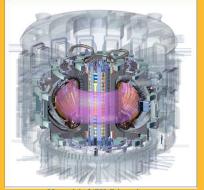
Introduction

ITER is an international project located in the South of France which aims to demonstrate the feasibility of nuclear fusion for power generation. The ITER superconducting magnet system is composed of 18 Nb3Sn Toroidal Field (TF) coils, 6 NbTi Poloidal Field (PF) coils, 6 Nb3Sn Central Solenoid (CS) Coils and 18 NbTi Correction Coils (CC).



3D model of ITER Tokamak.

The TF coils generate the toroidal field needed for plasma confinement; the PF and CS coils generate the poloidal field required for plasma control and shaping; the CS also acts as the primary of a transformer inducing and maintaining the plasma current. The CC correct field errors due to positioning or manufacturing tolerances of the main coils.

All these coils are cooled with supercritical Helium flowing at a temperature of ~4.5 K. The thermal loads in the magnet system have to be strictly controlled in order for the coils not to quench. These thermal loads are mainly the static heat loads (radiation from the thermal shield and conduction through the supports), neutron heating coming from the fusion plasma and the AC loss (in the superconducting cables and due to eddy currents in the surrounding metallic structures), each representing about a third during a normal plasma scenario.

AC loss in the coil conductors is the topic of this presentation.



Development of an AC Loss Model for the ITER Central Solenoid during a Plasma Scenario

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AC Loss in CS Coils

The coupling and hysteresis losses in the ITER CS coils average out to ~6 kW over the 1800 sec reference plasma pulse (which includes ~500 s of fusion burn). This power, which does not include the eddy current heat load in the passive structures (another 1 kW), needs to be compared to the 75 kW @ 4.2 K installed cryo-cooling capacity, For a regular plasma pulse the total AC loss represents $1/3^{rd}$ of the heat load on the ITER superconducting magnet system.

AC Loss Model

The AC loss power discussed in the following is per unit volume of superconducting strand (pure Copper strands are not counted). To obtain the total power the power density needs to be multiplied by the number of superconducting strands, the strand cross-section and the length of conductor considered. Also to be noted is that the models for the two types of AC losses are somehow related, as the hysteresis loss model is needed to obtain the parameters of the couping loss from the conductor sample tests, as they are always appear together.

Input Parameters

The following therefore describes analytical approaches in which considerable simplification is achieved by using AC loss parameters measured in standardized tests on conductor- (for coupling loss) or strand- (for hysteresis loss) samples. This phenomenological approach allows to bypass the calculation of the complex in-situ current distributions present in the conductor. Further description of these input parameters as well as the way they are obtained is given and acquired parent to its conference. A Tore "Review of experimental results and models for AC losses in the ITER PF and CS conductors".

Coupling Loss The coupling loss power density per volume depends on the square of the local ("inner") field variation dB/dt, and the L/R (inductance over resistance represented by the time

insert tests as discussed later). The second equ. On the left gives the phenomenological fit used here.

Hysteresis Loss

 $p = \frac{n\tau}{\mu_0} \left(\frac{dB_i}{dt}\right)^2 \qquad \left[\frac{W}{m^3}\right]$

 $n\tau = n\tau_0 + \gamma \times B_e(T) + \lambda \left(B_e(T) \times I(kA) \right) [s]$

$$\begin{split} & f_{cnoncu(B_{\theta},T,\epsilon)} = f_{cnoncumag(B_{\theta})} \times \frac{f_{cnoncurrang(B_{\theta},T,\epsilon)}}{f_{cnoncurrang(B_{\theta},T_{\theta}-0.00022)}} \left[\frac{A}{m^2}\right] \\ & p_{hyst.full} = \frac{2}{3\pi(1+c_{UNCu})} f_{cnoncu(B_{\theta},T,\epsilon)} d_{eff} \frac{dB}{dt} = \frac{W}{m^2} \end{split}$$

 $p_{hyst,part} = \frac{\pi \frac{dB_c}{dt} \Delta B^2}{2(1 + CuNCu) d_0^2 k_f^2 a_{f,anoncu}(B_e, T, \varepsilon) d_{eff}} \left(1 - \frac{\pi \Delta B}{3 \mu_0 k_{fil} J_{c,noncu}(B_e, T, \varepsilon) d_{eff}}\right)$

applied field \mathcal{B}_{e} . The instantaneous loss power is then as in $p_{hytat} = \frac{1}{2}\Delta M(\overline{\mathcal{B}_{e}}, T_{e})\frac{\partial d_{e}}{dt}\left[\frac{W}{m^{2}}\right]$, where ΔW is the width of the hysteresis loop from up- to down-field branch (the factor ½ assumes the hysteresis loop is up(down symmetric). The practical approach chosen is to measure the magnetization loop of a single strand exposed to an application loop to calculate the hysteresis dos of a single strand exposed to an application loop to calculate the hysteresis dos of a calbe where it is assumed that the applied field and use this strands in the cable. Alternatively magnetization measurements are less common). As also before of the coupling loss the magnetization is the response to the internal field \mathcal{B}_{e} , which is not the same s the applied field β_{e} because of the shielding effect. To the single strand case it is relatively straightforward to estimate \mathcal{B}_{e} from \mathcal{B}_{e} .

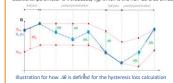
constant t) of the different coupling loops (first equ, on the left). The temporal behavior of B, the "internal" locally applied field after taking into account the shielding by the

eddy currents, is obtained from the externally applied field B_e(t) from the solution of the simple ODE B_i = B_e - τ * dB/dt. Here it is simply calculated from the previous B_i at time

The most straight forward way to calculate the hysteresis loss consists of integrating the magnetization M as a function of the

step (+.\t) by the finite differences method (but other more or less elaborate ways to calculate it exist). The externally applied field B (t) along the conductor inside the coils is obtained from numerical analysis with models implementing the entire magnet system (incl. the so-called passive structures) and plasma. The parameter describing the eddy current response of the CICC is represented by nr, which is typically obtained from AC loss tests on conductor samples as discussed in the companion paper noted above (or the comparison of the CICC is represented by nr, which is typically obtained from AC loss tests on conductor samples as discussed in the companion paper noted above (or the comparison of the CICC is represented by nr, which is typically obtained from AC loss tests on conductor samples as discussed in the companion paper noted above (or the comparison of the CICC is represented by nr, which is typically obtained from AC loss tests on conductor samples as discussed in the companion paper noted above (or the comparison of the CICC is represented by nr, which is typically obtained from AC loss tests on conductor samples as discussed in the companion paper noted above (or the comparison of the CICC is represented by nr, which is typically obtained from AC loss tests on conductor samples as discussed in the companion of the comparison of the c

Two susses complicate the computation of the hysteresis loss, however. The first is that the superconductor magnetization depends not only on magnetic field, but also on temperature 7 and in ND35n on strain a. Measurements of the magnetization for varying temperature and strain is experimentally complex. The work-around is provided by the homon dependence of the magnetization on J.*vd.g.*, the product of the critical current density and the "effective" filament diameter. J, is usually measured in strains as a function of 7 and a [and a], in a transport measurement setting (recording the super-to-monal-conduciting transition during a current tamp at field field). The transport type of measurement underestimates the J, at low field, however, because the higher transport current a) produces a non-registive field measurement is the field effective." File equation above gives the J, function finally used here, which combines the field dependence obtained with J, dependence



The second issue is that the magnetization response may not be fully developed, i.e. one is between the two branches of the magnetization loop (commonly referred to as the "partially penetrated" case). This condition is reached after a reversal of the sign of the *dg*/dt as long as *dg*, bias not allowed to reach the cohort side of the magnetization loop price by definition is 2*g*, wide with *B_i*_{*i*µ}/*i*_{*k*_{*j*}/*i*_{*i*}/*i*_{*k*}, *dg*, *f*_{*j*}/*i*_{*k*}, *f*_{*j*}, *f*_{*j*}, *b*, be not allowed the evolution of *G*, can be very complex due to the plasma partition tends due to rescan the use branches. This kines the dg/dt as allowed the dg/dt as allowed the evolution of *G*, can be very complex due to the plasma partition tendsads. (incl. noise) and the compared to fact and forth of the field produces very complex shielding current partment sing signs raised the sc s (ill ametry) and a simplification is needed here also. The key parameter which will be simplified is *AB*, which is the change of applied field accumulated since the last sign reveral of *CB*/dt. There the simplification the *AB* and *B_{aux}*, and *B_{aux}*, the first being the up-field branch and second the dwore branch. The difference *B_{aux}*, *B_{aux}*, *B_{aux}* effection. As a simplification the *AB* are averefined until the next sign reversal occurs. The formulas used for the fully operatially penetrated case is reached and *B_{aux}*, are redefined that the *AB* and *B_{aux}* and *B_{aux} and B_{aux} and <i>B_{aux}* and *B_{aux}* and *B_{aux} and B_{aux} and <i>B_{aux}* and *B_{aux}* and *B_{aux}* and *B_{aux} and B_{aux}* and *B_{aux}* and *B_{aux}* and *B_{aux} and B_{aux} and <i>B_{aux}* and *B_{aux}* and *B_{aux}* and *B_{aux} and <i>B_{aux}* and *B_{aux}* and *B_{aux} and <i>B_{aux}* and *B_{aux}*}

Rotating Field

$$\begin{split} &B_{e,mod}(t) = \left|\overline{B}_{e}\right| = \sqrt{B_{e,r}(t)^{2} + B_{e,s}(t)^{2}} \quad [T] \\ &B_{e,rot}(t) = \sum dB_{e,rot}(t) = \sum \overline{B_{e,mod}(t)} \; \partial \; \alpha(t) \quad [T] \\ &\partial \; \alpha(t) = \alpha \cos\left(\frac{B_{e,t}(t)}{B_{e,t}(t)B_{e,t}(-\alpha t)}\right) \quad [rad] \end{split}$$

For the case of the ITER CS (and PF) coils the applied field can be decomposed into the two projections perpendicular to conductor, i.e. vertical and radial (there is no field component along the conductor). It is assumed here that the coupling loss can be calculated separately for the two components and added. This assumes that the shielding currents excited in the two planes do not interfere with each other and the respective applied fields. In the case of the hysteresis loss this assumption does not hold, however, because the shielding current distribution in the superconducting filaments will always arrange in such away as to cancel the applied field, i.e. ir rotates with the field in the case of a rotating field. Assuming that the loss is driven by the applied field modulus would generate a hysteresis loss this to overstimated. A good example is the case of a purely rotating field. Assuming that the loss is driven by the applied field modulus would generate a hysteresis loss that is rotated and applied her approach, which strictly speaking applies only to anil rotations, projects the field onto a non-rotating and a rotating component. Only the rotating component produces loss in purely rotating field. The components of this applied field decomposition for which the separately domputed hysteresis loss has a rotating component. Only the rotating component produces loss in purely rotating field. The components is a diven to this spice field field decomposition for which the separately computed hysteresis losses are added are a given on the left. They are **Equilary and B**₀₀m. The case is no accurate the spice is a divent on the left. They are **Equilary and B**₀₀m. The rotating field field composition for which the separately computed hysteresis losses are added are a given on the left. They are **Equilary and B**₀₀m. The rotate field is a divent on the left. They are **Equilary and B**₀₀m. The rotate field field the separately applied field the dis a spice on the spice field field the divention of t

AC Loss General Explanation

Coupling Loss:

The coupling losses are the result of induced currents flowing between filaments in the multi-filamentary strands, and between strands within various cable stages as well as between sub-cables. The loss causing coupling currents partially flow inside superconductor, which enhances their time constants and thus their strength as compared to a normal conductor*. Therefore, in pulsed superconducting magnets as in ITER, all means possible to suppress these coupling currents are applied. This includes twisting of the filaments, strands and sub-cables to cut down on the inductance and increase the contact resistance with resistive coating on the strands (Cr or Ni). Finally stainless steel foil wraps over the sub-cables are also effective. But the effect cannot be suppressed entirely, as it would lead to electrically insulated strands (or filaments), preventing the transfer of currents from saturated strands and can thus lead to degraded cable performance and quench margin.

Hysteresis Loss:

Hysteresis (or "magnetization") loss is caused by the diamagnetic shielding effect inside the superconductor in which persistent currents are induced to expel the applied magnetic field. The loss causing mechanism is the re-arrangement of screening currents and pinned magnetic field vortices due to a change of applied field. The hysteresis loss depends on the so called "effective" filament diameter and the critical current density in the superconductor. The reduced filament sizes typical for LTS type strands (order 10 microns) cannot fully suppress this loss as the filaments are coupled through the Cu matrix (needed for electro-thermal stability of the strands) and thus behave like a larger filament. The critical current density should obviously be high enough to allow a large transport current to limit the superconductor cost. Since the superconductor is designed for the highest field region it can be over-designed for the low field regions in non-graded coil designs (as in ITER), additionally enhancing the hysteresis loss. *but then at comparable performance the superconducting cable have much smaller cross-

section, the AC loss in the superconductor, however, occurs at low temperature for which a penalty applies due to the limited Carnot efficiency.

Summary

- Simple analytical AC loss models are needed for fast turn-around calculations for the large superconducting coils in the ITER device.
- Once validated on conductor and coil measurements such models can be used to simulate the AC loss heating inside the superconducting coils for the ITER device during the plasma pulses.
- Such models have been presented here for the CS type ITER coils. In these coils the magnetic field transients are always transverse to the conductor, resulting in the simplest possible models.

