

Design and Assessment of a Robust Voltage Amplifier with 2.5 GHz GBW and >100 kGy Total Dose Tolerance

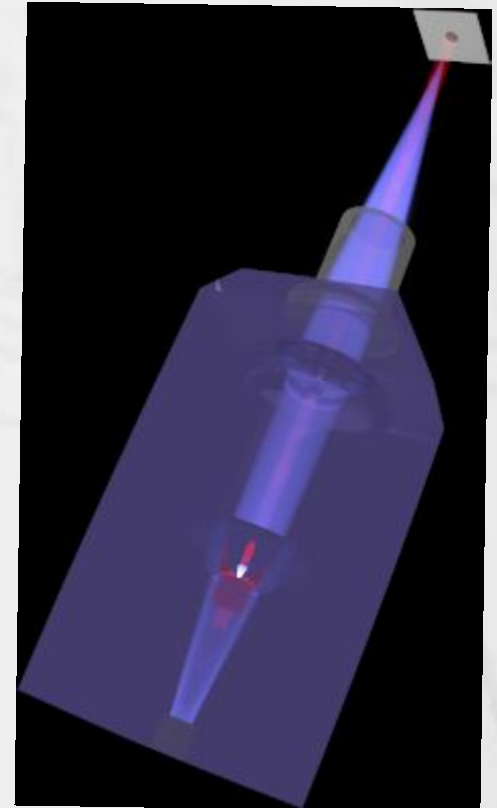
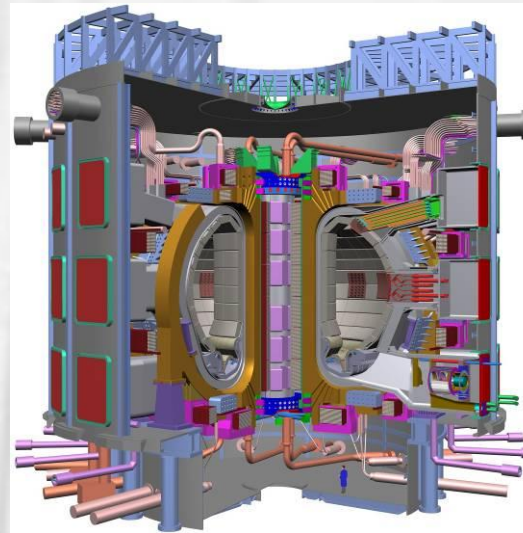
Jens Verbeeck
TWEPP 2010

- Introduction
- Radiation and temperature effects
- Constant g_m amplifier
- Matching structure
- Conclusions

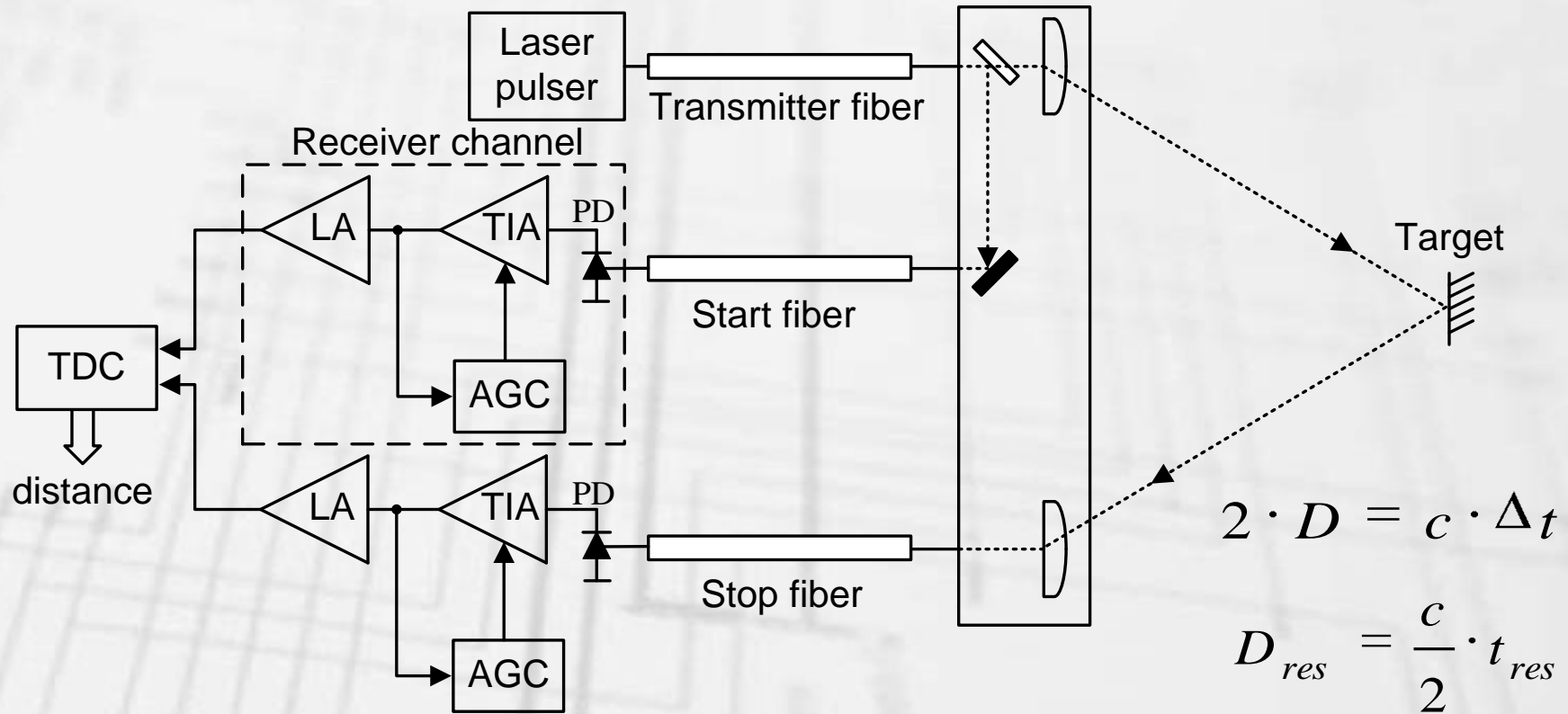
- **Introduction**
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Introduction

- Development of optical instrumentation and communication for nuclear and high temperature environments
- Applicable for
 - MYRRHA
 - ITER
 - LHC
 - Space



Optical instrumentation



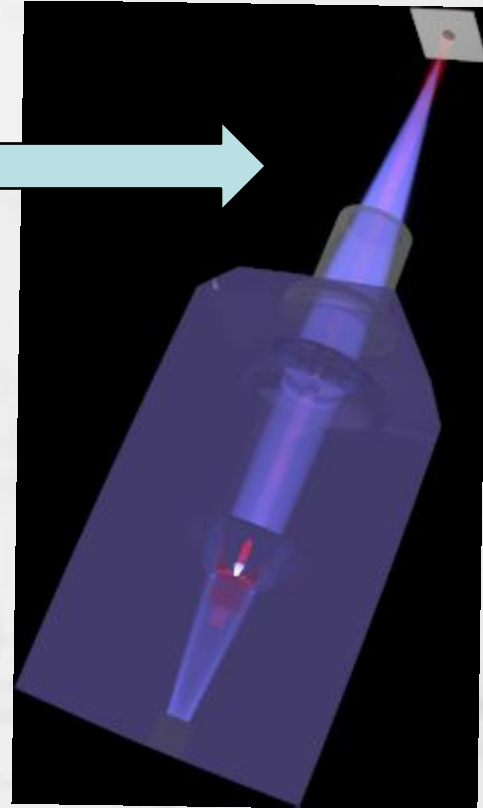
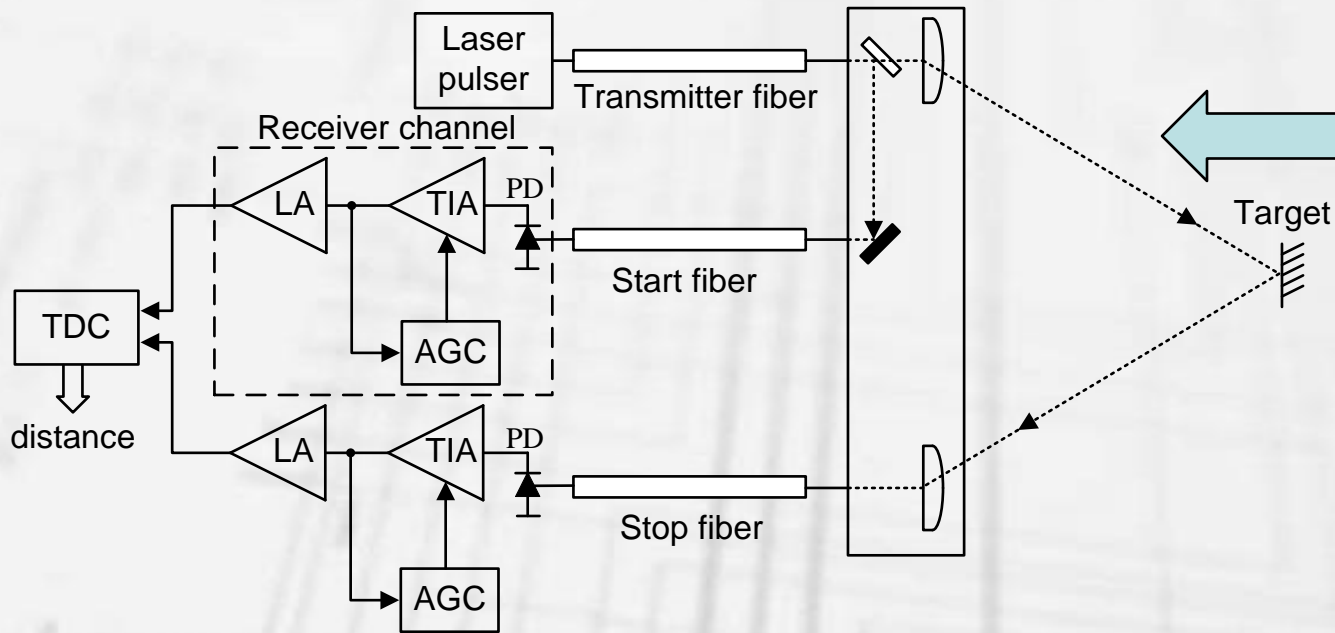
PD: Photo diode

AGC: Automatic gain controller

TIA: Transimpedance amplifier

TDC: Time to digital converter

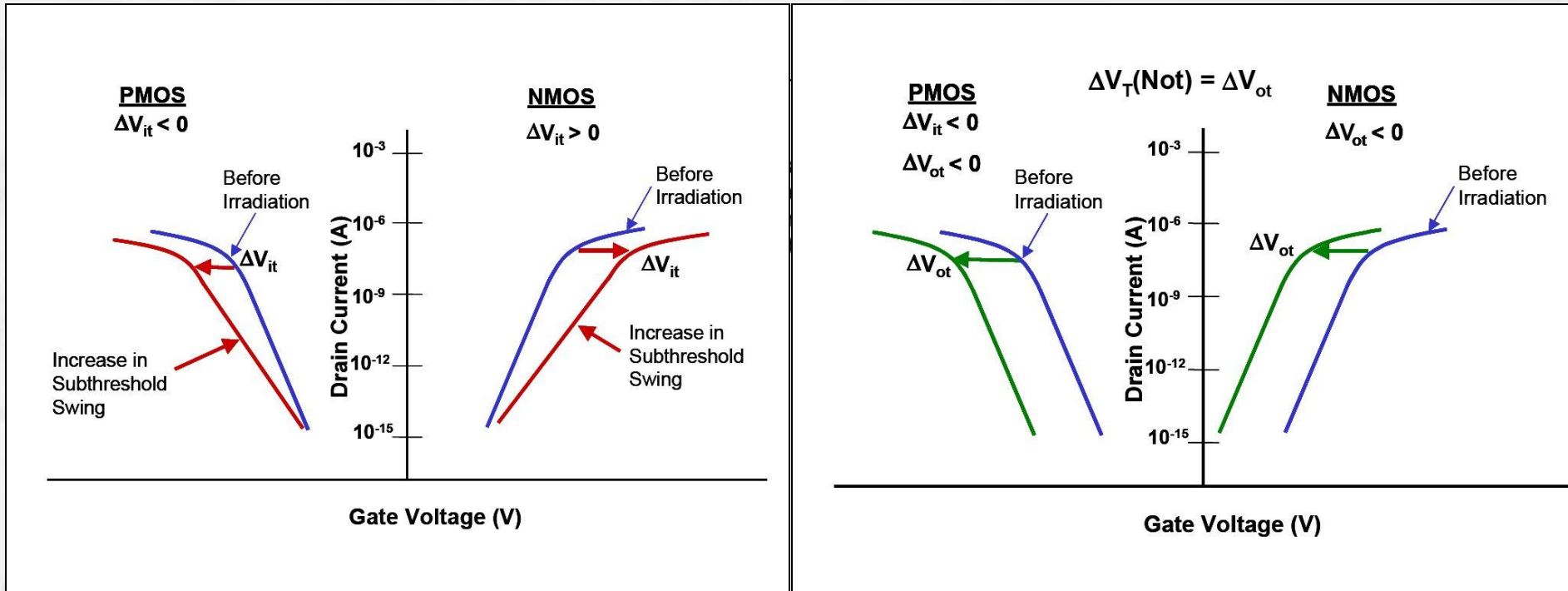
LA: Limiting amplifier



- Goal
 - LIDAR with ps accuracy, distance = 10m
- Current status
 - TDC with ps resolution developed and irradiated up to 5MGy (Ying Cao)
 - November tape-out receiver channel

- Introduction
- **Radiation and temperature effects**
- Constant g_m amplifier
- Matching as a function of radiation dose
- Conclusions

Radiation : Introduction TID

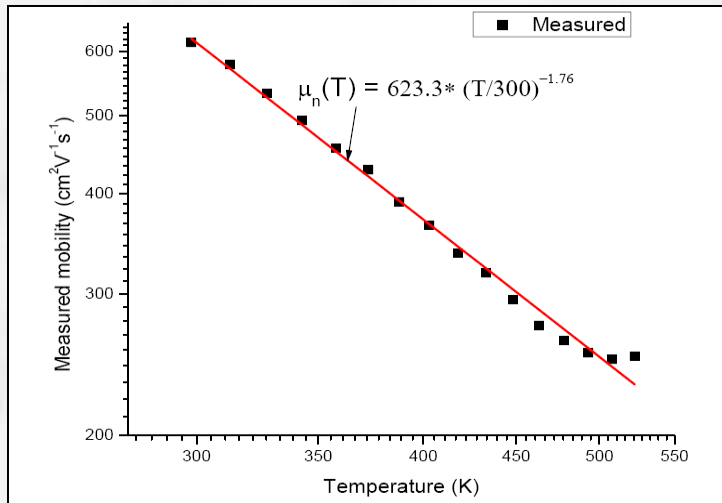


[Hugh J. Barnaby]

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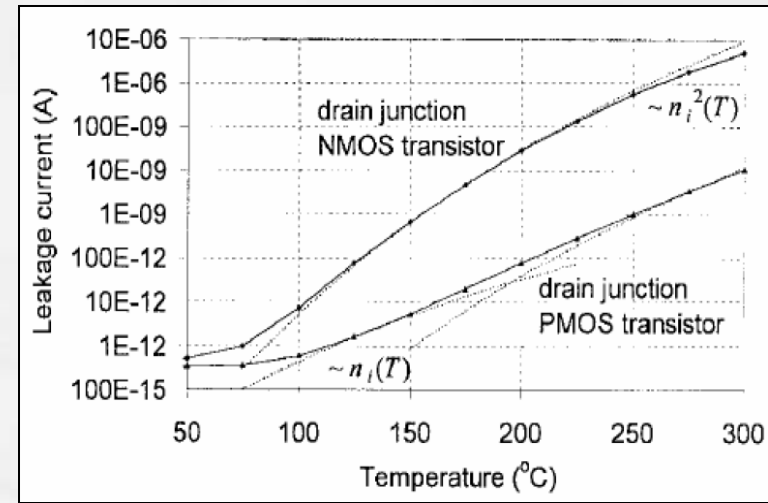
Temperature: effects

Decrease of mobility



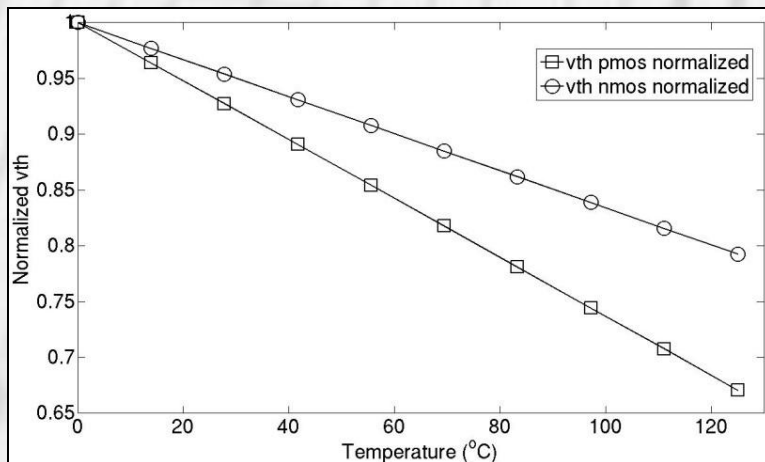
[M. Willander]

Increase of leakage current

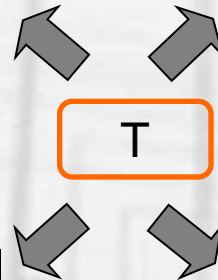
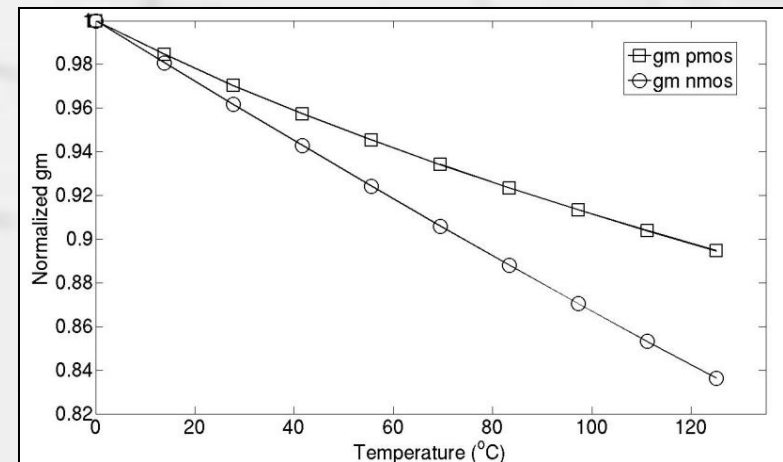


[P. C. de Jong et. al.]

Decrease of V_{TH}



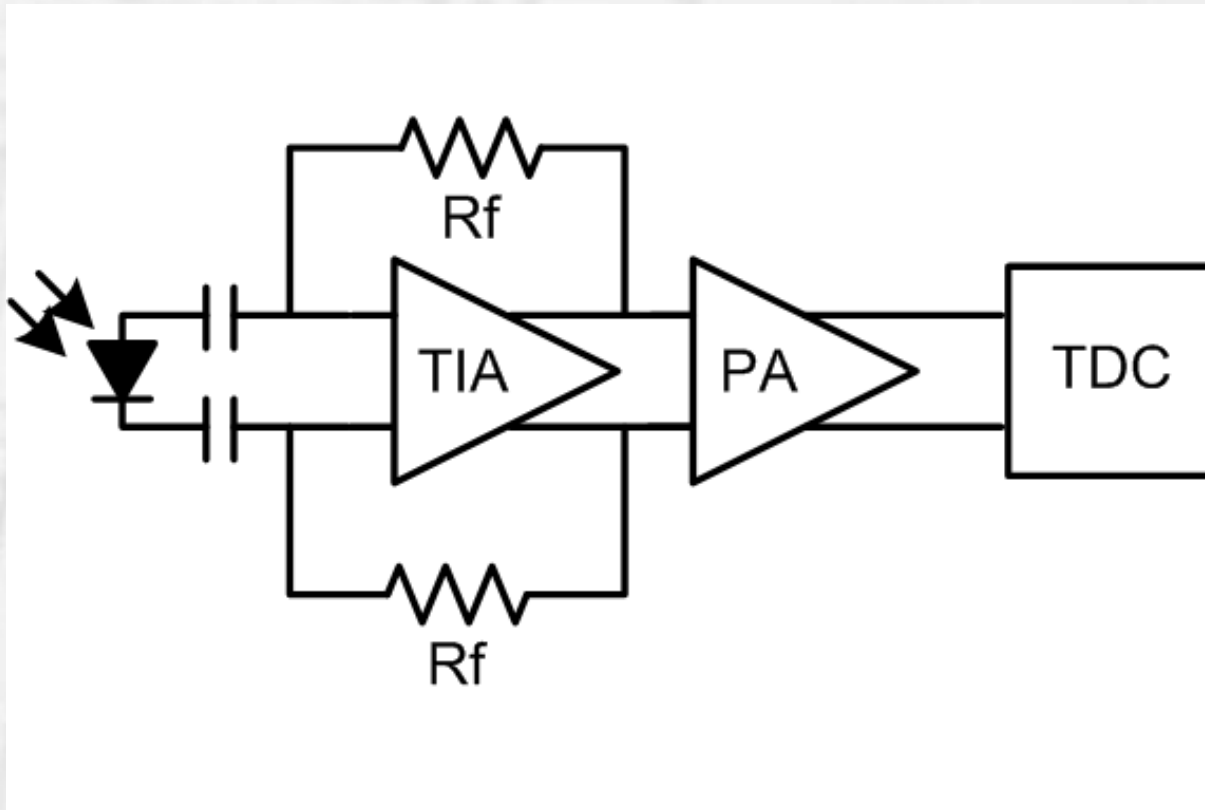
Decrease of g_m



- Introduction
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- Constant g_m amplifier
- Matching as structure
- Conclusions

- Classical building blocks

- Transimpedance amplifier (TIA)
- Post amplifier (PA)
- Time to digital convertor (TDC)

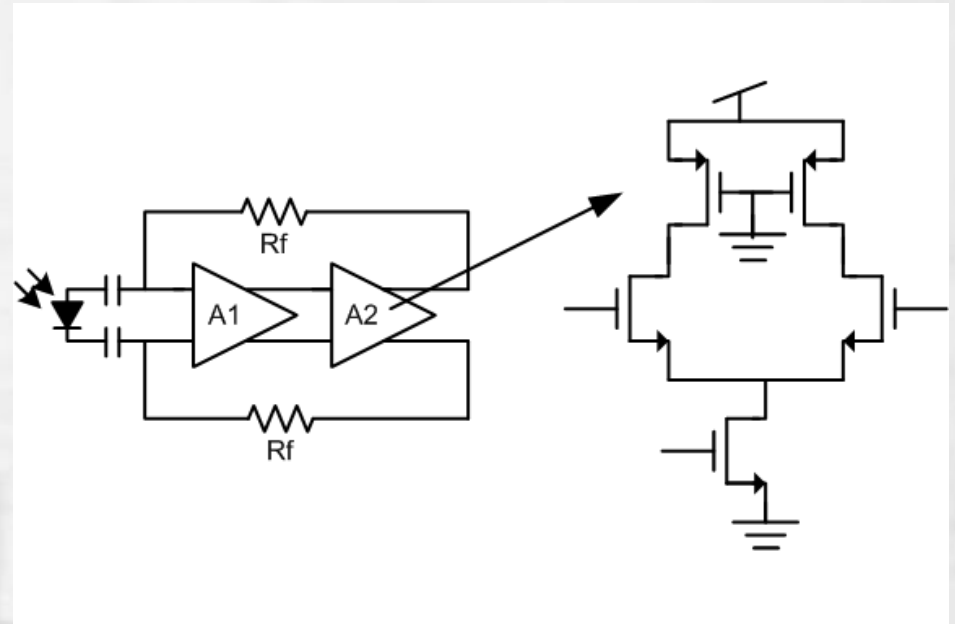


Constant g_m amplifier: Why?

Radiation/temperature causes variation in TIA parameters

$$Z_{TIA} = \frac{R_f \cdot A_0}{A_0 + 1} \quad BW = \frac{A_0}{2 \cdot \pi \cdot R_f \cdot C_{in}}$$

$$f_{nd} = \frac{1}{2 \cdot \pi \cdot R_f \cdot C_{out}} \quad GBW = \frac{A_0}{2 \cdot \pi \cdot C_{in}}$$



$$A_0 = g_{m1} \cdot r_{out1} \cdot g_{m2} \cdot r_{out2}$$

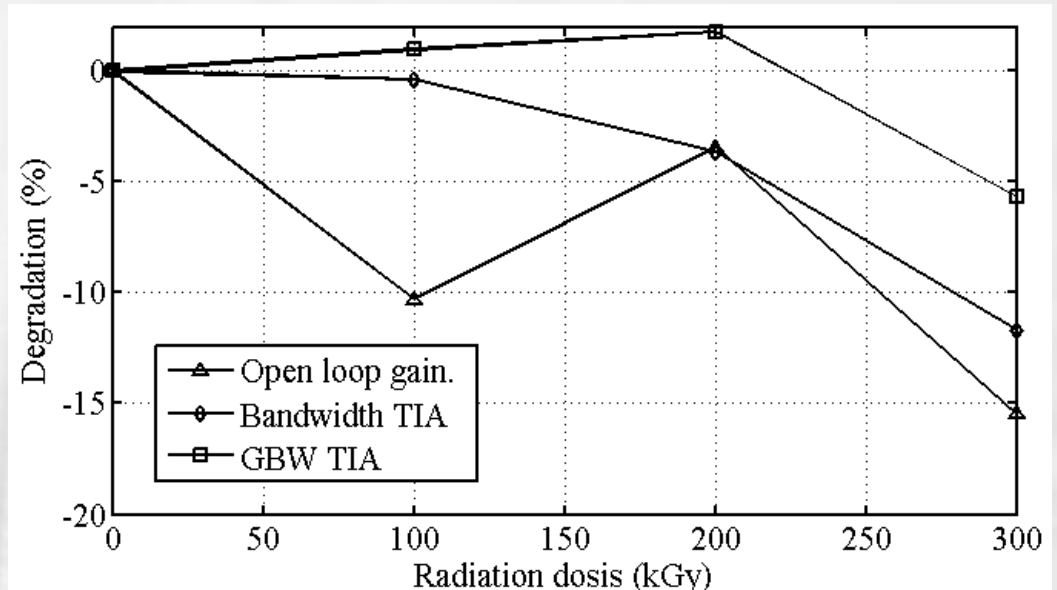


- g_m and r_{out} variable in formula A_0
 - Make gain independent of g_m and r_{out}

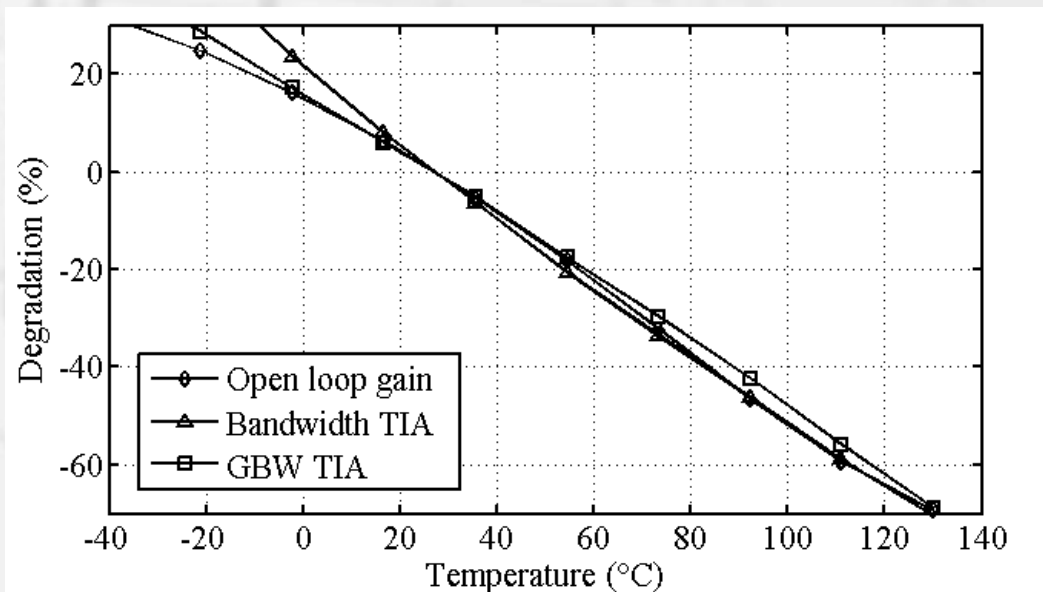
Constant g_m amplifier: Why?

Radiation
up to 300 kGy
(M. Manghisoni et al.)

0.18 μ m CMOS	100kGy	300kGy
NMOS	+7mV	+20mV
PMOS	-12mV	-60mV



Temperature
(-40 up to 130°C)



Constant g_m amplifier: Operation

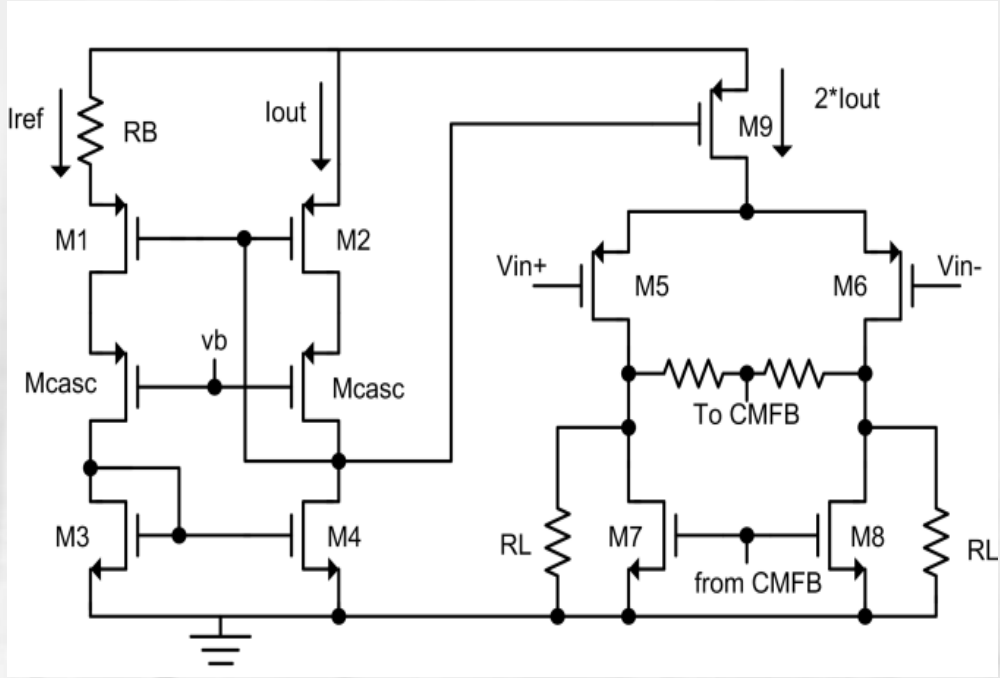
$$I_{OUT} = I_{REF}$$

$$V_{GS2} = V_{GS1} + I_{OUT} * R_b$$

$$I_{OUT} = \frac{2}{\mu_n C_{ox} \left(\frac{W}{L}\right)} \cdot \frac{1}{R_b^2} \left(1 - \frac{1}{\sqrt{B}}\right)^2$$

$$g_m = \sqrt{2 \cdot \mu_n \cdot C_{ox} \cdot (W/L)_5 \cdot I_{D5}}$$

$$g_{mx} = \frac{2}{R_B} \left(1 - \frac{1}{\sqrt{B}}\right) \sqrt{\frac{(W/L)_5 \mu_x}{(W/L)_1 \mu_n}}$$



$$A \approx 2 \frac{R_L}{R_B} \left(1 - \frac{1}{\sqrt{B}}\right)$$

[B.Razavi]

[Sean Nicalson et al.]

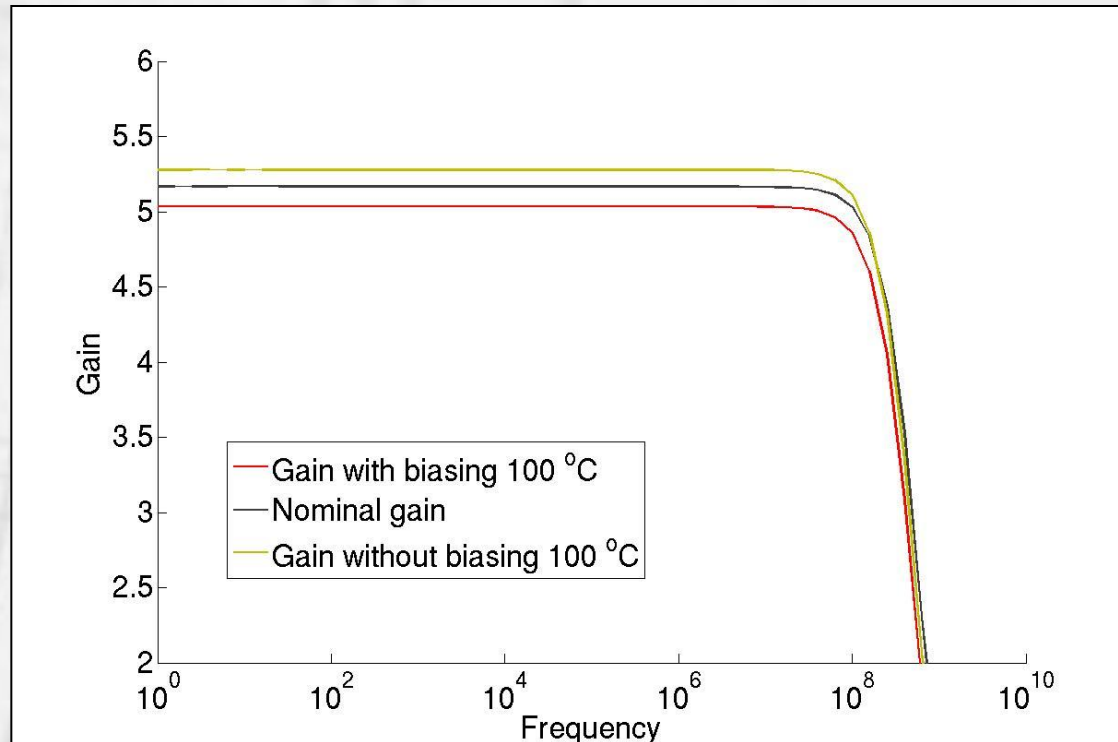
- $R_{DSTOT} \gg R_L$
 - Transistor length \uparrow
 - Trade-off: BB \Leftrightarrow temp/RAD tolerance

Temperature simulations

- Degradation with bias : 2.5 %

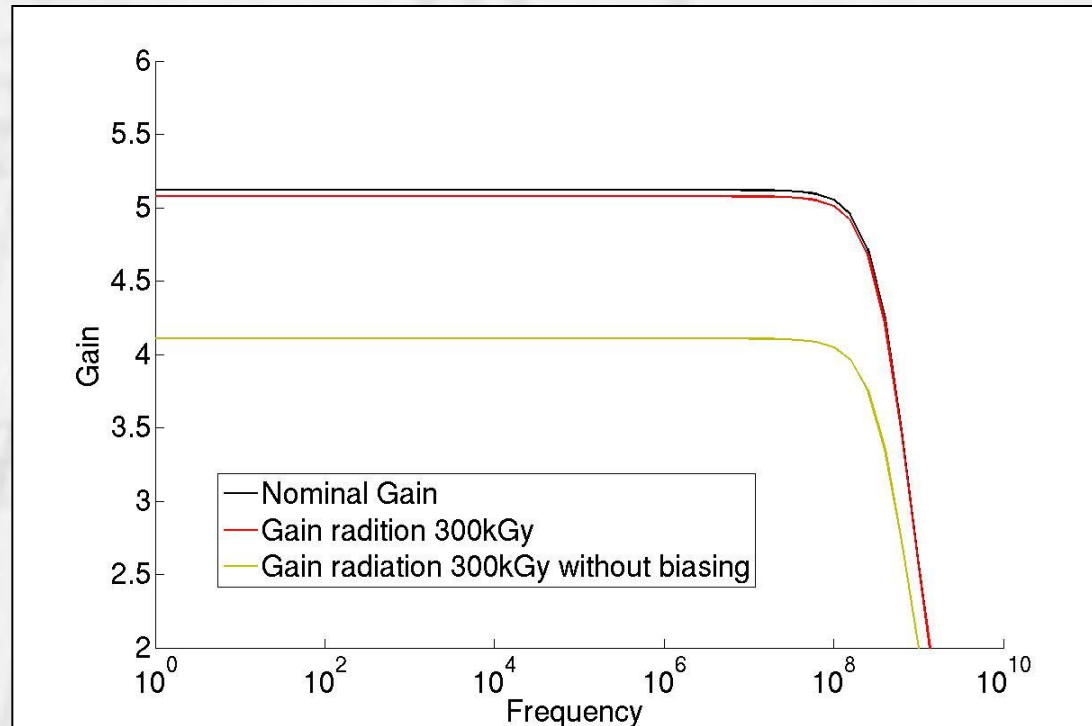
Temperature sweep from 0°C to 100°C

Gain = 14dB
GBW = 2.5GHz
BB = 500MHz



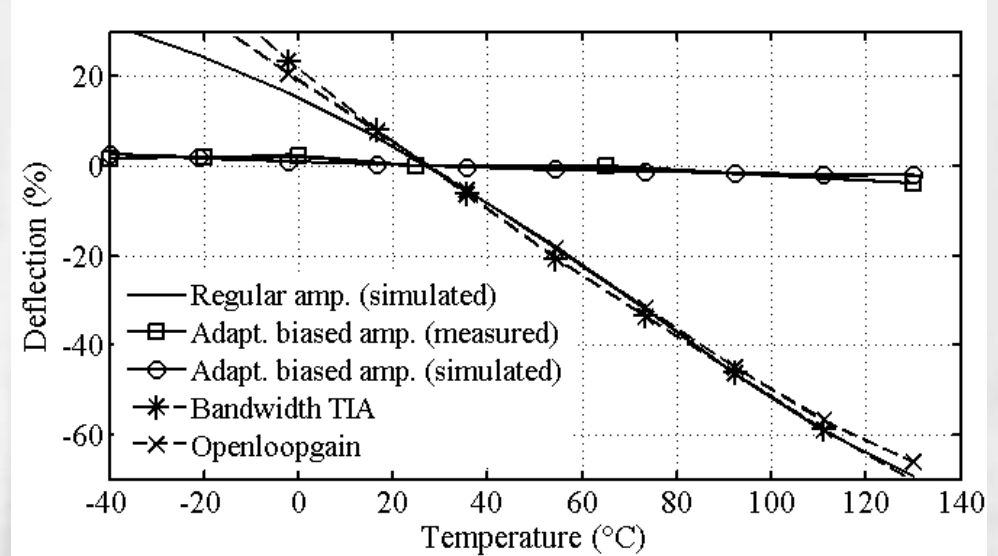
- Degradation with bias : 0.85 %

Amplifier irradiated up to 300kGy

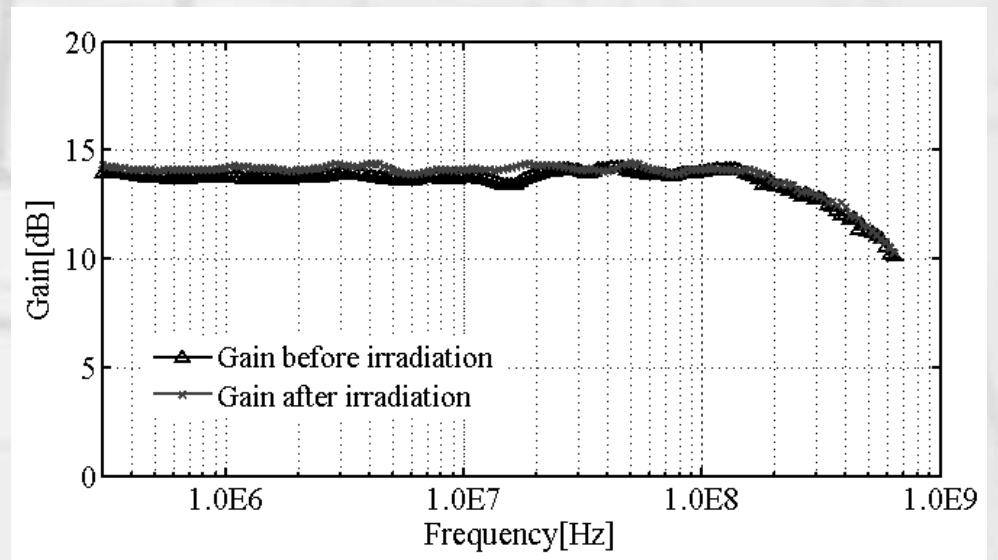


Results

Temperature drift 5.6 % or 343 ppm/°C



Radiation up to 100 kGy
Gain degradation 4.5%

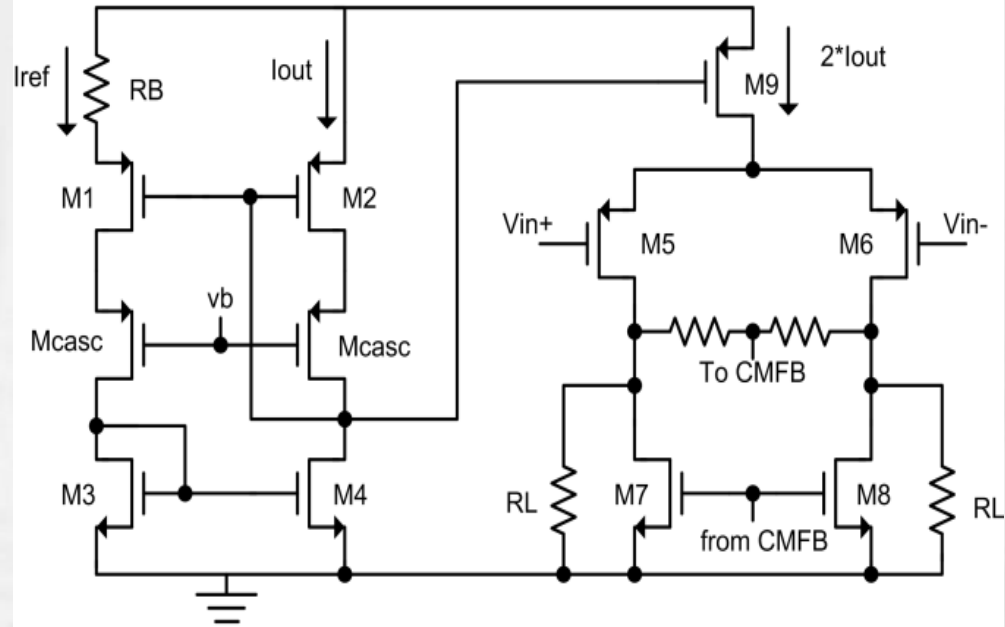


Results: Radiation

Cause degradation?

$$\frac{\Delta g_m}{g_m} = \frac{1}{V_{GSTH} \left(1 - \frac{1}{\sqrt{B}} + \frac{V_{THsens}}{V_{GSTH}}\right)} \Delta V_{THsens}$$

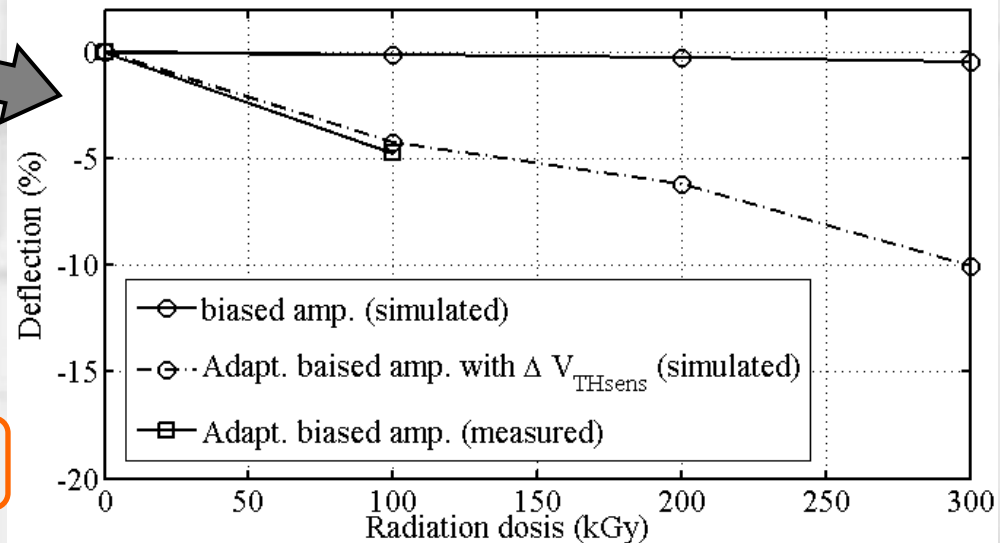
- V_{THsens} = Standard deviation before radiation of M1 and M2
- ΔV_{THsens} = Variation of V_{THsens}



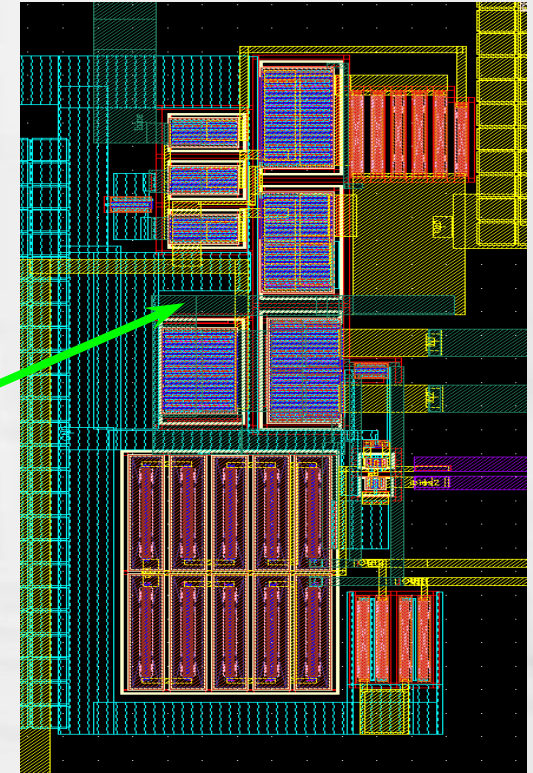
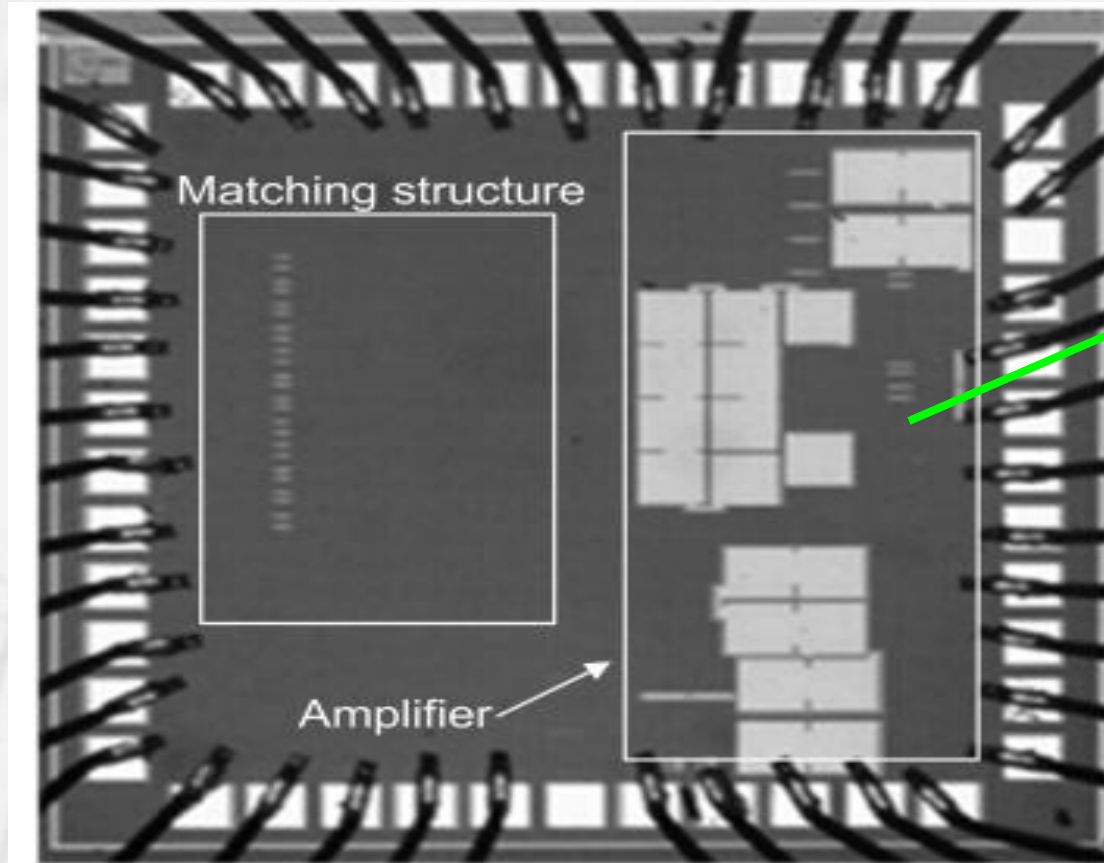
Impact of a ΔV_{THsens} of 10% on the gain of the amplifier.

Large $V_{GS} - V_{TH}$ for high radiation tolerance

$$g_m = 2 \cdot I_{DS} / (V_{GS} - V_{TH}) \Rightarrow I_{DS} \uparrow \uparrow \ \& \ V_{DSsat} \uparrow \uparrow$$



Layout



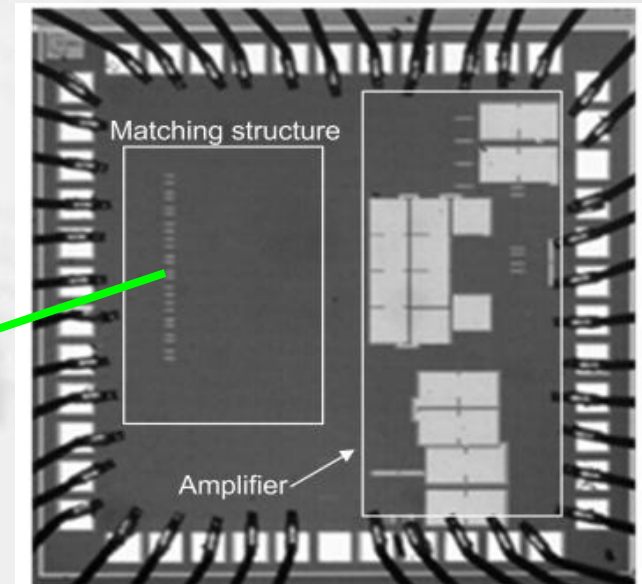
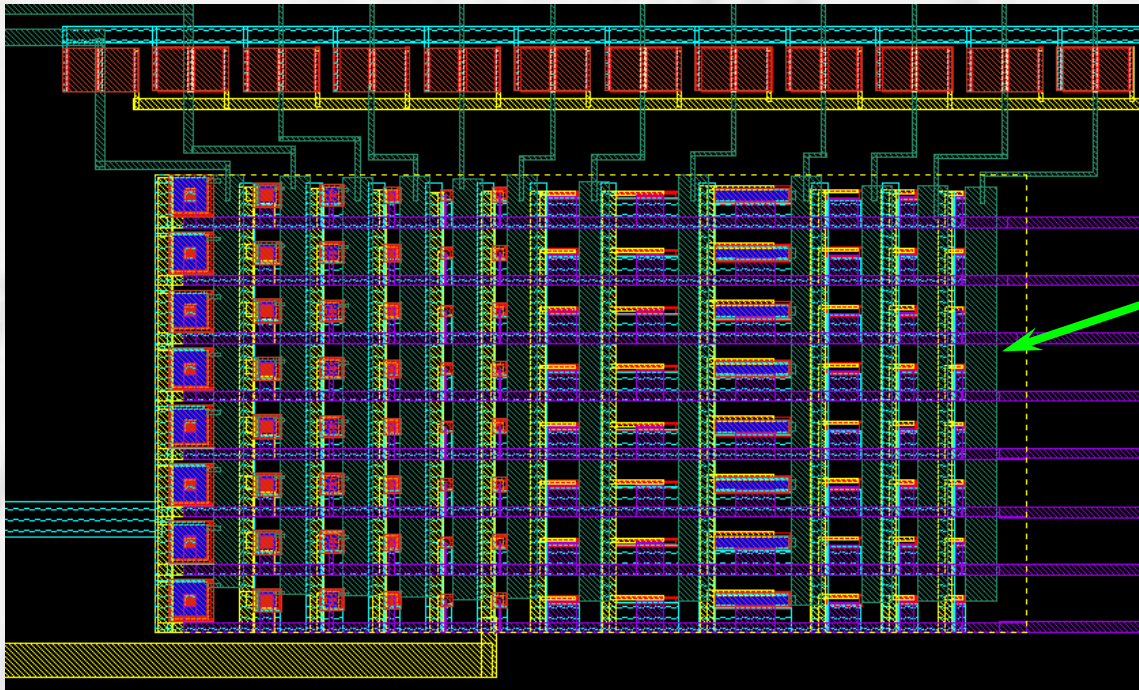
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- Standard deviation between identical components fabricated on the same chip
=>small statistic variations = MISMATCH
- Good matching
 - good PSRR, CMRR
 - Important for replica biasing
 - Reduces offset
- Matching of V_{th} depends strongly on area transistor

$$\sigma_{vth} = \frac{A_{vth}}{\sqrt{W * L}}$$

- Layout transistors on chip

- Drains connected in each column
- Gates connected in each row
- Shared source and bulk
- 6 regular transistors, 6 Enclosed layout transistors [G. Annelie et al.]



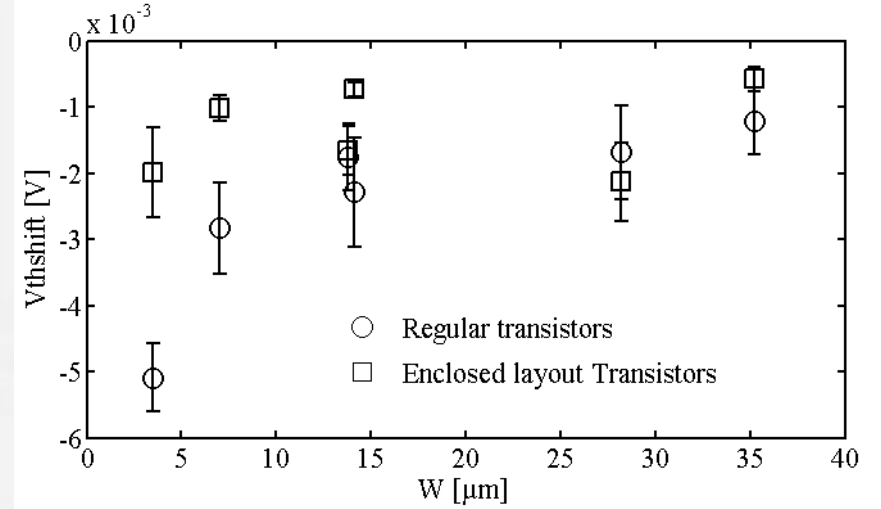
Regular NMOS transistors

- Decrease of V_{TH}
 - Rebound effect
- RINCE effect [F. Faccio]
 - Different effect for large gates
 - Radiation Induced Narrow Channel Effect

- Standard deviation V_{TH} -shift! =>

ELT transistors

- No significant effects



$$\frac{\Delta gm}{gm} = \frac{1}{V_{GSTH} \left(1 - \frac{1}{\sqrt{B}} + \frac{V_{THsens}}{V_{GSTH}}\right)} \Delta V_{THsens}$$

- Introduction
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- Transistor measurements
- **Conclusions**

- Effects of radiation and temperature
 - Change of transistor parameters
 - Varying gain
- Gain can be held stable with constant g_m amplifier
 - BW and GBW of optical receiver guaranteed
 - Open loop control
 - Generate bias voltages for whole chip with g_m biasing
 - Trade off: bandwidth \Leftrightarrow temperature tolerance
 - Trade off: current consumption & voltage headroom \Leftrightarrow rad. Tolerance
 - Larger $V_{GS} - V_{TH} \Rightarrow V_{DSsat} \uparrow \uparrow \Leftrightarrow$ difficult to keep transistors in saturation at high temperatures.
- Matching results
 - Decrease of V_{TH}
 - V_{TH} -shift depending on transistor width
 - Standard deviation V_{TH} -shift

Acknowledgments to:



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6. Sean Nicalson and Khoman Phang, "Improvements in biasing and compensation of cmos opamps", ISCAS 2004
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8. G. Anelli et al., " Radiation tolerant VLSI circuits in standard deep submicron CMOS technologies for the LHC experiments: Practical design aspects," *IEEE Trans. Nucl. Sci.*, vol. 46, no. 6, pp. 1690–1696, Dec.1999.