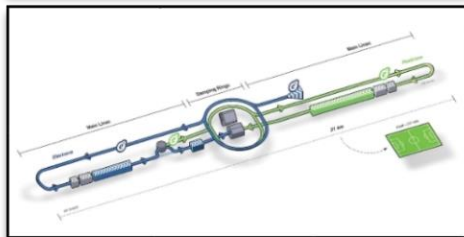
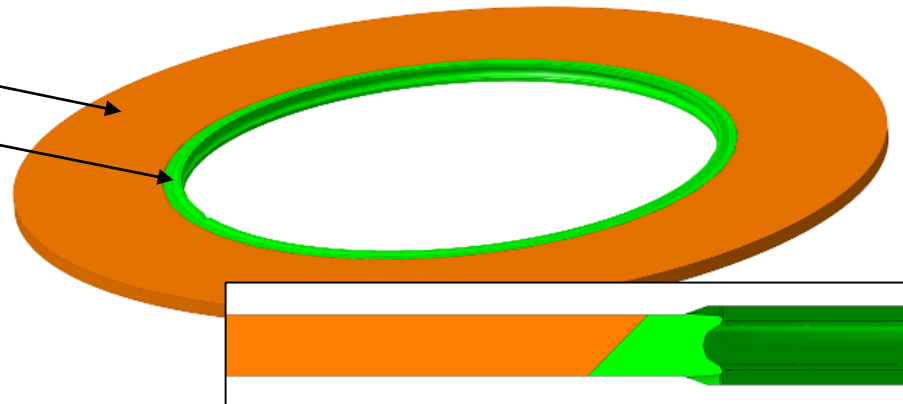
$$P = \int H^2 R_s dA$$

Dynamic Heat Loads

- Dynamic heat load on gaskets **could** be reduced by: Superconducting seals*.
- Resulting in more 'manufacturable' cavities and couplers.

Material	R_s @ 2 K [Ohms]
Copper	$\sim 1 \text{ e-3}$
Niobium	$\sim 20 \text{ e-9}$

We are trialing several options for tests!



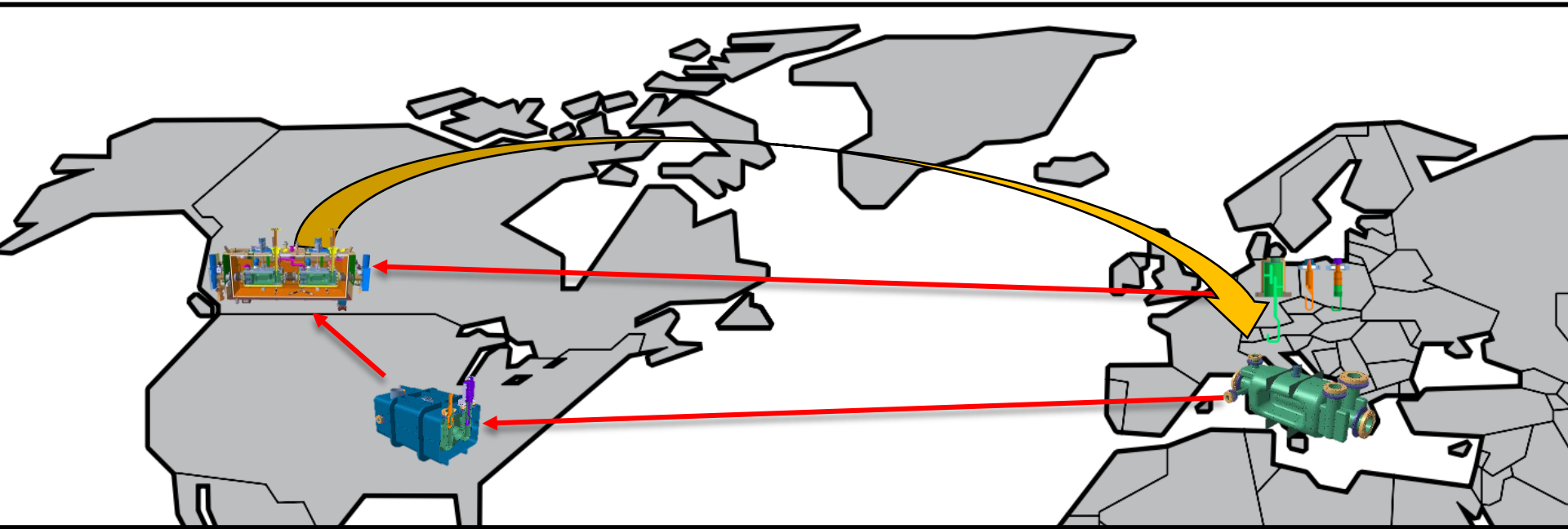
On DQW the copper gasket heat-load is 2-20 W

Future high energy machines → mass production.

Do superconducting seals allow simplification of cavity and coupler geometries?

*Existing literature on the use of superconducting seals in back-up slides.

Changing Z_0 : Manufacture and Transport



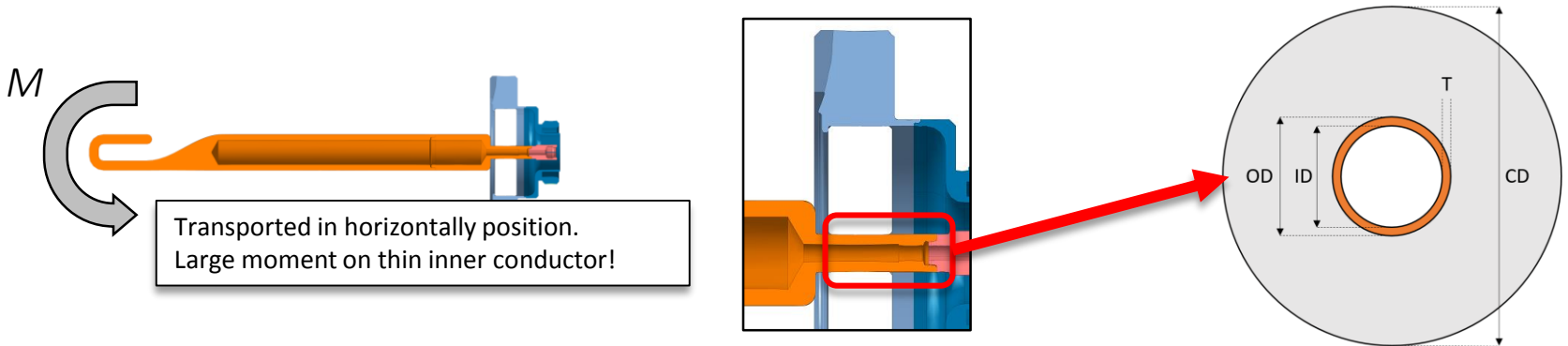
$$MI = \frac{\pi(OD^4 - ID^4)}{64}$$

$$\text{Deflection} = \frac{L^3 F}{3E \cdot MI}$$

$$\text{Bending Stress} = \frac{FL}{MI/(0.5h)}$$

Changing Z_0 : $50 \Omega \rightarrow 25 \Omega$

- ... concerns over thin diameter in feedthrough.
- Changed to $Z_0 = 25 \Omega$.

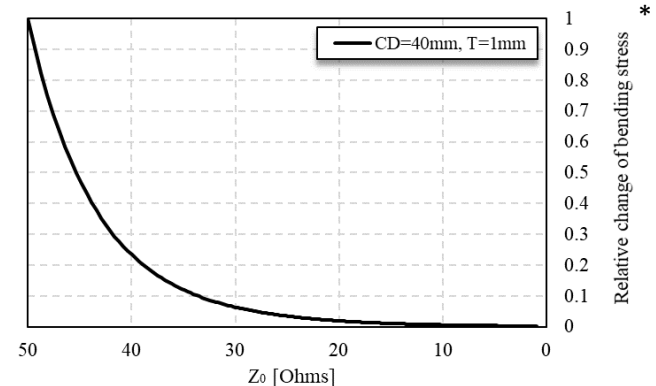


Since $Z \propto \log(OD/ID)$, diameter increases by factor of 3.7 if we move to 25 Ω .

Bending stress reduces by at least factor of 30.

*

Z_0 [Ω]	ID [mm]
75	0.78
50	2.90
<u>25</u>	<u>10.77</u>



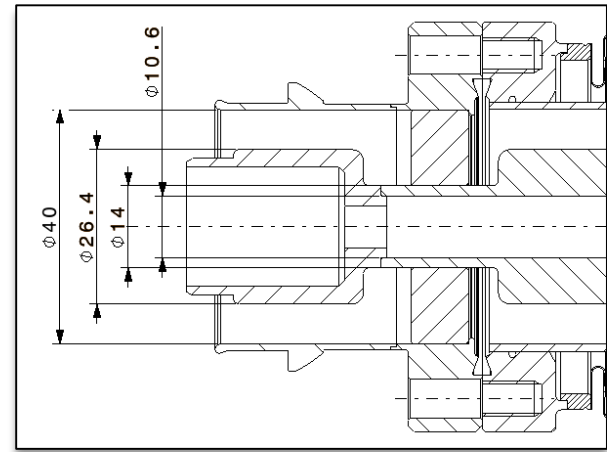
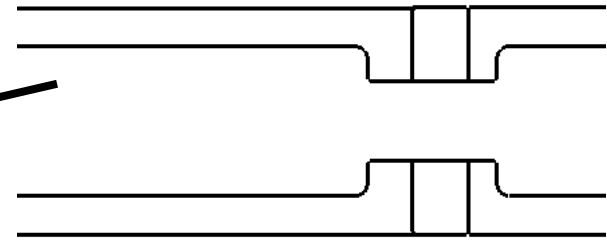
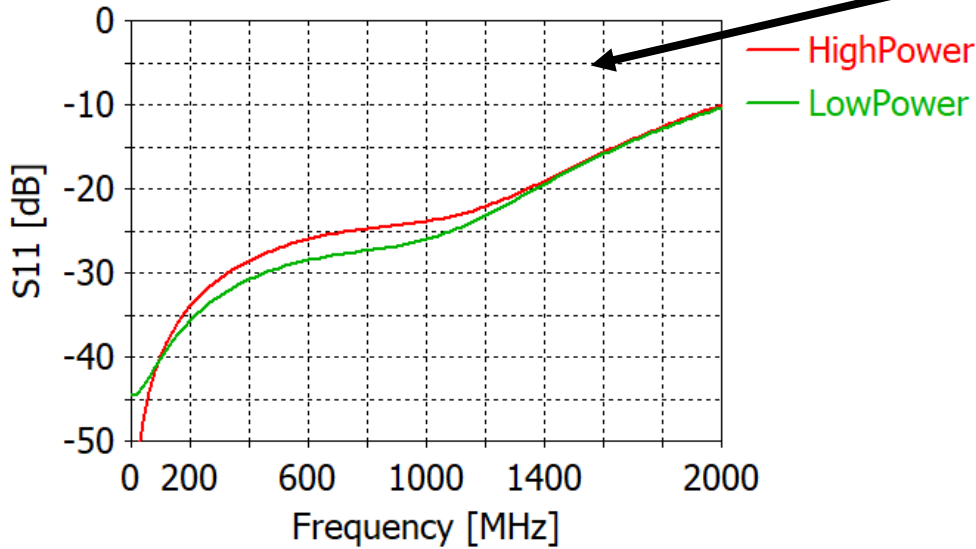
* Approximation without taking into account the boundary conditions of the ceramic.

Shock test results and videos in back-up slides!

Changing Z_0 : $50 \Omega \rightarrow 25 \Omega$

- Chose 25Ω , because $25 \Omega = 50 \Omega \parallel 50 \Omega$
- Feedthroughs designed.

With the steps and clearances needed for manufacture.
Inner diameter of 14 mm gave best broadband matching!

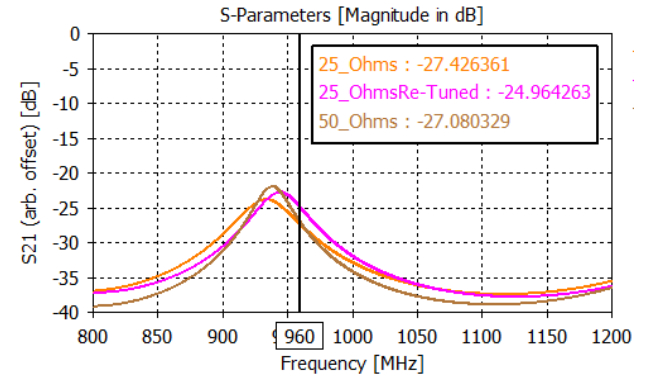
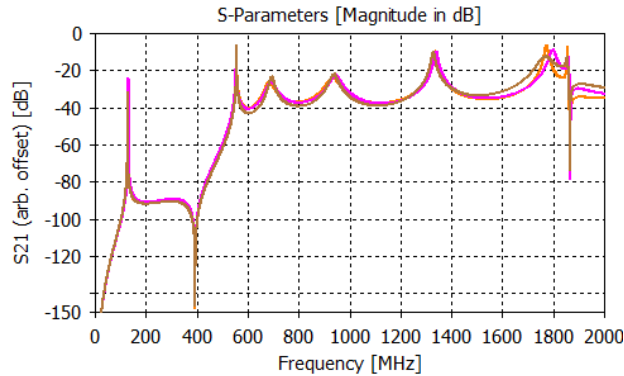
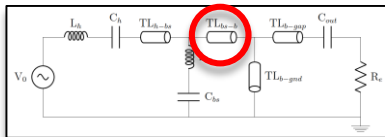
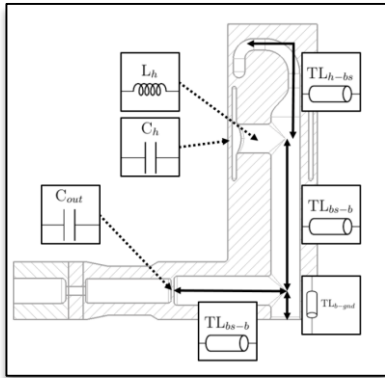


There is now a common 25Ω feedthrough validated with thermal shock and 'drop test'.

Changing Z_0 : $50 \Omega \rightarrow 25 \Omega$



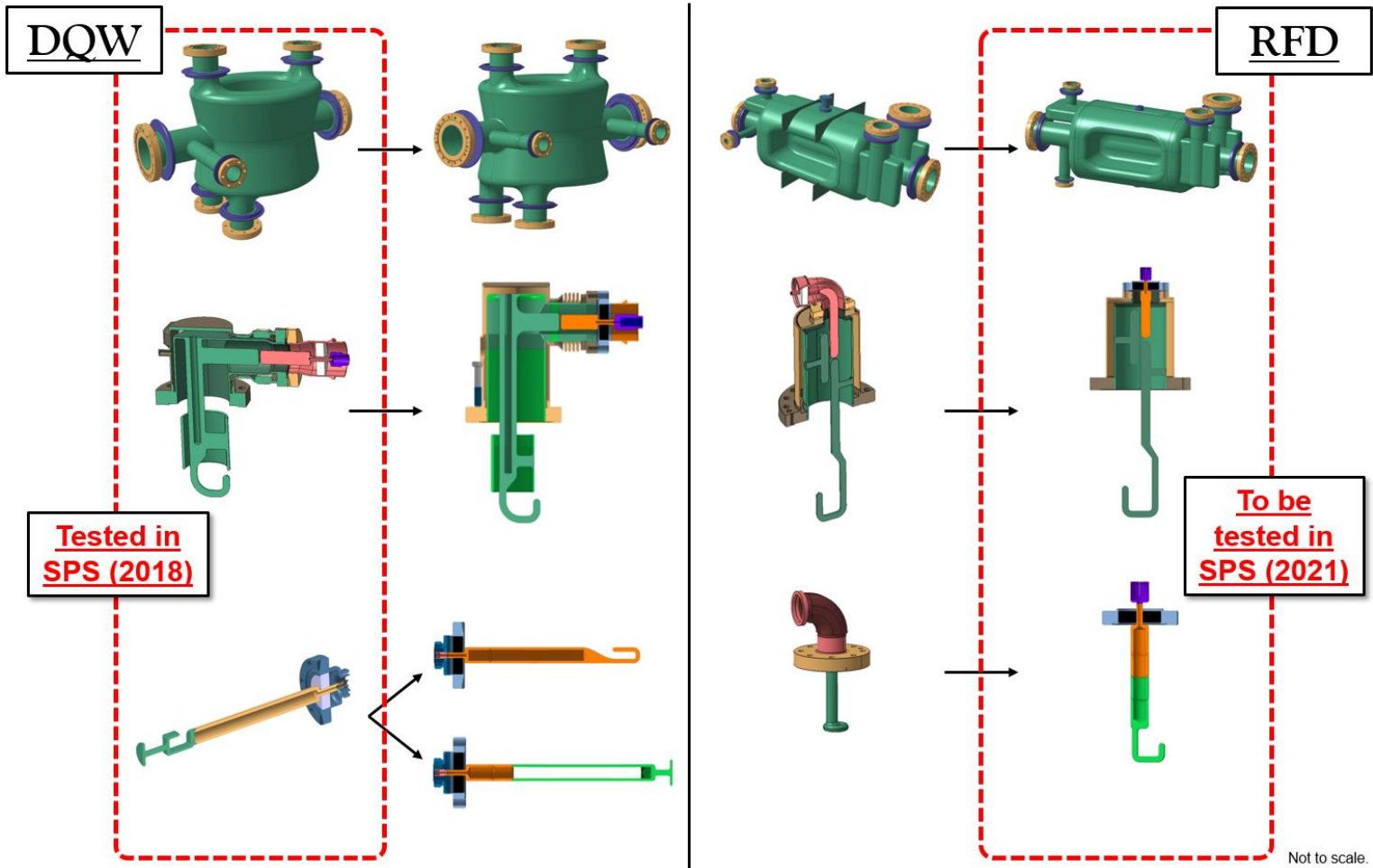
- DQW HOM coupler
 - Decrease in transmission at high power mode frequency (960 MHz).
 - Re-tuned - now impedance for this mode is the lowest it has been!



Design thresholds are met with 25 Ohm matching for both cavities!

Cavity impedances in back-up slides.

Changing Z_0 : $50 \Omega \rightarrow 25 \Omega$



Changing Z_0 : $50 \Omega \rightarrow 25 \Omega$

- Infrastructure and measurement challenges.
 - 25 ohm cables and loads are not standard: **Make cables of match in parallel?**
 - Using a 50 Ω VNA: **port re-normalization**, **de-embedding**, 50 $\Omega \rightarrow 25 \Omega$ **adapters**.

matrix S, that is, with respect to the port impedance matrix

$$Z_0 = \begin{bmatrix} Z_{01} & 0 & \dots & 0 \\ 0 & Z_{02} & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & Z_{0n} \end{bmatrix}$$

we, first, transform S to an impedance matrix using (2.21)

$$Z = Z_0^{1/2} (U - S)^{-1} (U + S) Z_0^{1/2}$$

Next the impedance matrix is transformed into the S-parameter matrix S'

$$S' = Z_0'^{-1/2} (Z - Z_0') (Z + Z_0')^{-1} Z_0'^{1/2}$$

where now a reference impedance matrix of Z_0' is used

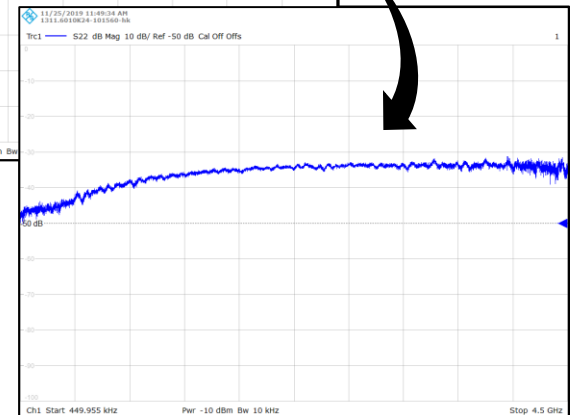
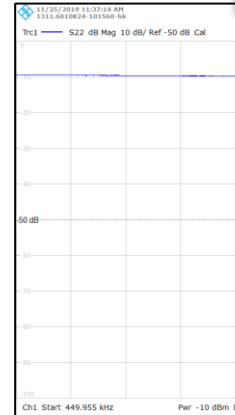
$$Z_0' = \begin{bmatrix} Z'_{01} & 0 & \dots & 0 \\ 0 & Z'_{02} & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & Z'_{0n} \end{bmatrix}$$

S' is then with respect to Z_0' . Of course, if $Z_0' = 50 \Omega$ in which case all the diag

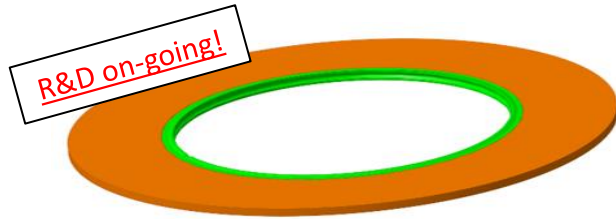
Port renormalization built into new VNAs!



WARNING:
Must de-embed adapter to 25 ohm section!

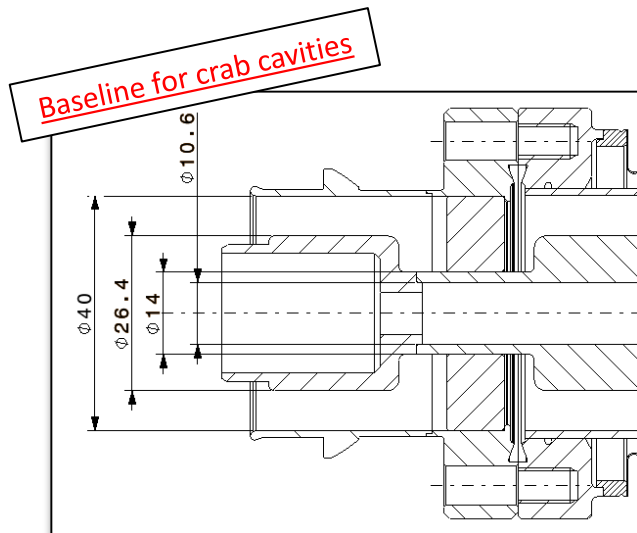


Conclusions and Discussion points



Dynamic heat loads

- We have reduced H-field.
- Could reduce Rs.
- Could this lead to simpler structures?
- Is there a want for this in the accelerator community?



Z0 = 25 Ω

- Inner conductor is too thin for transport.
- 25 Ohm infrastructure designed.
- Impedance thresholds met.
- Challenges: infrastructure and measurements.

Thank you for listening!

Thanks to BE-RF-PM and HL-LHC WP4 for the contribution and support.

Back-up slides

Fundamental mode RF parameters

Parameter	Unit	DQWCC	RFDCC
Frequency, f_0	MHz	400.44	400.75
Loaded Quality Factor, Q_l	-	—	—
r/Q_{\perp} [†]	Ω/m	429	432
Deflecting Voltage, V_{\perp}	MV	3.4	3.4
E_{pk}	MV/m	38	35
B_{pk}	mT	73	60
Accelerating Voltage, V_{\parallel}	kV	13.9	1.9
Stored Energy	J	10.72	10.62
$\Re\{b_3\}$	mT/m ²	—	—

[†] Accelerator definition.

Ancillary	P (Coax) @ VT = 3.4 MV [W]		
DQW HOMC(1,2,3)	< 0.01	< 0.01	< 0.01
DQW HF-Damper	0.16		
DQW FA	1.02		
RFD H-HOMC	0.13		
RFD V-HOMC	0.42		
RFD FA	0.97		

Fundamental mode heat loads

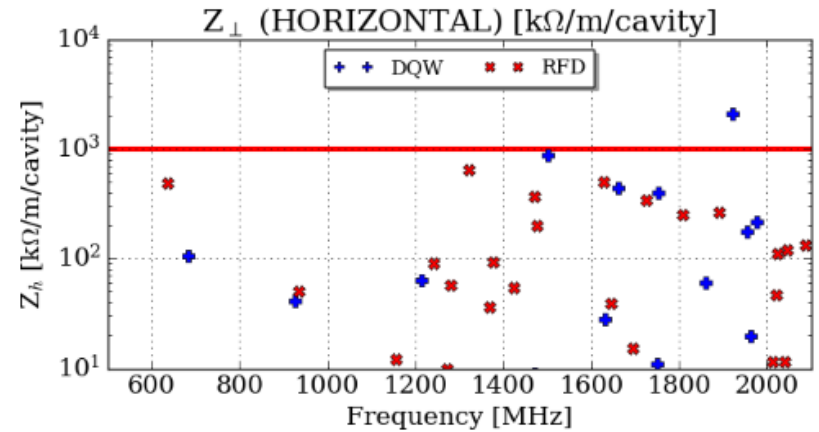
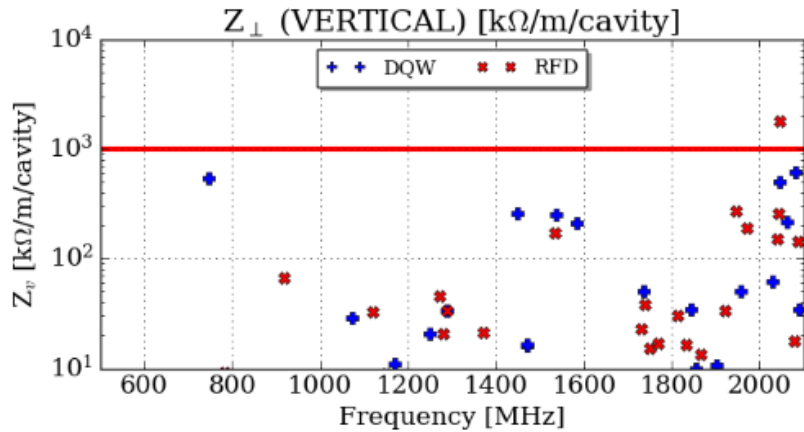
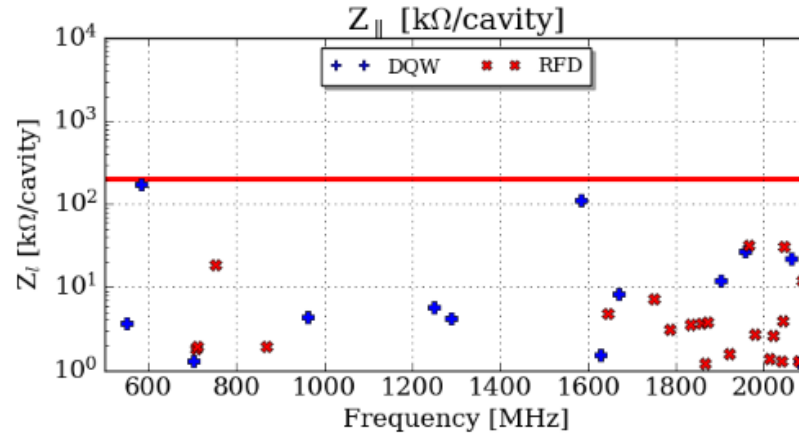
DQW

Component	Material	U = 1J			VT = 3.4 MV	VT = 5.0 MV
		P _{loss} [W]	Q0	P _{loss} [mW]	P _{loss} [W]	P _{loss} [W]
Cavity body	Nb	0.59	4.26E+09	592.34	6.35	13.73
HOMC1	Nb	0.00	1.35E+12	1.86	0.02	0.04
HOMC1 gasket	Cu	0.00	3.67E+12	0.69	0.01	0.02
HOMC2	Nb	0.00	1.11E+12	2.26	0.02	0.05
HOMC2 gasket	Cu	0.00	3.07E+12	0.82	0.01	0.02
HOMC3	Nb	0.00	1.12E+12	2.26	0.02	0.05
HOMC3 gasket	Cu	0.00	2.92E+12	0.86	0.01	0.02
FA	Cu	0.00	3.14E+12	0.80	0.01	0.02
FA gasket	Cu	0.00	6.18E+16	0.00	0.00	0.00
HF-Damper (Nb)	Nb	0.00	7.51E+14	0.00	0.00	0.00
HF-Damper (Cu)	Cu	0.00	6.94E+13	0.04	0.00	0.00
HF-Damper gasket	Cu	0.00	4.16E+17	0.00	0.00	0.00
Total [W]					<u>6.45</u>	<u>13.95</u>

RFD

Component	Material	U = 1J			VT = 3.4 MV	VT = 5.0 MV
		P _{loss} [W]	Q0	P _{loss} [mW]	P _{loss} [W]	P _{loss} [W]
Cavity body	Nb	0.50	5.02E+09	501.07	5.32	11.52
H-HOMC (Nb)	Nb	0.00	2.48E+16	0.00	0.00	0.00
H-HOMC (Cu)	Cu	0.00	6.37E+14	0.00	0.00	0.00
H-HOMC gasket 1	Cu	0.02	1.53E+11	16.41	0.17	0.38
H-HOMC gasket 2	Cu	0.00	2.39E+15	0.00	0.00	0.00
V-HOMC (Nb)	Nb	0.00	1.04E+13	0.24	0.00	0.01
V-HOMC (Cu)	Cu	0.00	1.57E+12	1.61	0.02	0.04
V-HOM gasket	Cu	0.00	5.99E+16	0.00	0.00	0.00
Field Antenna	Cu	0.00	1.14E+13	0.22	0.00	0.01
Field Antenna gasket	Cu	0.00	2.69E+16	0.00	0.00	0.00
Total [W]					<u>5.52</u>	<u>11.94</u>

Impedance spectra



P. Kneisel *et al.*, “Development of a Superconducting Connection for Niobium Cavities,” in *PAC’07*, Albuquerque, New Mexico, USA, 2007, pp. 2484–2486. [Online]. Available: <https://ieeexplore.ieee.org/document/4441291> 157

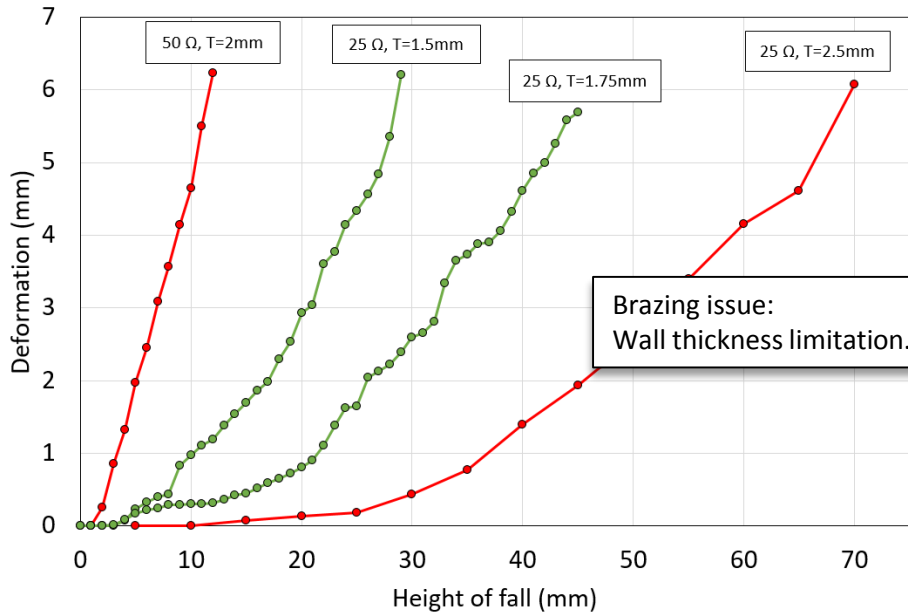
—, “Progress on the Development of a Superconducting Connection for Niobium Cavities,” *IEEE Trans. Appl. Supercond.*, vol. 19, no. 3, pp. 1416–1418, 2008. [Online]. Available: <https://ieeexplore.ieee.org/document/5109618>

R. Sundelin *et al.*, “Application of Superconducting RF Accelerating Sections to an Electron Synchrotron - A Progress Report,” Tech. Rep., 1974. [Online]. Available: <http://inspirehep.net/record/94004/files/HEACC74{-}149-153.pdf>

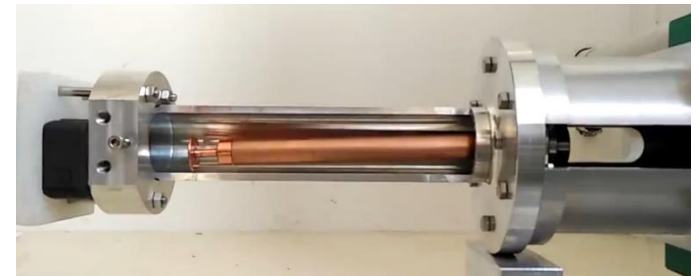
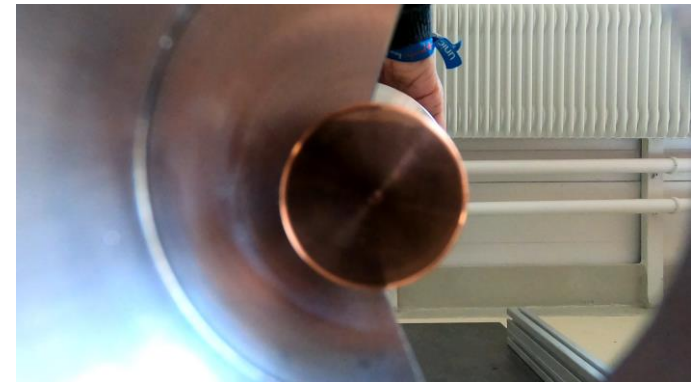
K. Saito, “Application of Mo Sealing for SRF Cavities,” in *IPAC’10*, Kyoto, Japan, 2010, pp. 3359–3361. [Online]. Available: <http://accelconf.web.cern.ch/AccelConf/IPAC10/papers/wepe009.pdf> 157

Changing Z_0 : $50 \Omega \rightarrow 25 \Omega$

- ‘Shock tests’ on-going.



25 ohm feedthrough was shown to be more resistant to a shock!



High Power HOMs

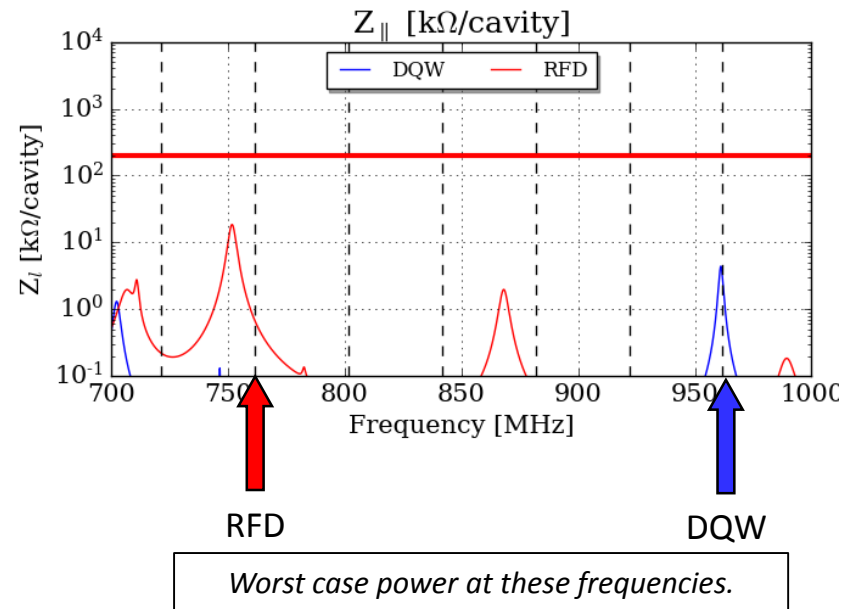
- Coaxial line power is becoming comparable to FPC power.

Using HL-LHC beam parameters:

	$P_{\text{worst-case}} (P_{\text{average}})$ [kW]	
	Gaussian bunch	Binomial bunch
DQW	1.0 (0.2)	0.5 (0.1)
RFD	7.4 (0.8)	5.9 (0.7)

HOM power from 10,000 stochastic simulations.

Parameters: bunch length, bunch form coefficient, mode frequencies and mode Q-factors.



- Infrastructure for HOM couplers is becoming larger and more difficult to assemble/replace.
- What will be the best damping method for future machines?

ImageGen

